



mitsubishi 1987 **SEMICONDUCTORS**

GENERAL PURPOSE ICs

OPERATIONAL AMPLIFIER

COMPARATOR

VOLTAGE REGULATOR

TIMER

MOTOR DRIVER

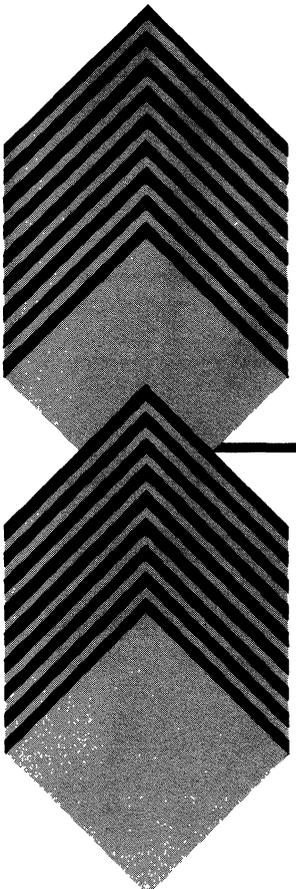
ANALOG SWITCHES

LEVEL INDICATOR

TRANSISTOR ARRAY

MISCELLANEOUS

**DATA
BOOK**



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All values shown in this catalogue are subject to change for product improvement.

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OPERATIONAL AMPLIFIER, PRE AMPLIFIER

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended [Features]
Single op amp	741	M51802P/M5F741P	●M51802L/P [Single in line package]
Dual op amp	4557, 4558, 4559	M5218P/M5R4558P	●M5218L [Compatible with 4557, 4558, 4559] [Single in line package]
	358/2904	M5223P/M5N358P	●M5223L [Single in line package]
	324/2902	M5224P/M5N324P	●M5216P/L [High current output] [equivalent to 4556]
J.FET Input op amp	072, 082	M5221P/M5T082P	●M5221L [Single in line package]
Switching op amp			△M5201
Dual High Voltage op amp			△M5209
			△M5210
Dual Low noise pre-amp			△M5219
Quad low noise op amp			△M5220
			△M5228
Dual low noise J-FET Input op amp			△M5238
			△M5240P

Note : △ Mitsubishi original, ● Functional equivalent

VOLTAGE COMPARATOR

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended [Features]
Single comparator			△M51201L [Low input current high output drive, including reference with hysteresis, V _{CC1} 7~6.5V]
			△M51202L [Low input current high output drive, V _{CC1} 7~6.5V]
			△M51203L [Low input current high output drive, V _{CC3} ~28V including reference with hysteresis]
			△M51204L [Low input current high output drive, V _{CC} 2.5~28V]
			△M51205L [Low input current stabilized power supply terminal including reference with hysteresis]
			△M51206L [Low input current stabilized power supply terminal]
			△M5123TL
Dual comparator			△M51200P
			△M51207L
			△M51922L
			△M51923P/FP
			△M5233
Quad comparator			△M51209P
			△M51924P/FP
			△M5234

Note : △ Mitsubishi original, ● Functional equivalent

VOLTAGE REGULATOR

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended (Features)
Variable Dual			△M5230L
Variable Single			△M5231TL
3-Terminal Variable			△M5235L
			△M5236
	78L	M5278LXX	
SW Control	494	M5T494P	

Note : △ Mitsubishi original, ● Functional equivalent

DISPLAY DRIVER

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended (Features)
5-Step bar indicator			△M51903L [Linear mode operation]
6-Step bar indicator			△M51906P [Including AC-DC converter log scale]
			△M51911L [Including AC-DC converter low current log scale]
			△M51912L [Including AC-DC converter low current linear scale]
8-Step bar indicator			△M51907P [Including AC-DC converter low current log scale]
			△M51909P [Including AC-DC converter low current linear scale]
9-Point level indicator			△M51910P [2 inputs indicator]
12-Point level indicator			△M51901P [23-Mode indicator]

Note : △ Mitsubishi original, ● Functional equivalent

TIMER

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended (Features)
Single timer	555	M51841P/M5E555P	● M51848P/M5E555AP [High speed guaranty reset terminal TTL compatible]
			● M51848L [High speed guaranty reset terminal TTL compatible single in line package]
			△ M51843P [Large supply voltage application]
Dual timer	556		● M51847P/M5E556AP [High speed guaranty reset terminal TTL compatible]
Counter timer			△ M51849L [Long time capability~50hours]
			△ M52051P
			△ M58479P [Low power dissipation superior noise immunity, extremely broad time-delay range (50ms~480h)]
			△ M58482P [Low power dissipation superior noise immunity, extremely broad time-delay range (50ms~480h)]
Long timer			△ M51845L

Note : △ Mitsubishi original, ● Functional equivalent

MOTOR DRIVER

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended (Features)
Motor control			△ M51660L [Radio control servo]
Position control			
F-G Servo			△ M51970L [Single in line]
PLL Servo			△ M51728L [High stability single in line]
			△ M51971L,FP
2-Phase brushless			△ M51781SP
			△ M51781FP
3-Phase brushless			△ M51724P
			△ M51714P
Bi-directional driver			△ M54540AL [$I_{o(max)}/I_o \pm 0.6A/\pm 0.1A$]
			△ M54541L [$I_{o(max)}/I_o \pm 0.8A/\pm 0.2A$]
			△ M54542L [$I_{o(max)}/I_o \pm 1.2A/\pm 0.3A$]
			△ M54543L [$I_{o(max)}/I_o \pm 1.2A/\pm 0.3A$ Motor braking function]
			△ M54544L [$I_{o(max)}/I_o \pm 1.2A/\pm 0.3A$ Motor braking function]
			△ M54545L [$I_{o(max)}/I_o \pm 1.2A/\pm 0.2A$ Motor braking function]
			△ M54546L [$I_{o(max)}/I_o \pm 0.7A/\pm 0.15A$ Motor braking function]
			△ M54547P [$I_{o(max)}/I_o \pm 0.6A/\pm 0.15A$ Built in error amp, motor braking function]
			△ M54548L [$I_{o(max)}/I_o \pm 1.2A/\pm 0.3A$ Motor braking function op amp and decoder]
			△ M54549L [$I_{o(max)}/I_o \pm 1.2A/\pm 0.3A$ Motor braking function dual bi-directional, thermal shutdown]

Note : △ Mitsubishi original, ● Functional equivalent

MOTOR DRIVER

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended (Features)
Bi-directional driver			△M54543AL $\left[\begin{array}{l} I_{Q(\max)}/I_Q \pm 1.2A/\pm 0.3A \\ \text{Motor braking function} \\ \text{thermal shutdown} \end{array} \right]$
			△M54544AL $\left[\begin{array}{l} I_{Q(\max)}/I_Q \pm 1.2A/\pm 0.3A \\ \text{Motor braking function} \\ \text{thermal shutdown} \end{array} \right]$
			△M54549AL
			△M54641L
			△M54642L
			△M54648L
			△M54648AL
			△M54649L
			△M51714P
F-V Servo			△M51722P,FP
			△M51723P,FP

Note : △ Mitsubishi original, ● Functional equivalent

ANALOG SWITCH

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended (Features)
Analog SW			△M51320P [Video SW, 2 inputs 3ch]
Bipolar SW			△M51326P [Video SW, 2 inputs 3ch]
			△M51321P [Video SW, 3 inputs 3ch]
			△M51327P [Video SW, 3 inputs 3ch]
			△M51329P
			△M51551P/FP [Audio SW, 2 inputs 2ch]
CMOS SW	4016B	M4016BP	
	4066B	M4066BP	
	4051B	M4051BP	
	4052B	M4052BP	
	4053B	M4053BP	

Note : △ Mitsubishi original, ● Functional equivalent

TRANSISTOR ARRAY

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended [Features]	
4-Unit Tr array	2064		△M54512L [50mA Single in line package]	
			△54532P [1.5A darlington, with clamp diode]	
			△M54567P [1.5A darlington with clamp diode] Input "L" active.	
			△M54568L [30mA PNP single in line package]	
5-Unit Tr array			△M54516P [500mA darlington]	
			△M54521P [500mA darlington]	
			△M54529P/AP [320mA with strobe]	
6-Unit Tr array			△M54527P [150mA with clamp diode]	
			△M54533P [320mA with clamp diode and strobe]	
			△M54534P [320mA with clamp diode and strobe]	
			△M54578P [700mA with clamp diode and strobe]	
			△M54539P [700mA with clamp diode]	
			△M54571P [350mA printer driver]	
7-Unit Tr array			△M54514AP [50mA NPN driver]	
			△M54515P [16mA NPN driver]	
			△M54517P [400mA darlington]	
			△M54519P [400mA darlington]	
	2003	●M54523P		
	2001	●M54524P		
	2002	●M54525P		
	2004	●M54526P		
				△M54528P [150mA darlington with clamp]
				△M54530P [400mA darlington with clamp]
				△M54531P [400mA darlington with clamp]
				△M54535P [150mA with clamp and strobe]
				△M54536P [150mA with clamp and strobe]
△M54537P [320mA non-darlington driver]				
△M54538P [350mA and motor driver]				
△M54560P [150mA source darlington]				
△M54561P [300mA source darlington]				
△M54566P [400mA darlington] Input "L" active				

Note : △ Mitsubishi original, ● Functional equivalent

TRANSISTOR ARRAY

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended (Features)
7-Unit Tr array			△M54576P/FP [30mA input "L" active]
			△M54577P/FP [30mA non-darlington driver]
			△M54580P [100mA source darlington for FL]
8-Unit Tr array			△M54513P [50mA NPN driver]
			△M54522P [400mA darlington with clamp diode]
	2982	●M54562P	
	2981	●M54563P	
			△M54564P [500mA source darlington for FL]
			△M54565P [50mA input "L" active]
			△M54569P [30mA PNP array]
			△M54581P
			△M54583P
			△M54584P
			△M54585P
		△M54586P	

Note : △ Mitsubishi original, ● Functional equivalent

MISCELLANEOUS

Function	Generic No.	Mitsubishi Pin For Pin Replacement	Mitsubishi Recommended [Features]
Electronic Volume Control			△M5222
			△M5241L
Voltage Defector			△M5232L
Volume Control			△M50601P
Disk Drive			△M51017AP
Servo Motor Position Control			△M51660L
Dual Channel Electronic ATT			△M51523AL
Zero volt trigger ckt Flame defector			△M5172L
			△M51743P
			△M5174P
Warning driver			△M51920P
Car cooler control			△M51950L
Flasher control			△M51961L
4-BIT A-D Converter			△M52670P,FP
Dual differential amp	3054	M5109P	
8-Digit display driver			△M54940P
2-Digit BCD-7 segment Decoder/driver			△M54847AP
8-BIT Latched driver			△M54975P
			△M54976P
9-BIT Latched driver			△M54970P
12-BIT Latched driver			△M54977P
Single chip CMOS speech synthesizer			△M50800-XXXSP/M50802-XXXXP
			△M50805-XXXXP,FP/M50806-XXXXP,FP

Note : △ Mitsubishi original, ● Functional equivalent

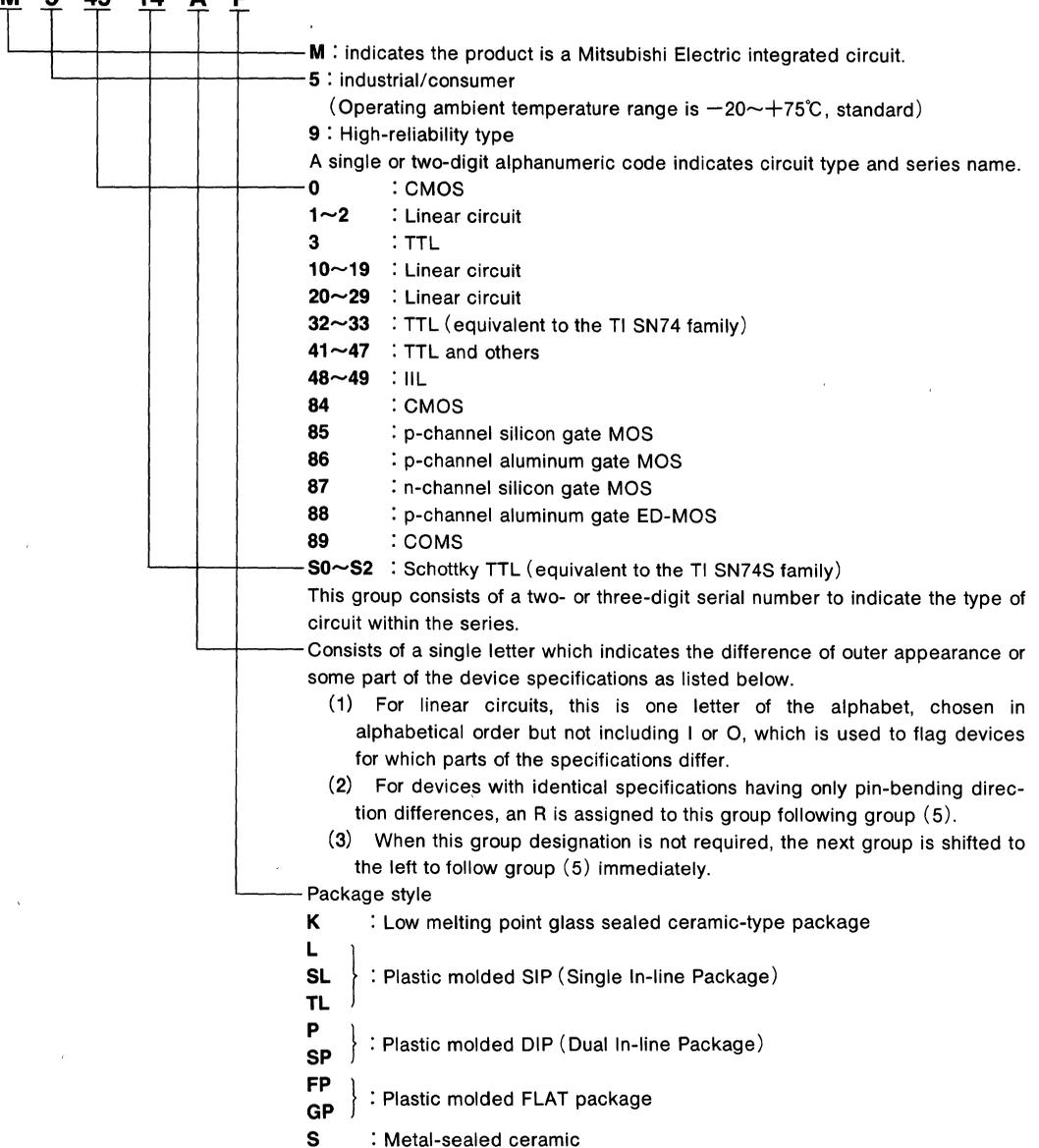
MITSUBISHI GENERAL PURPOSE ICs ORDERING INFORMATION

ORDERING INFORMATION

Function and style may be specified by using the following simplified alphanumeric code.

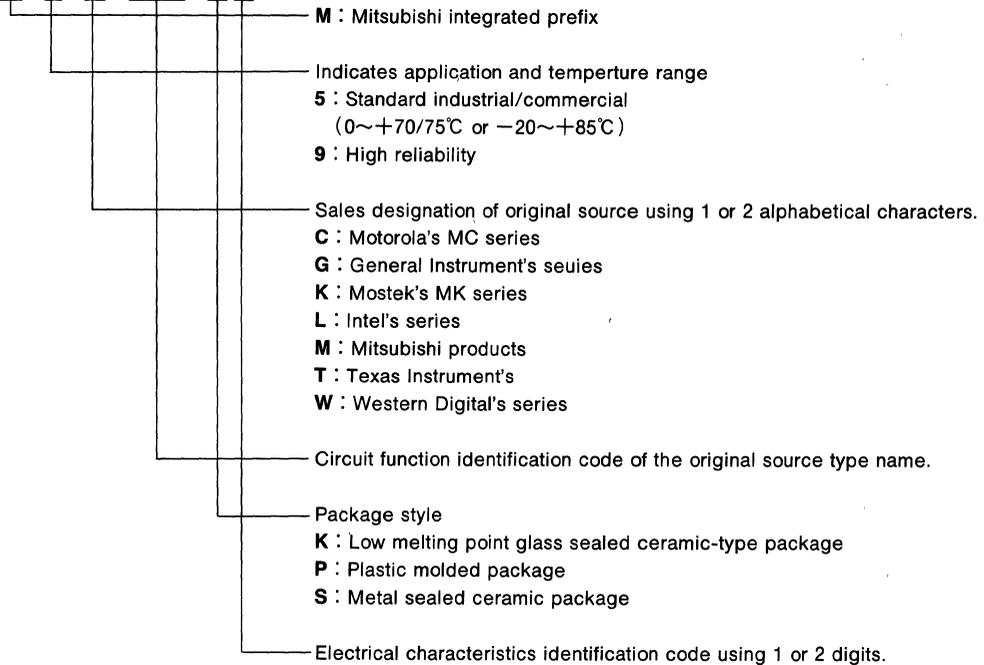
1. FOR MITSUBISHI ORIGINAL PRODUCTS

Example : **M 5 45 14 A P**



2. FOR SECOND SOURCE PRODUCTS

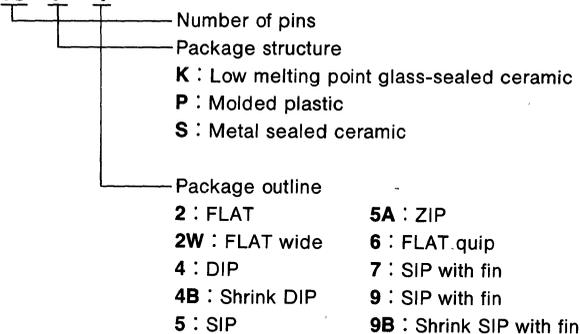
Example : **M 5 K 4116 S-2**



PACKAGE CODE

Package style may be specified by using the following simplified alphanumeric code

Example : **16 P 4**



QUALITY ASSURANCE AND RELIABILITY TESTING

1 INTRODUCTION

IC & LSI have made rapid technical progress in electrical performances of high integration, high speed, and sophisticated functionality. And now they have got boundless wider applications in electronic systems and electrical appliances.

To meet the above trend of expanding utilization of IC & LSI, Mitsubishi considers that it is extremely important to supply stable quality and high reliable products to customers.

Mitsubishi Electric places great emphasis on quality as a basic policy "Quality First", and has striven always to improve quality and reliability.

Mitsubishi has already developed the Quality Assurance System covering design, manufacturing, inventory and delivery for IC & LSI, and has supplied highly reliable products to customers for many years. The following articles describe the Quality Assurance System and examples of reliability control for Mitsubishi General Purpose ICs.

2. QUALITY ASSURANCE SYSTEM

The Quality Assurance System places emphasis on built-in reliability in designing and built-in quality in manufacturing. The System from development to delivery is summarized in Figure 1.

2.1 Quality Assurance in Designing

The following steps are applied in designing stage for new product.

- (1) Setting of performance, quality and reliability target for new product.
- (2) Discussion of performance and quality for circuit design, device structure, process, material and package.
- (3) Verification of design by CAD system to meet standardized design rule.
- (4) Functional evaluation for bread-board device to confirm electrical performance.
- (5) Reliability evaluation for TEG (Test Element Group) chip to detect basic failure mode and investigate failure mechanism.
- (6) Reliability test (In-house qualification) for new product to confirm quality and reliability target.
- (7) Decision of pre-production from the standpoint of performance, reliability, production flow/conditions, production capability, delivery and etc.

2.2 Quality Assurance in Manufacturing

The quality assurance in manufacturing are performed as follows.

- (1) Environment control such as temperature, humidity and dust as well as deionized water and utility gases.
- (2) Maintenance and calibration control for automatized manufacturing equipments, automatic testing equipments, and measuring instruments.
- (3) Material control such as silicon wafer, lead frame, packaging material, mask and chemicals.

- (4) In-process inspections in wafer-fabrication, assembly and testing.
- (5) 100% final inspection of electrical characteristics, visual inspection and burn-in, if necessary.
- (6) Quality assurance test
 - Electrical characteristics and visual inspection, lot by lot sampling
 - Environment and endurance test, periodical sampling.
- (7) Inventory and shipping control, such as storage environment, date code identification, handling and ESD (Electro Static Discharge) preventive procedure.

2.3 Reliability Test

To verify the reliability of a product as described in the Mitsubishi Quality Assurance System, reliability tests are performed at three different stages of new product development, pre-production, and mass-production.

At the development of new product the reliability test plan is fixed corresponding to quality and reliability target of each product, respectively. The test plan includes in-house qualification test, and TEG evaluation, if necessary. TEG chips are designed and prepared for new device structure, new process and new material.

After the proto-type product has passed the in-house qualification test, the product advances to the pre-production. In the pre-production stage. The specific reliability tests are programmed and performed again to verify the quality of pre-production product.

In the mass production, the reliability tests are performed periodically to confirm the quality of mass production product according to quality assurance test program.

Table 1 shows an example of reliability test program for plastic encapsulated IC & LSI.

Table 1 TYPICAL RELIABILITY TEST PROGRAM FOR PLASTIC ENCAPSULATED IC & LSI

Group	Test	Test condition
1	Solderability	230°C, 5sec Rosin flux
	Soldering heat	260°C, 10sec.
2	Thermal shock	-55°C, 125°C, 15cycles
	Temperature cycling	-65°C, 150°C, 100cycles
3	Lead fatigue	250gr, 90°, 2arcs
	Shock	1500G, 0.5msec.
4	Vibration	20G, 100~2000Hz X, Y, Z direction 4min./cycle, 4cycles/direction
	Constant acceleration	20000G, Y direction, 1min
5	Steady-state operation life	T _a =T _{oprmax} , V _{ccmax} 1000hours
6	High temperature storage life	T _a =150°C, 1000hours
7	High temperature and high humidity	85°C, 85%, 1000hours
	Pressure cooker	121°C, 100%, 100hours

QUALITY ASSURANCE AND RELIABILITY TESTING

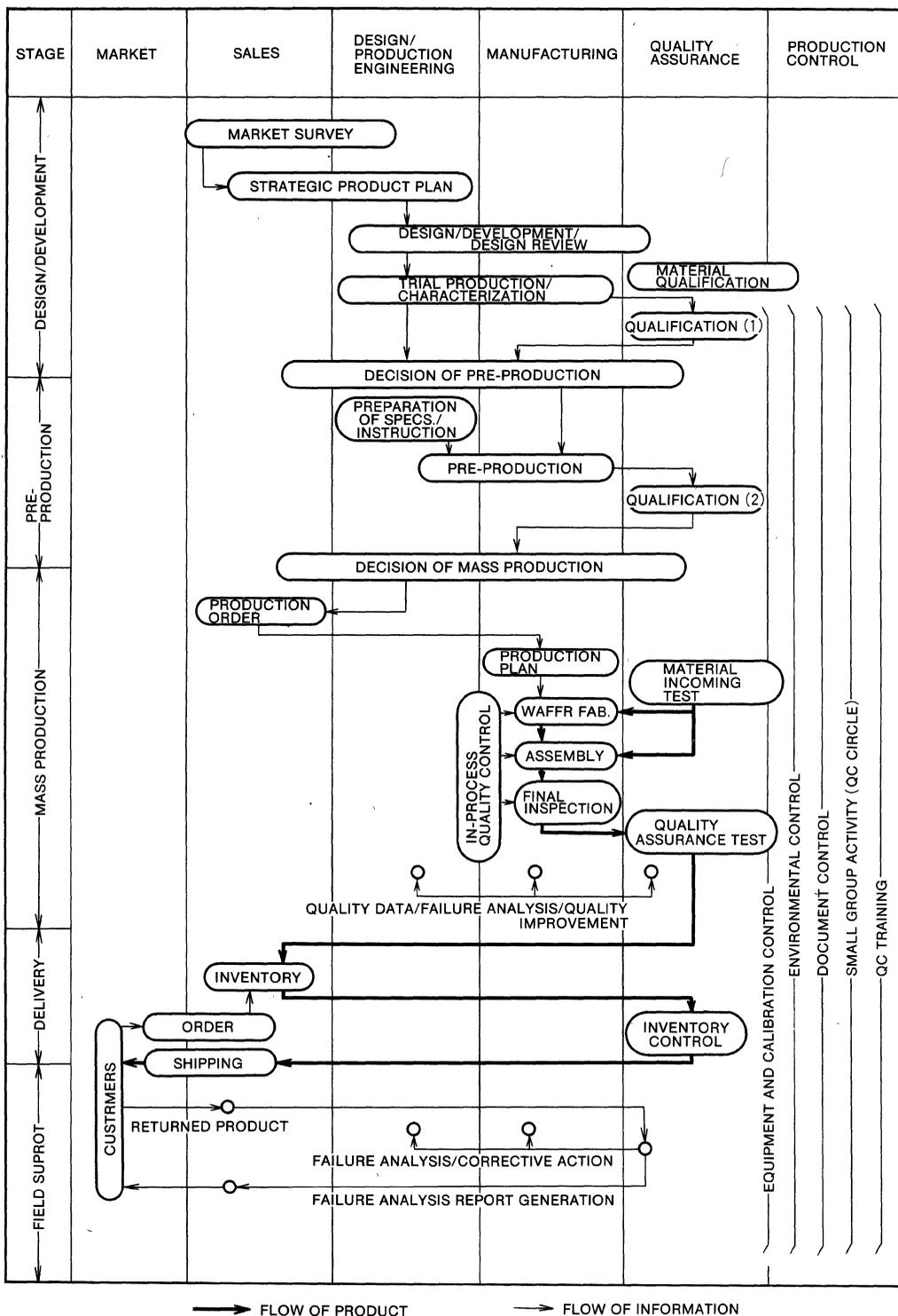


Fig.1 FLOW CHART OF QUALITY ASSURANCE SYSTEM

QUALITY ASSURANCE AND RELIABILITY TESTING

2.4 Returned Product Control

When failure analysis is requested by a customer, the failed devices are returned to Mitsubishi Electric via the sales office of Mitsubishi using the form of "Analysis Request of Returned Product"

Mitsubishi provides various failure analysis equipments to analyze the returned product. A failure analysis report is

generated to the customer upon completion of the analysis. Failure analysis result enforces to take corrective action for the design, fabrication, assembly or testing of the product to improve reliability and realize lower failure rate.

Figure 2 shows the procedure of returned product control from customer.

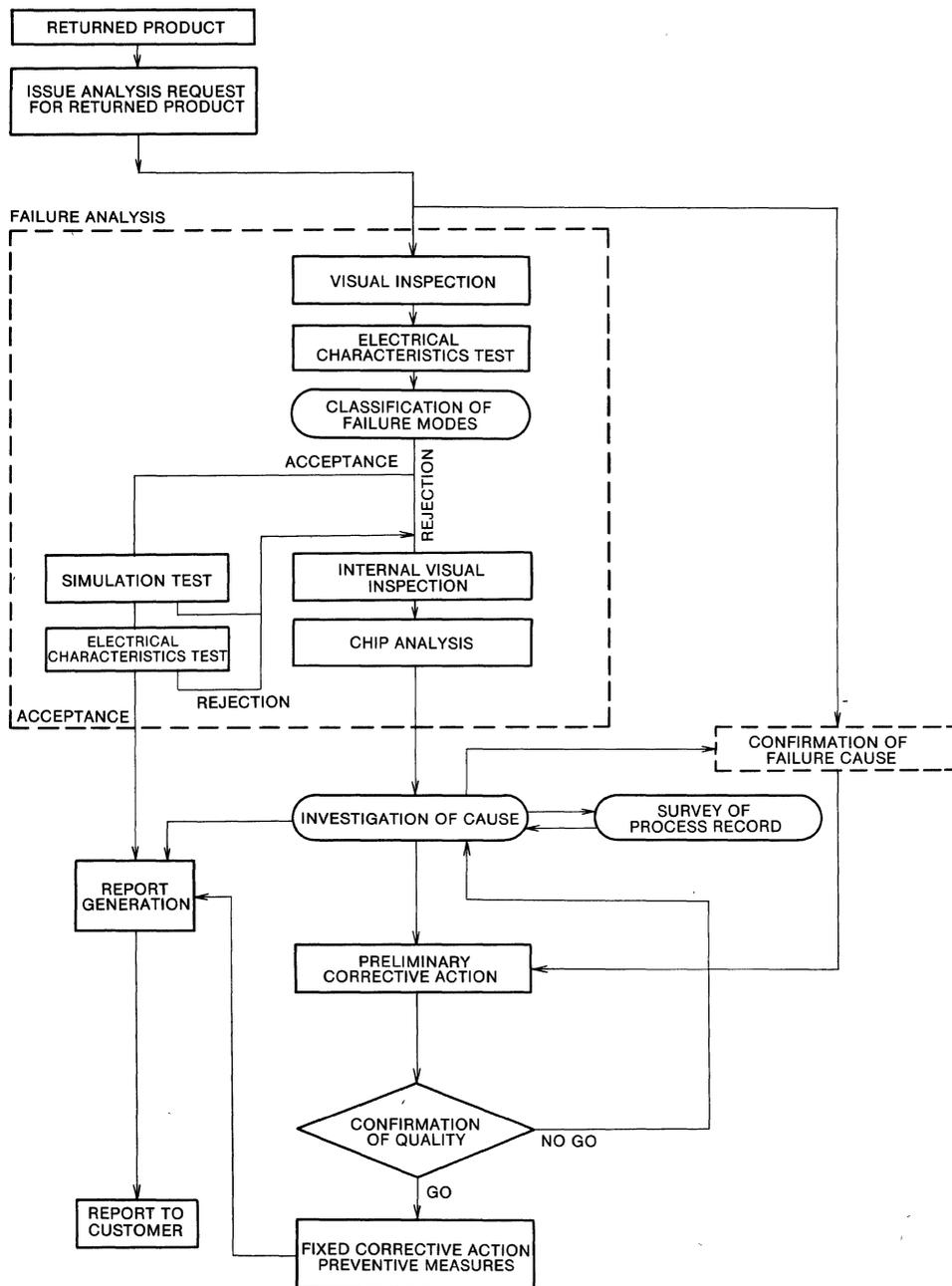


Fig.2 PROCEDURE OF RETURNED PRODUCT CONTROL

QUALITY ASSURANCE AND RELIABILITY TESTING

3 RELIABILITY TEST RESULTS

The reliability test results for Mitsubishi General Purpose ICs are shown in Table 2, Table 3 and Table 4.

Table 2 shows the result of endurance tests of steady-state operation life and high temperature storage life test for representative types of General Purpose ICs, Operational Amplifier, Voltage Comparator, Timer, Motor Driver, Voltage Regulator and Transistor Array. From Table 2, the combined

failure rate of Mitsubishi General Purpose ICs is calculated 0.045%/1000hours (60% confidence level) at maximum rating of operating condition.

Table 3 shows the result of environment test of temperature cycling, high temperature/high humidity and pressure cooker test for the same type of products as of endurance tests.

Table 4 shows the results of mechanical tests for representative products of various package types.

Table 2 ENDURANCE TEST RESULTS

Test		Steady-State Operation Life			High Temperature Storage Life		
Test Condition		$T_a = T_{oprmax}, V_{cc} \leq V_{ccmax}$			$T_a = 150^\circ\text{C}$		
Application	Type Number	Number of Samples	Device Hours	Number of Failures	Number of Samples	Device Hours	Number of Failures
Operational Amplifier	M5216P	22	22,000	0	22	22,000	0
	M5216L	22	22,000	0	22	22,000	0
	M5218P	66	66,000	0	44	44,000	0
	M5218L	66	66,000	0	44	44,000	0
	M5223P	44	44,000	0	22	22,000	0
	M5223L	44	44,000	0	22	22,000	0
	M5224P	22	22,000	0	22	44,000	0
	M5228P	44	44,000	0	22	44,000	0
	M5209P	22	22,000	0	22	22,000	0
	M5209L	22	22,000	0	22	22,000	0
	M5210P	44	44,000	0	22	22,000	0
	M5210L	22	22,000	0	22	22,000	0
	M5219P	44	44,000	0	44	44,000	0
	M5219L	44	44,000	0	22	22,000	0
	M5220P	22	22,000	0	22	22,000	0
	M5220L	22	22,000	0	22	22,000	0
	M5221P	22	22,000	0	22	22,000	0
	M5221L	22	22,000	0	22	22,000	0
	M5238P	22	22,000	0	22	22,000	0
	M5238L	22	22,000	0	22	22,000	0
M5240P	22	22,000	0	22	22,000	0	
M5201P	22	22,000	0	22	22,000	0	
M5201L	22	22,000	0	22	22,000	0	
M51802P	48	48,000	0	50	50,000	0	
M51802L	48	48,000	0	50	50,000	0	
Voltage Comparator	M5233P	44	44,000	0	44	44,000	0
	M5233L	44	44,000	0	22	22,000	0
	M5234P	22	22,000	0	22	22,000	0
	M51204TL	66	132,000	0	50	50,000	0
	M51209P	22	22,000	0	22	22,000	0
Timer	M51848P	70	158,000	0	22	22,000	0
	M51848L	70	158,000	0	22	22,000	0
Motor Driver	M51970L	22	22,000	0	22	22,000	0
	M54543L	44	88,000	0	22	44,000	0
Voltage Regulator	M5230L	66	66,000	0	44	44,000	0
	M5231TL	40	40,000	0	22	22,000	0
	M5236L	44	44,000	0	22	22,000	0
	M5278L	44	44,000	0	44	44,000	0
Transistor Array	M54519P	40	40,000	0	44	44,000	0
	M54523P	44	88,000	0	22	44,000	0
LED Level Indicator	M51903L	22	22,000	0	22	22,000	0
	M51906P	22	22,000	0	22	22,000	0
Voltage Control Amplifier	M5222P	22	22,000	0	22	22,000	0
	M5222L	22	22,000	0	22	22,000	0
	M5241L	22	22,000	0	22	22,000	0
Voltage Detector	M5232L	22	22,000	0	22	22,000	0

QUALITY ASSURANCE AND RELIABILITY TESTING

Table 3 ENVIRONMENTAL TEST RESULTS

Application	Type Number	Soldering Heat Thermal Shock Temperature Cycling		High Temperature/High Humidity		Pressure Cooker	
		260°C, 10sec		85°C, 85% 1000hours		121°C, 100%RH 240hours	
		-55°C, 125°C, 15cycles -65°C, 150°C, 100cycles					
		Number of Samples	Number of Failures	Number of Samples	Number of Failures	Number of Samples	Number of Failures
Operational Amplifier	M5216P	22	0	22	0	22	0
	M5216L	22	0	22	0	22	0
	M5218P	44	0	66	0	44	0
	M5218L	44	0	44	0	44	0
	M5223P	22	0	22	0	22	0
	M5223L	22	0	22	0	22	0
	M5224P	22	0	22	0	22	0
	M5228P	22	0	22	0	22	0
	M5209P	22	0	22	0	22	0
	M5209L	22	0	22	0	22	0
	M5210P	22	0	22	0	22	0
	M5210L	22	0	22	0	22	0
	M5219P	22	0	44	0	22	0
	M5219L	22	0	22	0	22	0
	M5220P	22	0	22	0	22	0
	M5220L	22	0	22	0	22	0
	M5221P	22	0	22	0	22	0
	M5221L	22	0	22	0	22	0
	M5238P	22	0	22	0	22	0
	M5238L	22	0	22	0	22	0
M5240P	22	0	22	0	22	0	
M5201P	22	0	22	0	22	0	
M5201L	22	0	22	0	22	0	
M51802P	22	0	44	0	22	0	
M51802L	22	0	44	0	22	0	
Voltage Comparator	M5233P	44	0	44	0	22	0
	M5233L	22	0	22	0	22	0
	M5234P	22	0	22	0	22	0
	M51204TL	22	0	44	0	22	0
	M51209P	22	0	22	0	22	0
Timer	M51848P	22	0	44	0	22	0
	M51848L	22	0	44	0	22	0
Motor Driver	M51970L	22	0	22	0	22	0
	M54543L	22	0	44	0	22	0
Voltage Regulator	M5230L	44	0	44	0	22	0
	M5231TL	22	0	40	0	22	0
	M5236L	22	0	22	0	22	0
	M5278L	22	0	44	0	22	0
Transistor Array	M54519P	22	0	22	0	22	0
	M54523P	22	0	44	0	22	0
LED Level Indicator	M51903L	22	0	22	0	22	0
	M51906P	22	0	22	0	22	0
Voltage Control Amplifier	M5222P	22	0	22	0	22	0
	M5222L	22	0	22	0	22	0
	M5241L	22	0	22	0	22	0
Voltage Detector	M5232L	22	0	44	0	22	0

QUALITY ASSURANCE AND RELIABILITY TESTING

Table 4 MECHANICAL TEST RESULTS

Package Pin Count	Test Condition	Solderability		Lead Fatigue		Shock Vibration Constant Acceleration	
		See Table 1		See Table 1		See Table 1	
	Type Number	Number of Samples	Number of Failures	Number of Samples	Number of Failures	Number of Samples	Number of Failures
5P5	M51204TL	22	0	15	0	22	0
8P4	M5218P	44	0	15	0	0	0
	M51802P	22	0	15	0	22	0
8P5	M5218L	44	0	15	0	0	0
	M51848L	22	0	15	0	22	0
9P9	M54543L	22	0	15	0	22	0
10P5	M5241L	22	0	15	0	0	0
14P4	M5224P	44	0	15	0	0	0
	M51906P	22	0	15	0	22	0
16P4	M5240P	44	0	15	0	0	0
	M54519P	22	0	15	0	22	0
TO-92L	M5278L	44	0	15	0	22	0

4 FAILURE ANALYSIS

Accelerated reliability tests are applied to observe failures caused by temperature, voltage, humidity, current, mechanical stress and those combined stresses on chips and packages.

Examples of typical failure modes are shown below.

(1) Wire Bonding Failure by Thermal Stress

Figure 3, Figure 4 and Figure 5 are example of a failure occurred by temperature storage test of 225°C , 1000hours.

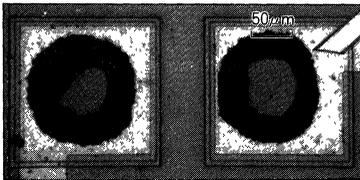


Fig.3 Micrograph of lifted Au ball trace on Al bonding pad

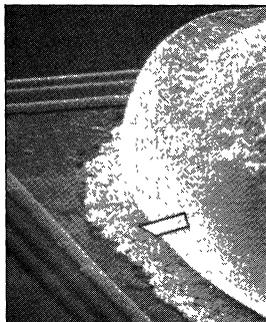


Fig.4 Au-Al plague formation on bonding pad

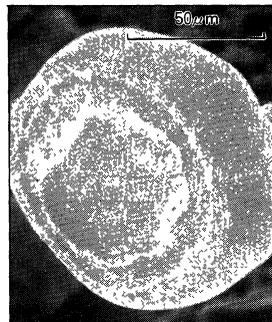


Fig.5 Lifted Au wire ball base

Au-Al intermetallic formation so-called "Purple plague" by thermal overstress makes Au wire lift off from aluminum metallization. The activation energy of this failure mode is estimated approximately 1.0eV and no failure has been observed so far in practical uses.

(2) Aluminum Corrosion Failure by Temperature/Humidity Stress.

Figure 6, Figure 7 and Figure 8 are an example of corroded failure of aluminum metallization of plastic encapsulated IC after accelerated temperature/humidity storage test (pressure cooker test) of 121°C , 100%RH, 1000hours duration.

Aluminum bonding pad is dissolved by penetrated water from plastic package, and chlorine concentration is observed on corroded aluminum bonding pad as shown in Figure 8.

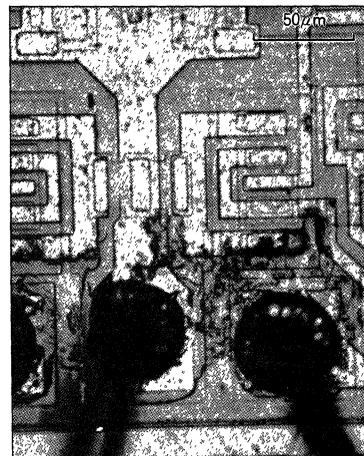


Fig.6 Micrograph of corroded Aluminum metallization

QUALITY ASSURANCE AND RELIABILITY TESTING

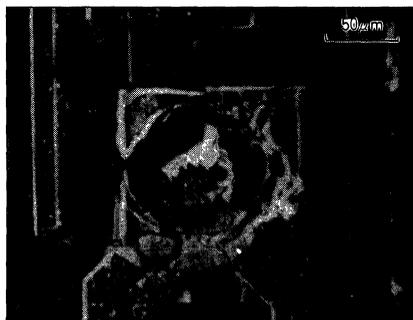


Fig.7
Enlarged micrograph of corroded Aluminum bonding pad



Fig.8
Cl distribution on corroded Aluminum bonding pad

(3) Destructive Failure by Electrical Overstress
Surge voltage marginal tests have been performed to reproduce the electrical overstress failure in field uses. Figure 9 and Figure 10 are an example of failure observed by surge voltage test. The trace of destruction is verified as the aluminum bridge by X ray micro analysis.

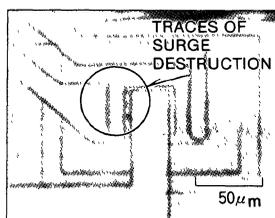


Fig.9
Micrograph of surge voltage destruction

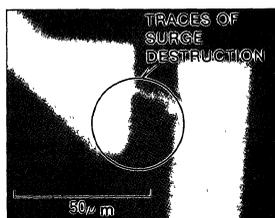


Fig.10
Aluminum trace of destructive spot

(4) Aluminum Electromigration
Figure 11 shows an open circuit of aluminum metallization in high current density region caused by accelerated operation life test. This failure is due to aluminum electromigration. Voids and hillock have been formed in aluminum metallization by high current density operation.

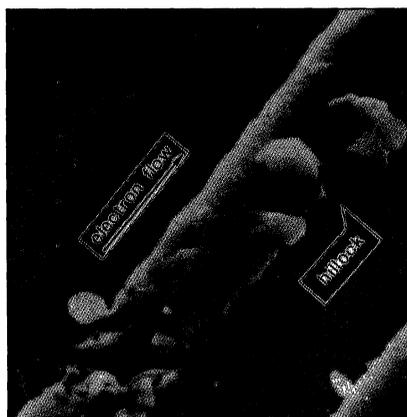


Fig.11
Voids and hillocks formation by Aluminum electromigration

5 SUMMARY

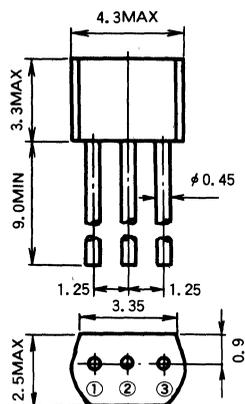
The Mitsubishi quality assurance system and examples of reliability control have been discussed. Customer's interest and requirement for high reliable IC & LSI are increasing significantly. To satisfy customer's expectancy, Mitsubishi as an IC vendor, would like to make perpetual efforts in the following areas.

- (1) Emphasis on built-in reliability at design stage and reliability evaluation to investigate latent failure modes and acceleration factors.
- (2) Execution of periodical endurance, environment and mechanical test to verify reliability target and realize higher reliability.
- (3) Focus on development of advanced failure analysis techniques. Detail failure analysis, intensive corrective action, and quick response to customer's analysis request.
- (4) Collection of customer's quality data in qualification, incoming inspection, production and field use to improve PPM, fraction defective and FIT, failure rate.

Mitsubishi would highly appreciate if the customer would provide quality and reliability data of incoming inspection or field failure rate essential to verify and improve the quality/reliability of IC & LSI.

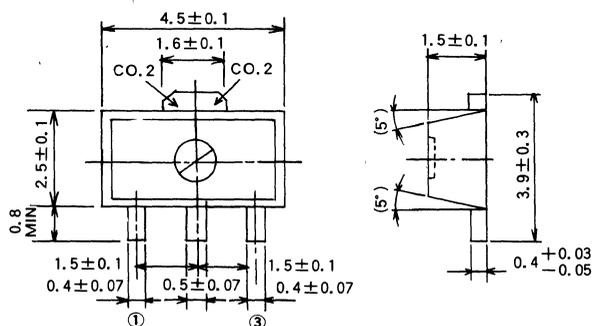
TYPE 3P5 3-PIN MOLDED PLASTIC SIP

Dimension in mm



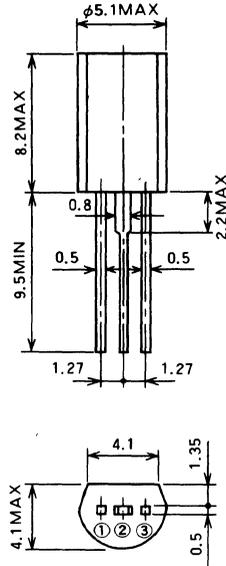
TYPE SOT-89L 3-PIN TRANSISTOR

Dimension in mm



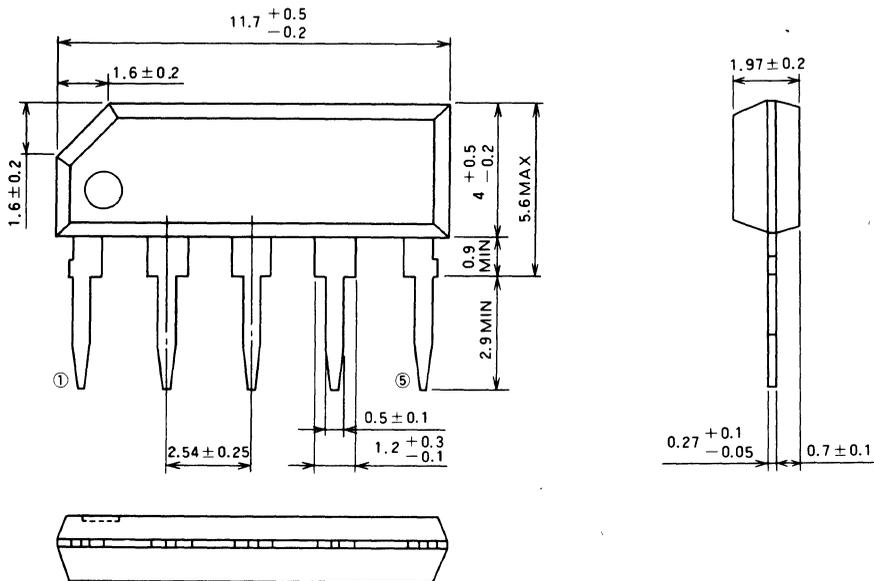
TYPE TO-92L 3-PIN TRANSISTOR

Dimension in mm



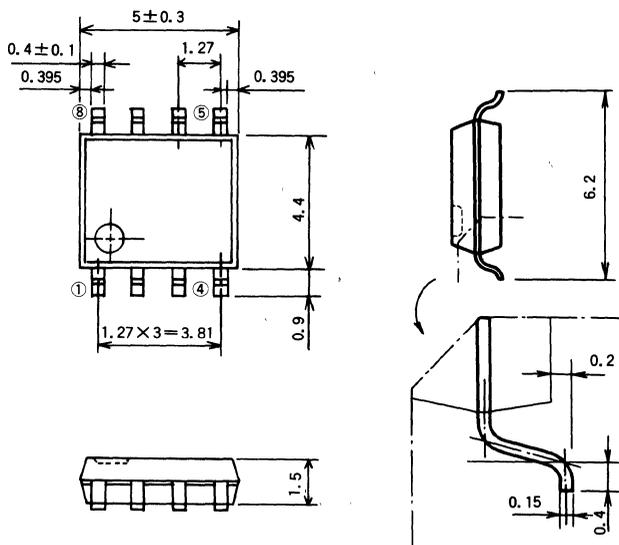
TYPE 5P5 5-PIN MOLDED PLASTIC SIP

Dimension in mm



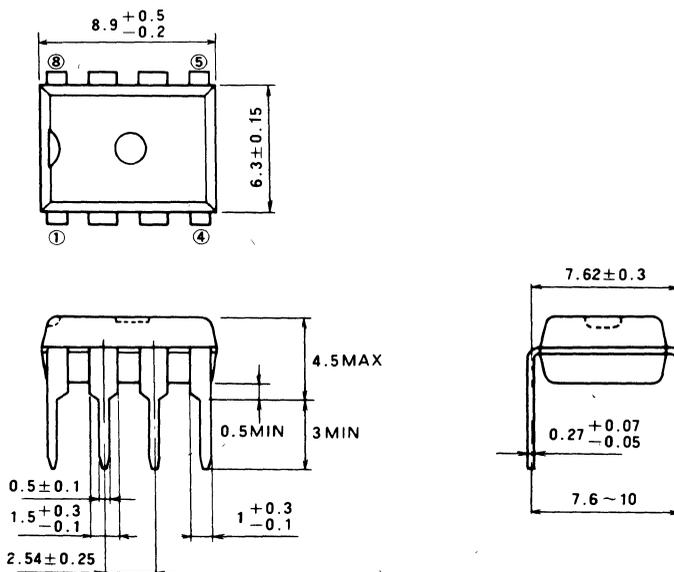
TYPE 8P2S 8-PIN MOLDED PLASTIC MINI FLAT

Dimension in mm



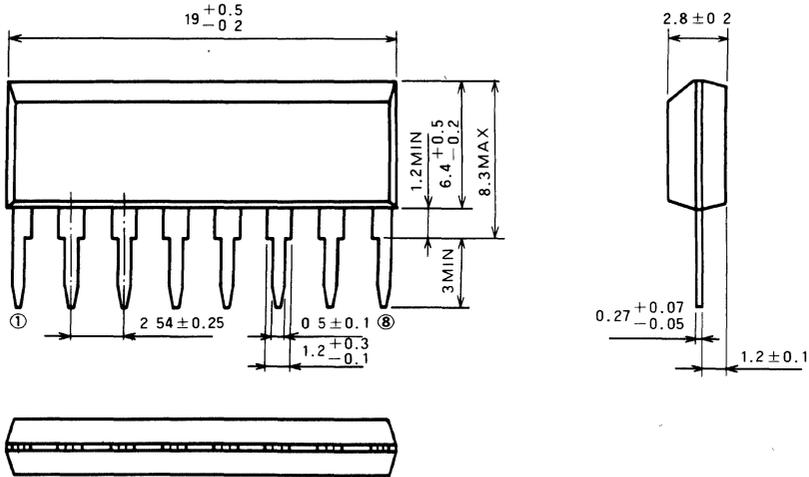
TYPE 8P4 8-PIN MOLDED PLASTIC DIP

Dimension in mm



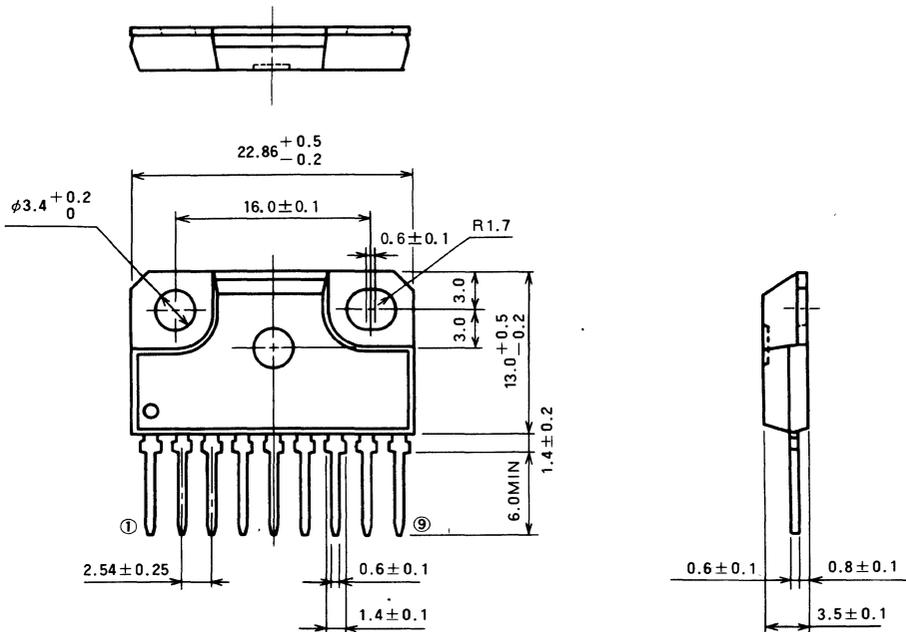
TYPE 8P5 8-PIN MOLDED PLASTIC SIP

Dimension in mm



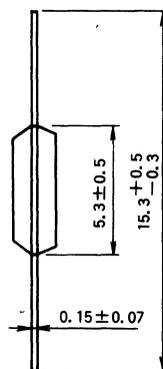
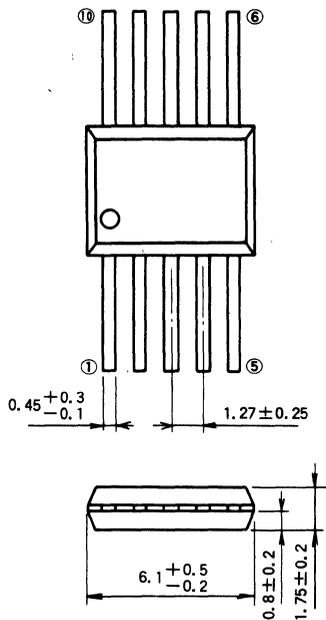
TYPE 9P9 9-PIN MOLDED PLASTIC SIP

Dimension in mm



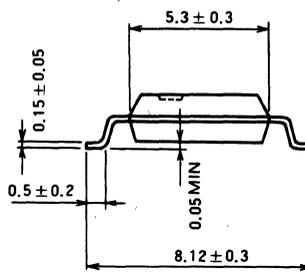
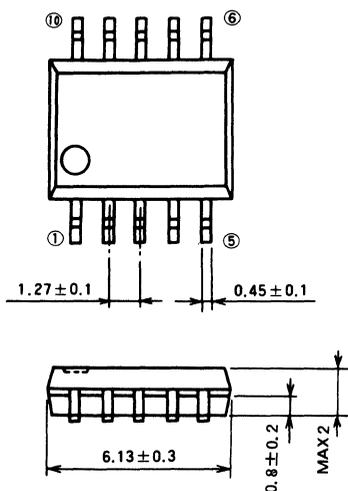
TYPE 10P2 10-PIN MOLDED PLASTIC FLAT

Dimension in mm



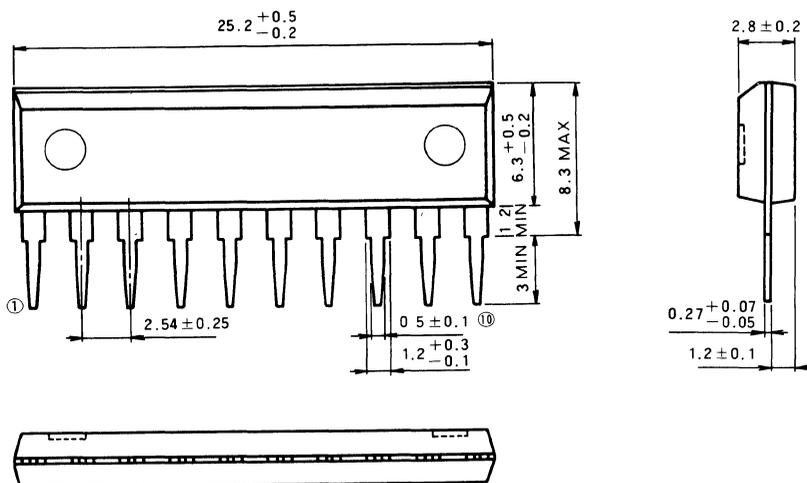
TYPE 10P2-C 10-PIN MOLDED PLASTIC FLAT

Dimension in mm



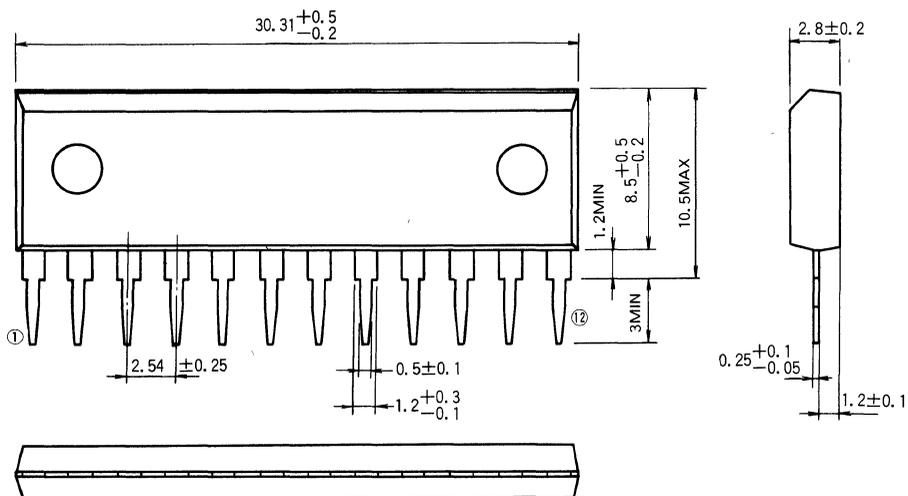
TYPE 10P5 10-PIN MOLDED PLASTIC SIP

Dimension in mm



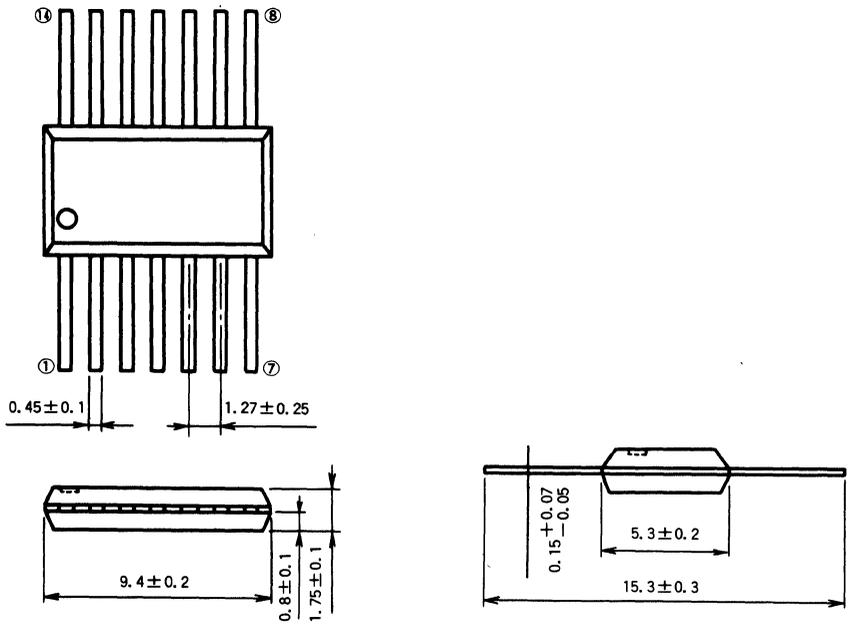
TYPE 12P5 12-PIN MOLDED PLASTIC SIP

Dimension in mm



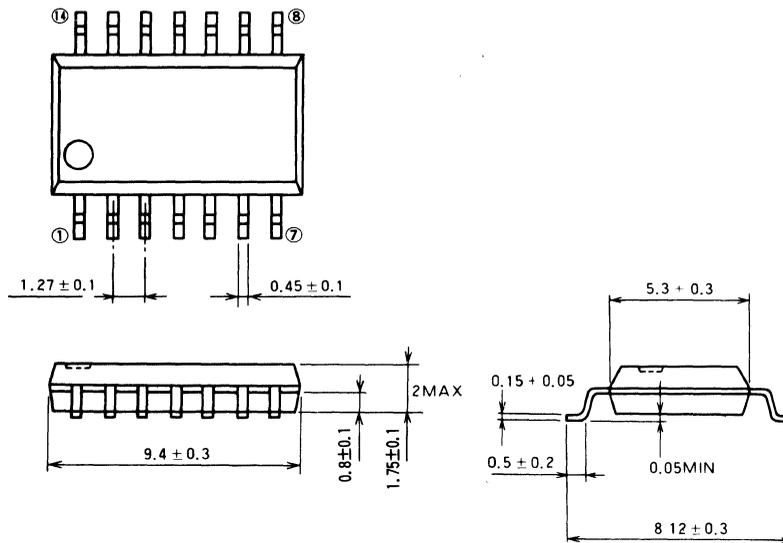
TYPE 14P2 14-PIN MOLDED PLASTIC FLAT

Dimension in mm



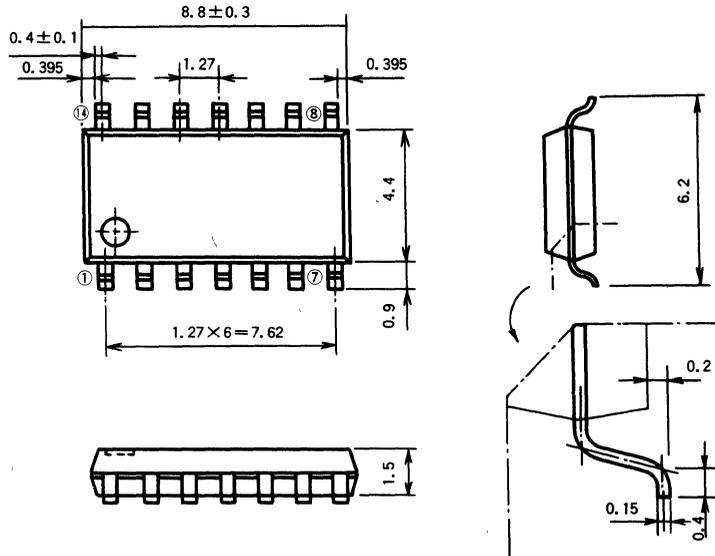
TYPE 14P2-C 14-PIN MOLDED PLASTIC FLAT

Dimension in mm



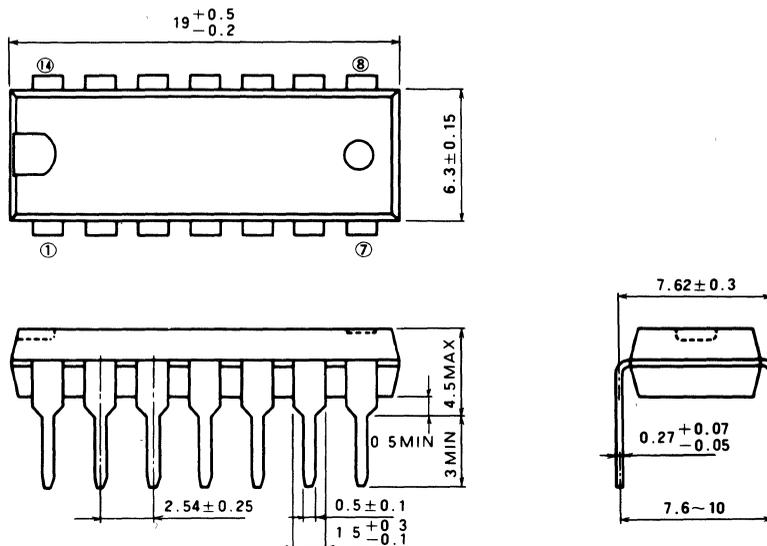
TYPE 14P2S 14-PIN MOLDED PLASTIC MINI FLAT

Dimension in mm



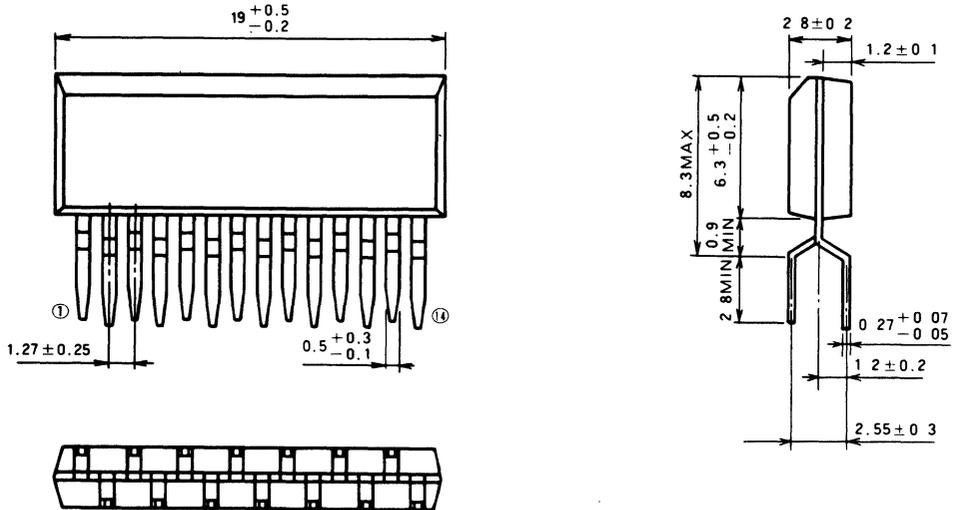
TYPE 14P4 14-PIN MOLDED PLASTIC DIP

Dimension in mm



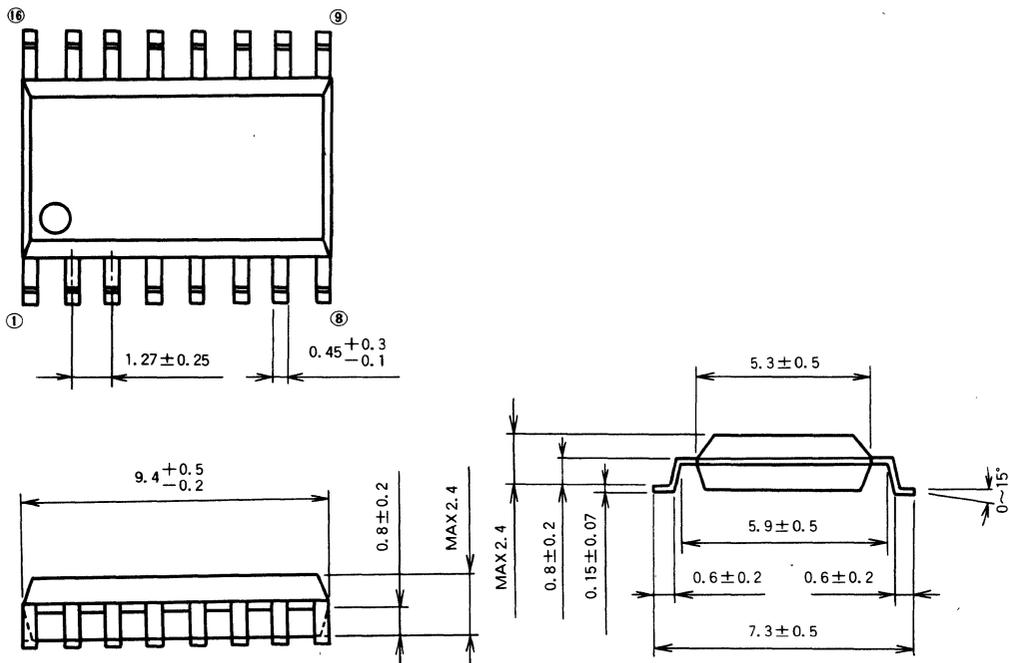
TYPE 14P5A 14-PIN MOLDED PLASTIC ZIP

Dimension in mm



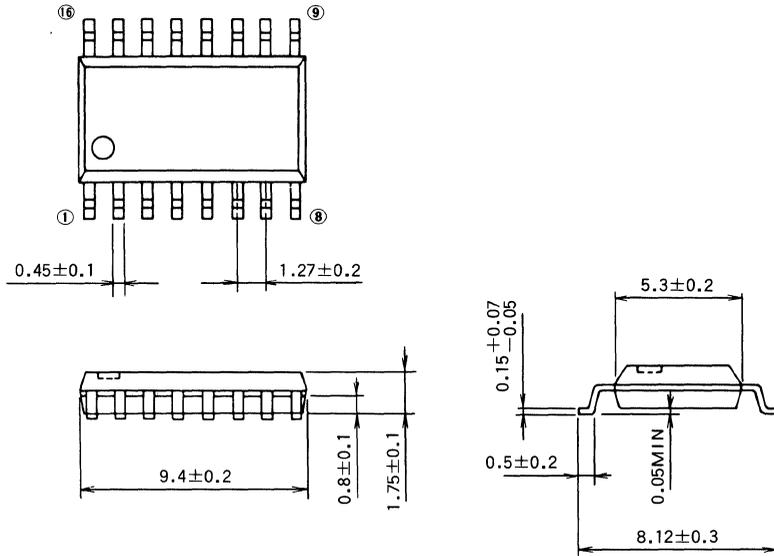
TYPE 16P2-B 16PIN MOLDED PLASTIC FLAT

Dimension in mm



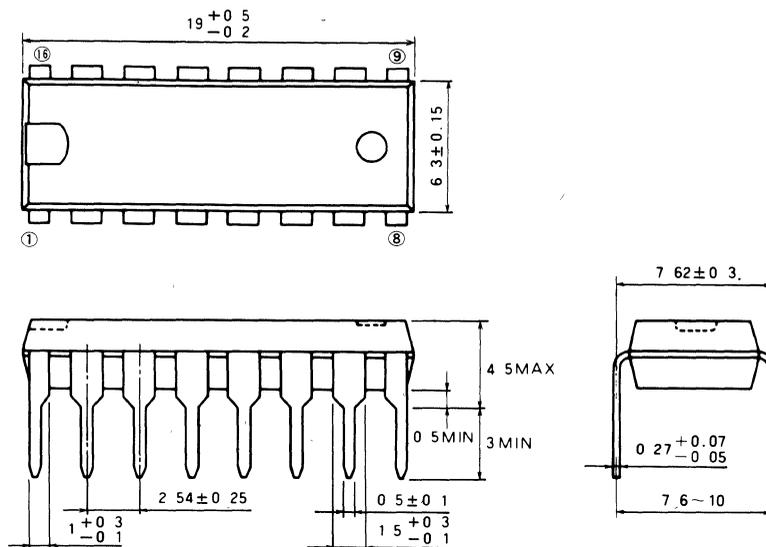
TYPE 16P2-C 16-PIN MOLDED PLASTIC FLAT

Dimension in mm



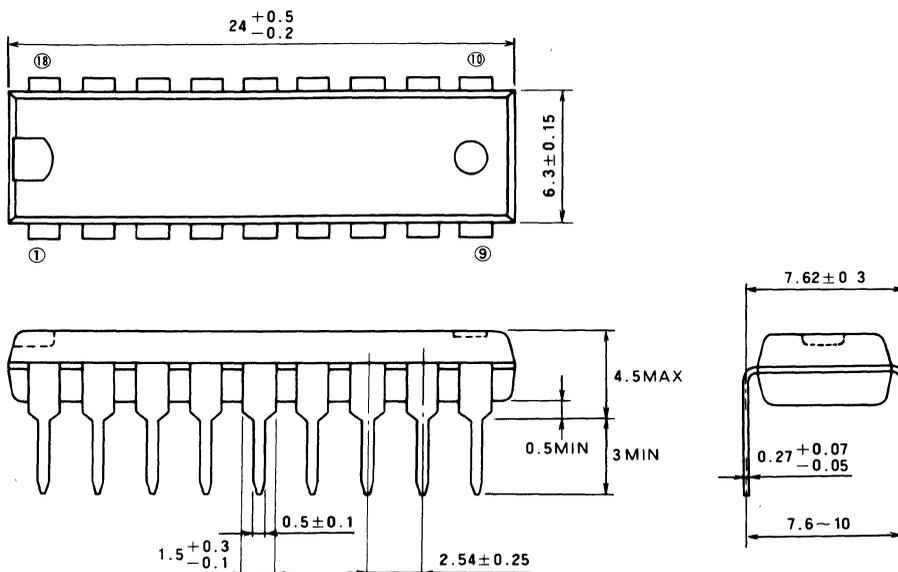
TYPE 16P4 16-PIN MOLDED PLASTIC DIP

Dimension in mm



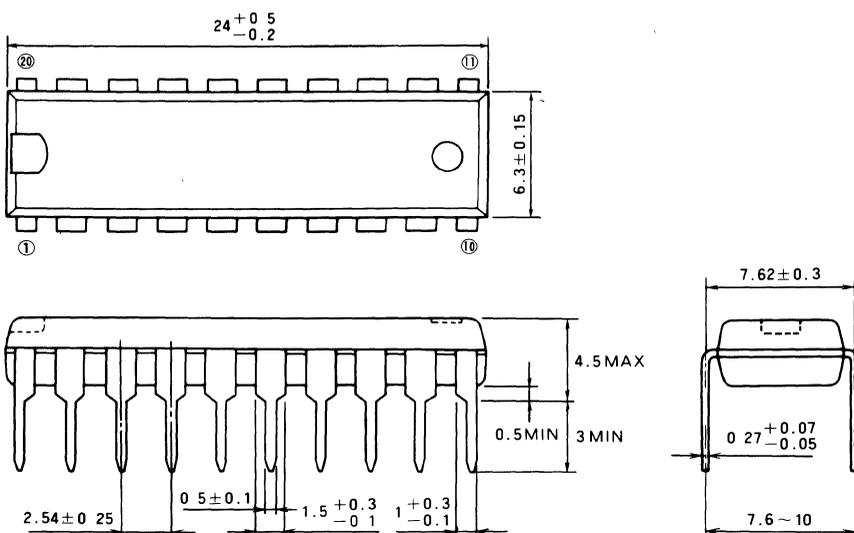
TYPE 18P4 18-PIN MOLDED PLASTIC DIP

Dimension in mm



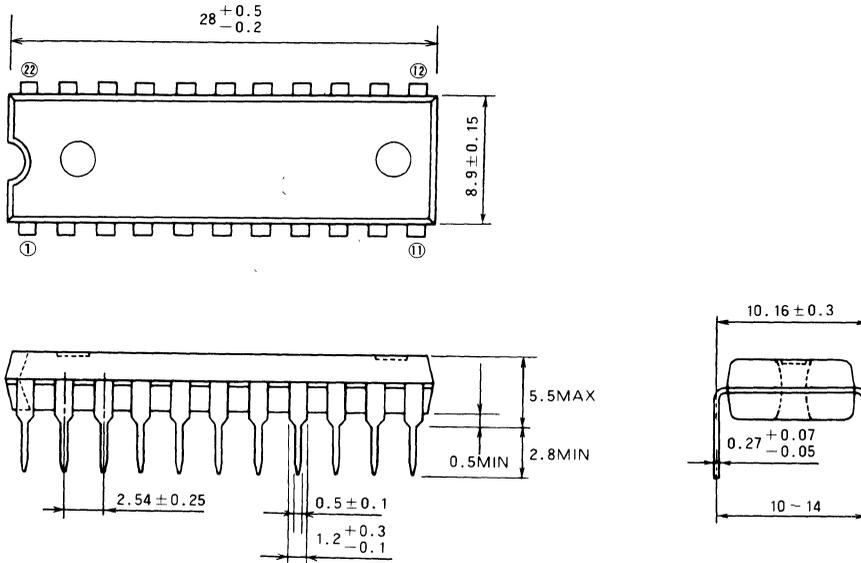
TYPE 20P4 20-PIN MOLDED PLASTIC DIP

Dimension in mm



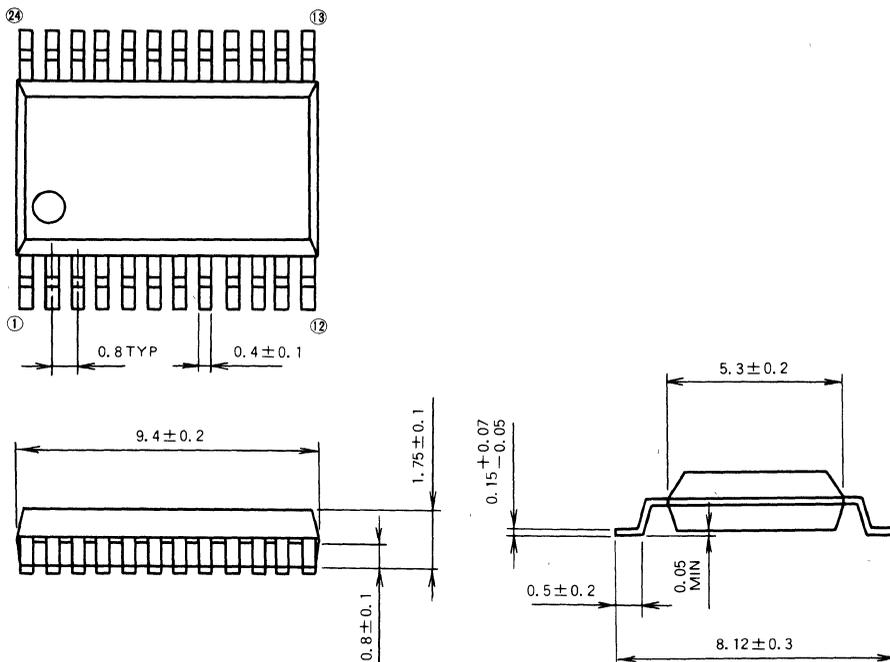
TYPE 22P4 22-PIN MOLDED PLASTIC DIP

Dimension in mm



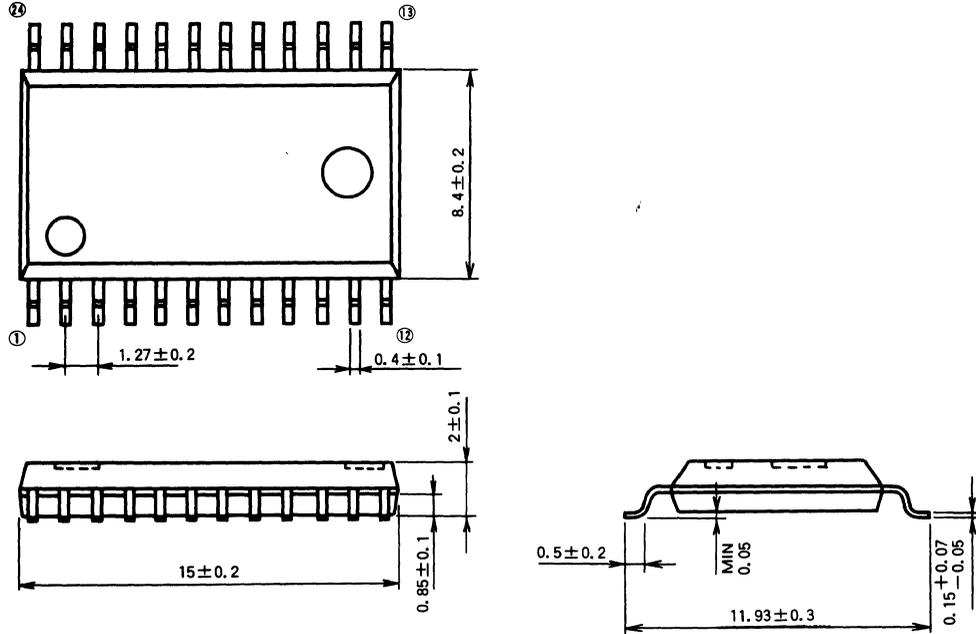
TYPE 24P2T-C 24-PIN MOLDED PLASTIC FLAT

Dimension in mm



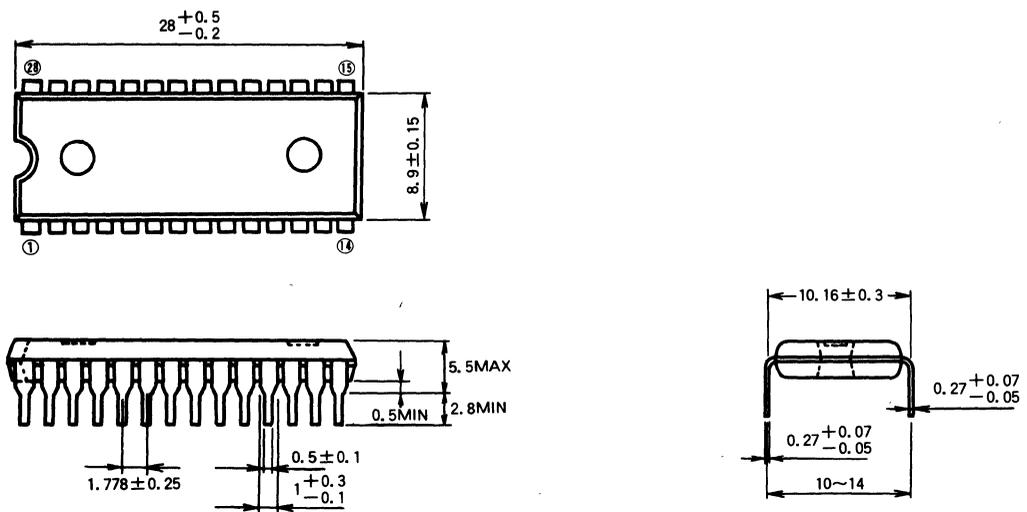
TYPE 24P2W 24-PIN MOLDED PLASTIC FLAT

Dimension in mm



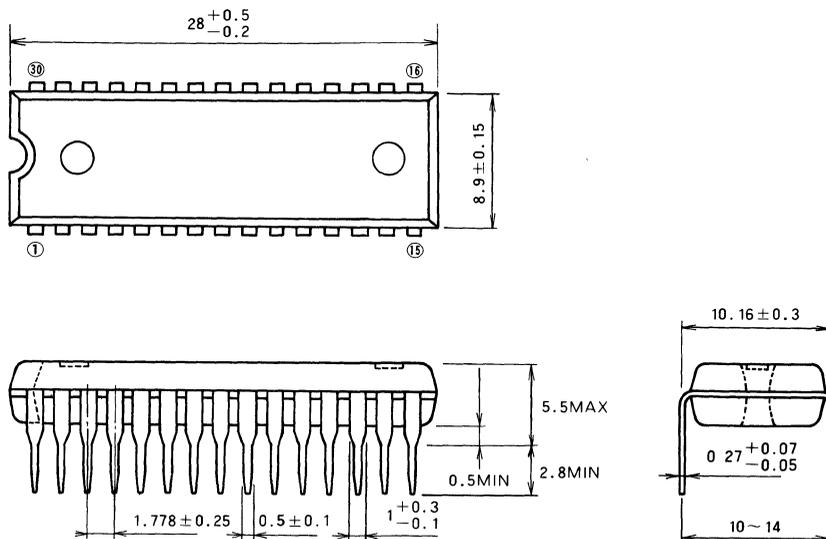
TYPE 28P4B 28-PIN MOLDED PLASTIC DIP (LEAD PITCH 1.778mm)

Dimension in mm



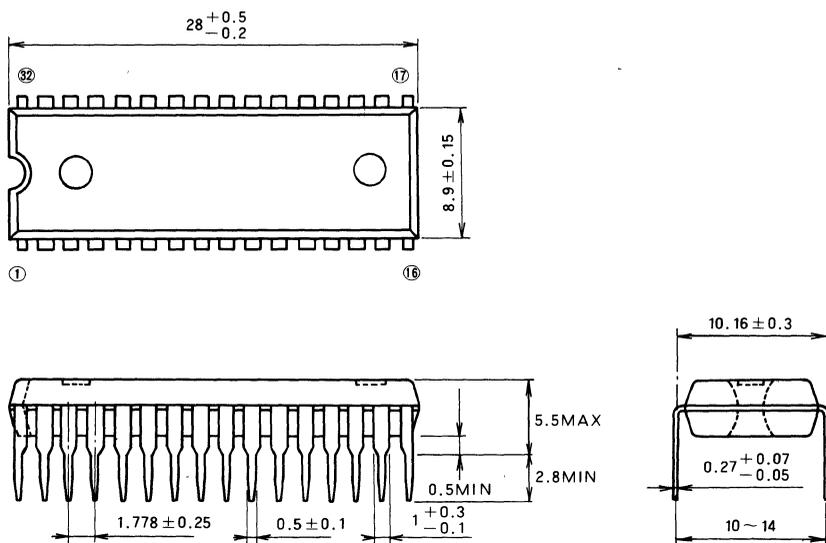
TYPE 30P4B 30-PIN MOLDED PLASTIC DIP(SHRINK)

Dimension in mm



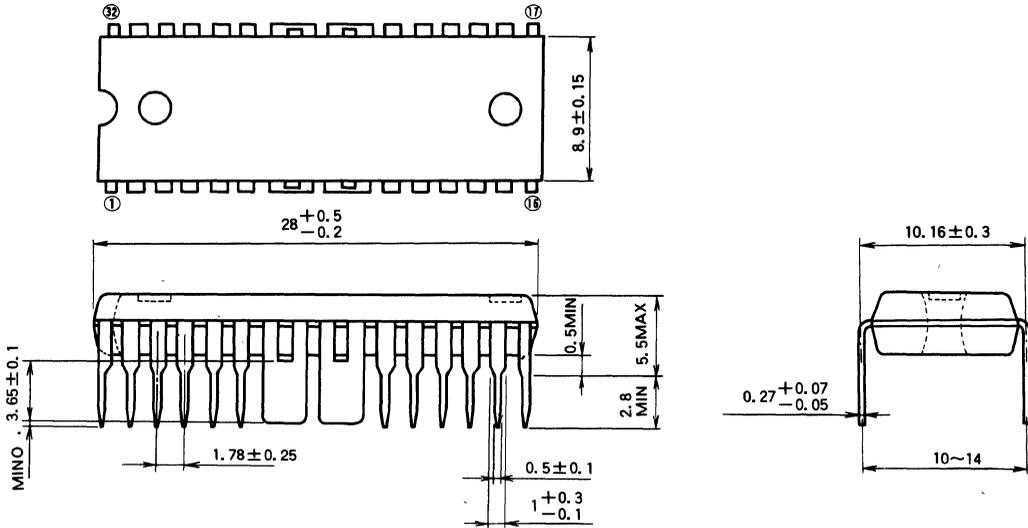
TYPE 32P4B 32-PIN MOLDED PLASTIC DIP (SHRINK)

Dimension in mm



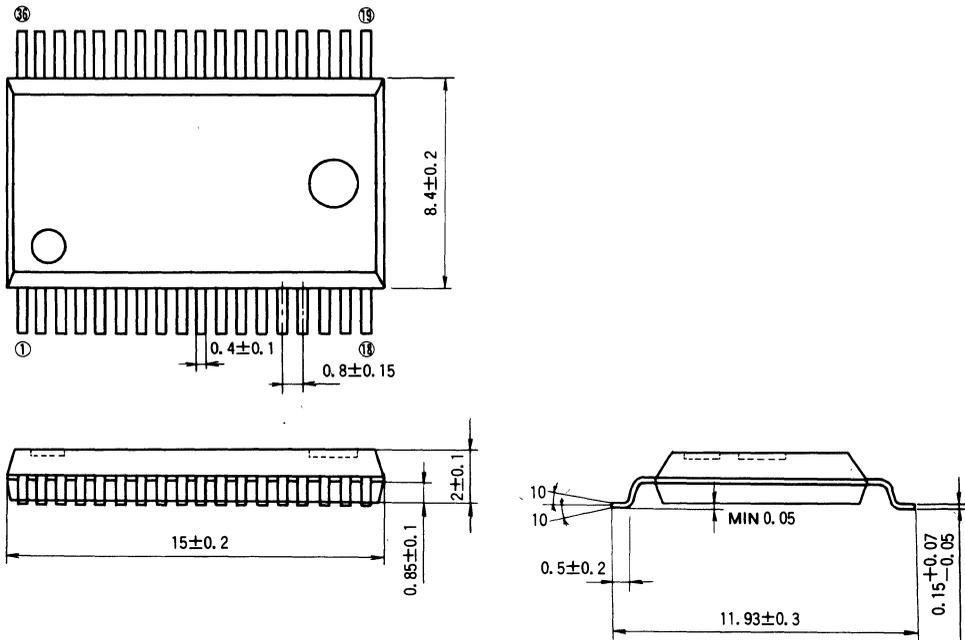
TYPE 32P4B-A 32-PIN PLASTIC DIP(SHRINK)

Dimension in mm



TYPE 36P2R-A 32-PIN MOLDED PLASTIC FLAT

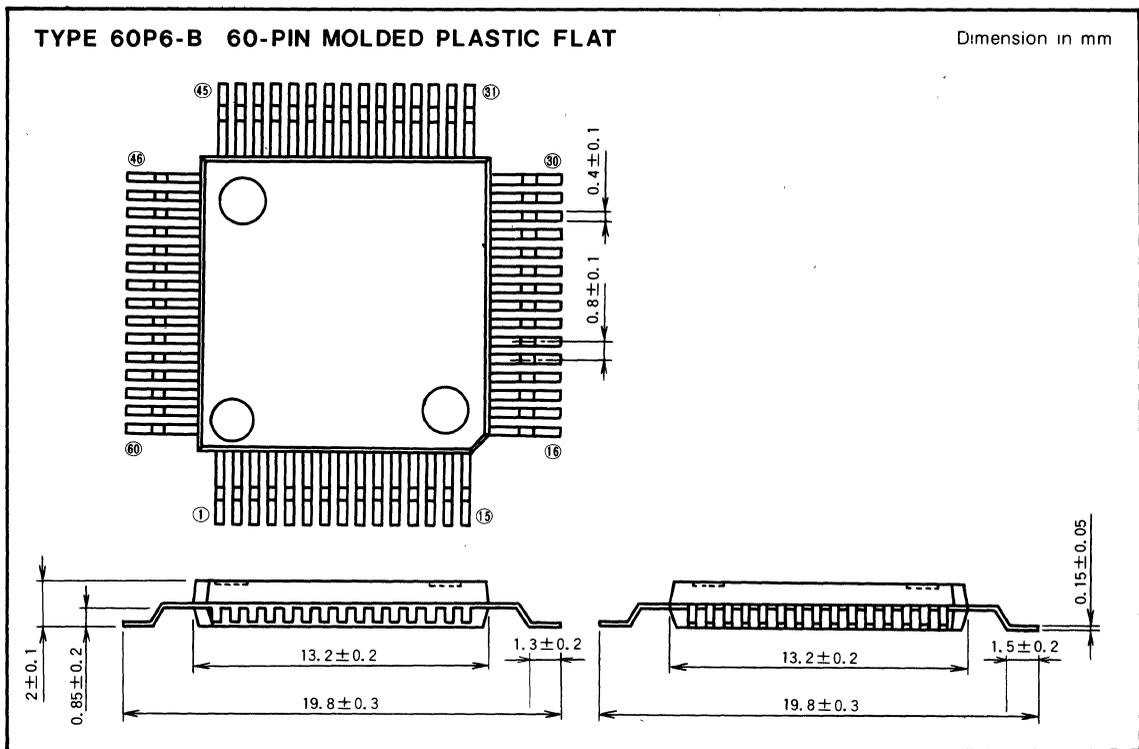
Dimension in mm



PACKAGE OUTLINES

TYPE 60P6-B 60-PIN MOLDED PLASTIC FLAT

Dimension in mm



OPERATIONAL AMPLIFIER

2

QUICK REFERENCE TABLE OF OPERATIONAL AMPLIFIERS

Application	Type No	Circuit Function	Features	Maximum Ratings			Electrical Characteristics (Typical characteristics)							Package Outline	Inter-changeable products
				Operating supply voltage range V_{CC} (V)	Load current I_L (mA)	Allowable power dissipation P_d (mW)	Circuit current I_{CC} (mA)	Input offset voltage V_{IO} (mV)	Input bias current I_B (nA)	Slew rate SR (V/ μ s)	Gain bandwidth product f_T (MHz)	Input-referred noise voltage V_{NI} (μ Vrms)			
General-purpose	M51802	Single with offset adjusting pin	<ul style="list-style-type: none"> No latch-up Built-in protection circuit against output short circuit Easy offset adjustment 	$\pm 2 \sim \pm 18$	—	360(L) 625(P)	1.7	1	150	0.3	0.7	—	L: 8-pin P. 8-pin SIP DIP	741	
	M5216	Dual large-current (Headphone amplifier)	<ul style="list-style-type: none"> Large-current, high output High-speed 	$\pm 2 \sim \pm 18$	± 100	(L)(P)(FP) 800/625/440	4.5	0.5	180	3.0	10	1.8 ($R_S=1k\Omega$) Flat	L: 8-pin SIP	4556	
	M5218/ M5R4558P	Dual low-noise general-purpose	<ul style="list-style-type: none"> Large-current High-speed 	$\pm 2 \sim \pm 18$	± 50	800/625/440	3.0	0.5	100	2.2	7	2.0 ($R_S=1k\Omega$) Flat	P: 8-pin DIP	4557 4558 4559	
	M5223/ M5N358P	Dual single power supply	<ul style="list-style-type: none"> Both input and output pins can be operated from 0V (GND) 	3~36	± 50	800/625/440	0.7	2.0	30	0.6	1	—	FP: 8-pin mini FLAT	358 2904	
	M5224/ M5N324P	Quad single power supply	<ul style="list-style-type: none"> High voltage Low power dissipation 	3~36	± 50	700/550	1.2	2.0	30	0.6	1	—	P: 14-pin DIP	324 2902	
	★ M5228	Quad low-noise general-purpose	<ul style="list-style-type: none"> Large-current High-speed 	$\pm 2 \sim \pm 18$	± 50	700/550	6.0	0.5	100	2.2	7	2.0 ($R_S=1k\Omega$) Flat	FP: 14-pin FLAT	4136 6554 4741	
Low-noise	★ M5209	Dual high withstand voltage	<ul style="list-style-type: none"> High withstand voltage $V_{CC}=\pm 25V$ Low-noise, high-speed 	$\pm 2 \sim \pm 25$	± 50	800/625/440	3.5	0.5	300	6.5	20	1.2 ($R_G=2.2k\Omega$) Flat	L: 8-pin SIP	4562	
	★ M5210	Dual high S/N (microphone amplifier)	<ul style="list-style-type: none"> High withstand voltage $V_{CC}=\pm 25V$ Low noise, high-speed 	$\pm 2 \sim \pm 25$	± 50	800/625/440	4.0	0.5	700	6.5	20	1.0 ($R_G=2.2k\Omega$) Flat	P: 8-pin DIP	2043	
	M5219	Dual 77dB S/N	<ul style="list-style-type: none"> Low noise, high-speed High withstand voltage $V_{CC}=\pm 25V$ 	$\pm 2 \sim \pm 25$	± 50	800/625/440	3.5	0.5	300	6.5	20	0.9 ($R_G=2.2k\Omega$) RIAA	FP: 8-pin mini FLAT	4562	
	M5220	Dual 83dB S/N	<ul style="list-style-type: none"> Low noise, high speed High withstand voltage $V_{CC}=\pm 25V$ 	$\pm 2 \sim \pm 25$	± 50	800/625/440	4.0	0.5	700	6.5	20	0.75 ($R_G=2.2k\Omega$) RIAA		2043	
Bi-FET	M5221/ M5T082P	Dual general-purpose	<ul style="list-style-type: none"> High input impedance (1000MΩ) High-speed 	$\pm 5 \sim \pm 18$	± 50	800/625/440	3.0	5.0	30PA	13	3	2.2 ($R_G=100k\Omega$) Flat	L: 8-pin SIP P: 8-pin DIP	072 082	
	★ M5238	Dual 73dB S/N	<ul style="list-style-type: none"> Low noise, high-speed High input impedance (1000MΩ) 	$\pm 5 \sim \pm 18$	± 50	800/625/440	4.5	5.0	30PA	20	6	1.6 ($R_G=2.2k\Omega$) RIAA	FP: 8-pin mini FLAT	—	
	★ M5240P	Dual 82dB S/N	<ul style="list-style-type: none"> Low noise, high-speed External phase compensation pin 	$\pm 5 \sim \pm 18$	± 50	1000	10	10.0	200PA	40	18	0.9 ($R_G=2.2k\Omega$) RIAA	16-pin DIP	—	
Complex	★ M5201	Dual input switching (general-purpose switching operational amplifier)	<ul style="list-style-type: none"> Operational amplifier equivalent to the M5218 is built-in Gain can be set independently Small switching shock noise 	$\pm 2.5 \sim \pm 18$	± 50	800/625/440	2.3	0.5	100	2.2	7	2.0 ($R_S=1k\Omega$) Flat	L: 8-pin SIP P: 8-pin DIP FP: 8-pin mini FLAT	—	

★: New product

INTERCHANGEABLE PRODUCTS OF OPERATIONAL AMPLIFIERS

Application	Mitsubishi		Interchangeable products of other manufacturers											
	Circuit function	Type No.	JRC	NEC	Matsushita	Toshiba	Hitachi	Tokyo Sanyo	ROHm	TI	NSS	FC	MOTO	
Operational amplifier	Dual large-current operational amplifier	M5216L	NJM4556S(9L)											
		M5216P	NJM4556D	μ PC4556C										
		★M5216FP	NJM4556M	μ PC4556G										
	Dual general-purpose low-noise operational amplifier	M5218L	NJM4558S(9L) NJM4559S(9L)			AN6551(9L) AN6555(9L)	TA75557S(9L) TA75558S(9L) TA75559S(9L)		LA6458S(9L)	BA715(9L)				
		M5218P M5R4558P	NJM4558D NJM4559D	μ PC4557C μ PC4558C μ PC4559C	AN4558 AN6552 AN6553, 6556	TA75557P TA75558P TA75559P	(HA17458)	LA6458D	BA4558	TL4558P RM4558 RC4558		μ A4558 μ A4558TC	(MC1458)	
		★M5218FP	NJM4558M NJM4559M	μ PC4558G	AN4558S AN6553S AN6556S	TA75558F TA75559F		LA6458M	BA4558F					
	Dual low-noise operational amplifier	M5219L	NJM4562S(9L) NJM4560S(9L)	μ PC1224(8L)										
		M5219P	NJM4562D NJM4560D								LM833			
	Dual low-noise operational amplifier	M5220L	NJM2041S(9L) NJM2043S(9L)			AN6557(9L)								
		M5220P	NJM2041D NJM2043D			AN6558					LM833			
	Dual J-FET input operational amplifier	M5221L	NJM072S(9L) NJM082S(9L)			AN6581								
		M5221P M5T082P	NJM072D NJM082D	μ PC803C μ PC4082C	AN1082		HA17082		BA082	TL072 TL082	LF353			
		★M5221FP	NJM072M NJM082M	μ PC4082G	AN1082S									
	Dual single power supply operational amplifier	M5223L	NJM2904S(9L)			AN6561(9L)			LA6358S(9L)	BA718(9L)				
		M5223P M5N358P	NJM2904D	μ PC358C μ PC1251C	AN1358 AN6562	TA75358P	HA17904PS	LA6358	BA728 BA10358	LM358P	LM358	LM358 LM2904	MLM358	
		★M5223FP	NJM2904M	μ PC358G	AN1358S	TA75358F			BA728F BA10358F					

L: SIP (8-pin), P: DIP(8-pin), (9L): SIP(9-pin), FP: mini FLAT (8-, 14-pin)

★: New product

INTERCHANGEABLE PRODUCTS OF OPERATIONAL AMPLIFIERS (CONTINUED)

Applica- tion	Mitsubishi		Interchangeable products of other manufacturers											
	Circuit function	Type No.	JRC	NEC	Matsushita	Toshiba	Hitachi	Tokyo Sanyo	ROHm	TI	NSS	FC	MOTO	
Operational amplifier	Quad single power supply operational amplifier	M5224P	NJM2902D	μ PC324	AN1324	TA75902P	HA17902PS	LA6324	BA10324	LM2902	LM324	μ A324	MLM2902	
		M5N324P		μ PC451C	AN6564	TA75324P				LM324N	LM2902	μ A2902		
	Quad general-purpose low-noise operational amplifier	M5224FP	NJM2902M	μ PC324G	AN1324NS	TA75902F		LA6324M	BA10324F					
		M5228P	NJM2058 NJM2059	μ PC4741	AN6554				BA14741	RC4136		μ A4136		
	Dual J-FET input, low-noise operational amplifiers	M5228FP	NJM2058M NJM2059M	μ PC4741G	AN6554NS									
		★ M5238L												
	Dual J-FET input, High S/N operational amplifier	★ M5238P												
		M5240P												
Dual input switching general-purpose operational amplifier	★ M5201L													
	★ M5201P													

L: SIP (8-pin), P: DIP (8-pin), (9L): SIP (9-pin), FP: mini FLAT (8-, 14-pins)

★: New product

HOW TO CHOOSE OPERATIONAL AMPLIFIERS

How to choose operational amplifiers

Choose operational amplifiers according to the following categories of usage.

- By basic circuit configuration of operational amplifiers
 - (1) Dual power supply operational amplifiers
M5218/M5R4558P, M5219, M5220...
 - (2) Single power supply operational amplifier
M5223/M5N358P, M5224P/M5N324P
- By functional type of operational amplifiers
 - (1) High voltage operational amplifiers
M5209, M5219, M5210, M5220 ($V_{CC} \sim \pm 25V$)
 - (2) Low-noise operational amplifiers
M5219, M5220, M5240 (S/N=77~83dB)
 - (3) High-speed operational amplifiers
M5209, M5210, M5219, M5220,
M5221/M5T082P, M5238
(SR=6.5~13V/ μ s)
 - (4) Large-current operational amplifiers
M5216 ($I_{LP} = \pm 100mA$)
 - (5) High input impedance operational amplifiers
M5221/M5T082P, M5238
(Bi-FET operational amplifier $R_i = 1000M\Omega$)

Method of choosing operational amplifiers according to the features of each type shown above is explained in the following.

1. Dual power supply and single power supply operational amplifiers

A dual power supply operational amplifier is one of the most popular operational amplifiers, represented by old types 709 or 741 or the more recent type 4558 (Mitsubishi's M5218/M5R4558P) and is usually used at a supply voltage of $V_{CC} \pm 15V$. As shown in Fig. 1, two power supplies of positive and negative voltages are applied to GND, and the device is used mainly at near GND level. In this case, V_x and V_y are residual voltage, generated inside an operational amplifier. Fig. 2 shows the relationship between the input and output of amplifiers, and voltage gain A_v , i.e., ratio of output V_o against input V_i , is determined by ratio of externally connected resistors R_2/R_1 . On the other hand a single power supply operational amplifier is used by applying a single power supply and amplification operation for GND level input is made possible. As shown in Fig. 3, the signal can be activated up to GND level.

Residual voltage V_x of $+V_{CC}$ is generated as in dual power supply type operational amplifiers. Fig. 4 shows the input and output characteristics. (Refer to operational logic in the individual data of M5223/M5N358P, M5224/M5N324P for details.)

Care must be taken with a single power supply operational amplifier that only a single power supply is applied. The output DC bias level is not determined automatically and an operational amplifier is activated only to GND level. It operates in the same way as a dual power supply operational amplifier operates within the

activated range. To sum up, basic DC output levels for both types of amplifiers must be determined by the voltage of non-inverted input (+input). In other words, a single power supply operational amplifier can be used with level-shifted voltages just as a dual operational amplifier used GND level as its basic DC level.

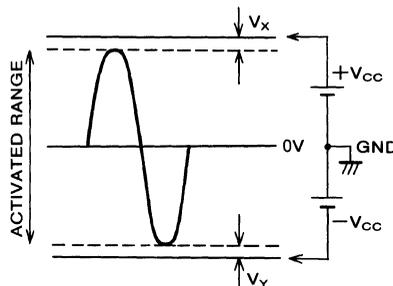


Fig. 1

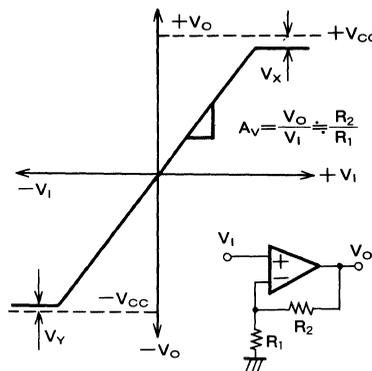


Fig. 2

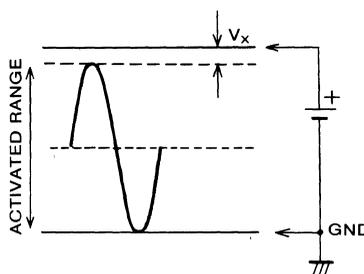


Fig. 3

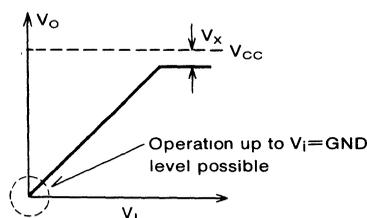


Fig. 4

HOW TO CHOOSE OPERATIONAL AMPLIFIERS

2. High voltage operational amplifiers

Ordinary amplifiers accept maximum supply voltage of $V_{CC} \pm 18V$ and are usually used at $V_{CC} \pm 15V$, but, $V_{CC} \pm 20V$ or above is used for audio amplifiers for a wider dynamic range. Mitsubishi has a variety of products for the M52XX series of operational amplifiers such as the M5209, M5210, M5220, which guarantee maximum supply voltage of $V_{CC} \pm 25V$ suitable for audio preamplifiers.

3. Low-noise operational amplifiers

To satisfy low-noise specifications required in audio equipment, the M52XX series includes the M5209, M5210, M5219 and M5220. The M5209 and M5210 are suitable for high S/N microphone or line amplifiers. The M5219 and M5220 have suppressed $1/f$ noise, pulse-related noise and low-wave noise and are suitable for EQ amplifiers of stereo equipment and tape decks. The M5238, M5240 are low-noise Bi-FET operational amplifiers.

4. High-speed operational amplifiers

The M52XX series includes the M5209, M5210, M5219 and M5220, which have characteristics of high-speed as well as high-voltage and low-noise, and $SR = 6.5V/\mu s$ typ. The SR stands for slew rate and is a parameter to indicate high speed in an amplifier and is related to high frequency distortion such as 20kHz in processing fast-rising signals and in audio circuits.

Gain bandwidth product f_T is also a high-frequency parameter of an operational amplifier as well as slew rate SR. This parameter indicates the limit of amplification capacity of an operational amplifier in small signal and indicates the frequency at which voltage gain is 1 ($G_V = 0$). It is an important parameter for determining the gain and frequency characteristics of an operational amplifier. For example, when an amplifier of $G_V = 40dB$ is designed, the M5209 of $f_T = 20MHz$ becomes an amplifier with a bandwidth of 200kHz and the M5218/M5R4558P becomes an amplifier with a bandwidth of 70kHz. Choose the most suitable operational amplifier according to usage.

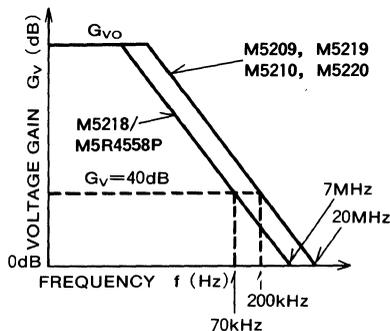


Fig. 5

5. Large-current operational amplifier

The M5216 is a large-current operational amplifier with load current $I_{LP} = \pm 100mA$ and is suitable for applications that require large-current, for example, for headphone drive, relay circuit or LED.

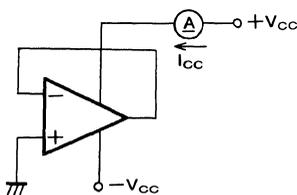
6. High input impedance operational amplifier

The M5221/M5T082P, M5238 Bi-FET operational amplifiers are high input impedance operational amplifiers, and have characteristic of input impedance $R_i = 1000 M\Omega$ typ. The device is suitable for application that is related to high input impedance, such as sample-hold circuit or various types of filters.

TEST METHOD OF OPERATIONAL AMPLIFIERS

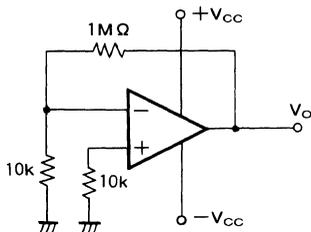
■ Test method of operational amplifiers

(1) Circuit current I_{CC}



* If output pin current is maintained at 0 and in quiescent state, the circuit needs not be a voltage follower as shown above.

(2) Input offset voltage V_{IO} , supply voltage rejection ratio SVRR



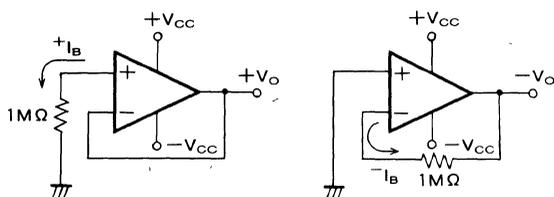
- $V_{IO} = \frac{V_O}{100}$ ($V_{CC} \pm 15V$)

- $SVRR = \frac{V_{O1} - V_{O2}}{100 \times 5}$ ($V_{O1}: V_{CC} \pm 17.5V$
 $V_{O2}: V_{CC} \pm 12.5V$)

* SVRR

Measure change of offset V_{IO} when $V_{CC} = \pm 12.5$ and ± 17.5 (where $R_S = 10k\Omega$).

(3) Input bias current I_B , input offset current I_{IO}



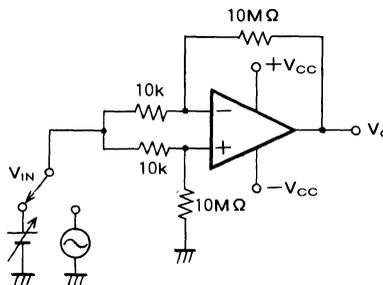
* I_B : Base current of first stage L-PNP

- $+I_B = \frac{+V_O}{1M\Omega}$

- $-I_B = \frac{-V_O}{1M\Omega}$

- $I_{IO} = | +I_B - -I_B |$

(4) Common phase input voltage width V_{CM} , common phase signal rejection ratio CMR

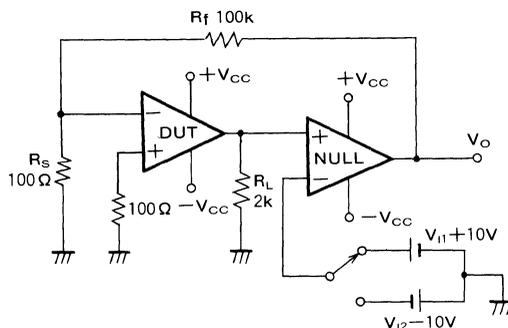


- V_{CM} DC input voltage when output V_O becomes $+10V$ or $-10V$ by applying DC input voltage to V_{IN} .

- CMR Measure V_O by applying AC input signal 0dBm to V_{IN} .

$$CMR = 20 \log \frac{V_{IN}}{\frac{V_O}{1000}}$$

(5) Voltage gain



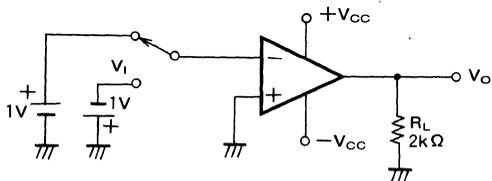
- $G_{VO} = 20 \log A_{vd}$
 $A_{vd} = \frac{R_S + R_f}{R_S} \times \frac{V_{I1} - V_{I2}}{V_{O1} - V_{O2}}$
 $= 10^3 \times \frac{20}{V_{O1} - V_{O2}}$
 $= \frac{2 \times 10^4}{V_{O1} - V_{O2}}$

Example $V_{O1} - V_{O2} = 0.2V \rightarrow 100dB$

* Measure DC voltage gain by using NULL amplifier. DUT output is maintained at $\pm 10V$ by varying (-) input of NULL amplifier between $+10V$ and $-10V$, and output V_O is the DUT input voltage, multiplied by $R_S + R_f/R_S$ times.

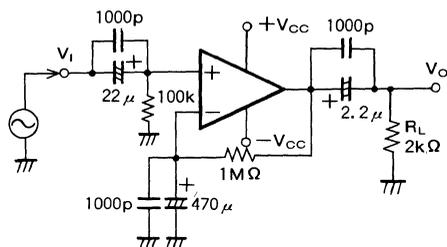
TEST METHOD OF OPERATIONAL AMPLIFIERS

(6) Maximum output voltage V_{OM}

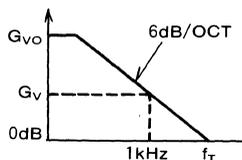


* DUT is used in open status without feedback and output V_O is measured by applying +1V to an input. In this case, the output is saturated when an input is 1V as open gain is high.

(7) Gain bandwidth product f_T , voltage gain G_V

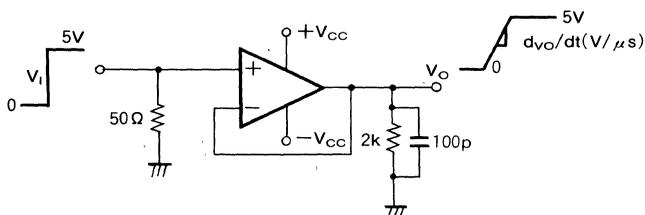


● f_T : Frequency where $G_{V0}=0\text{dB}$



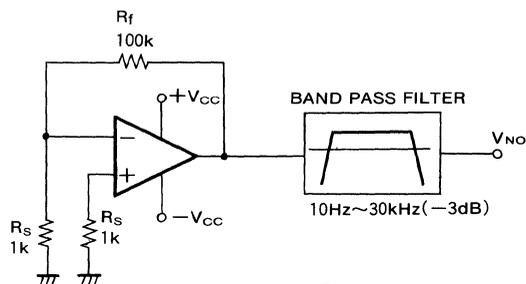
* Measure frequency characteristics of voltage gain G_V when a feedback of 100% ($1\text{M}\Omega$) is applied for DC and feedback is open ($1000\text{pF}/470\mu$) for AC. The f_T can be simply obtained by multiplying voltage gain of 1kHz by frequency.

(8) Slew rate SR



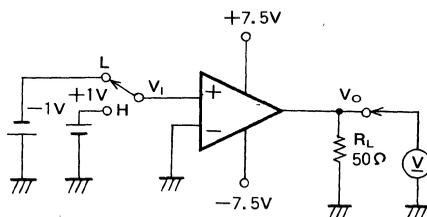
* Measure high-speed capacity of DUT by reading on an oscilloscope and output pulse response (rising speed) by inputting pulse step in a voltage follower circuit of 100% feedback.

(9) Input-referred noise voltage V_{NI}



$$V_{NI} = V_{NO} \cdot \frac{R_f + R_S}{R_S} \approx V_{NO} \cdot 10^{-2}$$

(10) I_{source} , I_{sink} (Single power supply operational amplifier)



V_i { H I_{source}
L I_{sink}

$$I_{source}, I_{sink} = \frac{V_O}{50\Omega}$$

M51802L,P

INTERNALLY PHASE COMPENSATED OPERATIONAL AMPLIFIERS

DESCRIPTION

The M51802L, P are semiconductor integrated circuits designed for a high-gain, high-stability operational amplifier. No phase compensation is required and the devices come with offset adjusting pins and are suitable for home use and industrial equipment.

FEATURES

- Built-in capacitor for phase compensation
- No latch-up
- Built-in protection circuit for output short circuit
- Easy offset adjustment

APPLICATION

Amplifier for measuring instrument, DC amplifier, various types of operational circuit, comparator, multi-vibrator, and oscillator.

RECOMMENDED OPERATING CONDITIONS

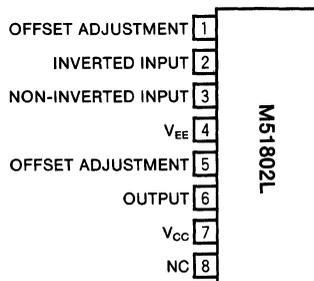
Supply voltage range..... $V_{CC}(+V)$, $V_{EE}(-V)$ 4~18V

Rated supply voltage..... $V_{CC}(+V)$, $V_{EE}(-V)$ 15V

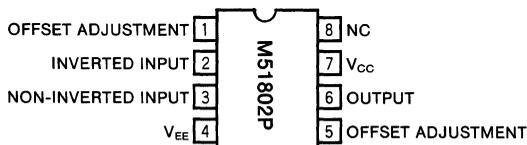
APPLICATION EXAMPLES

An operational amplifier employs feedback and has a variety of functions by the configuration of feedback circuits. Note (1) shows a method adjusting offset voltage.

PIN CONFIGURATION (TOP VIEW)



NC : NO CONNECTION

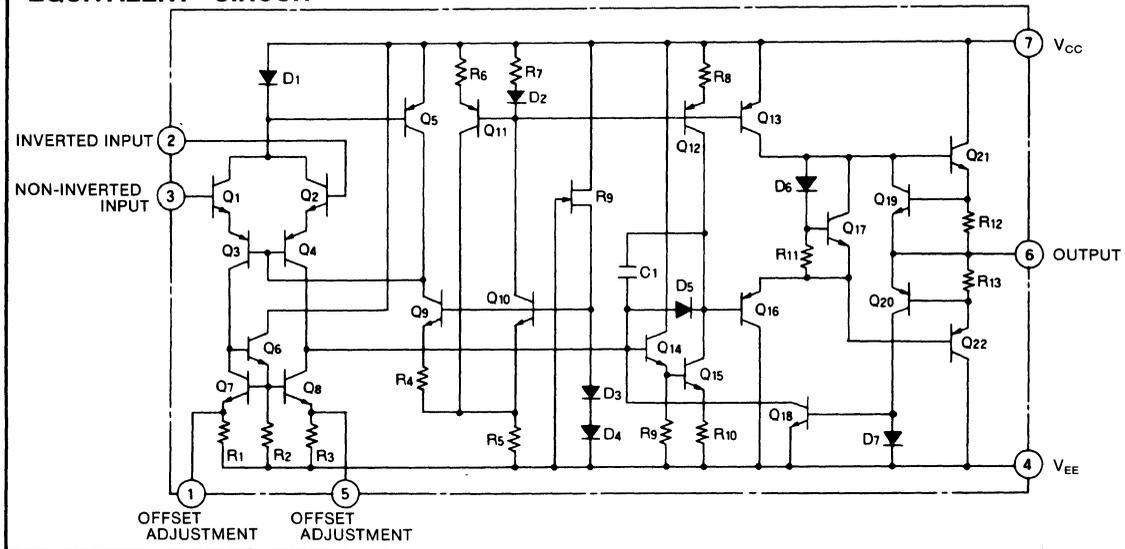


8-pin molded plastic SIP



8-pin molded plastic DIP

EQUIVALENT CIRCUIT



INTERNALLY PHASE COMPENSATED OPERATIONAL AMPLIFIERS

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
V_{EE}			-18	V
V_{id}	Differential input voltage		± 30	V
V_{ic}	Common phase input voltage		± 15	V
P_d	Power dissipation		360(M51802L)	mW
			625(M51802P)	
K_e	Thermal derating	$T_a \geq 25^\circ\text{C}$	3.6(M51802L)	mW/°C
			6.25(M51802P)	
T_{opr}	Operating temperature		-20~+75	°C
T_{stg}	Storage temperature		-40~+125	°C

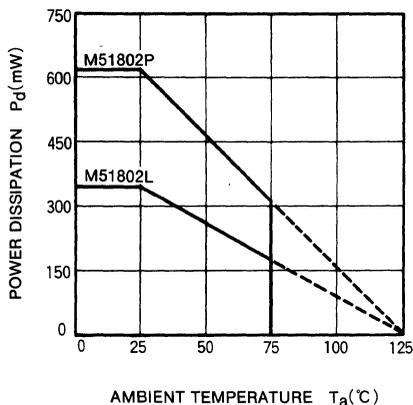
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=+15\text{V}$, $V_{EE}=-15\text{V}$)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{IO}	Input offset voltage	$R_g \leq 10\text{k}\Omega$		1.0	5.0	mV
I_{IB}	Input bias current			150	500	nA
I_{IO}	Input offset current			30	200	nA
R_{in}	Input resistance	Input frequency $f=1\text{kHz}$		1.0		M Ω
G_v	Voltage gain	$R_L \geq 2\text{k}\Omega$, $V_o = \pm 10\text{V}$		86	100	dB
		$R_L \geq 10\text{k}\Omega$	± 12	± 14		
V_{OPP}	Maximum output voltage width	$R_L \geq 2\text{k}\Omega$	± 10	± 13		V
			± 12	± 13		
V_{IC}	Common phase input voltage range		± 12	± 13		V
$CMRR$	Common phase signal rejection ratio	$R_g \leq 10\text{k}\Omega$	70	90		dB
$SVRR$	Supply voltage rejection ratio	$R_g \leq 10\text{k}\Omega$		80	150	$\mu\text{V/V}$
P_d	Power dissipation	$R_L = \infty$		50	85	mW
t_r	Transient response	Rising time		0.3		μs
		Overshoot	$V_{in}=20\text{mV}$, $R_L=2\text{k}\Omega$, $C_L \leq 100\text{pF}$		5.0	
SR	Slew rate	$R_L=2\text{k}\Omega$		0.5		V/ μs

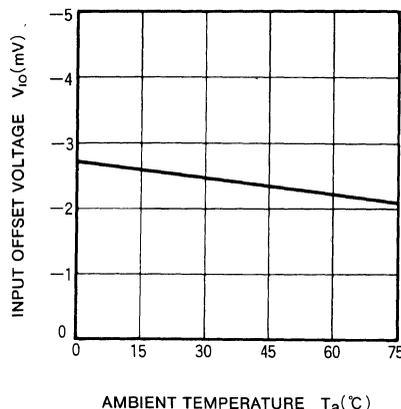
INTERNALLY PHASE COMPENSATED OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=+15\text{V}$, $V_{EE}=-15\text{V}$, unless otherwise noted)

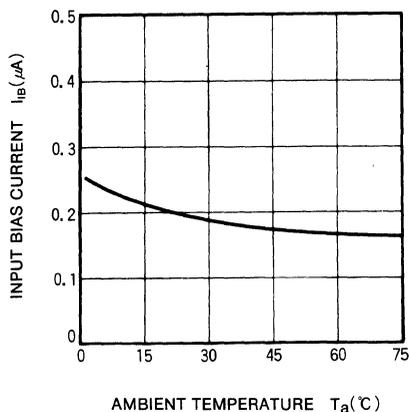
THERMAL DERATING
(MAXIMUM RATING)



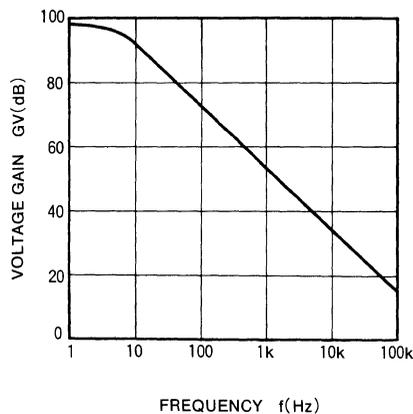
INPUT OFFSET VOLTAGE
VS. AMBIENT TEMPERATURE



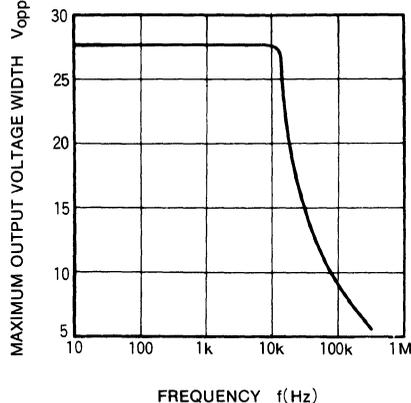
INPUT BIAS CURRENT
VS AMBIENT TEMPERATURE



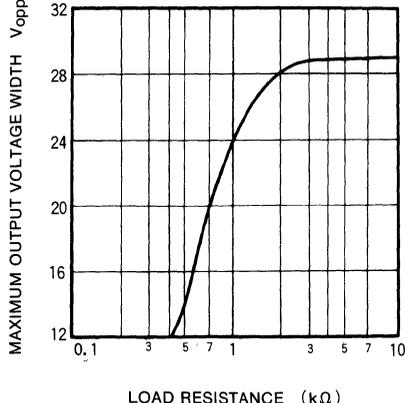
VOLTAGE GAIN
VS FREQUENCY RESPONSE



MAXIMUM OUTPUT VOLTAGE
WIDTH VS FREQUENCY

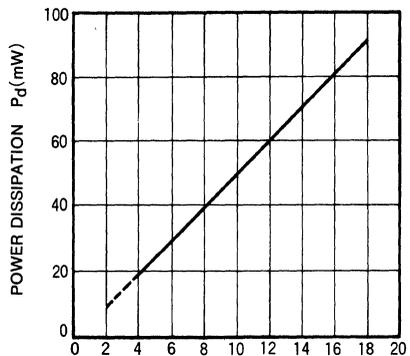


MAXIMUM OUTPUT VOLTAGE
WIDTH VS LOAD RESISTANCE



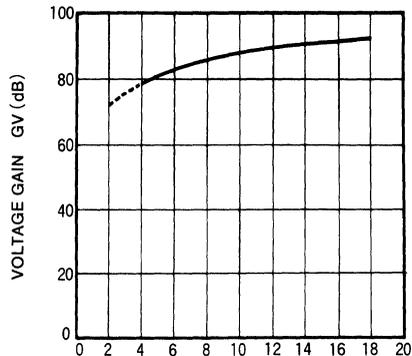
INTERNALLY PHASE COMPENSATED OPERATIONAL AMPLIFIERS

POWER DISSIPATION VS SUPPLY VOLTAGE



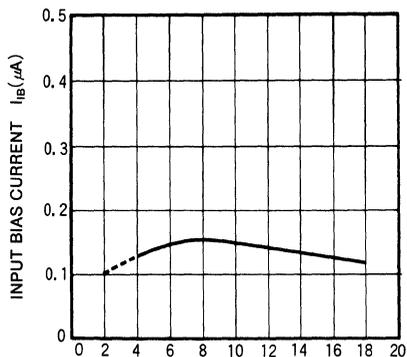
SUPPLY VOLTAGE $V_{CC}(+V), V_{EE}(-V)$

VOLTAGE GAIN VS SUPPLY VOLTAGE



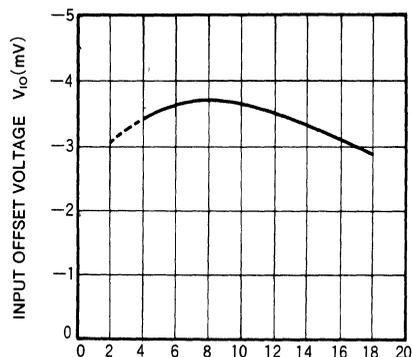
SUPPLY VOLTAGE $V_{CC}(+V), V_{EE}(-V)$

INPUT BIAS CURRENT VS SUPPLY VOLTAGE



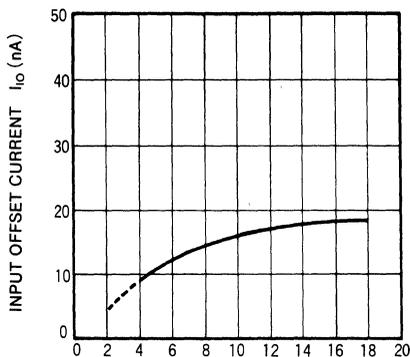
SUPPLY VOLTAGE $V_{CC}(+V), V_{EE}(-V)$

INPUT OFFSET VOLTAGE VS SUPPLY VOLTAGE



SUPPLY VOLTAGE $V_{CC}(+V), V_{EE}(-V)$

INPUT OFFSET CURRENT VS SUPPLY VOLTAGE

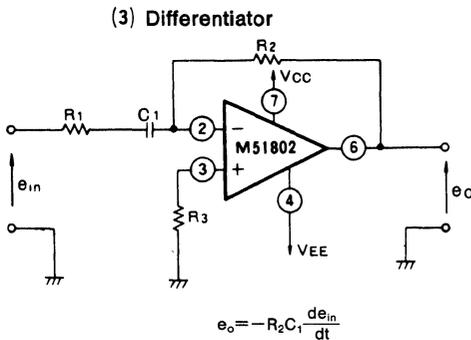
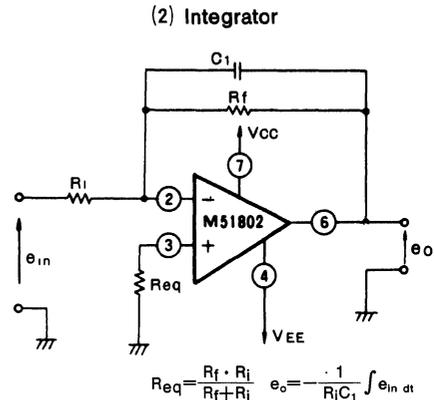
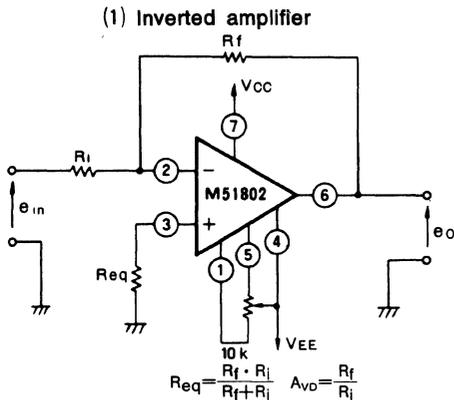


SUPPLY VOLTAGE $V_{CC}(+V), V_{EE}(-V)$

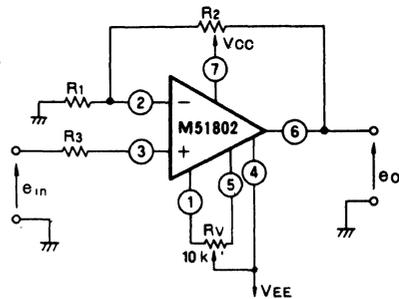
INTERNALLY PHASE COMPENSATED OPERATIONAL AMPLIFIERS

APPLICATION EXAMPLES

An operational amplifier employs feedback and has a variety of functions by the configuration of feedback circuits. Note (1) shows a method adjusting offset voltage.



Note (1) Method of adjusting input offset voltage



- (1) Set e_{in} to equal 0V.
- (2) Adjust $R_v = 10k\Omega$ so that $e_o = 0V$.

DEFINITION OF ABBREVIATED TERMS

- CMRR : Common phase signal rejection ratio. Ratio of output voltage against allowable common phase input voltage and ratio against differential gain.
- G_v : Voltage gain. Ratio of output voltage against input voltage required to change the output voltage from 0V to a fixed value in an open loop circuit.
- I_{IB} : Input bias current. Average current supplied to two input pins when output voltage is set at 0V.
- I_{IO} : Input offset current. Difference of current supplied to two input pins when output voltage is set at 0V.
- P_d : Power dissipation. DC power required to set no load output at 0V in quiescent state.
- R_{in} : Input resistance. Resistance of an integrated circuit measured at one differential input pin by grounding the other input pin.

- SR : Slew rate. Time in which output is increased from normal value of 10% to 90% against step input and ratio of the voltage (voltage between 10% to 90%).
- SVRR : Supply voltage rejection ratio. Ratio of input offset voltage change to supply voltage change.
- t_r, K_{OV} : Transient response. Response against pulse in an open loop amplifier circuit.
- V_{IC} : Common input voltage range. Common input voltage range in which an amplifier operates normally.
- V_{IO} : Input offset voltage. Voltage between input pins when output voltage is set at 0V.
- V_{OPP} : Maximum output voltage width. Maximum output voltage obtained without being clipped.

DUAL LARGE-CURRENT OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

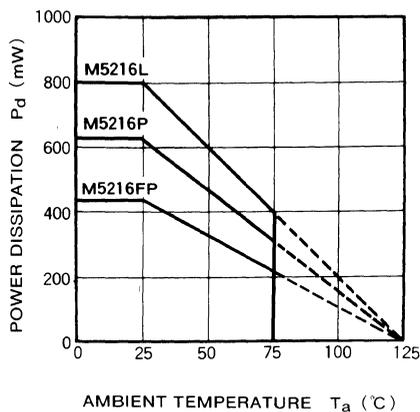
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		± 18	V
I_{LP}	Load current		± 100	mA
V_{id}	Differential input voltage		± 30	V
V_{ic}	Common input voltage		± 15	V
P_d	Power dissipation		800(SIP)/625(DIP)/440(FP)	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	8(SIP)/6.25(DIP)/4.4(FP)	mW/°C
T_{opr}	Ambient temperature		$-20 \sim +75$	°C
T_{stg}	Storage temperature		$-55 \sim +125$	°C

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 15\text{V}$)

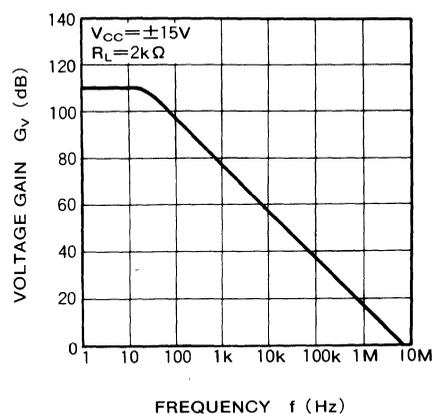
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{in}=0$		4.5	9.0	mA
V_{IO}	Input offset voltage	$R_s \leq 10\text{k}\Omega$		0.5	6.0	mV
I_{IO}	Input offset current			5	200	nA
I_{IB}	Input bias current			180	500	nA
R_{in}	Input resistance		0.3	5		MΩ
G_{VO}	Open loop voltage gain	$R_L \geq 2\text{k}\Omega$, $V_O = \pm 10\text{V}$	86	110		dB
V_{OM}	Maximum output voltage	$R_L \geq 10\text{k}\Omega$	± 12	± 13.5		V
		$R_L \geq 2\text{k}\Omega$	± 10.5	± 11		
V_{CM}	Common input voltage width		± 12	± 14		V
CMRR	Common mode rejection ratio	$R_s \leq 10\text{k}\Omega$	70	90		dB
SVRR	Supply voltage rejection ratio	$R_s \leq 10\text{k}\Omega$		30	150	$\mu\text{V}/\text{V}$
P_d	Power dissipation			135	270	mW
SR	Slew rate	$G_V=0\text{dB}$, $R_L=2\text{k}\Omega$		3.0		$\text{V}/\mu\text{s}$
f_T	Gain bandwidth product			10		MHz
V_{NI}	Input referred noise voltage	$R_s=1\text{k}\Omega$, $\text{BW}=10\text{Hz} \sim 30\text{kHz}$		1.8		μVrms

TYPICAL CHARACTERISTICS

**THERMAL DERATING
(MAXIMUM RATING)**

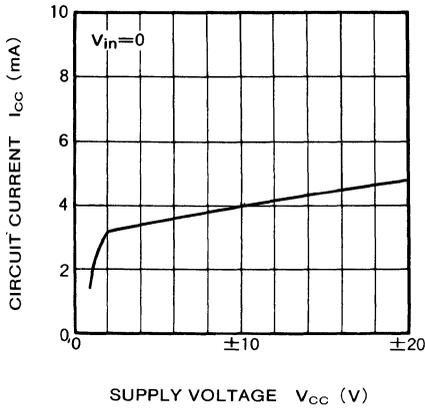


**VOLTAGE GAIN VS.
FREQUENCY RESPONSE**

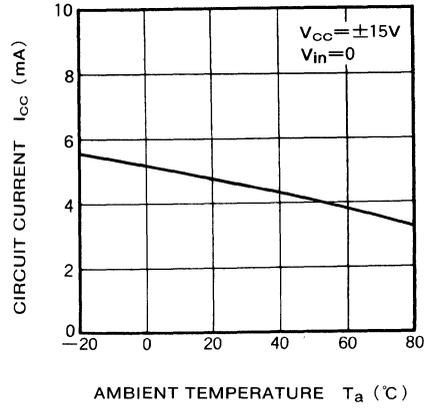


DUAL LARGE-CURRENT OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

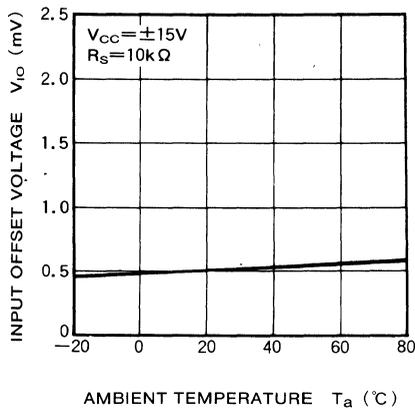
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



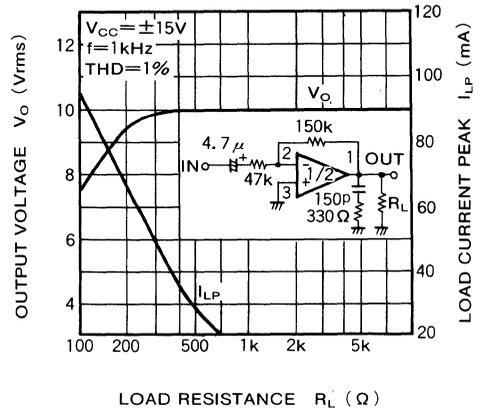
CIRCUIT CURRENT VS. AMBIENT TEMPERATURE



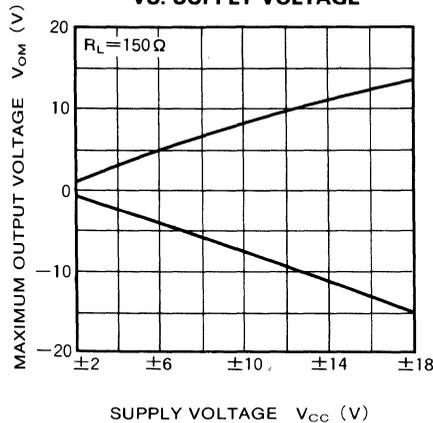
INPUT OFFSET VOLTAGE VS. AMBIENT TEMPERATURE



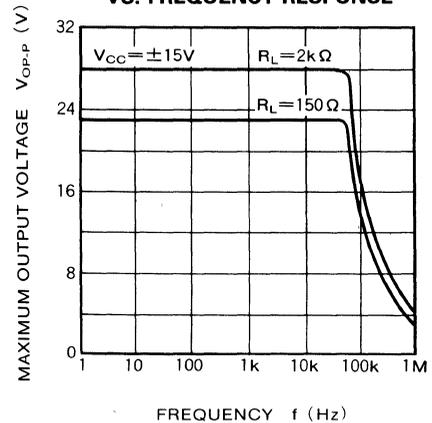
OUTPUT VOLTAGE / LOAD CURRENT PEAK VS. LOAD RESISTANCE



MAXIMUM OUTPUT VOLTAGE VS. SUPPLY VOLTAGE



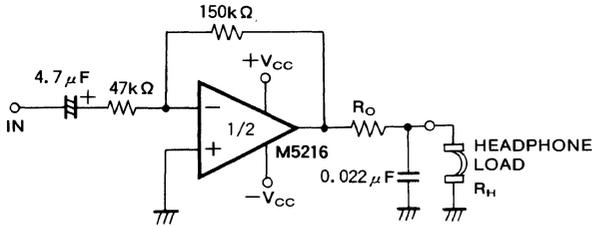
MAXIMUM OUTPUT VOLTAGE VS. FREQUENCY RESPONSE



DUAL LARGE-CURRENT OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

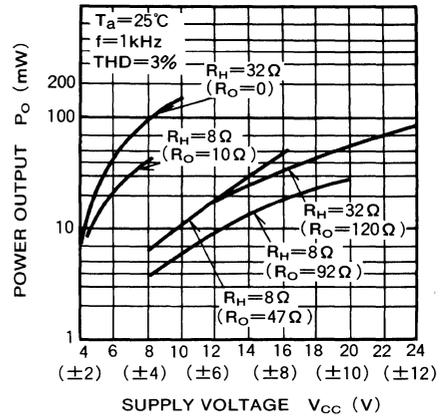
APPLICATION EXAMPLE FOR A HEADPHONE AMPLIFIER (DUAL POWER SUPPLY TYPE)

INVERTED INPUT TYPE



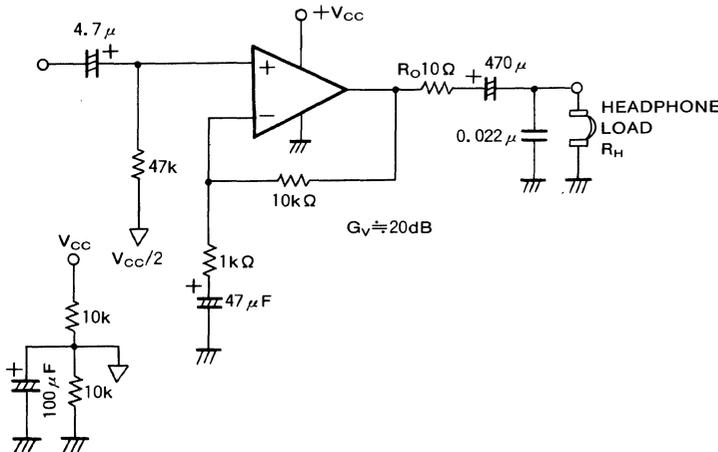
(Note) For a single power supply type, (+) input pin voltage level is shifted at $V_{CC}/2$ and output must be used by AC connection by means of a capacitor

**HEADPHONE AMPLIFIER CIRCUIT
 $P_O - V_{CC}$ CHARACTERISTICS**

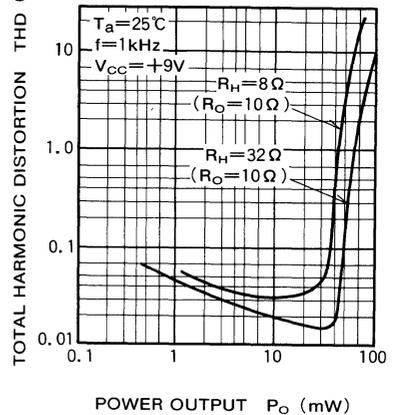


APPLICATION EXAMPLE FOR A HEADPHONE AMPLIFIER (SINGLE POWER SUPPLY TYPE)

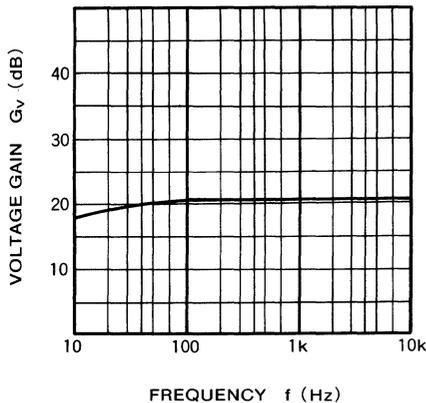
NON-INVERTED INPUT TYPE



**HEADPHONE AMPLIFIER CIRCUIT
 $THD - P_O$ CHARACTERISTICS**

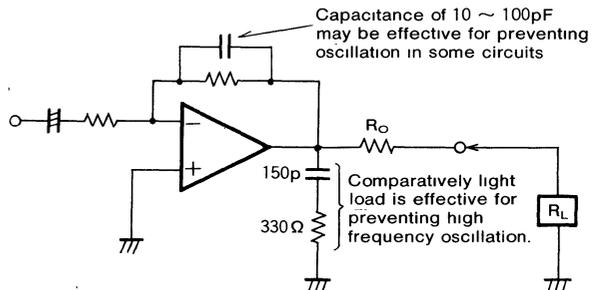


**VOLTAGE GAIN VS.
 FREQUENCY RESPONSE**



**COUNTERMEASURE AGAINST
 OSCILLATION**

If oscillation occurs due to load condition, substrate wiring condition, instability of power supply after the M5216 is mounted on the equipment, the following preventative circuit is recommended.



R_O is recommended because it is effective for preventing capacitive load oscillation and controlling current when load is shorted.

MITSUBISHI LINEAR ICs

M5218L, P, FP / M5R4558P

DUAL LOW-NOISE OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

DESCRIPTION

The M5218/M5R4558P are semiconductor integrated circuits designed for a low noise preamplifier in audio equipment and a general-purpose operational amplifier in other electronic equipment. Two low noise operational amplifier circuits displaying internal phase-compensated high gain and low distortion are contained in an 8-pin SIP, DIP or FP for application over a wide range as a general-purpose dual amplifier in general electronic equipment.

The devices have virtually the same characteristics as the 4557, 4558, 4559 and 741 operational amplifiers.

The units can also be used as a single power supply type and amplifier in portable equipment. It is also suitable as a headphone amplifier because of its high load current.

FEATURES

- High gain, low distortion
 $G_{VO}=110\text{dB}$, $\text{THE}=0.0015\%$ (typ.)
- High slew rate, high f_T
 $\text{SR}=2.2\text{V}/\mu\text{s}$, $f_T=7\text{MHz}$ (typ.)
- Low noise ($R_S=1\text{k}\Omega$) FLAT $V_{NI}=2\mu\text{Vrms}$ (typ.)
 RIAA $V_{NI}=1\mu\text{Vrms}$ (typ.)
- Operation with low supply voltage
 $V_{CC}\geq 4\text{V}(\pm 2\text{V})$
- High load current, high power dissipation
 $I_P=\pm 50\text{mA}$, $P_D=800\text{mW}$ (SIP)
 $P_D=625\text{mW}$ (DIP), $P_D=440\text{mW}$ (FP)

APPLICATION

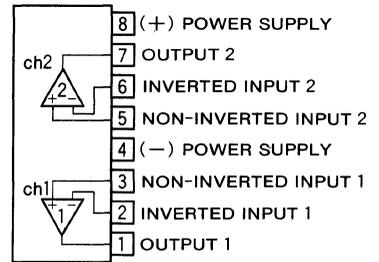
General-purpose amplifier in stereo equipment, tape decks, and radio stereo cassette recorders; active filters, servo amplifiers, operational circuits in other general electronic equipment.

RECOMMENDED OPERATING CONDITIONS

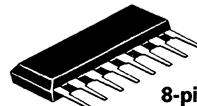
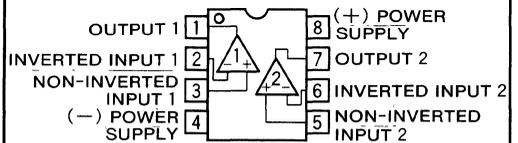
Supply voltage range $\pm 2\sim\pm 15\text{V}$
 Rated supply voltage $\pm 15\text{V}$

PIN CONFIGURATION (TOP VIEW)

SIP



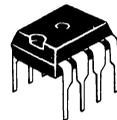
DIP, MINI FLAT



8-pin molded plastic SIP

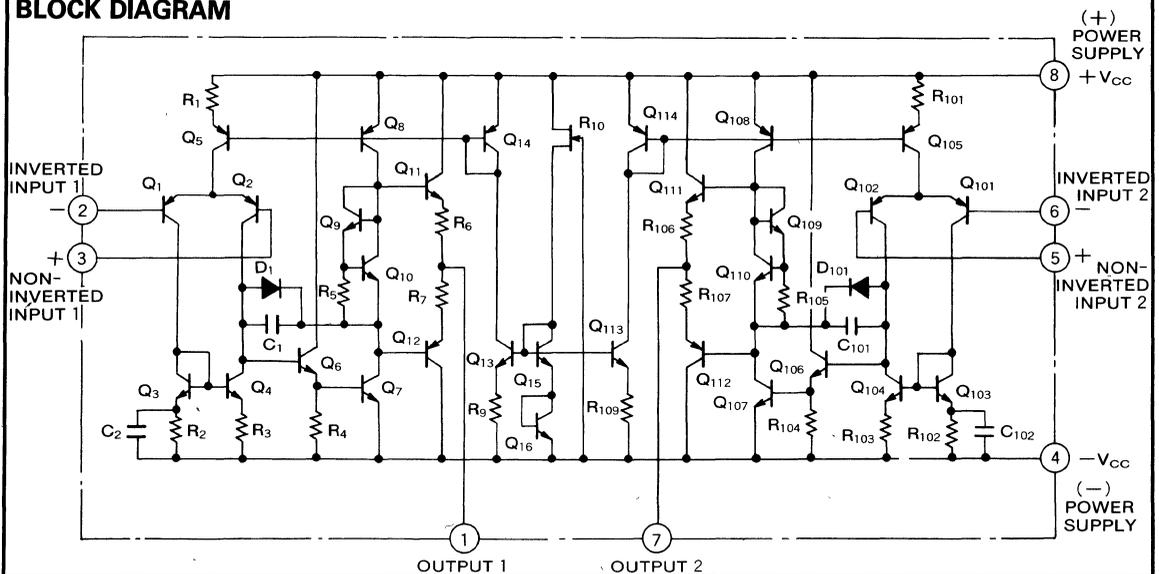


8-pin molded plastic FP (MINI FLAT)



8-pin molded plastic DIP

BLOCK DIAGRAM



MITSUBISHI LINEAR ICs

M5218L, P, FP/M5R4558P

DUAL LOW-NOISE OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

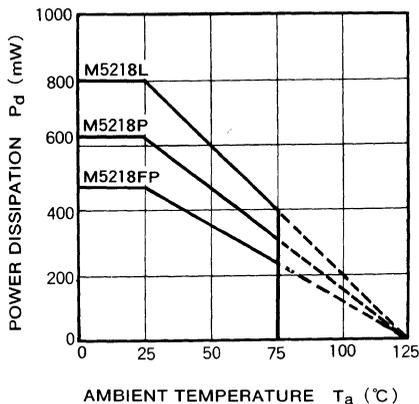
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		± 18	V
I_{LP}	Load current		± 50	mA
V_{id}	Differential input voltage		± 30	V
V_{ic}	Common input voltage		± 15	V
P_d	Power dissipation		800(SIP)/625(DIP)/440(FP)	mW
K_θ	Thermal dirating	$T_a \geq 25^\circ\text{C}$	8(SIP)/6.25(DIP)/4.4(FP)	mW/ $^\circ\text{C}$
T_{opr}	Ambient temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-55 \sim +125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 15\text{V}$)

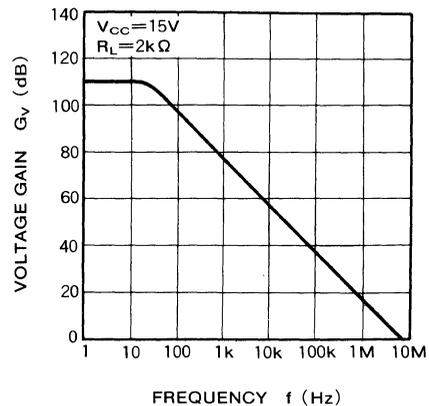
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{in}=0$		3.0	6.0	mA
V_{IO}	Input offset voltage	$R_s \leq 10\text{k}\Omega$		0.5	6.0	mV
I_{iO}	Input offset current			5	200	nA
I_{IB}	Input bias current				500	nA
R_{in}	Input resistance		0.3	5		M Ω
G_{VO}	Open loop voltage gain	$R_L \geq 2\text{k}\Omega$, $V_O = \pm 10\text{V}$	86	110		dB
V_{OM}	Maximum output voltage	$R_L \geq 10\text{k}\Omega$	± 12	± 14		V
		$R_L \geq 2\text{k}\Omega$	± 10	± 13		V
V_{CM}	Common input voltage range		± 12	± 14		V
CMRR	Common mode rejection ratio	$R_s \leq 10\text{k}\Omega$	70	90		dB
SVRR	Sypply voltage	$R_s \leq 10\text{k}\Omega$		30	150	$\mu\text{V}/\text{V}$
P_d	Power dissipation			90	180	mW
SR	Slew rate	$G_v=0\text{dB}$, $R_L=2\text{k}\Omega$		2.2		V/ μs
f_T	Gain bandwidth product			7		MHz
V_{NI}	Input referred noise voltage	$R_s=1\text{k}\Omega$, BW.10Hz \sim 30kHz		2.0		μVrms

TYPICAL CHARACTERISTICS

**THERMAL DERATING
(MAXIMUM RATING)**

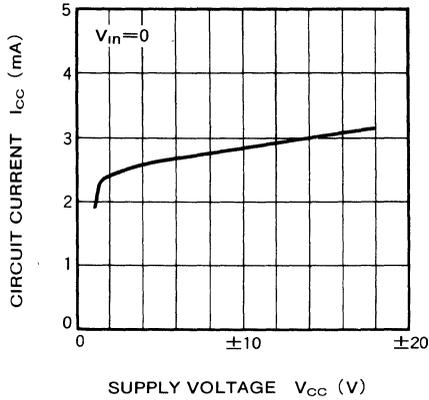


**VOLTAGE GAIN VS.
FREQUENCY RESPONSE**

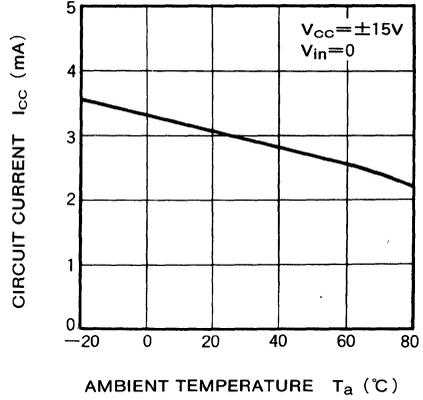


DUAL LOW-NOISE OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

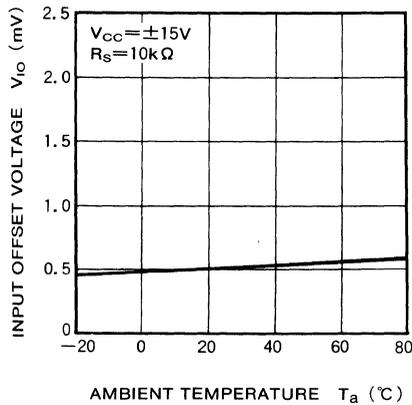
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



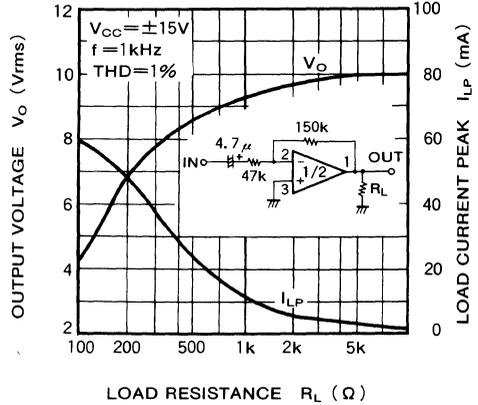
CIRCUIT CURRENT VS. AMBIENT TEMPERATURE



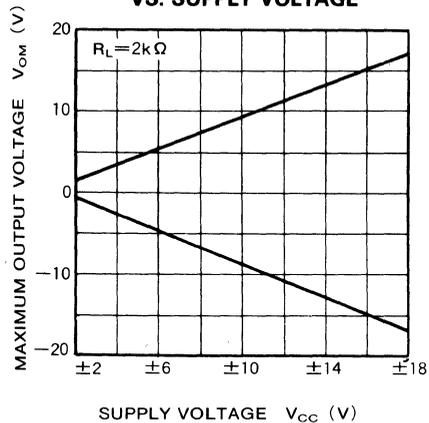
INPUT OFFSET VOLTAGE VS. AMBIENT TEMPERATURE



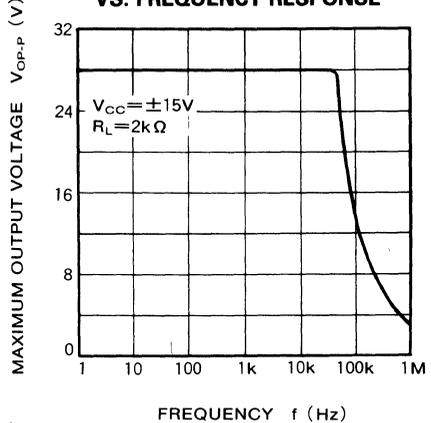
OUTPUT VOLTAGE / LOAD CURRENT PEAK VS. LOAD RESISTANCE



MAXIMUM OUTPUT VOLTAGE VS. SUPPLY VOLTAGE



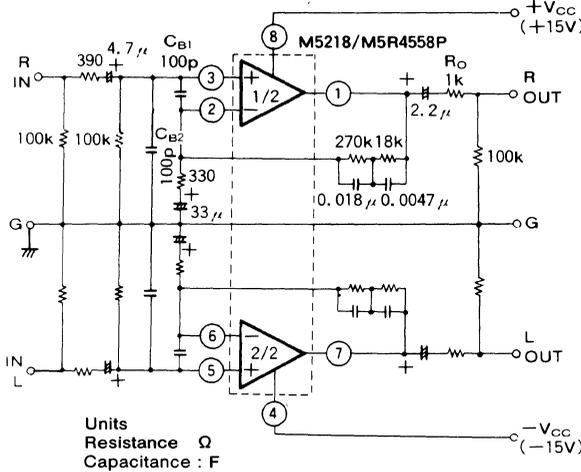
MAXIMUM OUTPUT VOLTAGE VS. FREQUENCY RESPONSE



DUAL LOW-NOISE OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

APPLICATION EXAMPLES

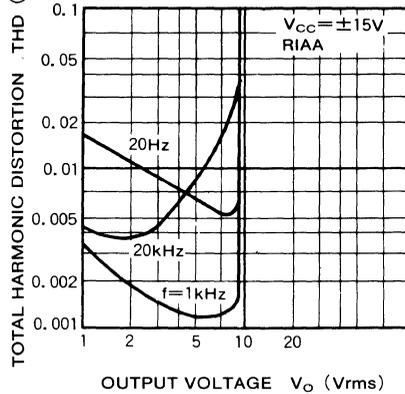
(1) Stereo Equalizer amplifier circuit



TYPICAL CHARACTERISTICS ($V_{CC}=\pm 15V$, RIAA)

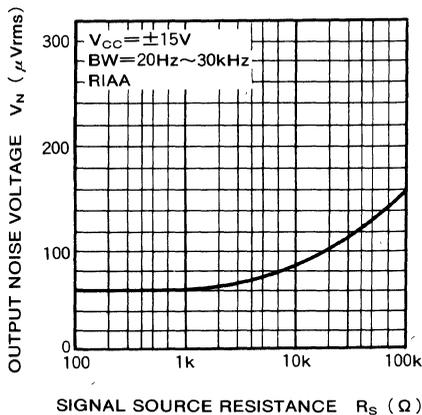
- $G_v=35.6\text{dB}$ ($f=1\text{kHz}$)
- $V_{NI}=1\mu\text{Vrms}$ ($R_S=1\text{k}\Omega$, $\text{BW}=20\text{Hz}\sim 30\text{kHz}$)
- $\text{Signal-to-noise}=72.5\text{dB}$ (IHF-A network, shorted input, 2.5mVrms input sensitivity)
- $\text{THD}=0.0015\%$ ($f=1\text{kHz}$, $V_O=3\text{Vrms}$)

TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE

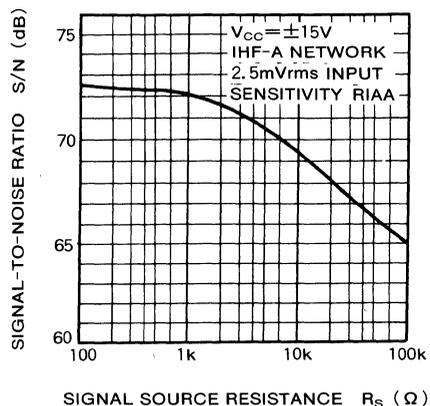


Left channel circuit constants are identical to those of right channel.
 C_{B1} , C_{B2} : Capacitors for buzz prevention, use if required.
 R_O : Resistor used to prevent parasitic oscillation for capacitive loads and current limiting with shorted and other abnormal load conditions.

OUTPUT NOISE VOLTAGE VS. SIGNAL SOURCE RESISTANCE



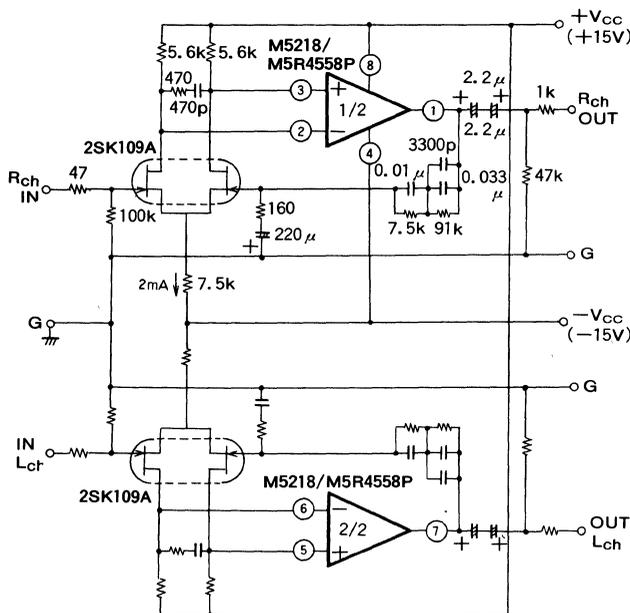
SIGNAL-TO-NOISE RATIO VS. SIGNAL SOURCE RESISTANCE



M5218L, P, FP / M5R4558P

DUAL LOW-NOISE OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

(2) High S / N stereo DC ICL equalizer

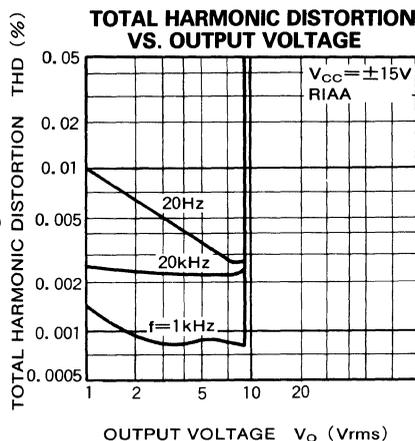


Left channel circuit constants are identical to those of right channel.

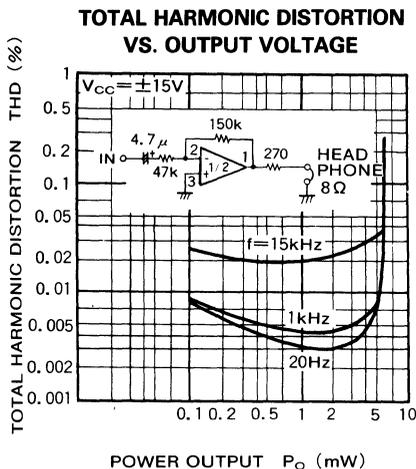
Units Resistance : Ω
Capacitance F

TYPICAL CHARACTERISTICS ($V_{CC} = \pm 15V$, RIAA)

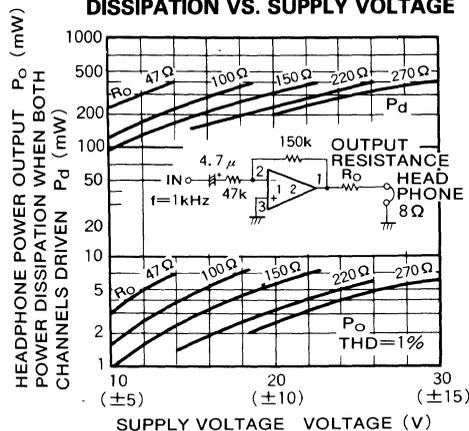
- Signal-to-noise = 72.5dB (IHF-A network, shorted input, 2.5mVrms input sensitivity)
- $V_{NI} = 0.77 \mu V_{rms}$ ($R_s = 5.1k\Omega$, BW = 5Hz ~ 100kHz)
- $G_V = 35.6dB$ ($f = 1kHz$)



(3) Headphone amplifier



(Output resistance R_o is made the parameter) POWER OUTPUT / POWER DISSIPATION VS. SUPPLY VOLTAGE



M5223L, P, FP/M5N358P

DUAL SINGLE POWER SUPPLY OPERATIONAL AMPLIFIERS

DESCRIPTION

The M5223/M5N358P are semiconductor integrated circuits designed as dual operational amplifiers which permit single power supply operation.

The devices come in a compact 8-pin SIP, DIP or FP and contain two circuits for yielding a high internal phase compensation and high performance. For both input and output, operation is possible from the GND level and this makes it possible for the device to be used widely as a general-purpose operational amplifier in the motor control circuits of such equipment as cassette decks, turntables, VTRs and digital audio disc players as well as in automotive electronic products and communications equipment. It can also be employed as a simple comparator.

FEATURES

- Wide common input voltage range and operation permitted with GND level input $V_I = -0.3V \sim +36V$
- Output voltage level can be reduced to near the GND level
- Wide operating supply voltage range and single power supply operation possible $V_{CC} = 3V \sim 36V (V_{CCmax})$
- High voltage gain $G_{VO} = 110dB (typ.)$
- High allowable power dissipation $P_d = 800mW (SIP)$
 $P_d = 625mW (DIP)$
 $P_d = 440mW (FP)$

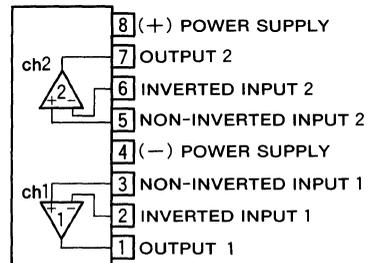
APPLICATION

General-purpose amplifier in control circuits of cassette decks, turntable, VTRs, video disc players and audio disc players; general-purpose amplifier in automotive electronic products, communications equipment and copying machines.

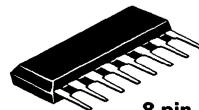
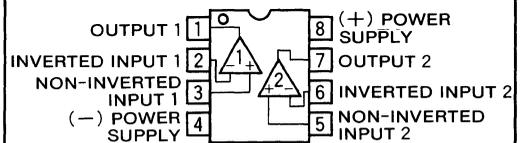
General-purpose amplifier in radio-controlled and electronic toys, and electronic games.

PIN CONFIGURATION (TOP VIEW)

SIP



DIP, MINI FLAT



8-pin molded plastic SIP

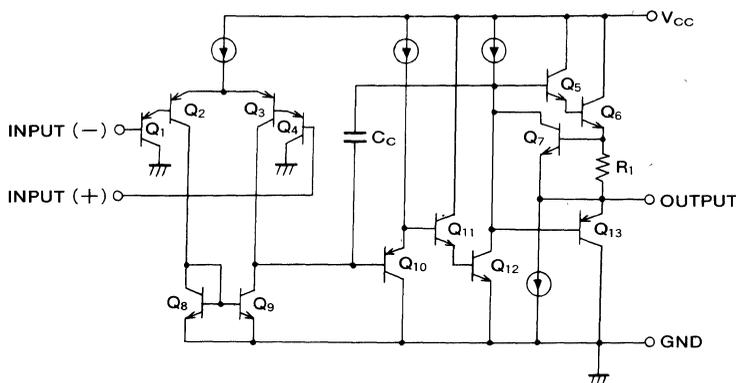


8-pin molded plastic FP (MINI FLAT)



8-pin molded plastic DIP

EQUIVALENT CIRCUIT



* Two circuits are featured in the circuit on the left

M5223 L, P, FP/M5N358P

DUAL SINGLE POWER SUPPLY OPERATIONAL AMPLIFIERS

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

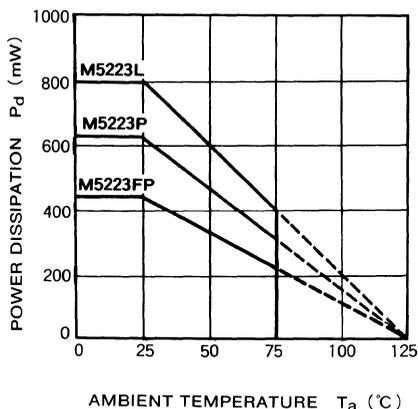
Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		36(±18)	V
I _{sink}	Load current		±50	mA
I _{source}				mA
V _{id}	Differential input voltage		±36	V
V _I	Input voltage		-0.3~+36	V
P _d	Power dissipation		800(SIP)/625(DIP)/440(FP)	mW
K _θ	Thermal derating	T _a ≥25°C	8(SIP)/6.25(DIP)/4.4(FP)	mW/°C
T _{opr.}	Ambient temperature		-20~+75	°C
T _{stg}	Storage temperature		-55~+125	°C

ELECTRICAL CHARACTERISTICS (T_a=25°C, V_{CC}=+5V)

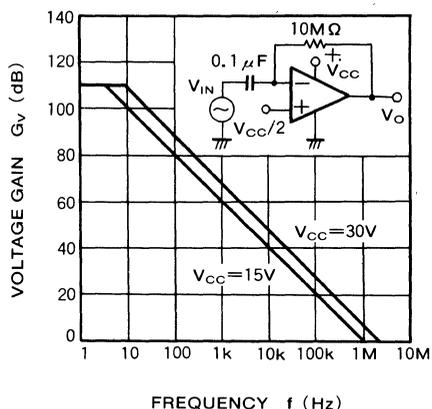
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I _{CC}	Circuit current	V _{in} =0		0.7	1.5	mA
V _{IO}	Input offset voltage	R _S =50Ω, V _{CM} =7.5V		2	7	mV
I _{IO}	Input offset current	I _{B(+)} -I _{B(-)} , V _{CM} =7.5V		5	50	nA
I _B	Input bias current	V _{CM} =7.5V		30	250	nA
CSR	Channel separation	f=1kHz		120		dB
G _{VO}	Open loop voltage gain	R _L ≥2kΩ, V _{CM} =7.5V	90	110		dB
V _{OH}	Maximum output voltage	R _L ≥2kΩ	12.0	13.5		V
V _{OL}		R _L ≥2kΩ		0.9	1.8	
V _{CM}	Common phase input voltage width	R _S =1kΩ, f=100Hz	-0.3		+13.5	V
CMRR	Common phase rejection ratio	R _S =50Ω, V _{CM} =7.5V		85		dB
SVRR	Supply voltage rejection ratio	R _S =50Ω		100		dB
P _d	Power dissipation			10.5	22.5	mW
SR	Slew rate	G _v =0dB, R _L =2kΩ		0.6		V/μs
I _{source}	Output source current	V _{IN(-)} =0V, V _{IN(+)} =1V	20	40		mA
I _{sink}	Output sink current	V _{IN(-)} =1V, V _{IN(+)} =0V	10	20		mA

TYPICAL CHARACTERISTICS

THERMAL DERATING (MAXIMUM RATING)

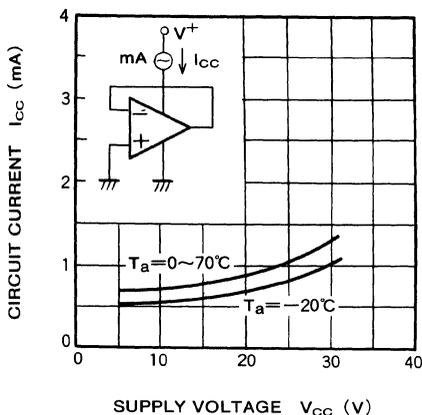


VOLTAGE GAIN VS. FREQUENCY RESPONSE

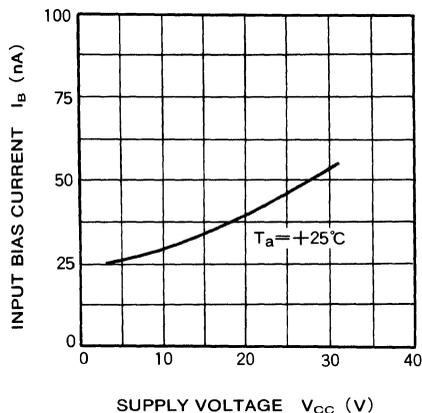


DUAL SINGLE POWER SUPPLY OPERATIONAL AMPLIFIERS

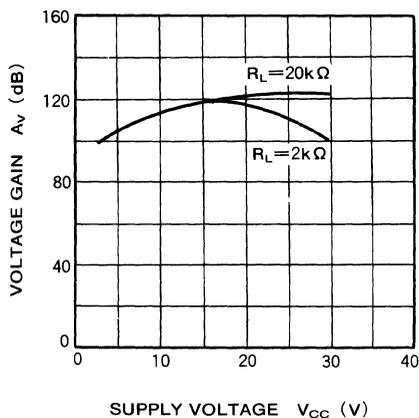
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



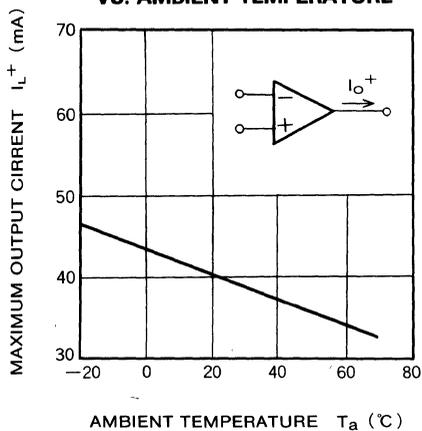
INPUT BIAS CURRENT VS. SUPPLY VOLTAGE



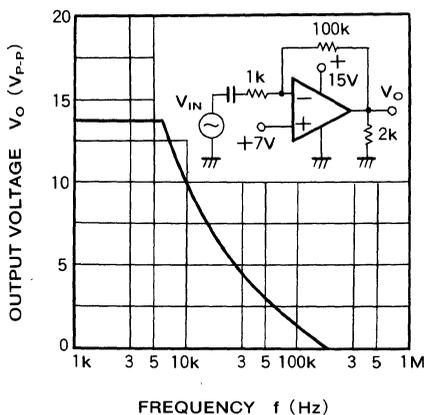
VOLTAGE GAIN VS. SUPPLY VOLTAGE



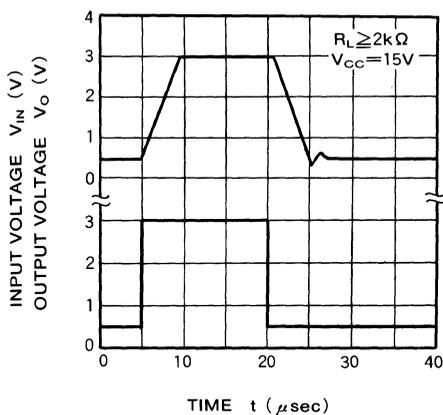
MAXIMUM OUTPUT CURRENT VS. AMBIENT TEMPERATURE



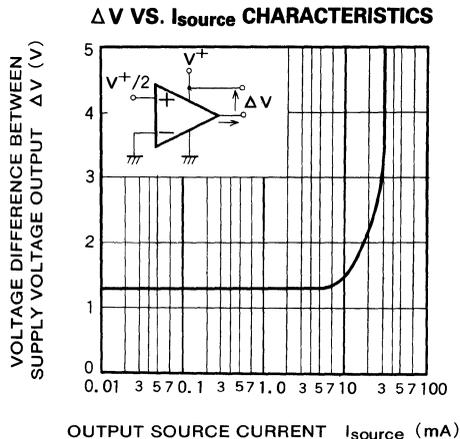
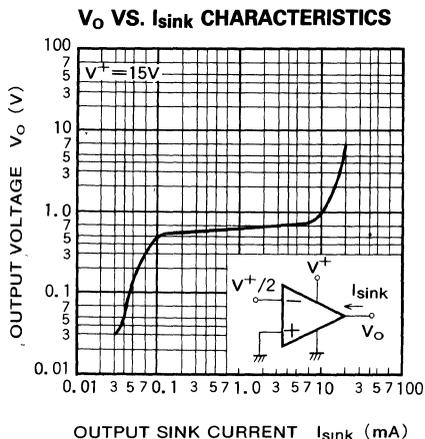
MAXIMUM OUTPUT VOLTAGE VS. FREQUENCY RESPONSE



INPUT VOLTAGE / OUTPUT VOLTAGE VS. TIME



DUAL SINGLE POWER SUPPLY OPERATIONAL AMPLIFIERS



BASIC OPERATION OF SINGLE POWER SUPPLY OPERATIONAL AMPLIFIER

The M5223/M5N358P is an operational amplifier that operates from 0V (GND) level for both input and output if used at a single-power supply voltage.

Basic operation of the device is explained in the following, comparing characteristics of the operational amplifier M5218/M5R4558P.

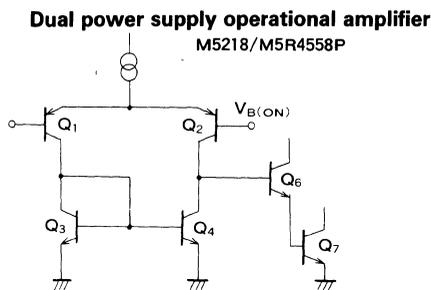


Fig. 1

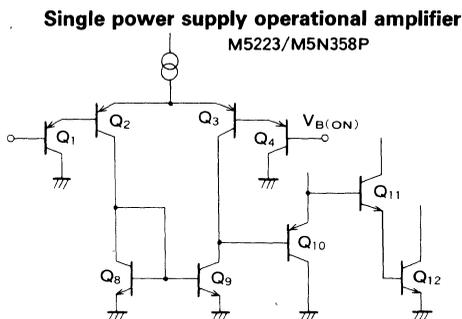


Fig. 2

Fig. 1 illustrates the dual power supply operational amplifier M5218/M5R4558P, and Fig. 2 illustrates an input differential circuit of the single power supply operational amplifier M5223/M5N358P. In this case, the input stage base voltage $V_{B(ON)}$ for operating an input differential circuit in the M5218/M5R4558P is as follow;

$$V_{B(ON)} = V_{BE7} + V_{BE6} + V_{CE2} - V_{BE2}$$

(If $V_{BE} \cong 0.6V$, $V_{CE} \cong 0.5V$)

$$\cong 1.1V$$

and, in the M5223/M5N358P,

$$V_{B(ON)} = V_{BE12} + V_{BE11} - V_{BE10} + V_{CE3} - V_{BE3} - V_{BE4}$$

$$\cong 1.1V$$

and, in the M5223/M5N358P, the differential circuit is activated even when the input level is 0V. The input and output characteristics are shown in Fig. 3.

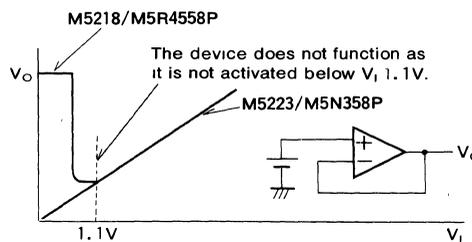


Fig. 3

Operation of an output stage is explained in the following.

DUAL SINGLE POWER SUPPLY OPERATIONAL AMPLIFIERS

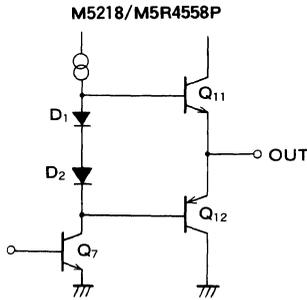


Fig. 4

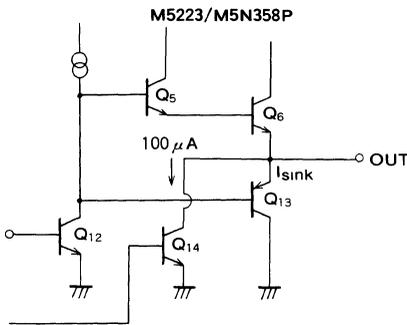


Fig. 5

Fig. 4 illustrates an output stage of the M5218/M5R4558P, which adopts an AB-class push-pull circuit of ordinary operational amplifier output type, where no crossover distortion occurs because idling current is running from D_1, D_2 bias to Q_{11}, Q_{12} in quiescent state. In this case, the output never goes below the level of $V_{BE12} + V_{CE7}$, and the device is activated up to a voltage of approximately 1.1V. Moreover, the voltage changes greatly according to conditions of load current.

Fig. 5 illustrates the M5223/M5N358P, to which rated current circuit of $I_C \approx 100 \mu A$ is connected by Q_{14} and the output can be reduced to near GND level as A-class bias output stage up to the current of $100 \mu A$. For a load in excess of this current, in case of the M5223/M5N358P driven by Q_{13} , no idling current is present because of C-class bias (where the base of output transistors Q_5, Q_6, Q_{13} is shorted), and crossover distortion occurs. Therefore, the device may not be suitable for audio signal amplifiers.

Both single and dual power supply amplifiers operate exactly with the same operating circuit logic (activation level of input and output and load driving methods are different), but bias must be set at $V_{CC}/2$ for output DC current in both single and dual power supply amplifiers.

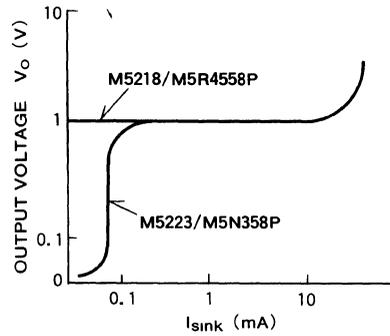


Fig. 6

REDUCTION OF DISTORTION IN A SINGLE POWER SUPPLY OPERATIONAL AMPLIFIER

As has been explained previously, a single power supply operational amplifier operates with low distortion as an A-class bias circuit up to a load condition that can be driven by current ($100 \mu A$) in a rated current circuit which is built in the output stage, and it can be used for audio signals but, if the load condition exceeds the value of current, the device is placed into a C-class bias condition, and crossover distortion occurs. To reduce this distortion, a pull-up resistor (e.g. $3k \Omega$) for running A-class bias current externally can be connected as shown in Fig. 7 to increase the A-class bias current and reduce distortion. (Refer to Fig. 8 DISTORTION VS OUTPUT VOLTAGE.)

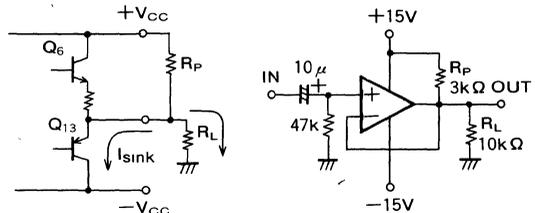


Fig. 7 (Explanation by dual power supply method)

TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE

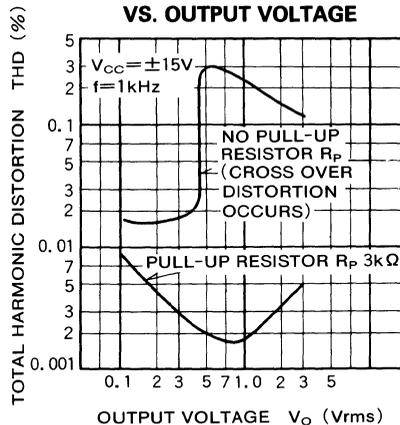


Fig. 8

M5224P, FP/M5N324P

QUAD SINGLE POWER SUPPLY OPERATIONAL AMPLIFIERS

DESCRIPTION

The M5224/M5N324P are semiconductor integrated circuits designed as quad operational amplifiers in which single power supply operation is possible.

The devices come in a standard 14-pin DIP, FP and contain four circuits for yielding a high internal phase compensation and high performance. For both input and output, operation is possible from the GND level and this makes it possible for the device to be used widely as a general-purpose operational amplifier in the motor control circuits of such equipment as cassette decks, turntables, VTRs and digital audio disc players as well as in automotive electronic products and communications equipment. It can be also employed as a simple comparator.

FEATURES

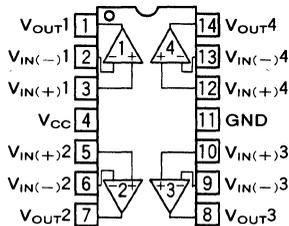
- Wide common input voltage range and operation permitted with GND level input $V_i = -0.3V \sim +36V$
- Output voltage level can be reduced to near the GND level
- Wide operating supply voltage range and single power supply operation possible $V_{CC} = 3V \sim +36V (V_{CCmax})$
- High voltage gain $G_{VO} = 110dB (typ.)$
- High allowable power dissipation
 $P_d = 700mW (M5224P/M5N324P)$
 $P_d = 550mW (M5224FP)$

APPLICATION

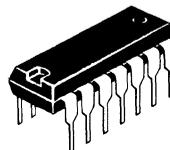
General-purpose amplifier in control circuits of cassette decks, turntable, VTRs, video disc players and audio disc players; general-purpose amplifier in automotive electronic products, communications equipment and copying machines.

General-purpose amplifier in radio-controlled and electronic toys, and electronic games.

PIN CONFIGURATION (TOP VIEW) DIP, MINI FLAT

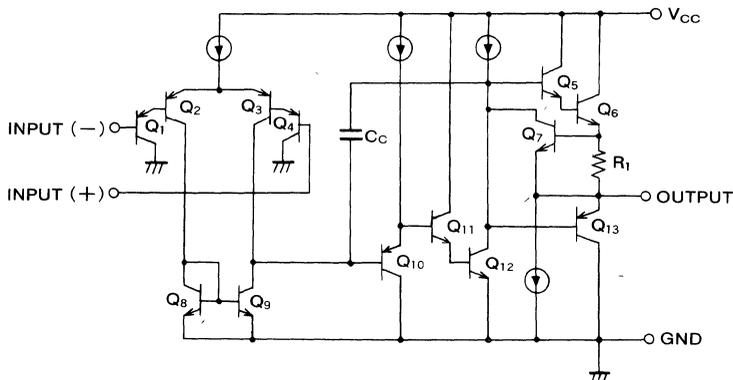


14-pin molded plastic FP
(MINI FLAT)



14-pin molded plastic DIP

EQUIVALENT CIRCUIT



* Four circuits are featured in the circuit on the left

QUAD SINGLE POWER SUPPLY OPERATIONAL AMPLIFIERS

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

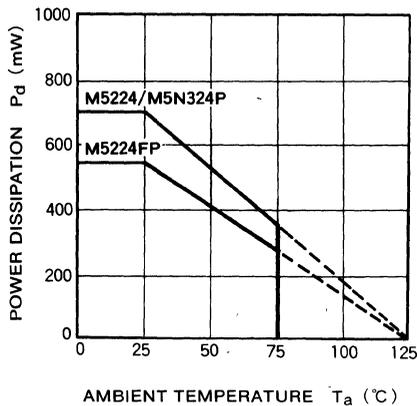
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		36(± 18)	V
I_{sink}	Load current		± 50	mA
I_{source}				mA
V_{id}	Differential input voltage		± 36	V
V_i	Input voltage		-0.3~+36	V
P_d	Power dissipation		700(DIP)/550(FP)	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	7(DIP)/5.5(FP)	mW/ $^\circ\text{C}$
T_{opr}	Ambient temperature		-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature		-55~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=+15\text{V}$)

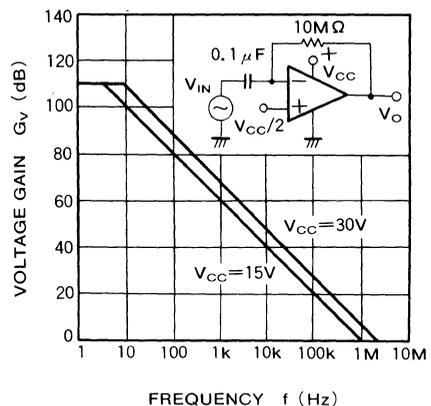
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{in}=0$		1.2	2.5	mA
V_{IO}	Input offset voltage	$R_S=50\Omega$, $V_{CM}=7.5\text{V}$		2	7	mV
I_{IO}	Input offset current	$ I_{B(+)} - I_{B(-)} $, $V_{CM}=7.5\text{V}$		5	50	nA
I_{IB}	Input bias current	$V_{CM}=7.5\text{V}$		30	250	nA
CSR	Channel separation	$f=1\text{kHz}$		120		dB
G_{VO}	Open loop voltage gain	$R_L \geq 2\text{k}\Omega$, $V_{CM}=7.5\text{V}$		90	110	dB
V_{OH}	Maximum output voltage	$R_L \geq 2\text{k}\Omega$		12.0	13.5	V
V_{OL}		$R_L \geq 2\text{k}\Omega$		0.9	1.8	
V_{CM}	Common phase input voltage width	$R_S=1\text{k}\Omega$, $f=100\text{Hz}$		-0.3	+13.5	V
CMRR	Common phase rejection ratio	$R_S=50\Omega$, $V_{CM}=7.5\text{V}$		85		dB
SVRR	Supply voltage rejection ratio	$R_S=50\Omega$		100		dB
P_d	Power dissipation			18	37.5	mW
SR	Slew rate	$G_V=0\text{dB}$, $R_L=2\text{k}\Omega$		0.6		V/ μs
I_{source}	Output source current	$V_{IN(-)}=0\text{V}$, $V_{IN(+)}=1\text{V}$		20	40	mA
I_{sink}	Output sink current	$V_{IN(-)}=1\text{V}$, $V_{IN(+)}=0\text{V}$		10	20	mA

TYPICAL CHARACTERISTICS

**THERMAL DERATING
(MAXIMUM RATING)**

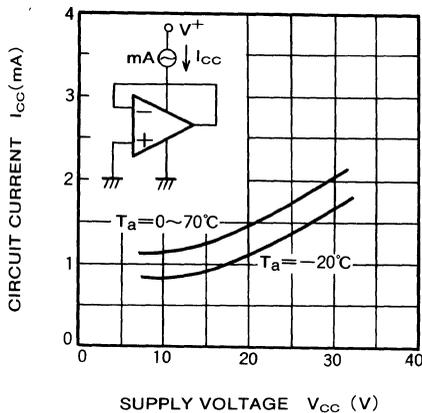


**VOLTAGE GAIN VS.
FREQUENCY RESPONSE**

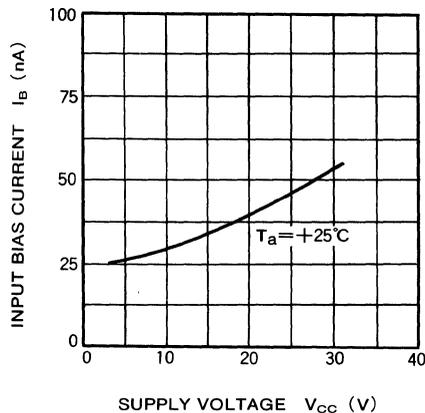


QUAD SINGLE POWER SUPPLY OPERATIONAL AMPLIFIERS

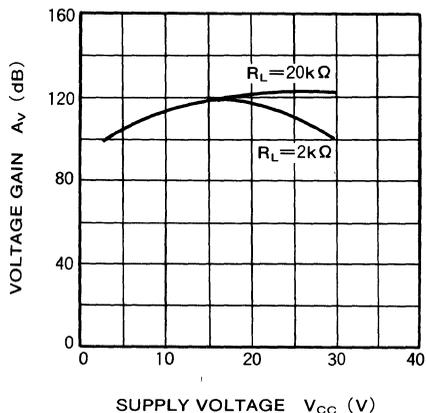
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



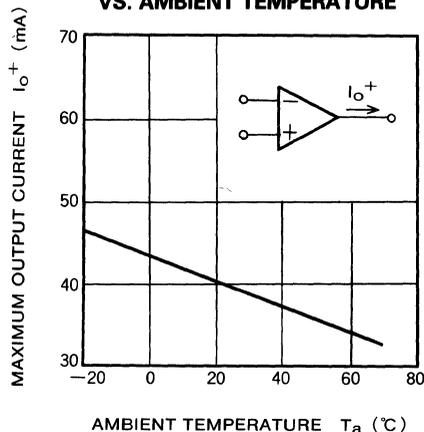
INPUT BIAS CURRENT VS. SUPPLY VOLTAGE



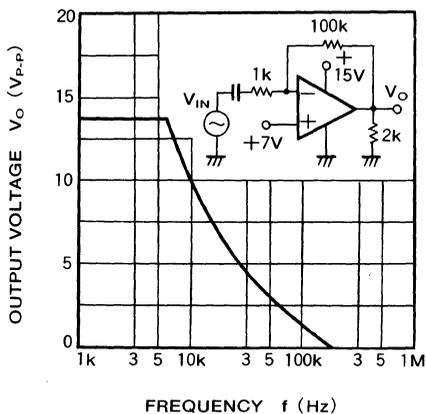
VOLTAGE GAIN VS. SUPPLY VOLTAGE



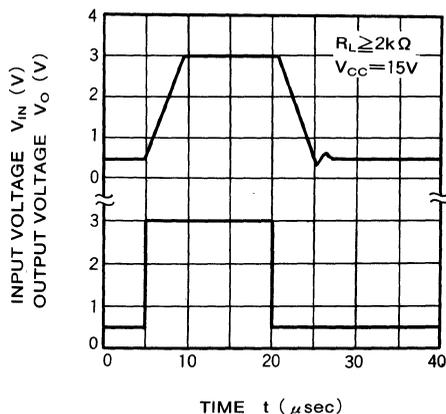
MAXIMUM OUTPUT CURRENT VS. AMBIENT TEMPERATURE



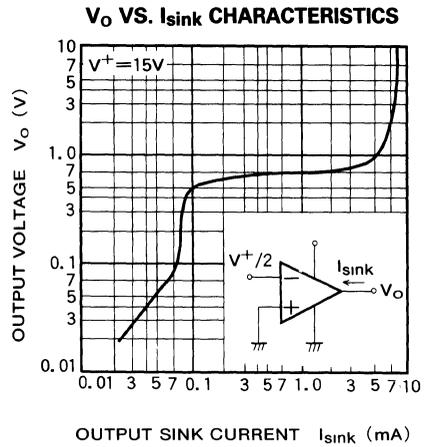
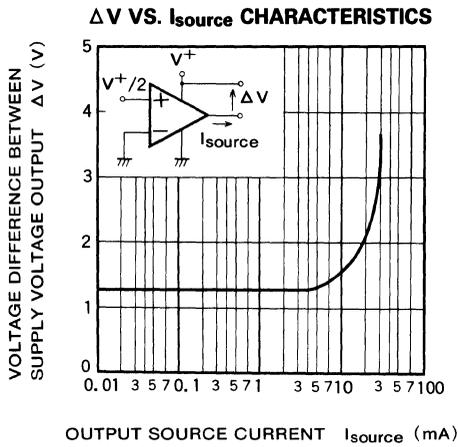
MAXIMUM OUTPUT VOLTAGE VS. FREQUENCY RESPONSE



INPUT VOLTAGE / OUTPUT VOLTAGE VS. TIME



QUAD SINGLE POWER SUPPLY OPERATIONAL AMPLIFIERS



BASIC OPERATION OF SINGLE POWER SUPPLY OPERATIONAL AMPLIFIER

The M5224/M5N324P is an operational amplifier that operates from 0V (GND) level for both input and output if used at a single power supply voltage.

Basic operation of the device is explained in the following, comparing characteristics of the operational amplifier M5218/M5R4558P.

Fig. 1 illustrates the dual power supply operational amplifier M5218/M5R4558P, and Fig. 2 illustrates an input differential circuit of the single power supply operational amplifier M5224/M5N324P. In this case, the input stage base voltage V_{B(ON)} for operating an input differential circuit in the M5218/M5R4558P is as follow;

$$V_{B(ON)} = V_{BE7} + V_{BE6} + V_{CE2} - V_{BE2}$$

(If $V_{BE} \cong 0.6V$, $V_{CE} \cong 0.5V$)
 $\cong 1.1V$

and, in the M5224/M5N324P,

$$V_{B(ON)} = V_{BE12} + V_{BE11} - V_{BE10} + V_{CE3} - V_{BE3} - V_{BE4}$$

$\cong -0.1V$

and, in the M5224/M5N324P, the differential circuit is activated even when the input level is 0V. The input and output characteristics are shown in Fig. 3.

Dual power supply operational amplifier
 M5218/M5R4558P

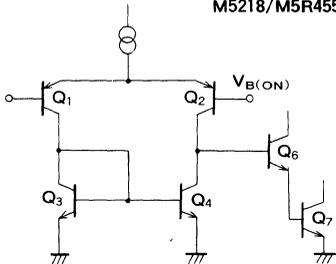


Fig. 1

Single power supply operational amplifier
 M5224/M5N324P

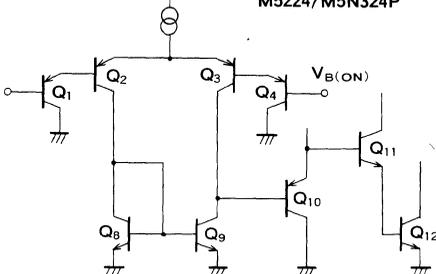


Fig. 2

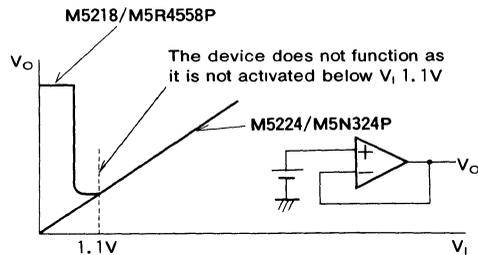


Fig. 3

Operation of an output stage is explained in the following.

QUAD SINGLE POWER SUPPLY OPERATIONAL AMPLIFIERS

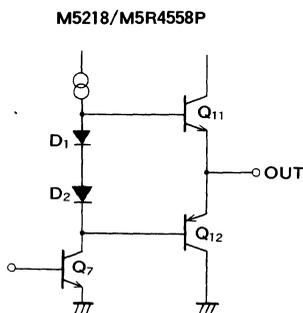


Fig. 4

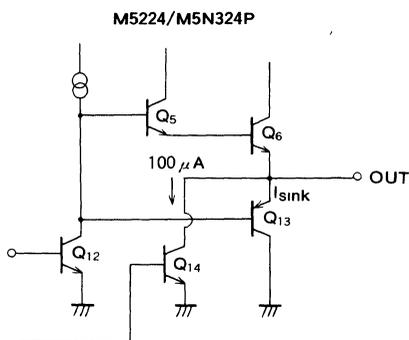


Fig. 5

Fig. 4 illustrates an output stage of the M5218/M5R4558P, which adopts an AB-class push-pull circuit of ordinary operational amplifier output type, where no crossover distortion occurs because idling current is running from D_1 , D_2 bias to Q_{11} , Q_{12} in quiescent state. In this case, the output never reaches below the level of $V_{BE12} + V_{CE7}$, and the device is activated up to voltage of approximately 1.1V. Moreover, the voltage changes greatly according to conditions of load current.

Fig. 5 illustrates the M5224/M5N324P, to which rated current circuit of $I_C \approx 100 \mu A$ is connected by Q_{14} and the output can be reduced to near GND level as A-class bias output stage up to the current of $100 \mu A$. For a load in excess of this current, is case of the M5224/M5N324P driven by Q_{13} , no idling current is present because of C-class bias (where the base of output transistors Q_5 , Q_6 , Q_{13} are shorted), and crossover distortion occurs. Therefore, the device may not be suitable for audio signal amplifiers.

Both single and dual power supply amplifiers operate exactly with the same operating circuit logic (activation level of input and output and load driving methods are different), but bias must be set at $V_{CC}/2$ for output DC current in both single and dual power supply amplifiers.

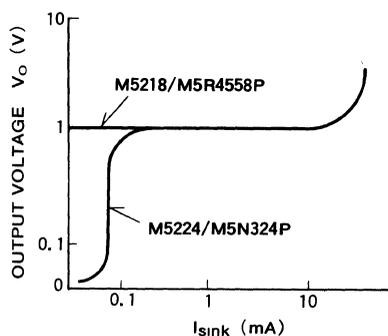


Fig. 6

REDUCTION OF DISTORTION IN A SINGLE POWER SUPPLY OPERATIONAL AMPLIFIER

As has been explained previously, a single power supply operational amplifier operates with low distortion as an A-class bias circuit up to a load condition that can be driven by current ($100 \mu A$) in a rated current circuit which is built in the output stage, and it can be used for audio signals but, if the load condition exceeds the value of current, the device is placed into a C-class bias condition, and crossover distortion occurs. To reduce this distortion, a pull-up resistor (e.g. $3k \Omega$) for running A-class bias current externally can be connected as shown in Fig. 7 to increase the A-class bias current and reduce distortion. (Refer to Fig. 8 DISTORTION VS OUTPUT VOLTAGE.)

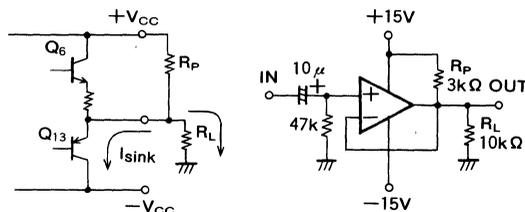


Fig. 7 (Explanation by dual power supply method)

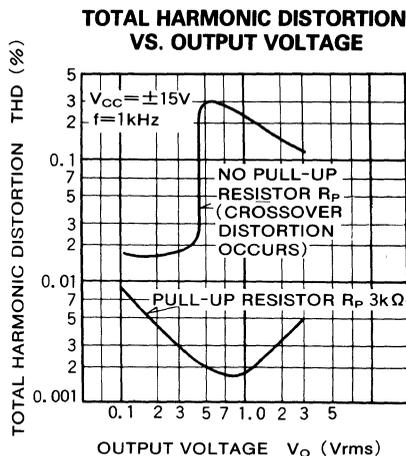


Fig. 8

MITSUBISHI LINEAR ICs

M5228P, FP

QUAD LOW-NOISE OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

DESCRIPTION

The M5228 is a semiconductor integrated circuit designed for a low-noise preamplifier in audio equipment and a general-purpose operational amplifier in other electronic equipment. Four low-noise operational amplifier circuits displaying internal phase-compensated high gain and low distortion are contained in a 14-pin standard DIP and 14-pin mini flat (FP) package for application over a side range as a general-purpose dual amplifier in general electronic equipment.

The device has virtually the same characteristics as the 4557, 4558, 4559 and 741 operational amplifiers. The unit can also be used as a single power supply type and amplifier in portable equipment. It is also suitable as a headphone amplifier because of its high load current.

FEATURES

- High gain, low distortion
 $G_{VO} = 110\text{dB}$, $\text{THD} = 0.0015\%$ (typ.)
- High slew rate, high f_T $\text{SR} = 2.2\text{V}/\mu\text{s}$, $f_T = 7\text{MHz}$ (typ.)
- Low noise ($R_G = 1\text{k}\Omega$) FLAT $V_{NI} = 2\mu\text{Vrms}$ (typ.)
 RIAA $V_{NI} = 1\mu\text{Vrms}$ (typ.)
- Operation with low supply voltage $V_{CC} \geq 4\text{V} (\pm 2\text{V})$
- High load current, high power dissipation
 $I_{LP} = \pm 50\text{mA}$, $P_d = 700\text{mW}$ (M5228P)
 $P_d = 550\text{mW}$ (M5228FP)

APPLICATION

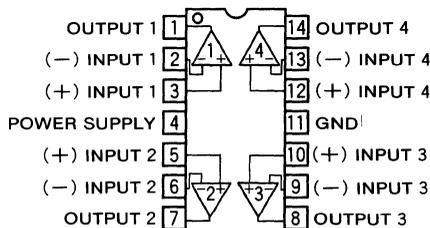
General-purpose amplifier in stereo equipment, tape decks and radio stereo cassette recorders; active filters, servo amplifiers, operational circuits in other general electronic equipment.

RECOMMENDED OPERATING CONDITIONS

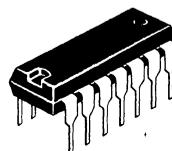
- Supply voltage range $\pm 2 \sim \pm 16\text{V}$
- Rated supply voltage $\pm 15\text{V}$

PIN CONFIGURATION (TOP VIEW)

DIP, MINI FLAT

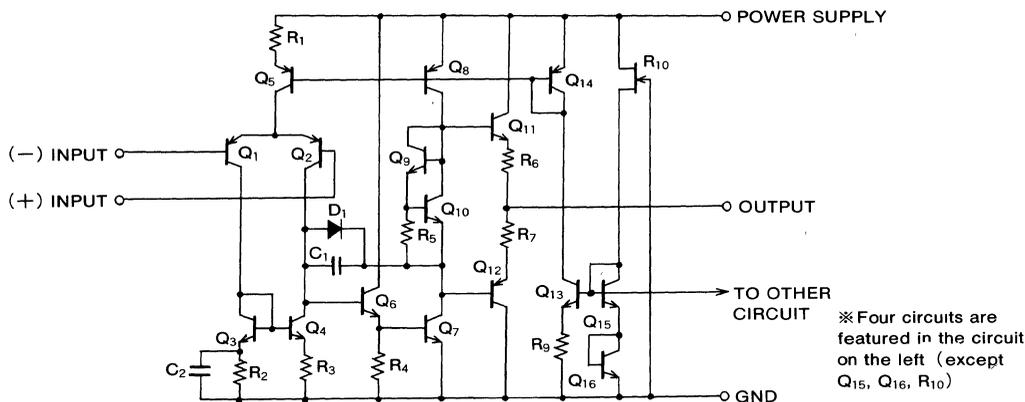


14-pin molded plastic FP
(MINI FLAT)



14-pin molded plastic DIP

EQUIVALENT CIRCUIT



QUAD LOW-NOISE OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

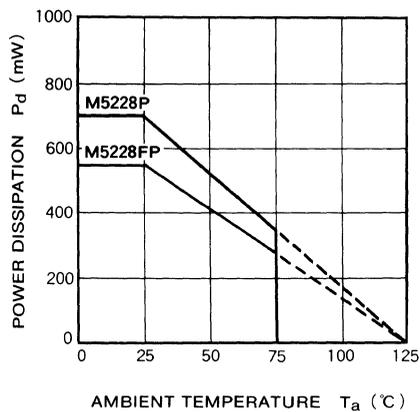
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		± 18	V
I_{LP}	Load current		± 50	mA
V_{id}	Differential input voltage		± 30	V
V_{ic}	Common input voltage		± 15	V
P_d	Power dissipation		700(DIP)/550(FP)	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	7(DIP)/5.5(FP)	mW/ $^\circ\text{C}$
T_{opr}	Ambient temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-55 \sim +125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 15\text{V}$)

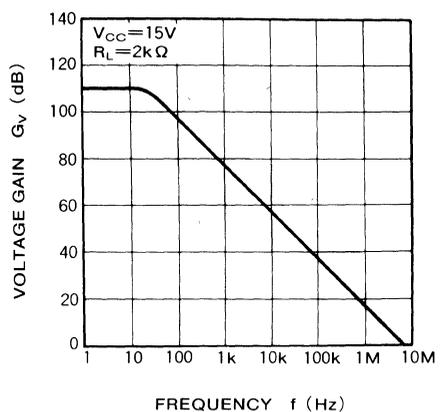
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{in}=0$		6.0	12.0	mA
V_{IO}	Input offset voltage	$R_s \leq 10\text{k}\Omega$		0.5	6.0	mV
I_{IO}	Input offset current			5	200	nA
I_{IB}	Input bias current				500	nA
R_{in}	Input resistance		0.3	5		M Ω
G_{VO}	Open loop voltage gain	$R_L \geq 2\text{k}\Omega, V_O = \pm 10\text{V}$	86	110		dB
V_{OM}	Maximum output voltage	$R_L \geq 10\text{k}\Omega$ $R_L \geq 2\text{k}\Omega$	± 12 ± 10	± 14 ± 13		V
V_{CM}	Common input voltage range		± 12	± 14		V
CMRR	Common mode rejection ratio	$R_s \leq 10\text{k}\Omega$	70	90		dB
SVRR	Supply voltage rejection ratio	$R_s \leq 10\text{k}\Omega$		30	150	$\mu\text{V/V}$
P_d	Power dissipation			180	360	mW
SR	Slew rate	$G_v=0\text{dB}, R_L=2\text{k}\Omega$		2.2		V/ μs
f_T	Gain bandwidth product			7		MHz
V_{NI}	Input referred noise voltage	$R_s=1\text{k}\Omega, \text{BW } 10\text{Hz} \sim 30\text{kHz}$		2.0		μVrms

TYPICAL CHARACTERISTICS

**THERMAL DERATING
(MAXIMUM RATING)**

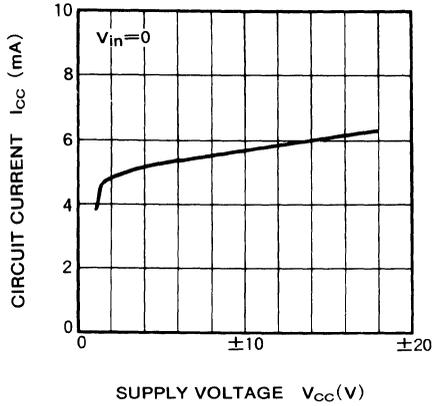


**VOLTAGE GAIN VS.
FREQUENCY RESPONSE**

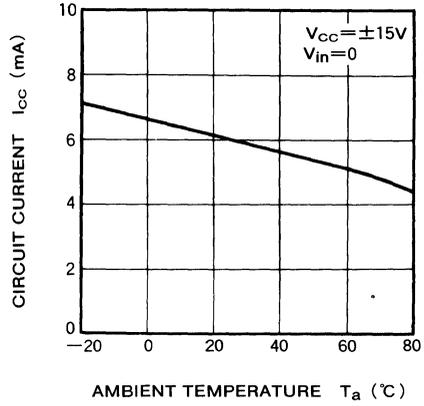


QUAD LOW-NOISE OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

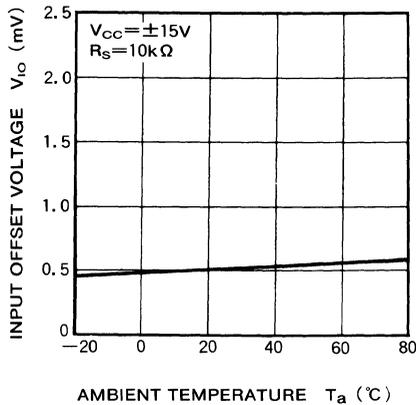
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



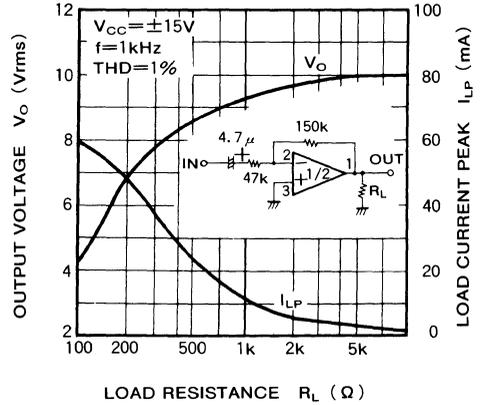
CIRCUIT CURRENT VS. AMBIENT TEMPERATURE



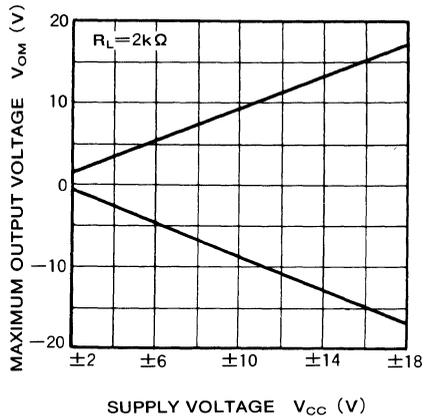
INPUT OFFSET VOLTAGE VS. AMBIENT TEMPERATURE



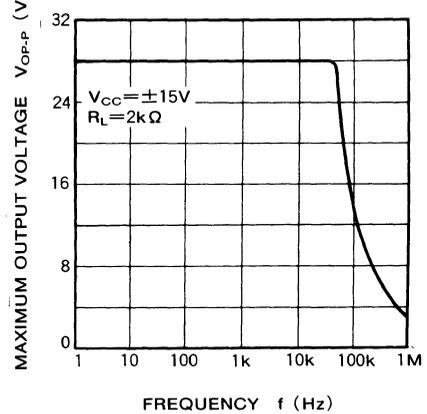
OUTPUT VOLTAGE / LOAD CURRENT PEAK VS. LOAD RESISTANCE



MAXIMUM OUTPUT VOLTAGE VS. SUPPLY VOLTAGE



MAXIMUM OUTPUT VOLTAGE VS. FREQUENCY RESPONSE



MITSUBISHI LINEAR ICs

M5209L, P, FP

DUAL HIGH-VOLTAGE, HIGH S/N OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

DESCRIPTION

The M5209 is a semiconductor integrated circuit designed for a preamplifier in audio equipment of stereo and cassette tape decks.

Two low-noise operational amplifier circuits displaying internal phase-compensated high gain and low distortion are contained in a 8-pin SIP, DIP or FP, suitable for application as a microphone and tone control amplifier of stereo equipment and cassette tape decks.

The unit can also be used as a general-purpose amplifier in portable equipment such as a stereo cassette tape recorder of a single power supply type as it operates at a low supply voltage.

FEATURES

- Low noise $V_{NI}=1.2\mu\text{Vrms typ.}$ ($R_G=2.2\text{k}\Omega$, FLAT)
S/N=61dB typ. ($R_G=600\Omega$, IHF-A network)
(microphone amplifier, reference input=-60dBm)
Higher S/N ratio by 4dB when compared to general operational amplifiers
- High voltage $V_{CC}=\pm 25\text{V}$ (50V)
- Low maximum input voltage $V_i=140\text{mVrms}$ (typ.)
($V_{CC}=\pm 22.5\text{V}$, $G_V=40\text{dB}$)
- High gain, low distortion
..... $G_{VO}=110\text{dB}$, THD=0.004% (typ.)
- High slew rate SR=6.5V/ μs (typ.)
- High load current, high power dissipation
..... $I_{LP}=\pm 50\text{mA}$, $P_d=800\text{mW}$ (SIP)
 $P_d=625\text{mW}$ (DIP), $P_d=440\text{mW}$ (FP)

APPLICATION

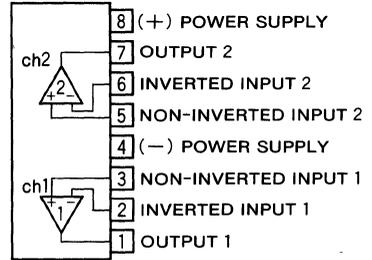
General-purpose preamplifier in stereo equipment, tape decks and radio stereo cassette recorders

RECOMMENDED OPERATING CONDITIONS

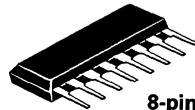
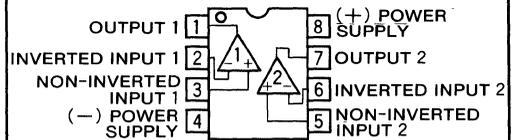
- Supply voltage range $\pm 2 \sim \pm 22.5\text{V}$
- Rated supply voltage $\pm 22.5\text{V}$

PIN CONFIGURATION (TOP VIEW)

SIP



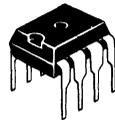
DIP, MINI FLAT



8-pin molded plastic SIP

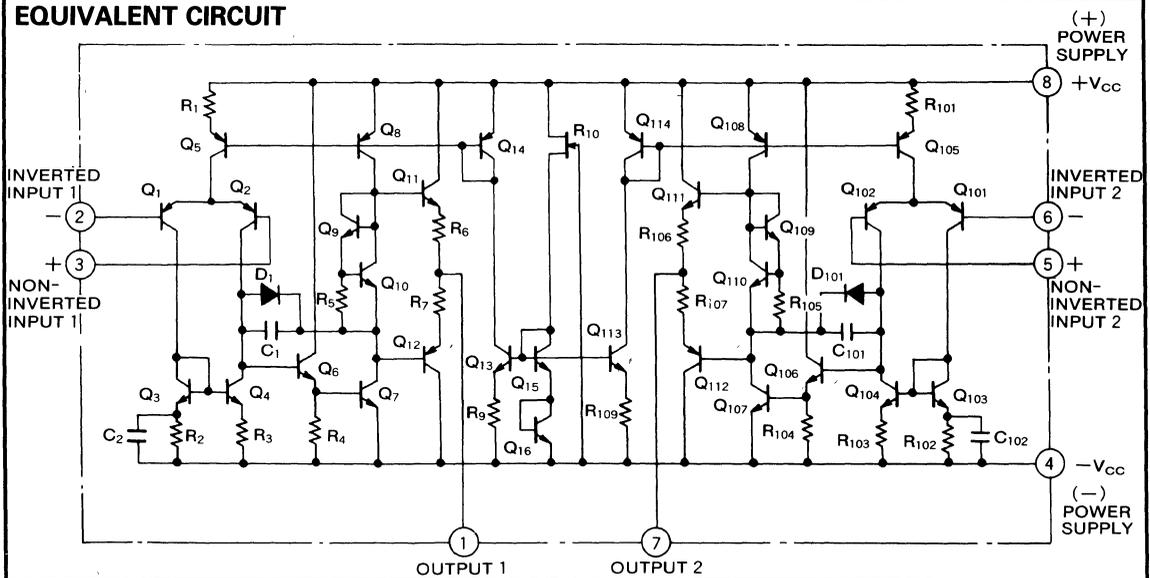


8-pin molded plastic FP
(MINI FLAT)



8-pin molded plastic DIP

EQUIVALENT CIRCUIT



**DUAL HIGH-VOLTAGE, HIGH S/N OPERATIONAL AMPLIFIERS
 (DUAL POWER SUPPLY TYPE)**

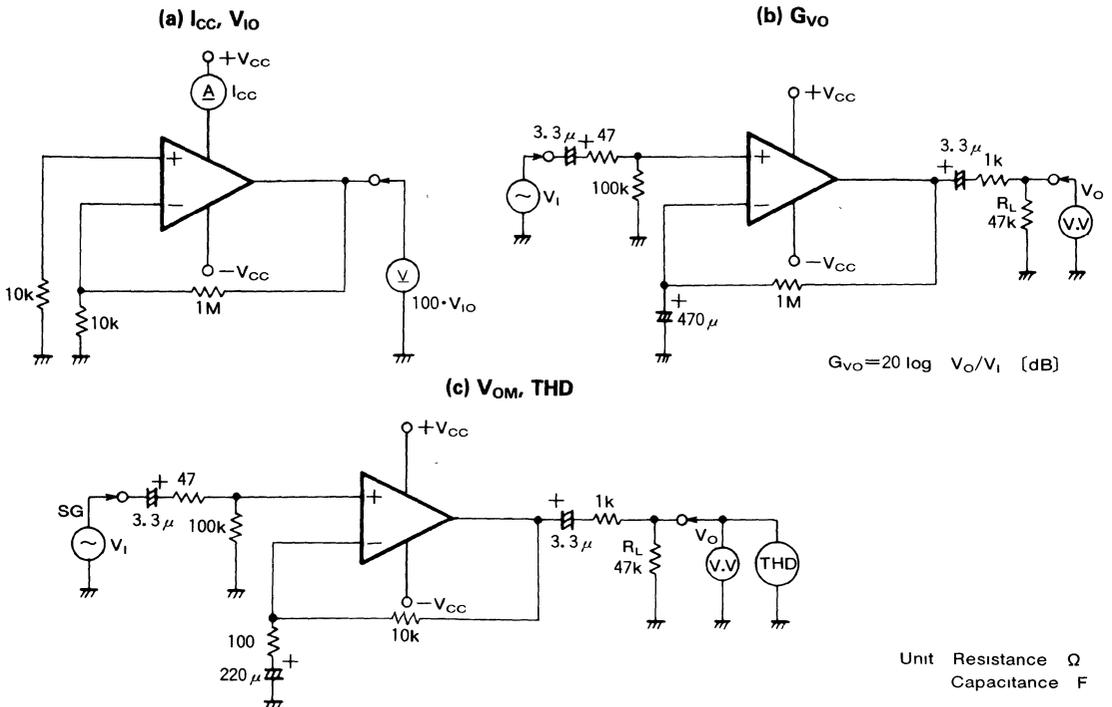
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		$\pm 25(50)$	V
I_{LP}	Load current		± 50	mA
V_{id}	Differential input voltage		± 30	V
V_{ic}	Common input voltage		± 22.5	V
P_d	Power dissipation		800(SIP)/625(DIP)/440(FP)	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	8(SIP)/6.25(DIP)/4.4(FP)	mW/°C
T_{opr}	Ambient temperature		$-20 \sim +75$	°C
T_{stg}	Storage temperature		$-55 \sim +125$	°C

ELECTICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 22.5\text{V}$)

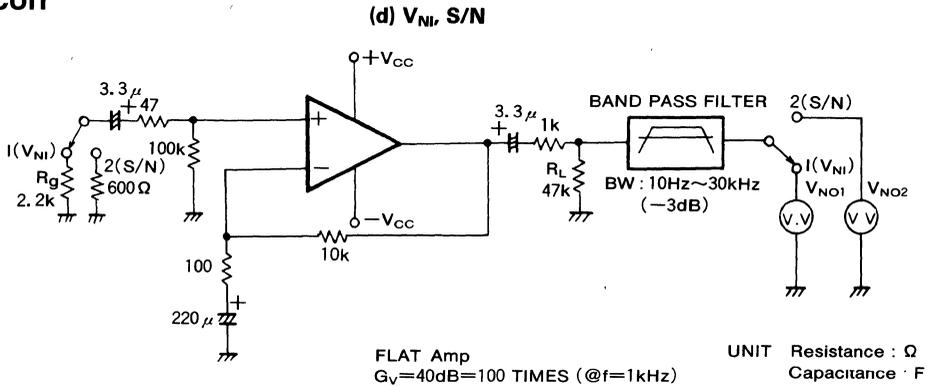
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{in}=0$		3.5	7.0	mA
V_{IO}	Input offset voltage	$R_S \leq 10\text{k}\Omega$		0.5	6.0	mV
I_{IB}	Input bias current			0.3		μA
G_{VO}	Open loop voltage gain	$f=100\text{Hz}$, $R_L=47\text{k}\Omega$, $C_{NF}=470\mu\text{F}$		90	110	dB
V_{OM}	Maximum output voltage	$f=1\text{kHz}$, $\text{THD}=0.1\%$, $R_L=47\text{k}\Omega$, FLAT	12.5	14.0		Vrms
THD	Total harmonic distortion	$f=1\text{kHz}$, $V_O=5\text{Vrms}$, $R_L=47\text{k}\Omega$, FLAT		0.004		%
V_{NI}	Input referred noise voltage	$R_g=2.2\text{k}\Omega$, $\text{BW}=10\text{Hz} \sim 30\text{kHz}$, FLAT		1.2	1.8	μVrms
S/N	Signal-to-noise ratio	$R_g=600\Omega$, $G_v=40\text{dB}$, IHF-A network Reference input -60dBm (microphone)		61		dB

TEST CIRCUITS



**DUAL HIGH-VOLTAGE, HIGH S/N OPERATIONAL AMPLIFIERS
 (DUAL POWER SUPPLY TYPE)**

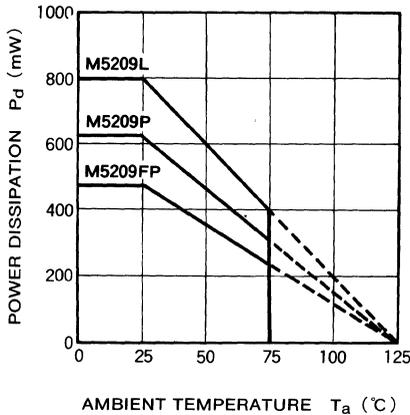
TEST CIRCUIT



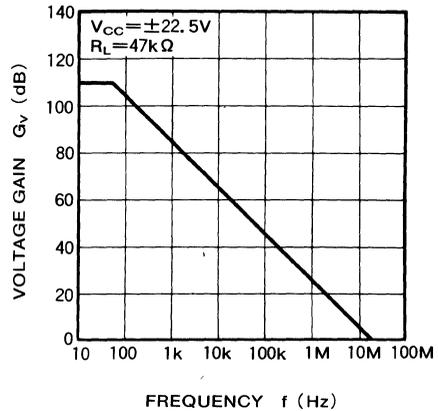
1. $V_{NI} = V_{NO1} / 100 (\mu\text{Vrms})$
 2. $S/N = 20 \log [775 \mu\text{Vrms} / (V_{NO2} / 100)]$ (dB) $775 \mu\text{Vrms} = -60\text{dBm}$ (Microphone reference input voltage)
- * An AC voltmeter V V with a built-in IHF-A network filter should be used for measuring the S/N ratio.

TYPICAL CHARACTERISTICS

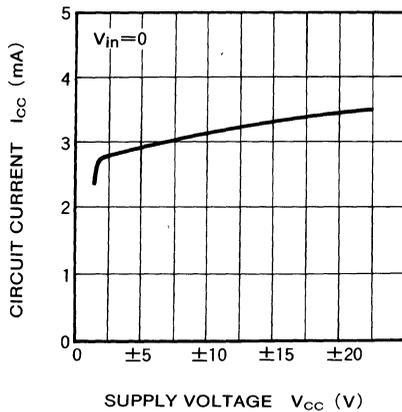
**THERMAL DERATING
 (MAXIMUM RATING)**



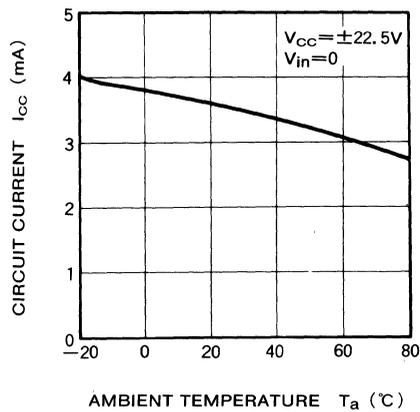
**VOLTAGE GAIN VS.
 FREQUENCY RESPONSE**



**CIRCUIT CURRENT VS.
 SUPPLY VOLTAGE**



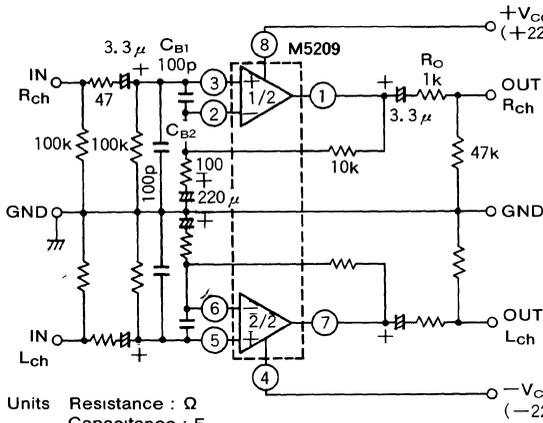
**CIRCUIT CURRENT VS.
 AMBIENT TEMPERATURE**



**DUAL HIGH-VOLTAGE, HIGH S/N OPERATIONAL AMPLIFIERS
 (DUAL POWER SUPPLY TYPE)**

APPLICATION EXAMPLES

(1) Stereo FLAT (microphone) amplifier circuit



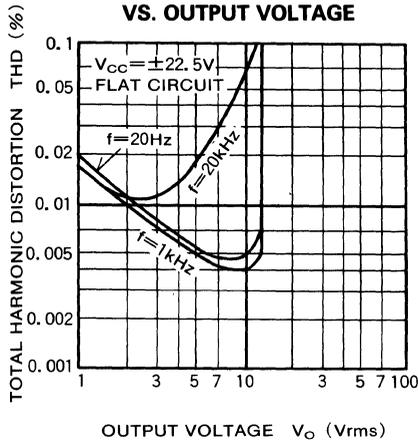
TYPICAL CHARACTERISTICS ($V_{CC} = \pm 22.5V$, FLAT)

- $G_v = 40dB$ ($f = 1kHz$)
- $V_{Ni} = 1.2 \mu V_{rms}$ ($R_g = 2.2k\Omega$, $BW = 10Hz \sim 30kHz$)
- $S/N = 61dB$ (IHF-A network, $R_g = 600\Omega$, $-600dBm$ input sensitivity)
- $THD = 0.004%$ ($f = 1kHz$, $V_o = 10V_{rms}$)

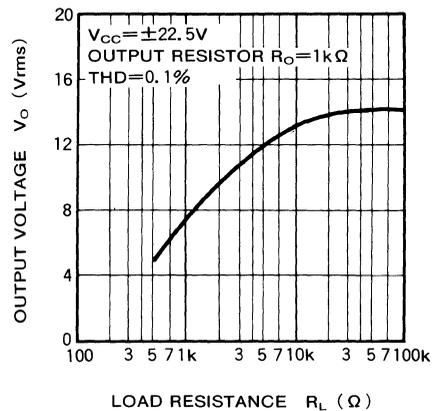
L_{ch} circuit constants are identical to those of R_{ch}
 C_{B1} , C_{B2} : Capacitors for buzz prevention, use if required
 R_o : Resistor used to prevent parasitic oscillation for capacitive loads and current limiting with shorted and other abnormal load conditions

Units Resistance : Ω
 Capacitance : F

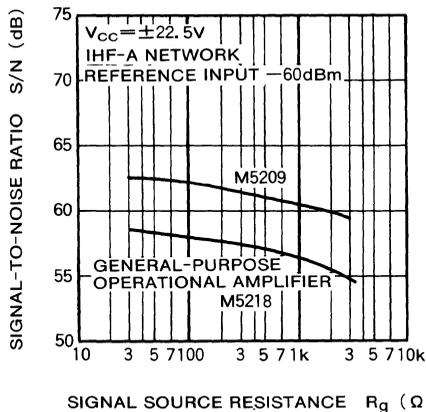
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



OUTPUT NOISE VOLTAGE VS. LOAD RESISTANCE

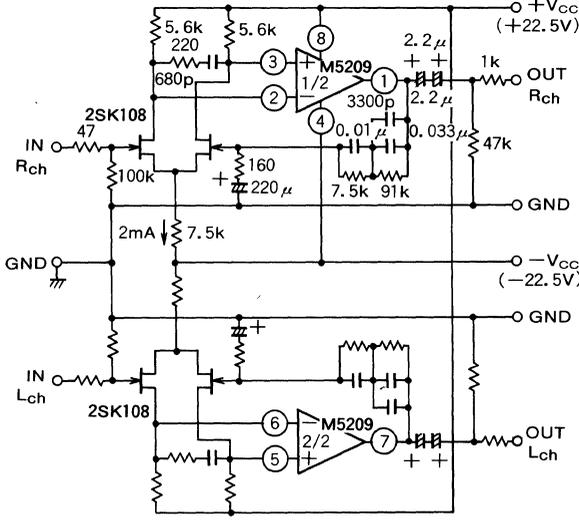


SIGNAL-TO-NOISE RATIO VS. SIGNAL SOURCE RESISTANCE



**DUAL HIGH-VOLTAGE, HIGH S/N OPERATIONAL AMPLIFIERS
 (DUAL POWER SUPPLY TYPE)**

(2) High S/N stereo DC ICL equalizer

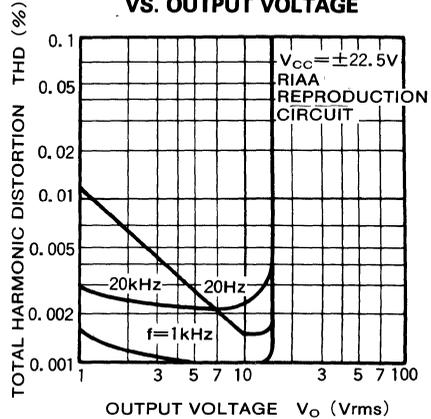


L_{ch} circuit constants are identical to those of R_{ch}
 Units Resistance : Ω
 Capacitance : F

TYPICAL CHARACTERISTICS ($V_{CC} = \pm 22.5V$, RIAA)

- S/N=85dB (IHF-A network, shorted input, 2.5mVrms input sensitivity)
- $V_{NI} = 0.77 \mu V_{rms}$ ($R_g = 5.1k\Omega$, BW=5Hz~100kHz)
- $G_v = 35.6dB$ ($f = 1kHz$)

**TOTAL HARMONIC DISTORTION
 VS. OUTPUT VOLTAGE**



MITSUBISHI LINEAR ICs

M5210L, P, FP

DUAL HIGH-VOLTAGE, HIGH S/N OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

DESCRIPTION

The M5210 is a semiconductor integrated circuit designed for a preamplifier in audio equipment of stereo and cassette tape decks.

Two low-noise operational amplifier circuits displaying internal phase-compensated high gain and low distortion are contained in a 8-pin (SIP, DIP), suitable for application as a microphone and tone control amplifier of stereo equipment and cassette tape decks.

The unit can also be used as a general-purpose amplifier in portable equipment such as a stereo cassette tape recorder of a single power supply type as it operates at a low supply voltage.

FEATURES

- Low noise $V_{NI}=1.0\mu\text{Vrms typ.}$ ($R_G=2.2\text{k}\Omega$, FLAT)
S/N=66dB typ. ($R_G=600\Omega$, IHF-A network)
(microphone amplifier, reference input=-60dBm)
Higher S/N ratio by 10dB when compared to ordinary operational amplifiers
- High voltage $V_{CC}=\pm 25\text{V}(50\text{V})$
- Low maximum input voltage $V_i=140\text{mVrms}(typ.)$
($V_{CC}=\pm 22.5\text{V}$, $G_V=40\text{dB}$)
- High gain, low distortion
..... $G_{VO}=113\text{dB}$, THD=0.002%(typ.)
- High slew rate SR=6.5V/ μs (typ.)
- High load current, high power dissipation
..... $I_{LP}=\pm 50\text{mA}$, $P_d=800\text{mW}$ (SIP)
 $P_d=625\text{mW}$ (DIP) $P_d=440\text{mW}$ (FP)

APPLICATION

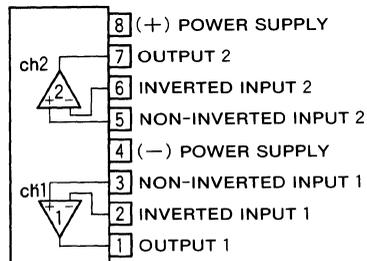
General-purpose preamplifier in stereo equipment, tape decks and radio stereo cassette recorders

RECOMMENDED OPERATING CONDITIONS

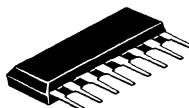
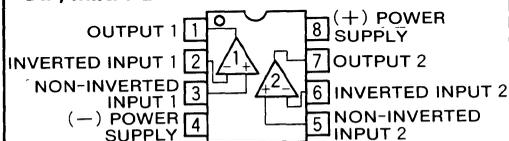
- Supply voltage range $\pm 2\sim\pm 22.5\text{V}$
Rated supply voltage $\pm 22.5\text{V}$

PIN CONFIGURATION (TOP VIEW)

SIP



DIP, MINI FLAT



8-pin molded plastic SIP

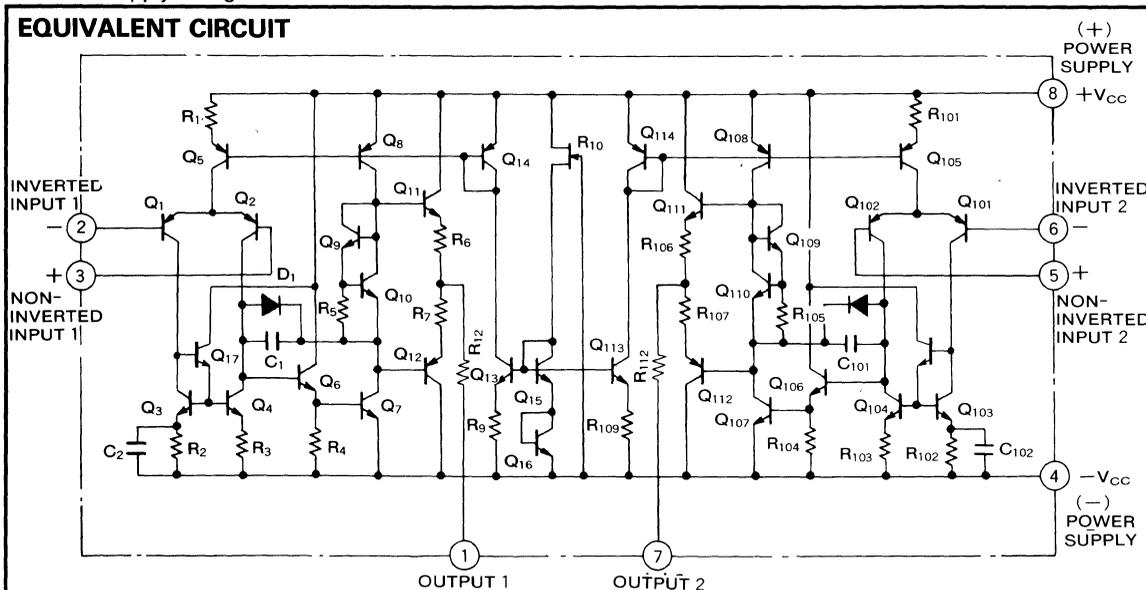


8-pin molded plastic FP
(MINI FLAT)



8-pin molded plastic DIP

EQUIVALENT CIRCUIT



MITSUBISHI LINEAR ICs M5210L, P, FP

DUAL HIGH-VOLTAGE, HIGH S/N OPERATIONAL AMPLIFIERS (DUAL POWER SUPPLY TYPE)

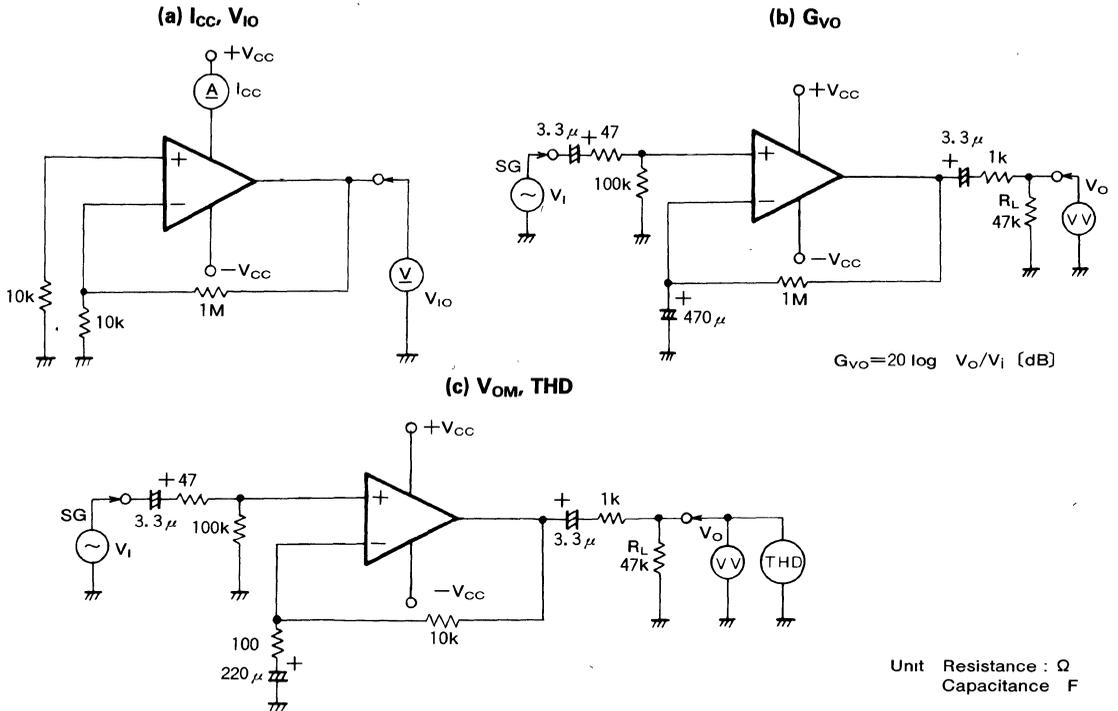
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		$\pm 25(50)$	V
I_{LP}	Load current		± 50	mA
V_{id}	Differential input voltage		± 30	V
V_{ic}	Common input voltage		± 22.5	V
P_d	Power dissipation		800(SIP)/625(DIP)/440(FP)	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	8(SIP)/6.25(DIP)/4.4(FP)	mW/ $^\circ\text{C}$
T_{opr}	Operating temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-55 \sim +125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 22.5\text{V}$)

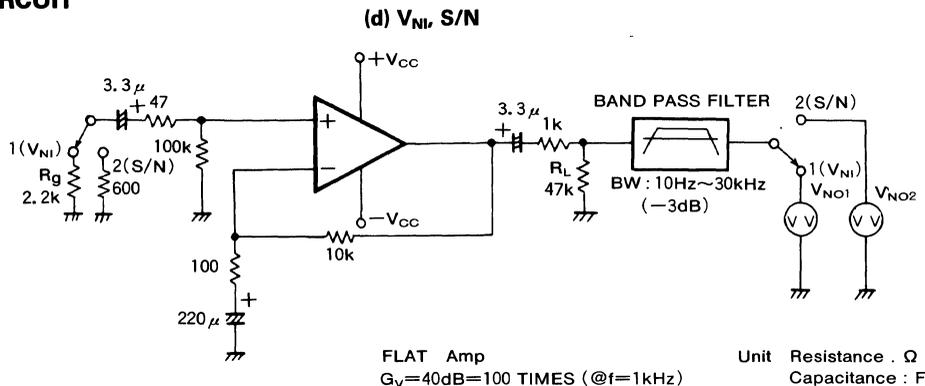
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{in}=0$	2.0	4.0	8.0	mA
V_{IO}	Input offset voltage	$R_S \leq 10\text{k}\Omega$		0.5	6.0	mV
I_{IB}	Input bias current			0.7		μA
G_{VO}	Open loop voltage gain	$f=100\text{Hz}$, $R_L=47\text{k}\Omega$, $C_{NF}=470\mu\text{F}$	90	113		dB
V_{OM}	Maximum output voltage	$f=1\text{kHz}$, $\text{THD}=0.1\%$, $R_L=47\text{k}\Omega$, FLAT	12.5	14.2		Vrms
THD	Total harmonic distortion	$f=1\text{kHz}$, $V_O=10\text{Vrms}$, $R_L=47\text{k}\Omega$, FLAT		0.002		%
V_{NI}	Input-referred noise voltage	$R_g=2.2\text{k}\Omega$, $\text{BW}=10\text{Hz} \sim 30\text{kHz}$, FLAT		1.0	1.5	μVrms
S/N	Signal to noise ratio	$R_g=600\Omega$, $G_v=40\text{dB}$, IHF-A network Reference input -60dBm (microphone)		66		dB

TEST CIRCUITS



**DUAL HIGH-VOLTAGE, HIGH S/N OPERATIONAL AMPLIFIERS
 (DUAL POWER SUPPLY TYPE)**

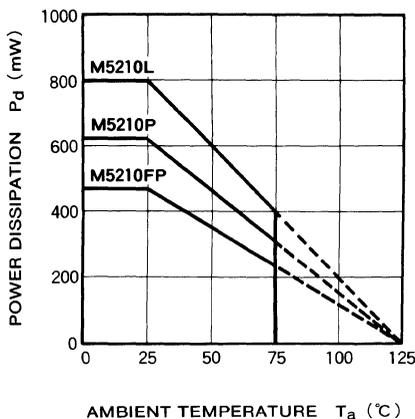
TEST CIRCUIT



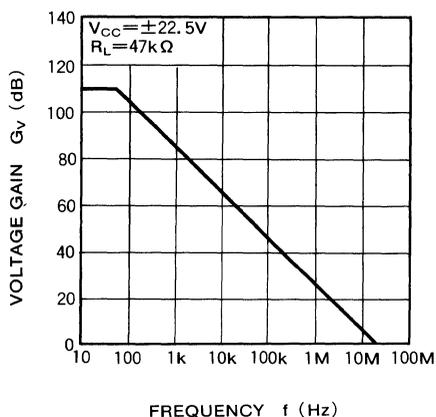
- $V_{Ni} = V_{NO1} / 100 (\mu\text{Vrms})$
 - $S/N = 20 \log [775 \mu\text{Vrms} / (V_{NO2} / 100)]$ (dB) $775 \mu\text{Vrms} = -60\text{dBm}$ (microphone reference input voltage)
- * An AC voltmeter V.V with a built-in IHF-A network filter should be used for measuring the S/N ratio.

TYPICAL CHARACTERISTICS

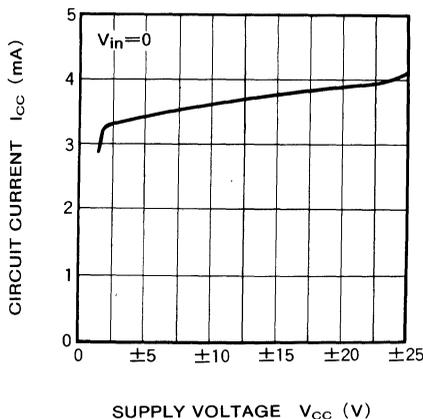
**THERMAL DERATING
 (MAXIMUM RATING)**



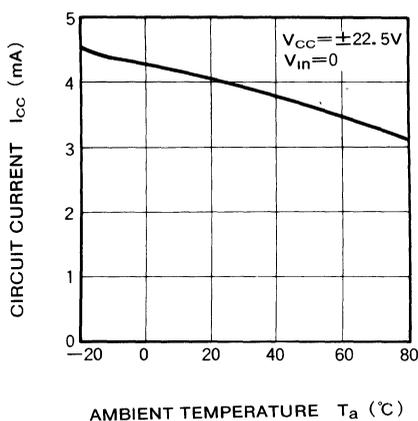
**VOLTAGE GAIN VS.
 FREQUENCY RESPONSE**



**CIRCUIT CURRENT VS.
 SUPPLY VOLTAGE**



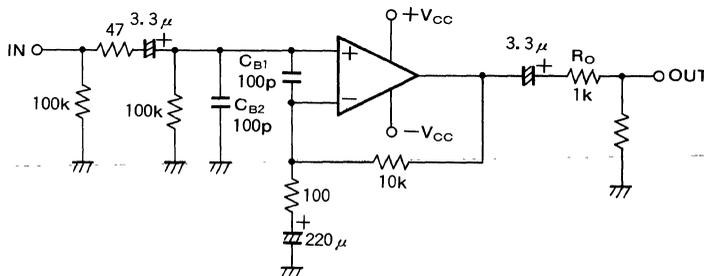
**CIRCUIT CURRENT VS.
 AMBIENT TEMPERATURE**



**DUAL HIGH-VOLTAGE, HIGH S/N OPERATIONAL AMPLIFIERS
 (DUAL POWER SUPPLY TYPE)**

APPLICATION EXAMPLES

(1) Stereo FLAT (microphone) amplifier circuit



Units Resistance : Ω
 Capacitance : F

TYPICAL CHARACTERISTICS ($V_{CC} = \pm 22.5V$, FLAT)

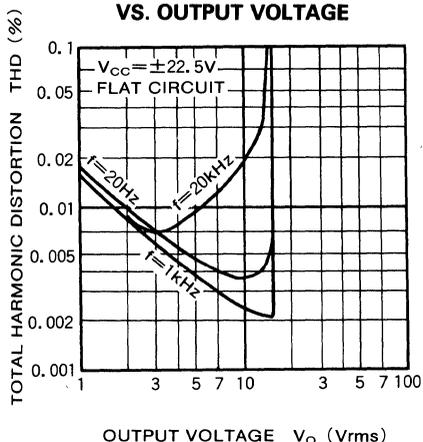
- $G_V = 40dB$ ($f = 1kHz$)
- $V_{NI} = 1.0 \mu V_{rms}$ ($R_g = 2.2k\Omega$, $BW = 10Hz \sim 30kHz$)
- $S/N = 66dB$ (IHF-A network, $R_g = 600\Omega$, $-60dBm$ input sensitivity)
- $THD = 0.002\%$ ($f = 1kHz$, $V_O = 10V_{rms}$)

Left channel circuit constants are identical to those of right channel

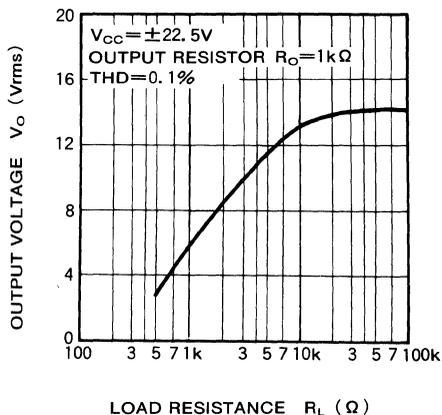
C_{B1} , C_{B2} : Capacitors for buzz prevention, use if required.

R_O : Resistor used to prevent parasitic oscillation for capacitive loads and current limiting with shorted and other abnormal load conditions.

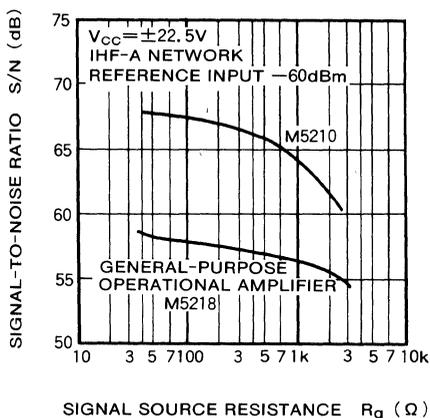
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



OUTPUT NOISE VOLTAGE VS. LOAD RESISTANCE



SIGNAL-TO-NOISE RATIO VS. SIGNAL SOURCE RESISTANCE



MITSUBISHI LINEAR ICs

M5219L, P, FP

DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

DESCRIPTION

The M5219 is a semiconductor integrated circuit designed for a preamplifier in audio equipment of stereo and cassette tape decks.

Two low-noise operational amplifier circuits displaying internal phase-compensated high gain and low distortion are contained in a 8-pin SIP, DIP or FP, suitable for application as an equalizer and tone control amplifier of stereo equipment and cassette tape decks.

The unit can also be used as a general-purpose amplifier in portable equipment such as a stereo cassette tape recorder of a single power supply type as it operates at a low supply voltage.

FEATURES

- Low noise $V_{NI}=0.9\mu\text{Vrms typ.}$ ($R_g=2.2\text{k}\Omega$, RIAA)
S/N=77dB typ. (Shorted input, IHF-A network)
(RIAA, PHONO=2.5mVrms)
- High voltage $V_{CC}=\pm 25\text{V}(50\text{V})$
- Low PHONO maximum input voltage $V_i=230\text{mVrms}(typ.)$
($V_{CC}=\pm 22.5\text{V}$, $f=1\text{kHz}$)
- High gain, low distortion $G_{VO}=110\text{dB}$, THD=0.001%(typ.)
- High slew rate SR=6.5V/ μs (typ.)
- High load current, high power dissipation $I_{LP}=\pm 50\text{mA}$, $P_d=800\text{mW}(SIP)$
 $P_d=625\text{mW}(DIP)$, $P_d=440\text{mW}(FP)$

APPLICATION

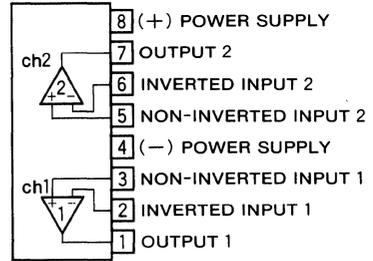
General-purpose preamplifier in stereo equipment, tape decks and radio stereo cassette recorders.

RECOMMENDED OPERATING CONDITIONS

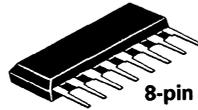
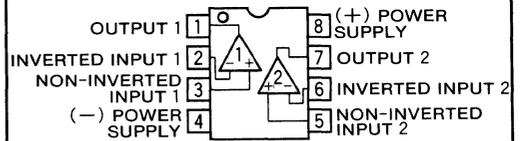
- Supply voltage range $\pm 2 \sim \pm 22.5\text{V}$
- Rated supply voltage $\pm 22.5\text{V}$

PIN CONFIGURATION (TOP VIEW)

SIP



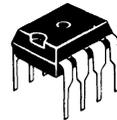
DIP, MINI FLAT



8-pin molded plastic SIP

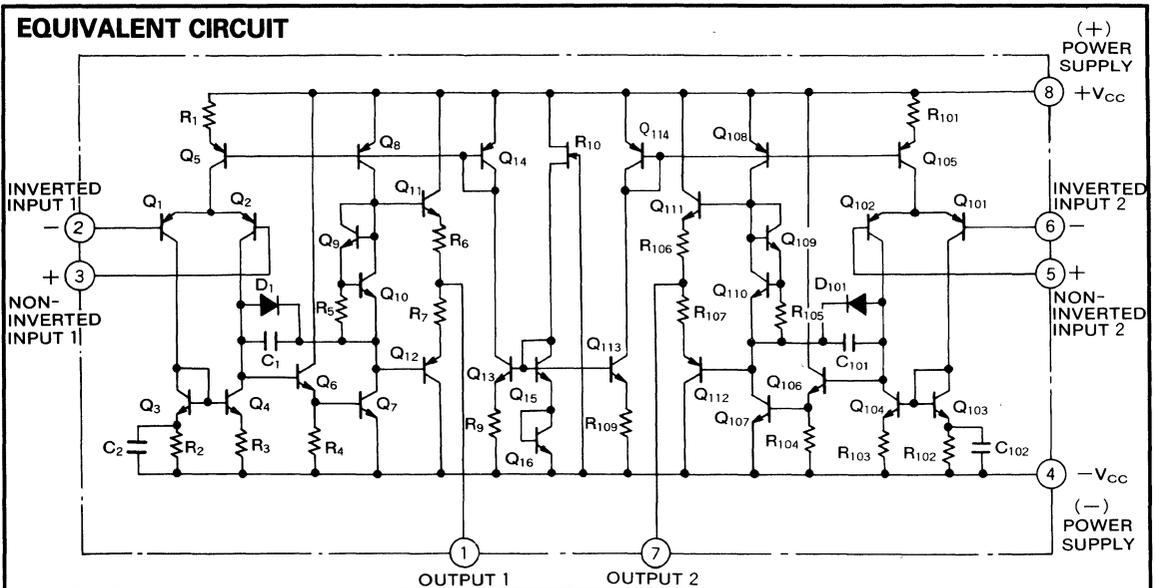


8-pin molded plastic FP
(MINI FLAT)



8-pin molded plastic DIP

EQUIVALENT CIRCUIT



DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

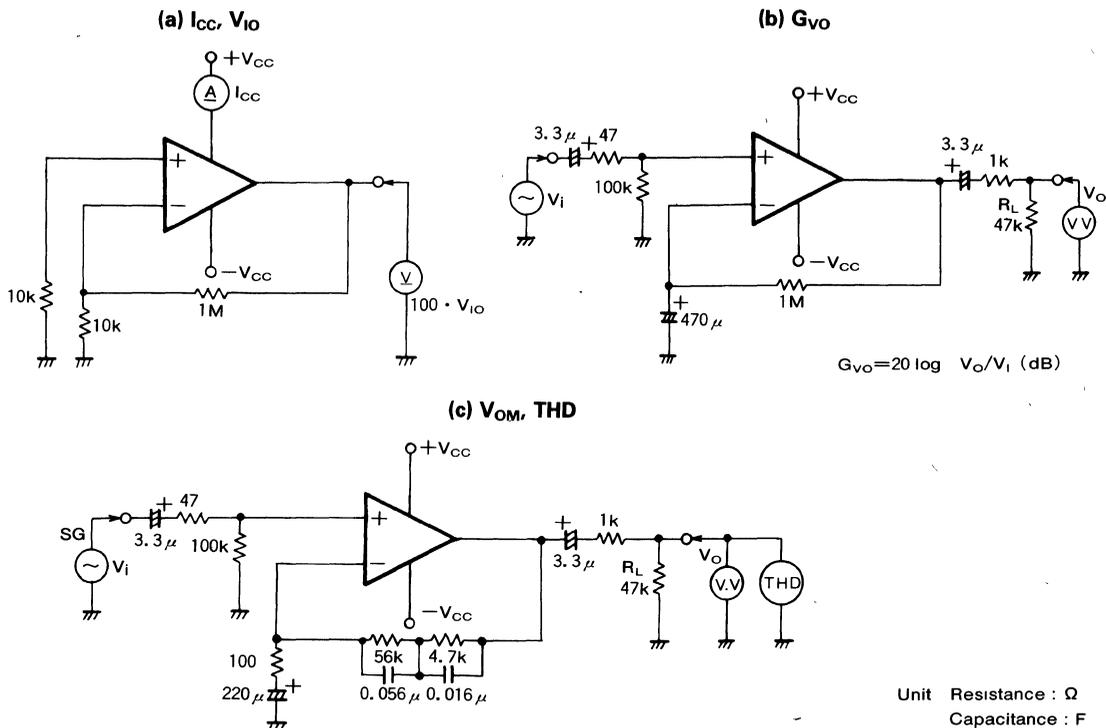
ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		±25(50)	V
I _{LP}	Load current		±50	mA
V _{id}	Differential input voltage		±30	V
V _{ic}	Common input voltage		±22.5V	V
P _d	Power dissipation		800(SIP)/625(DIP)/440(FP)	mW
K _θ	Thermal derating	T _a ≥25°C	8(SIP)/6.25(DIP)/4.4(FP)	mW/°C
T _{opr}	Ambient temperature		-20~+75	°C
T _{stg}	Storage temperature		-55~+125	°C

ELECTRICAL CHARACTERISTICS (T_a=25°C, V_{CC}=±22.5V)

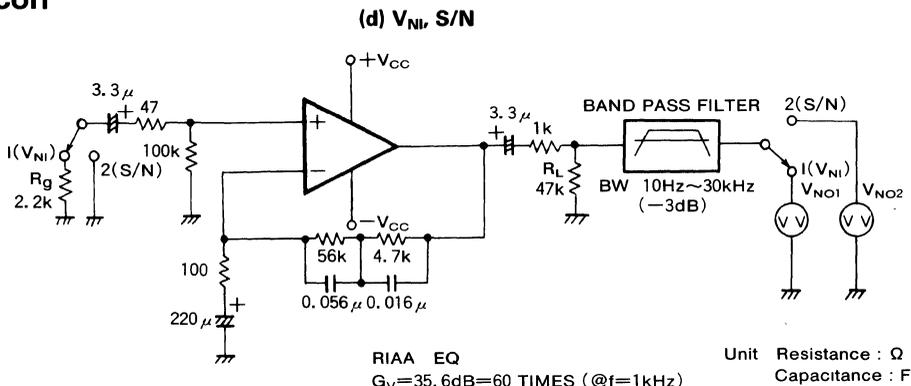
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I _{CC}	Circuit current	V _{in} =0		3.5	7.0	mA
V _{IO}	Input offset voltage	R _s ≤10kΩ		0.5	6.0	mV
I _{IB}	Input bias current			0.3		μA
G _{VO}	Open loop voltage gain	f=100Hz, R _L =47kΩ, C _{NF} =470μF	90	110		dB
V _{OM}	Maximum output voltage	f=1kHz, THD=0.1%, R _L =47kΩ, RIAA	12.5	14.0		V _{rms}
THD	Total harmonic distortion	f=1kHz, V _O =5V _{rms} , R _L =47kΩ, RIAA		0.001	0.03	%
V _{NI}	Input referred noise voltage	R _g =2.2kΩ, BW=10Hz~30kHz, RIAA		0.9	1.8	μV _{rms}
S/N	Signal-to-noise ratio	Shorted input (R _g =47Ω), IHF-A network PHONO=2.5mV _{rms} , RIAA		77		dB

TEST CIRCUITS



DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

TEST CIRCUIT

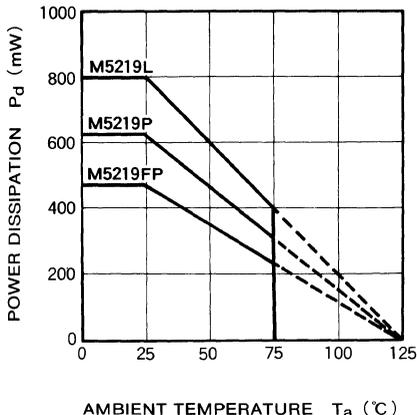


1. $V_{Ni} = V_{NO1} / 60 (\mu \text{Vrms})$
2. $S/N = 20 \log (2.5 \text{ mVrms} / (V_{NO2} / 60)) \text{ (dB)}$

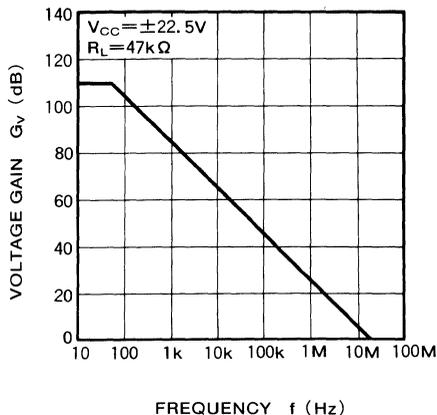
An AC voltmeter V.V with a built-in IHF-A network filter should be used for measuring the S/N ratio

TYPICAL CHARACTERISTICS

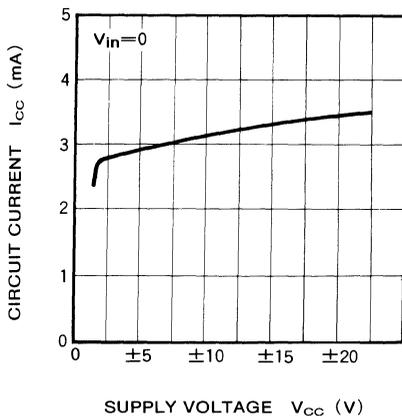
**THERMAL DERATING
 (MAXIMUM RATING)**



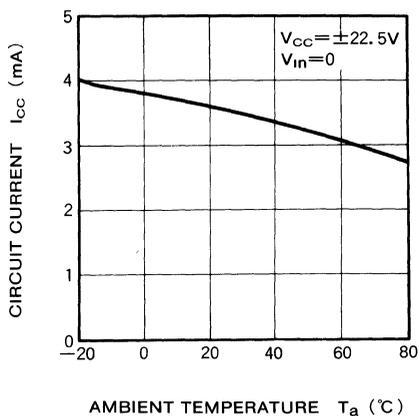
**VOLTAGE GAIN VS.
 FREQUENCY RESPONSE**



**CIRCUIT CURRENT VS.
 SUPPLY VOLTAGE**



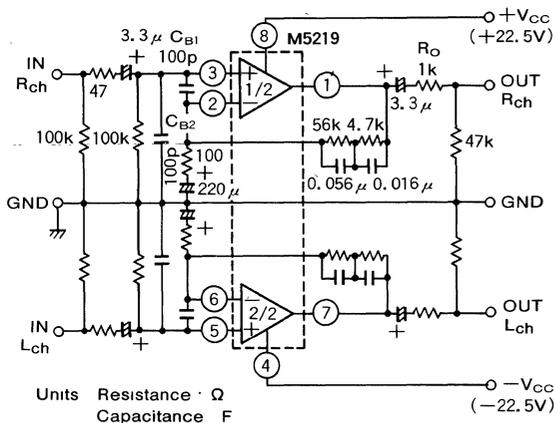
**CIRCUIT CURRENT VS.
 AMBIENT TEMPERATURE**



DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

APPLICATION EXAMPLES

(1) Stereo equalizer amplifier circuit

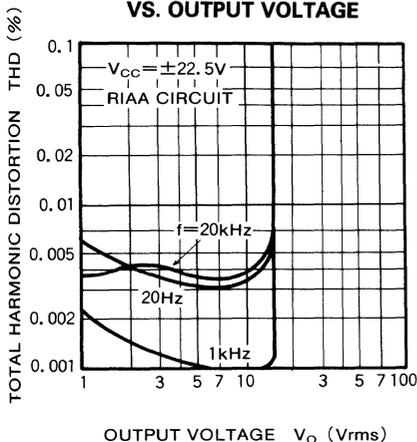


TYPICAL CHARACTERISTICS ($V_{CC} = \pm 22.5V$, RIAA)

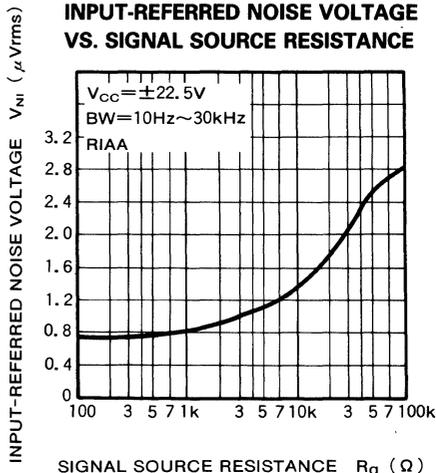
- $G_v = 35.6dB$ ($f = 1kHz$)
- $V_{NI} = 0.9 \mu V_{rms}$ ($R_g = 2.2k\Omega$, $BW = 10Hz \sim 30kHz$)
- $S/N = 77dB$ (IHF-A network, shorted input, $2.5mV_{rms}$ input sensitivity)
- $THD = 0.001\%$ ($f = 1kHz$, $V_o = 5V_{rms}$)

L_{ch} circuit constants are identical to those of R_{ch}
 C_{B1} , C_{B2} Capacitors for buzz prevention, use if required
 R_o : Resistor used to prevent parasitic oscillation for capacitive loads and current limiting with shorted and other abnormal load conditions.

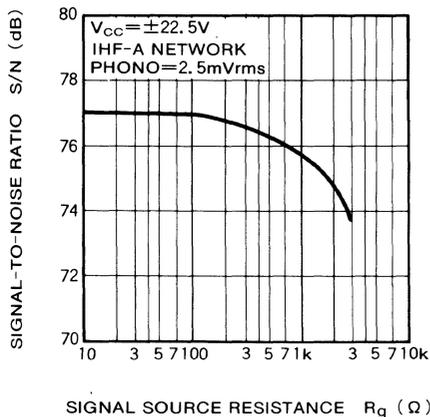
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



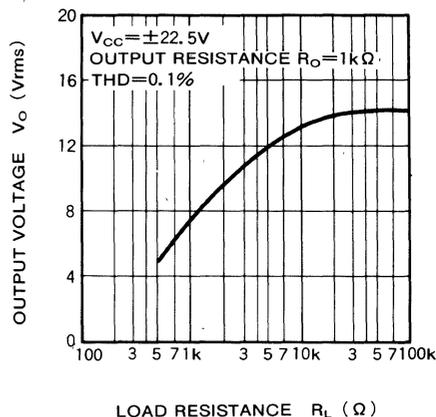
INPUT-REFERRED NOISE VOLTAGE VS. SIGNAL SOURCE RESISTANCE



SIGNAL-TO-NOISE RATIO VS. SIGNAL SOURCE RESISTANCE

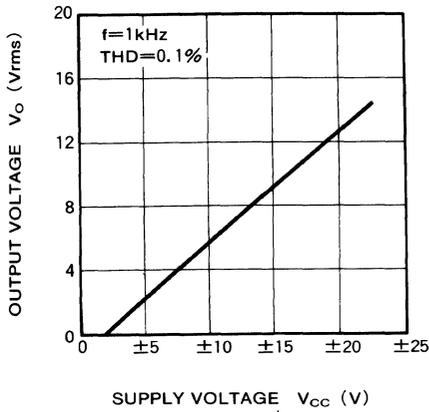


OUTPUT VOLTAGE VS. LOAD RESISTANCE

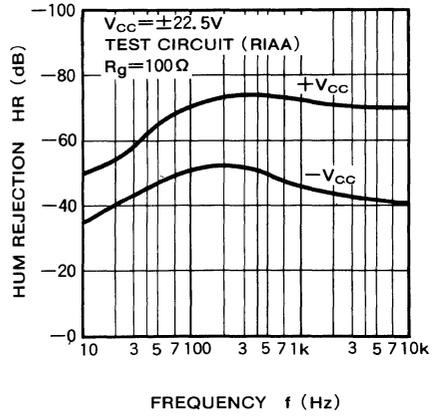


DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

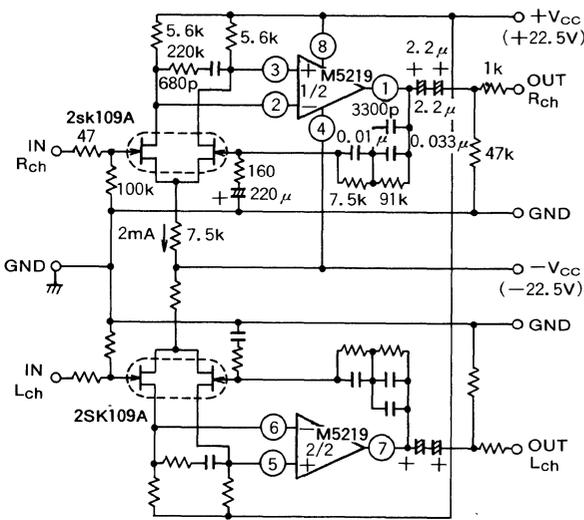
OUTPUT VOLTAGE VS. SUPPLY VOLTAGE



HUM REJECTION VS. FREQUENCY



(2) High S/N stereo DC ICL equalizer circuit



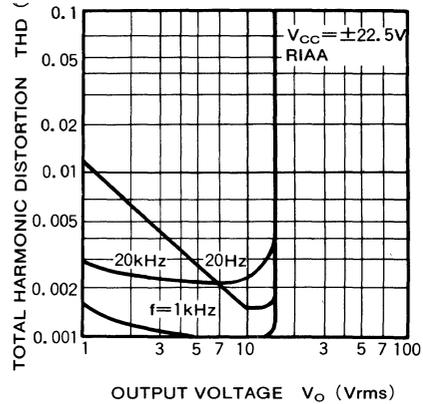
L_{ch} circuit constants are identical to those of R_{ch} .

Units Resistance Ω
 Capacitance F

TYPICAL CHARACTERISTICS ($V_{CC}=\pm 22.5V$, RIAA)

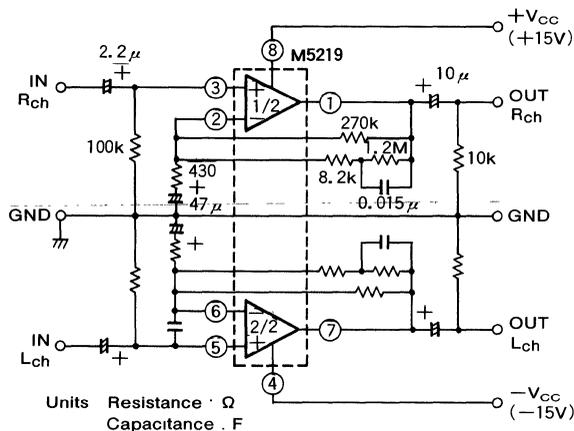
- $S/N=85dB$ (IHF-A network, shorted input, 2.5mVrms input sensitivity)
- $V_{NI}=0.77\mu Vrms$ ($R_g=5.1k\Omega$, $BW=5Hz\sim 100kHz$)
- $G_v=35.6dB$ ($f=1kHz$)

TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

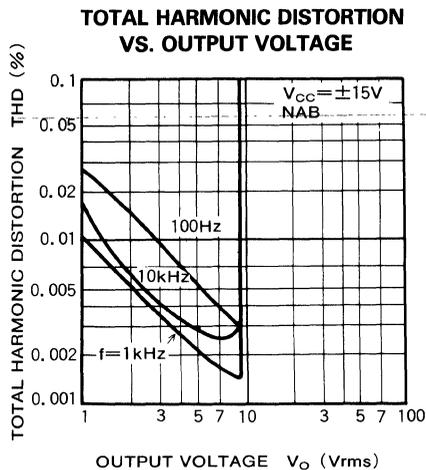
(3) Tape deck equalizer amplifier circuit



L_{ch} circuit constants are identical to those of R_{ch}

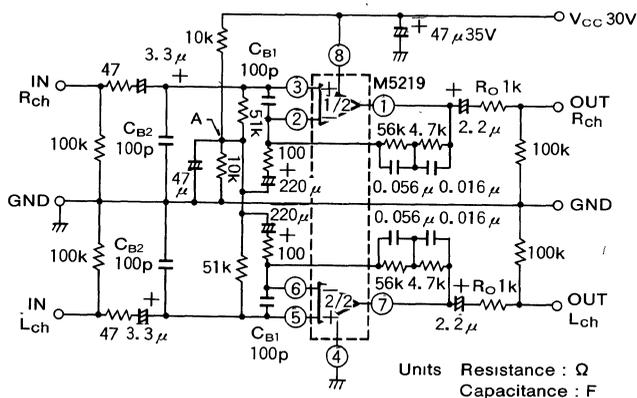
TYPICAL CHARACTERISTICS ($V_{CC}=\pm 15V$, NAB)

- $G_V=29.9dB(f=1kHz)$
- $V_{NI}=1.4\mu Vrms(R_G=2.2k\Omega, BW=20Hz\sim 15kHz)$
 ($-117dBv$)



(4) Typical single power supply application

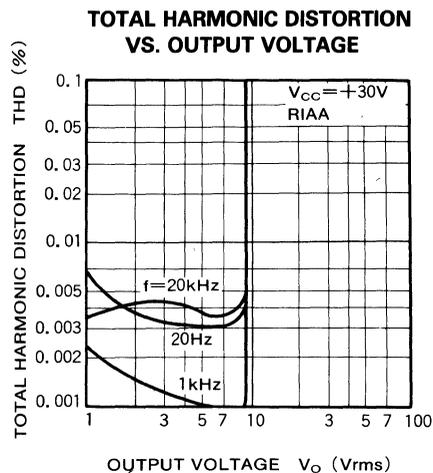
PHONO EQUALIZER AMPLIFIER (RIAA)



- → Point A is the $V_{CC}/2$ point in DC terms (virtual ground) when the device is used as a single power supply type
- C_{B1} , C_{B2} : Capacitor for buzz prevention, use if required
- R_O : Resistor used to prevent parasitic oscillation for capacitive loads and current limiting with shorted and other abnormal conditions.

TYPICAL CHARACTERISTICS ($V_{CC}=+30V$, RIAA)

- $G_V=35.6dB(f=1kHz)$
- $V_{NI}=0.9\mu Vrms(R_G=2.2k\Omega, BW=10Hz\sim 30kHz)$
- $S/N=77dB$ (1HF-A network, shorted input, 2.5mVrms input sensitivity)



MITSUBISHI LINEAR ICs M5220L, P, FP

DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

DESCRIPTION

The M5220 is a semiconductor integrated circuit designed for a preamplifier in audio equipment of stereo and cassette tape decks.

Two low-noise operational amplifier circuits displaying internal phase-compensated high gain and low distortion are contained in a 8-pin SIP, DIP or FP, suitable for application as an equalizer and tone control amplifier of stereo equipment and cassette tape decks. The unit can also be used as a general-purpose amplifier in portable equipment such as a stereo cassette tape recorder of a single power supply type as it operates at a low supply voltage.

FEATURES

- Low noise $V_{NI}=0.75\mu\text{Vrms typ.}$ ($R_G=2.2\text{k}\Omega$, RIAA)
S/N=83dB typ.(Shorted input, IHF-A network)
(RIAA, PHONO=2.5mVrms)
- High voltage $V_{CC}=\pm 25\text{V}(50\text{V})$
- Low PHONO maximum input voltage
..... $V_I=235\text{mVrms}(typ.)$
($V_{CC}=\pm 22.5\text{V}$, $f=1\text{kHz}$)
- High gain, low distortion
..... $G_{VO}=113\text{dB}$, THD=0.001%(typ.)
- High slew rate SR=6.5V/ μs (typ.)
- High load current, high power dissipation
..... $I_{LP}=\pm 50\text{mA}$, $P_d=800\text{mW}(SIP)$
 $P_d=625\text{mW}(DIP)$
 $P_d=440\text{mW}(FP)$

APPLICATION

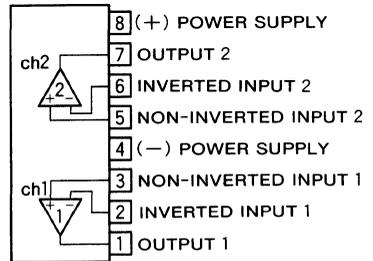
General-purpose preamplifier in stereo equipment, tape decks and radio stereo cassette recorders.

RECOMMENDED OPERATING CONDITIONS

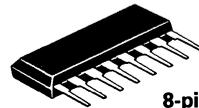
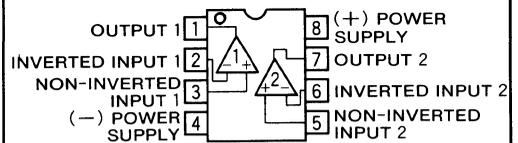
- Supply voltage range $\pm 2 \sim \pm 22.5\text{V}$
Rated supply voltage $\pm 22.5\text{V}$

PIN CONFIGURATION (TOP VIEW)

SIP



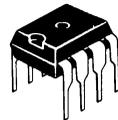
DIP, MINI FLAT



8-pin molded plastic SIP

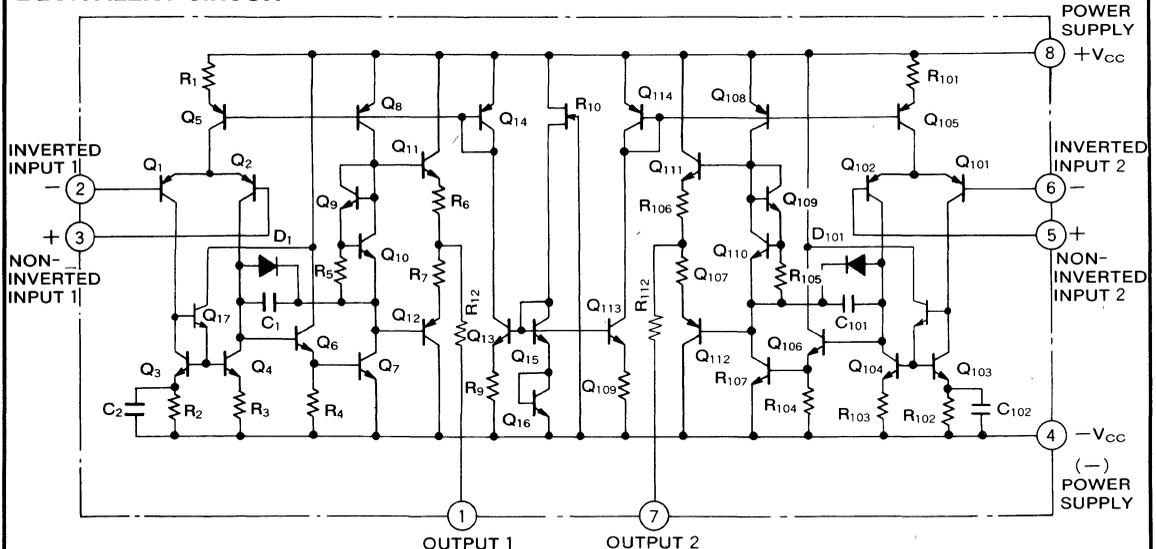


8-pin molded plastic FP
(MINI FLAT)



8-pin molded plastic DIP

EQUIVALENT CIRCUIT



DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

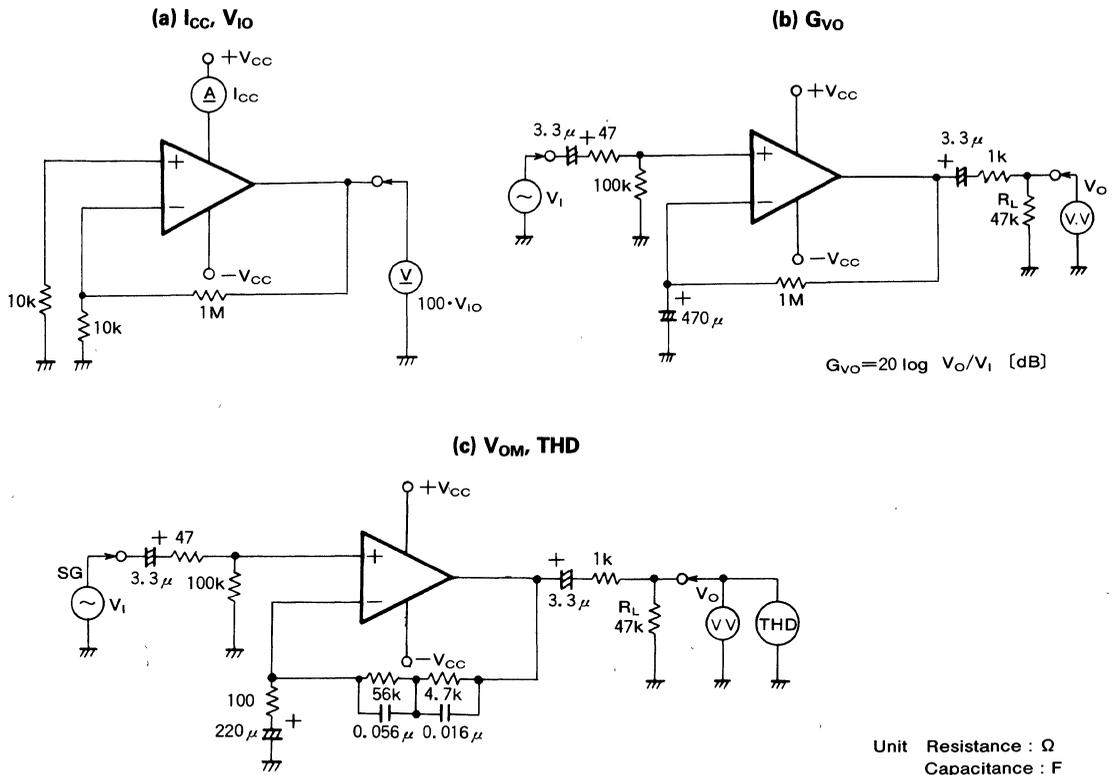
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		$\pm 25(50)$	V
I_{LP}	Load current		± 50	mA
V_{id}	Differential input voltage		± 30	V
V_{IC}	Common input voltage		± 22.5	V
P_d	Power dissipation		800(SIP)/625(DIP)/440(FP)	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	8(SIP)/6.25(DIP)/4.4(FP)	mW/ $^\circ\text{C}$
T_{opr}	Ambient temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-55 \sim +125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 22.5\text{V}$)

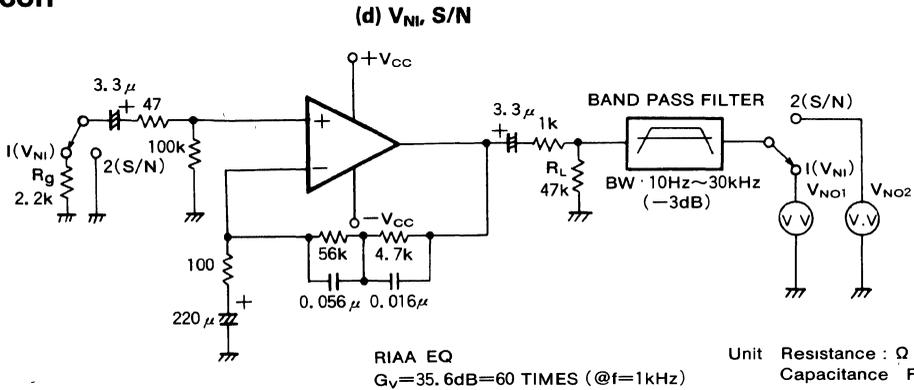
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{in}=0$		4.0	8.0	mA
V_{IO}	Input offset voltage	$R_s \leq 10\text{k}\Omega$		0.5	3.0	mV
I_{IB}	Input bias current			0.7		μA
G_{VO}	Open loop voltage gain	$f=100\text{Hz}$, $R_L=47\text{k}\Omega$, $C_{NF}=470\mu\text{F}$	90	113		dB
V_{OM}	Maximum output voltage	$f=1\text{kHz}$, THD=0.1%, $R_L=47\text{k}\Omega$, RIAA	12.5	14.2		Vrms
THD	Total harmonic distortion	$f=1\text{kHz}$, $V_O=5\text{Vrms}$, $R_L=47\text{k}\Omega$, RIAA		0.001	0.03	%
V_{NI}	Input referred noise voltage	$R_g=2.2\text{k}\Omega$, BW=10Hz~30kHz, RIAA		0.75	1.8	μVrms
S/N	Signal-to-noise ratio	Shorted input ($R_g=47\Omega$), IHF-A network PHONO=2.5mVrms, RIAA		83		dB

TEST CIRCUITS



DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

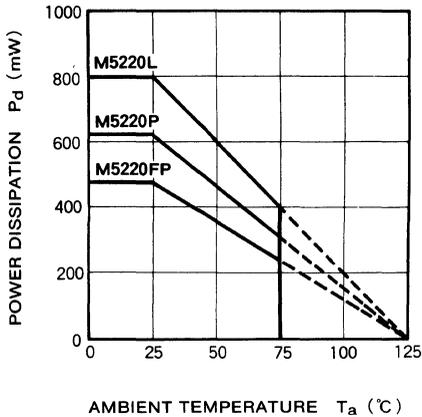
TEST CIRCUIT



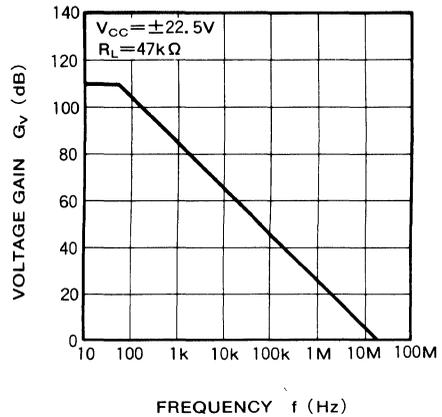
1. $V_{NI} = V_{NO1} / 60 (\mu \text{Vrms})$
 2. $S/N = 20 \log [2.5 \text{ mVrms} / (V_{NO2} / 60)] \text{ (dB)}$
- * An AC voltmeter V.V with a built-in IHF-A network filter should be used for measuring the S/N ratio.

TYPICAL CHARACTERISTICS

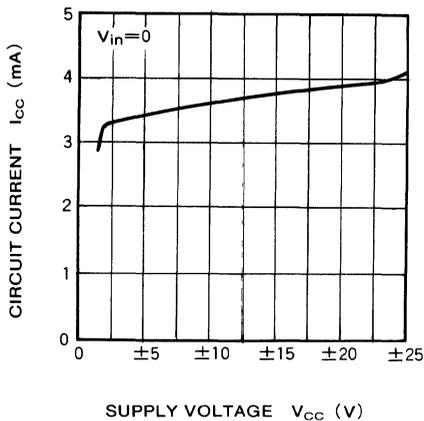
**THERMAL DERATING
 (MAXIMUM RATING)**



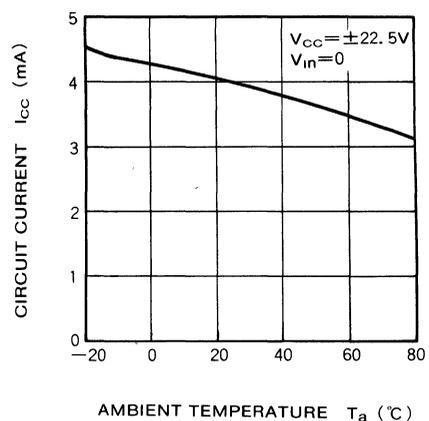
**VOLTAGE GAIN VS.
 FREQUENCY RESPONSE**



**CIRCUIT CURRENT VS.
 SUPPLY VOLTAGE**



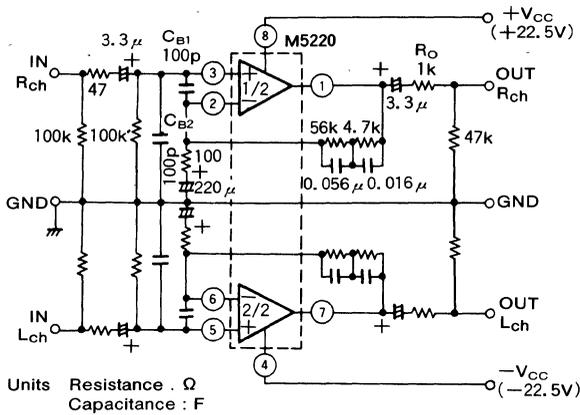
**CIRCUIT CURRENT VS.
 AMBIENT TEMPERATURE**



DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

APPLICATION EXAMPLES

(1) Stereo equalizer amplifier circuit

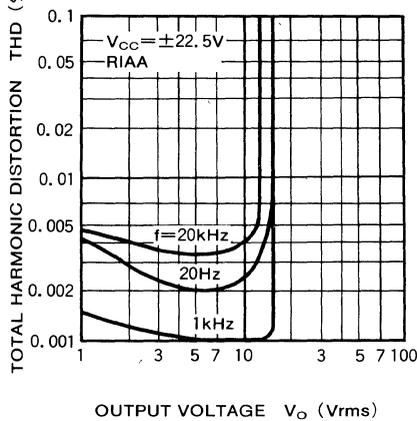


TYPICAL CHARACTERISTICS ($V_{CC}=\pm 22.5V$, RIAA)

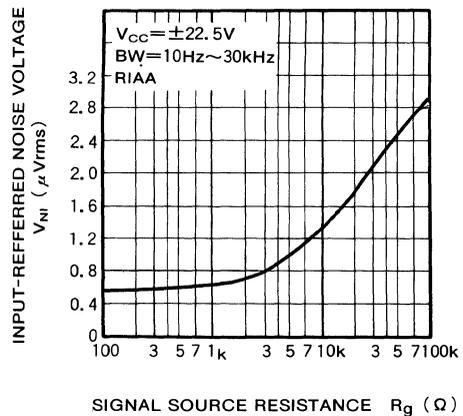
- $G_v=35.6\text{dB}(f=1\text{kHz})$
- $V_{Ni}=0.75\mu\text{Vrms}(R_g=2.2\text{k}\Omega, \text{BW}=10\text{Hz}\sim 30\text{kHz})$
- $S/N=83\text{dB}$ (IHF-A network, shorted input, 2.5mVrms input sensitivity)
- $\text{THD}=0.001\%(f=1\text{kHz}, V_o=5\text{Vrms})$

L_{ch} circuit constants are identical to those of R_{ch}
 C_{B1}, C_{B2} : Capacitors for buzz prevention, use if required.
 R_o : Resistor used to prevent parasitic oscillation for capacitive loads and current limiting with shorted and other abnormal load conditions

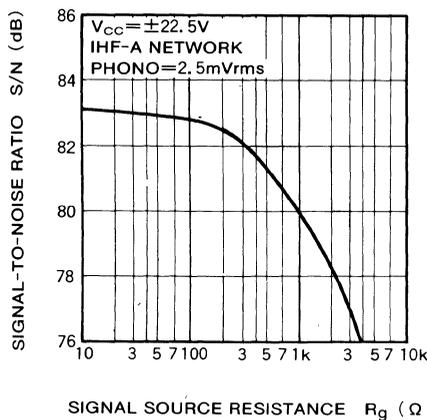
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



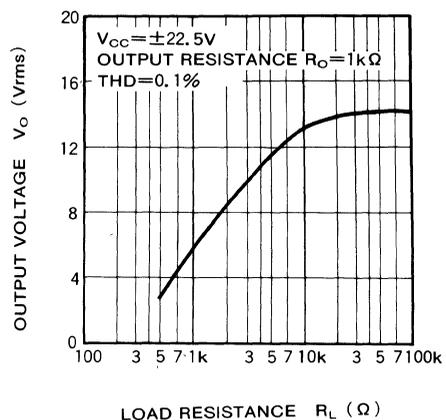
INPUT-REFERRED NOISE VOLTAGE VS. SIGNAL SOURCE RESISTANCE



SIGNAL-TO-NOISE RATIO VS. SIGNAL SOURCE RESISTANCE

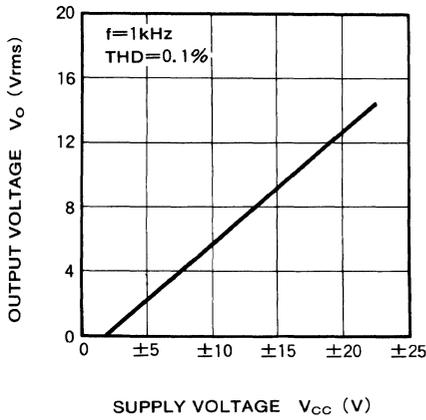


OUTPUT VOLTAGE VS. LOAD RESISTANCE

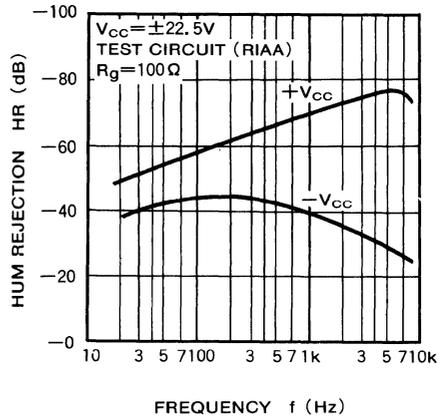


DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

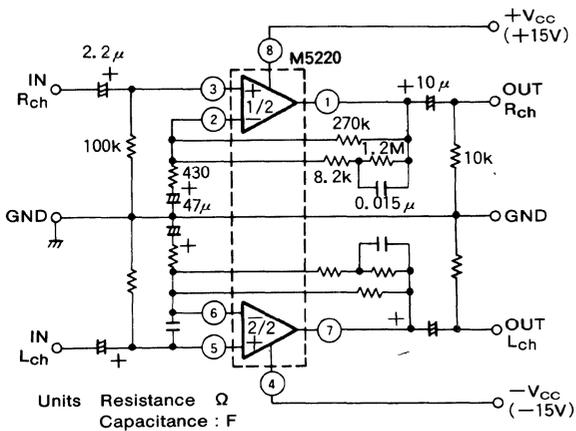
OUTPUT VOLTAGE VS. SUPPLY VOLTAGE



HUM REJECTION VS. FREQUENCY



(2) Stereo deck equalizer amplifier circuit

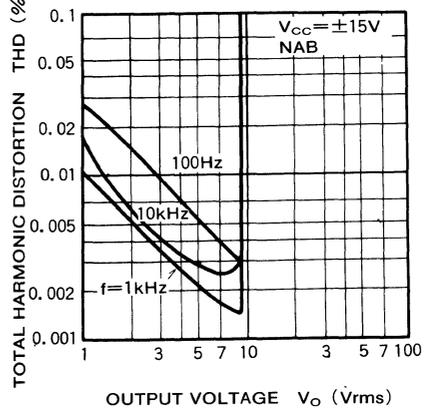


L_{ch} circuit constants are identical to those of R_{ch}

TYPICAL CHARACTERISTICS ($V_{CC}=\pm 15V$, NAB)

- $G_v=29.9dB$ ($f=1kHz$)
- $V_{NI}=1.0\mu V_{rms}$ ($R_g=2.2k\Omega$, $BW=20Hz\sim 15kHz$)
($-120dB_v$)

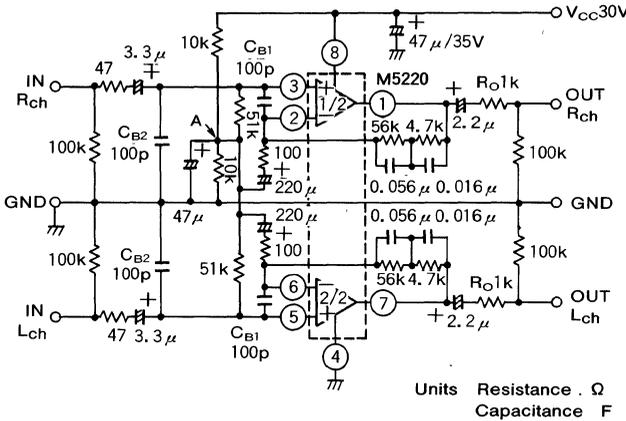
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



DUAL LOW-NOISE OPERATIONAL AMPLIFIERS(DUAL POWER SUPPLY TYPE)

(3) Typical single power supply application

PHONO EQUALIZER AMPLIFIER (RIAA)

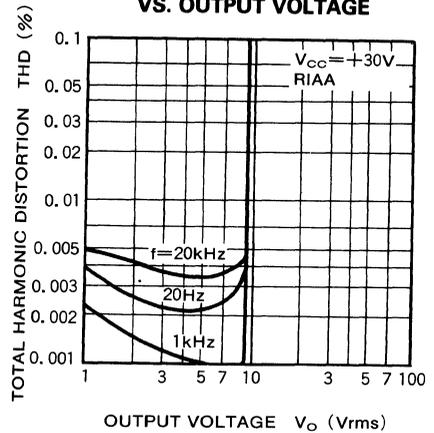


- → Point A is the $V_{CC}/2$ point in DC terms (virtual ground) when the device is used as a single power supply type.
- C_{B1} , C_{B2} : Capacitor for buzz prevention, used if required
- R_O : Resistor used to prevent parasitic oscillation for capacitive loads and current limiting with shorted and other abnormal conditions

TYPICAL CHARACTERISTICS ($V_{CC}=+30V$, RIAA)

- $G_v=35.6dB(f=1kHz)$
- $V_{NI}=0.75\mu V_{rms}(R_g=2.2k\Omega, BW=10Hz\sim 30kHz)$
- $S/N=83dB$ (IHF-A network, shorted input, $2.5\mu V_{rms}$ input sensitivity)

TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



M5221L, P, FP/M5T082P

DUAL J-FET INPUT OPERATIONAL AMPLIFIERS

DESCRIPTION

The M5221/M5T082P are semiconductor integrated circuits designed as high-performance dual operational amplifiers which adopt J-FETs in the input stage.

The devices come in an 8-pin SIP, DIP or FP and contain two circuits for yielding a high input impedance, high slew rate, low bias current and other excellent characteristics. They can be widely used as a general-purpose operational amplifiers in stereo equipment, tape decks, digital audio disc players and other similar products as well as in VTRs, video disc players and video related players.

FEATURES

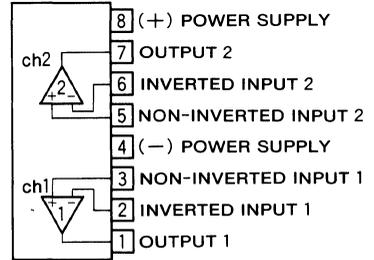
- High input impedance due to J-FET input
..... $R_i=1000M\Omega$ (typ.)
- High slew rate $SR=13V/\mu s$ (typ.)
- High gain, low distortion $G_{V0}=106dB$, $THD=0.007\%$
($G_V=40dB$ @ $f=1kHz$) (typ.)
- Large load current and allowable current
..... $I_{LP}=\pm 50mA$, $P_d=800mW$ (SIP)
 $P_d=625mW$ (DIP)
 $P_d=440mW$ (FP)

APPLICATION

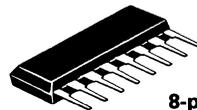
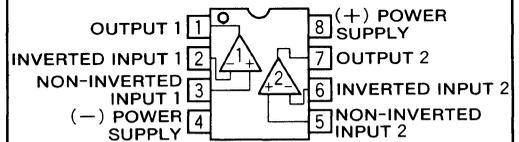
General-purpose preamplifier in stereo equipment, tape decks and digital audio disc players, VTRs and video disc players.

PIN CONFIGURATION (TOP VIEW)

SIP



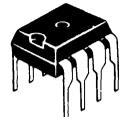
DIP, MINI FLAT



8-pin molded plastic SIP

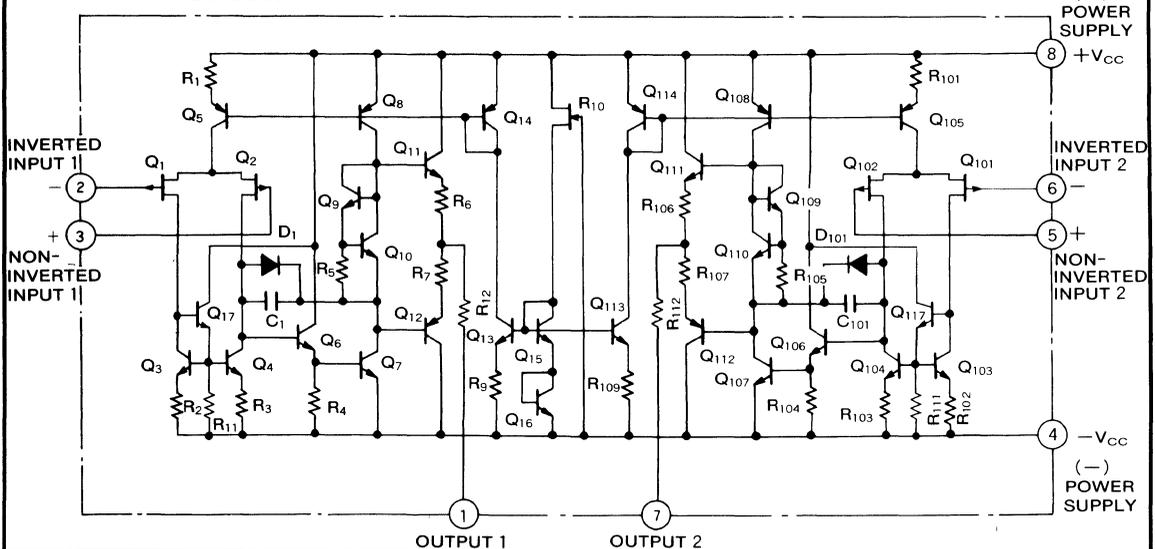


8-pin molded plastic FP (MINI FLAT)



8-pin molded plastic DIP

EQUIVALENT CIRCUIT



MITSUBISHI LINEAR ICs

M5221L, P, FP/M5T082P

DUAL J-FET INPUT OPERATIONAL AMPLIFIERS

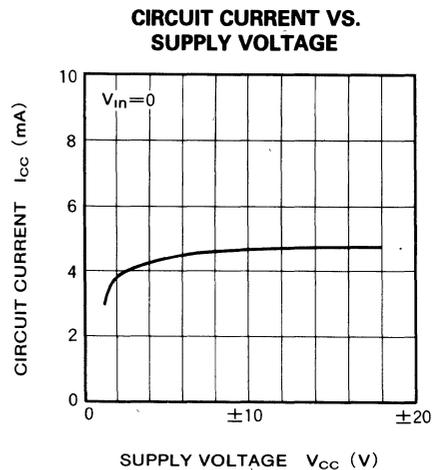
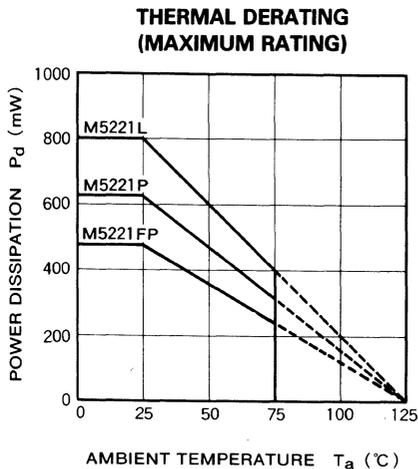
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		± 18	V
I_{LP}	Load current		± 50	mA
V_{id}	Differential input voltage		± 30	V
V_{ic}	Common input voltage		± 15	V
P_d	Power dissipation		800(SIP)/625(DIP)/440(FP)	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	8(SIP)/6.25(DIP)/4.4(FP)	mW/°C
T_{opr}	Operating temperature		$-20 \sim +75$	°C
T_{stg}	Storage temperature		$-55 \sim +125$	°C

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 15\text{V}$)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{in}=0$		3.0	6.0	mA
V_{IO}	Input offset voltage	$R_s \leq 10\text{k}\Omega$		5.0	15.0	mV
I_{IO}	Input offset current			5	200	pA
I_{IB}	Input bias current			30	400	pA
R_{in}	Input resistance			10^3		M Ω
G_{VO}	Open loop voltage gain	$R_L \geq 2\text{k}\Omega$, $V_o = \pm 10\text{V}$	86	106		dB
V_{OM}	Maximum output voltage	$R_L \geq 10\text{k}\Omega$	± 12	± 14		V
		$R_L \geq 2\text{k}\Omega$	± 10	± 13		
V_{CM}	Common input voltage width		± 10	12		V
CMRR	Common mode rejection ratio	$R_s \leq 10\text{k}\Omega$	70	76		dB
SVRR	Supply voltage rejection ratio	$R_s \leq 10\text{k}\Omega$		30	150	$\mu\text{V/V}$
P_d	Power dissipation			90	180	mW
SR	Slew rate	$G_v=0\text{dB}$, $R_L=2\text{k}\Omega$		13		V/ μs
f_T	Gauging bandwidth product			3		MHz
V_{NI}	Input-referred noise voltage	$R_s=100\Omega$, BW=10Hz~30kHz		2.2		μVrms

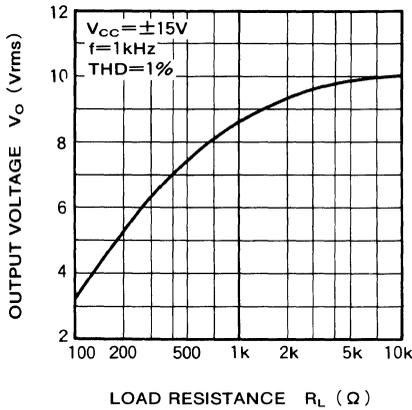
TYPICAL CHARACTERISTICS



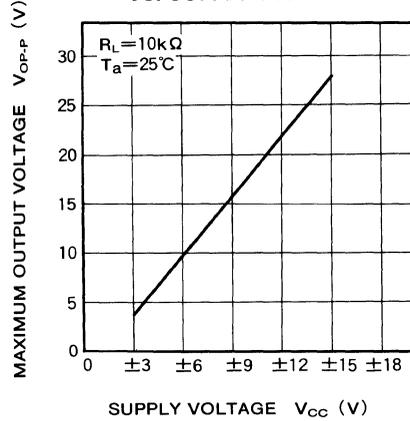
MITSUBISHI LINEAR ICs M5221L, P, FP / M5T082P

DUAL J-FET INPUT OPERATIONAL AMPLIFIERS

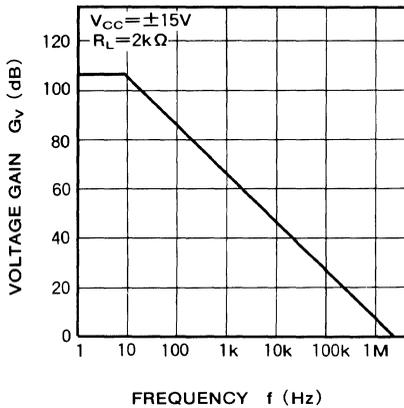
OUTPUT VOLTAGE VS. LOAD RESISTANCE



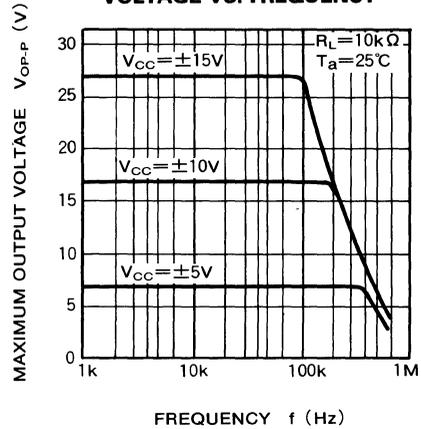
MAXIMUM OUTPUT VOLTAGE VS. SUPPLY VOLTAGE



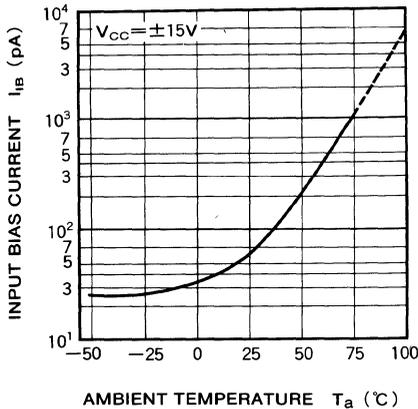
VOLTAGE GAIN VS. FREQUENCY RESPONSE



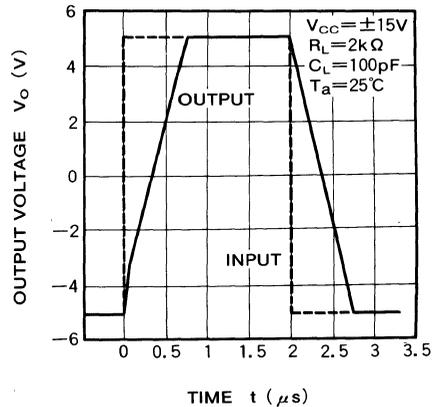
MAXIMUM OUTPUT VOLTAGE VS. FREQUENCY



INPUT BIAS CURRENT VS. AMBIENT TEMPERATURE



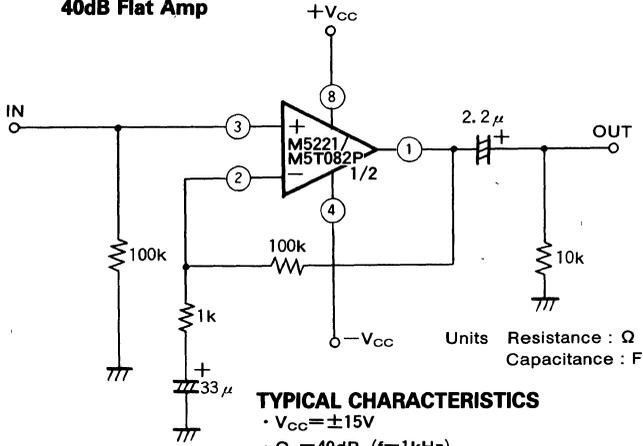
SLEW RATE (SR) CHARACTERISTICS



DUAL J-FET INPUT OPERATIONAL AMPLIFIERS

APPLICATION EXAMPLE

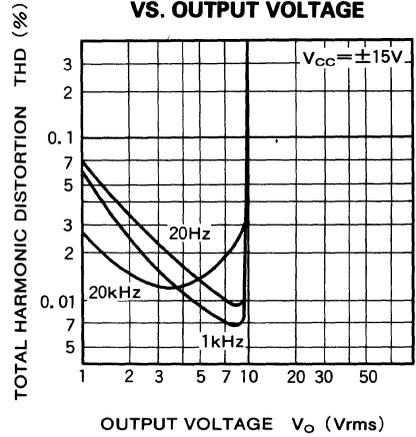
40dB Flat Amp



TYPICAL CHARACTERISTICS

- $V_{CC} = \pm 15V$
- $G_v = 40dB$ ($f = 1kHz$)
- $V_o = 9.5V_{rms}$ ($f = 1kHz$, $THD = 0.1\%$)
- $THD = 0.007\%$ ($f = 1kHz$, $V_o = 7V_{rms}$)

TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



MITSUBISHI LINEAR ICs M5238L, P, FP

DUAL LOW-NOISE J-FET INPUT OPERATIONAL AMPLIFIERS

DESCRIPTION

The M5238 is a semiconductor integrated circuit designed as a low-noise Bi-FET operational amplifier which adopts J-FETs in the input stage. Noise reduction characteristic in the input stage has been improved by 3 - 4dB, when compared with the M5221 general-purpose Bi-FET operational amplifier, and two circuits for yielding a high input impedance, high slew rate and low bias current and other excellent characteristics, are housed in an 8-pin SIP, DIP or FP.

It can be widely used as a general-purpose operational amplifier in stereo equipment, tape decks, digital audio disc players and other similar products as well as in VTRs, video disc players and video related players.

FEATURES

- Low noise, input-referred noise $V_{NI}=1.9\mu\text{Vrms}(\text{typ.})$
 $(R_S=100\text{k}\Omega \text{ BW}10\text{Hz}\sim 30\text{kHz FLAT})$
 $S/N=73\text{dB}(\text{typ.})$
 (Shorted input, RIAA, IHF-A network, PHONO 2.5mVrms)
- High input impedance due to J-FET input
 $R_i=1000\text{M}\Omega(\text{typ.})$
- High slew rate $S_R=20\text{V}/\mu\text{s}(\text{typ.})$
- High gain, low distortion
 $G_{VO}=100\text{dB}(\text{typ.})$, $\text{THD}=0.002\%$
 $(G_V=35.6\text{dB, RIAA, }V_O=5\text{Vrms})$
- Large load current and allowable current
 $I_{LP}=\pm 50\text{mA}$, $P_d=800\text{mW}(\text{SIP})$
 $P_d=625\text{mW}(\text{DIP})$, $P_d=440\text{mW}(\text{FP})$

APPLICATION

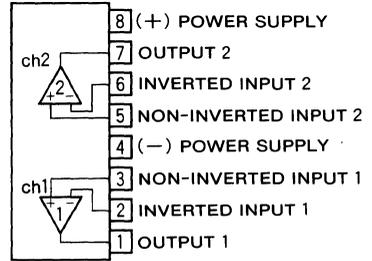
General purpose preamplifier in stereo equipment, tape decks and digital audio disc players, VTRs and video disc players.

RECOMMENDED OPERATING CONDITIONS

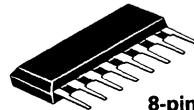
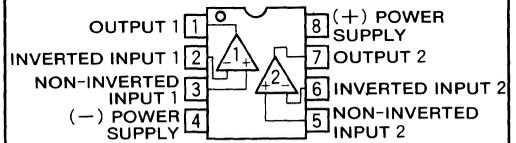
- Supply voltage range $\pm 5\sim\pm 15\text{V}$
- Rated supply voltage $\pm 15\text{V}$

PIN CONFIGURATION (TOP VIEW)

SIP



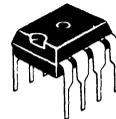
DIP, MINI FLAT



8-pin molded plastic SIP

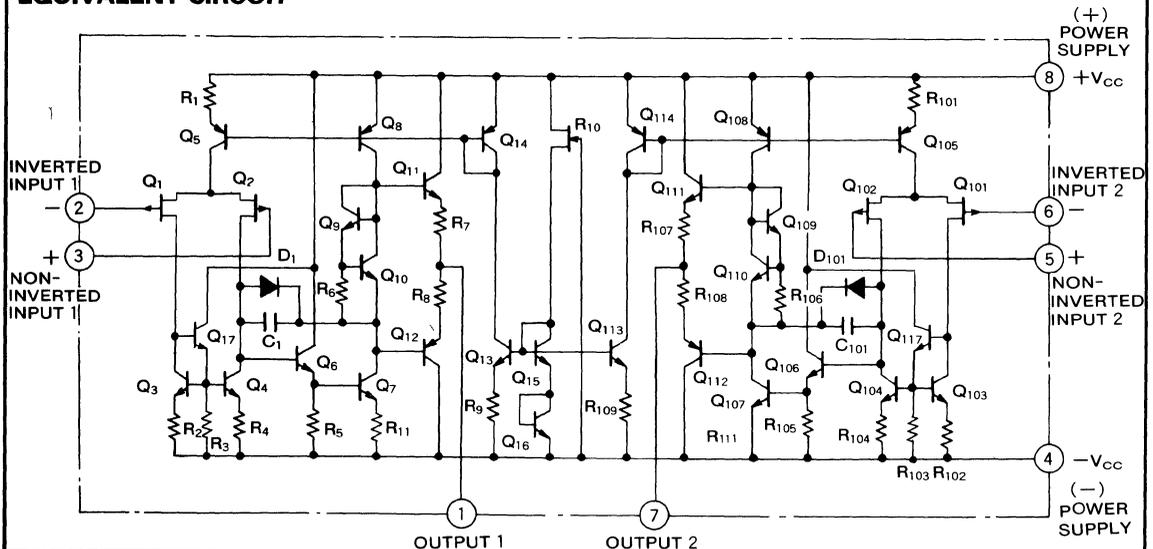


8-pin molded plastic FP
(MINI FLAT)



8-pin molded plastic DIP

EQUIVALENT CIRCUIT



DUAL LOW-NOISE J-FET INPUT OPERATIONAL AMPLIFIERS

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

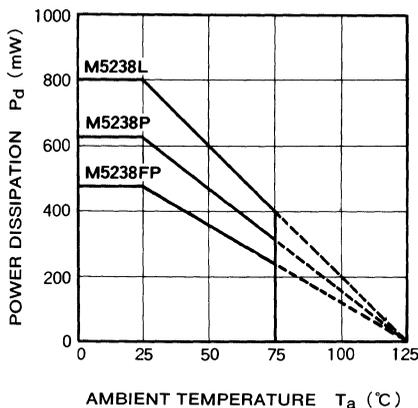
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		± 18	V
I_{LP}	Load current		± 50	mA
V_{id}	Differential input voltage		± 30	V
V_{ic}	Common input voltage		± 15	V
P_d	Power dissipation		800(SIP)/625(DIP)/440(FP)	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	8(SIP)/6.25(DIP)/4.4(FP)	mW/°C
T_{opr}	Ambient temperature		$-20 \sim +75$	°C
T_{stg}	Storage temperature		$-55 \sim +125$	°C

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 15\text{V}$)

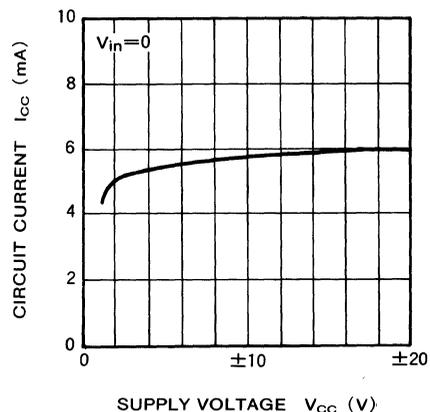
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{in}=0$		5.8	9.0	mA
V_{IO}	Input offset voltage	$R_s \leq 10\text{k}\Omega$		2.0	10.0	mV
I_{IO}	Input offset current			5	200	pA
I_{IB}	Input bias current			30	400	pA
R_{in}	Input resistance			10^9		MΩ
G_{VO}	Open loop voltage gain	$R_L \geq 2\text{k}\Omega$, $V_o = \pm 10\text{V}$	86	106		dB
V_{OM}	Maximum output voltage	$R_L \geq 10\text{k}\Omega$	± 12	± 14		V
		$R_L \geq 2\text{k}\Omega$	± 10	± 13		
V_{CM}	Common input voltage width		± 10	± 12		V
CMRR	Common mode rejection ratio	$R_s \leq 10\text{k}\Omega$	70	76		dB
SVRR	Supply voltage rejection ratio	$R_s \leq 10\text{k}\Omega$		30	150	$\mu\text{V/V}$
P_d	Power dissipation			174	270	mW
SR	Slew rate	$G_v=0\text{dB}$, $R_L=2\text{k}\Omega$		20		V/ μs
f_T	Gain bandwidth product			6		MHz
V_{NI}	Input referred noise voltage	$R_s=100\Omega$, BW=10Hz~30kHz		1.9		μVrms

TYPICAL CHARACTERISTICS

**THERMAL DERATING
(MAXIMUM RATING)**

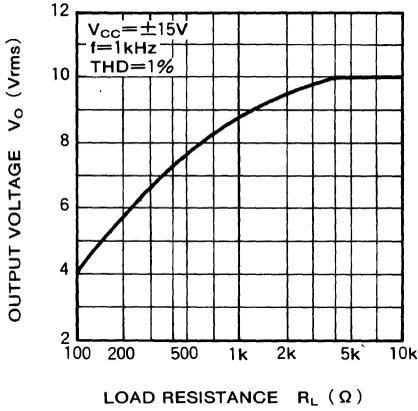


**CIRCUIT CURRENT VS.
SUPPLY VOLTAGE**

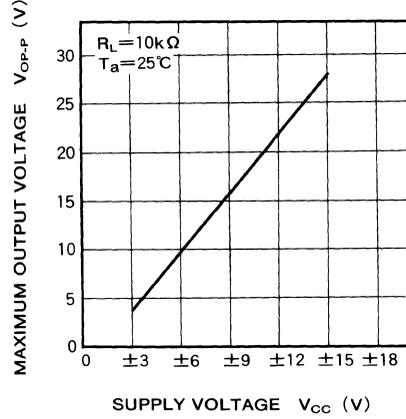


DUAL LOW-NOISE J-FET INPUT OPERATIONAL AMPLIFIERS

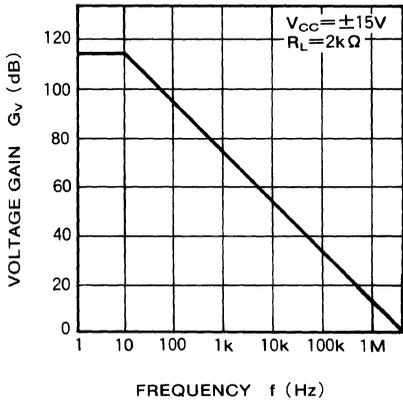
OUTPUT VOLTAGE VS. LOAD RESISTANCE



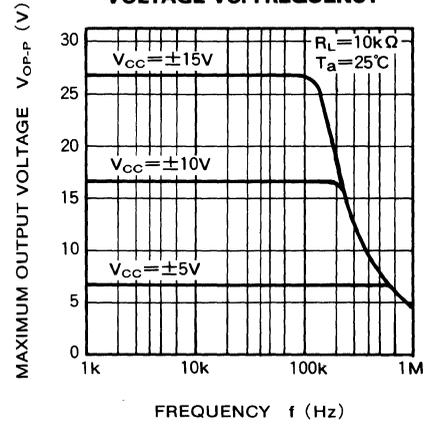
MAXIMUM OUTPUT VOLTAGE VS. SUPPLY VOLTAGE



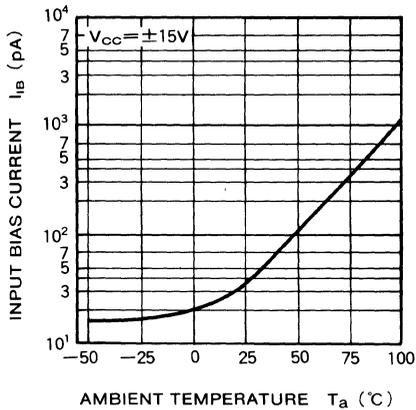
VOLTAGE GAIN VS. FREQUENCY RESPONSE



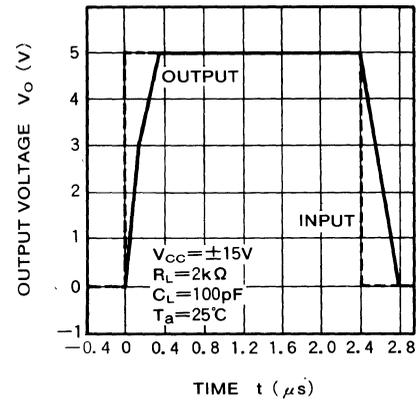
MAXIMUM OUTPUT VOLTAGE VS. FREQUENCY



INPUT BIAS CURRENT VS. AMBIENT TEMPERATURE



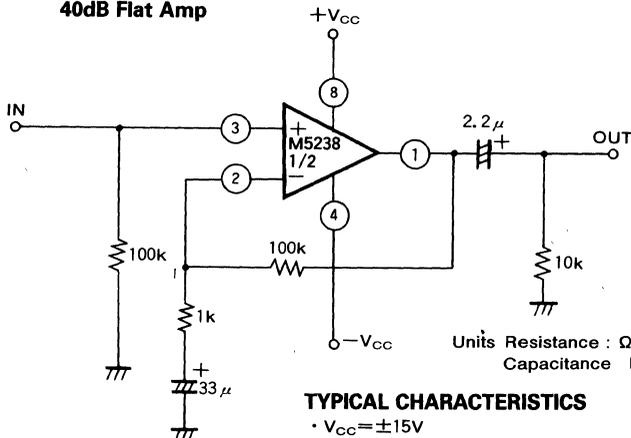
SLEW RATE (SR) CHARACTERISTICS



DUAL LOW-NOISE J-FET INPUT OPERATIONAL AMPLIFIERS

APPLICATION EXAMPLE

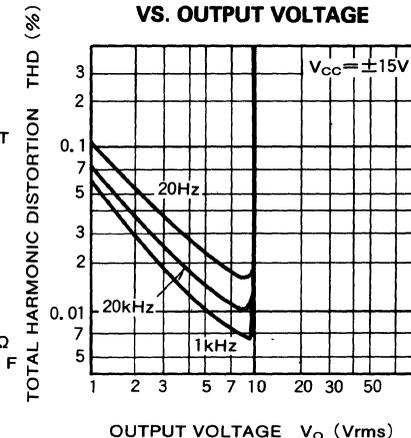
40dB Flat Amp



TYPICAL CHARACTERISTICS

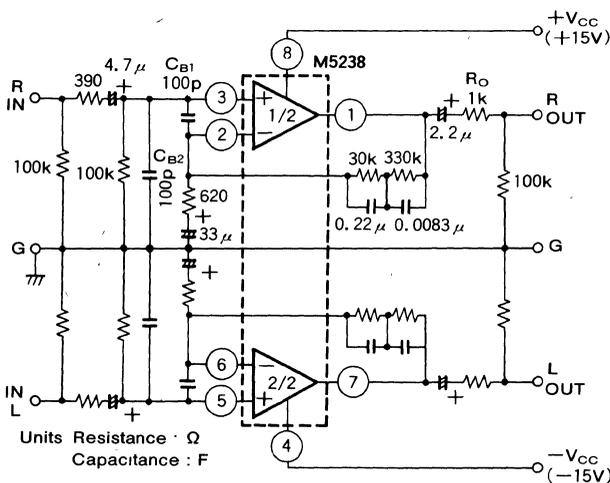
- $V_{CC} = \pm 15V$
- $G_V = 40dB (f = 1kHz)$
- $V_O = 9.5V_{rms} (f = 1kHz, THD = 0.1\%)$
- $THD = 0.007\% (f = 1kHz, V_O = 7V_{rms})$

TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



APPLICATION EXAMPLE

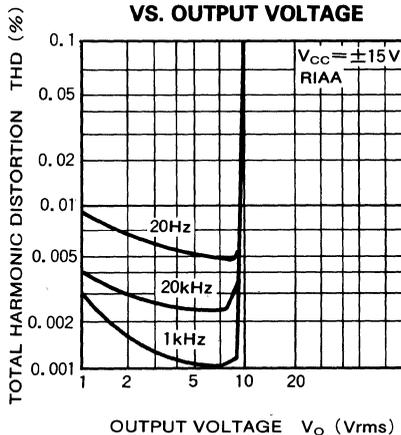
Stereo equalizer amplifier circuit



TYPICAL CHARACTERISTICS ($V_{CC} = \pm 15V, R_{IAA}$)

- $G_V = 35.6dB (f = 1kHz)$
- $V_{NI} = 1.9\mu V_{rms} (R_s = 100\Omega, BW = 20Hz \sim 30kHz)$
- $S/N = 73dB$ (IHF-A network, shorted input, 2.5mVrms input sensitivity)
- $THD = 0.001\% (f = 1kHz, V_O = 7V_{rms})$

TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



L_{ch} circuit constants are identical to those of R_{ch}

C_{B1}, C_{B2} : Capacitors for buzz prevention, use if required.

R_O : Resistor used to prevent parasitic oscillation for capacitive loads and current limiting with shorted and other abnormal load conditions.

M5240P

DUAL LOW-NOISE J-FET INPUT OPERATIONAL AMPLIFIERS

DESCRIPTION

The M5240P is a semiconductor integrated circuit designed as a low-noise Bi-FET operational amplifier which adopts J-FETs in the input stage. The device comes in a 16-pin DIP and contains two circuits for yielding high sound quality, high input impedance, high slew rate and low distortion characteristics. It can be widely used as an operational amplifier in stereo equipment, tape decks, and as preamplifier for digital audio disc players and other similar products, mixer for public announcement or LM, high fidelity VTRs.

FEATURES

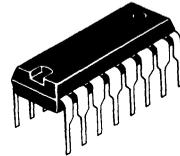
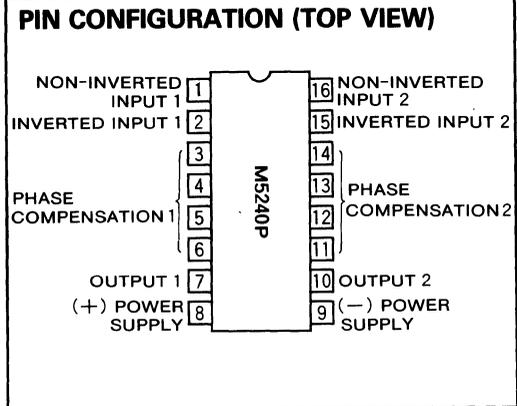
- Low noise S/N=82dB(typ.)
(shorted input, RIAA, IHF-A, PHONO 2.5mVrms)
- High slew rate SR=40V/ μ s(typ.)
(Variable by externally connected RC due to external phase compensation type)
- High input impedance $R_i=1000M\Omega$ (typ.)
- Low distortion rate THD=0.001%(typ.)
(f=1kHz, RIAA, $V_o=5V_{rms}$)
- Large load current and allowable current $I_{LP}=\pm 50mA$
 $P_d=1W$

APPLICATION

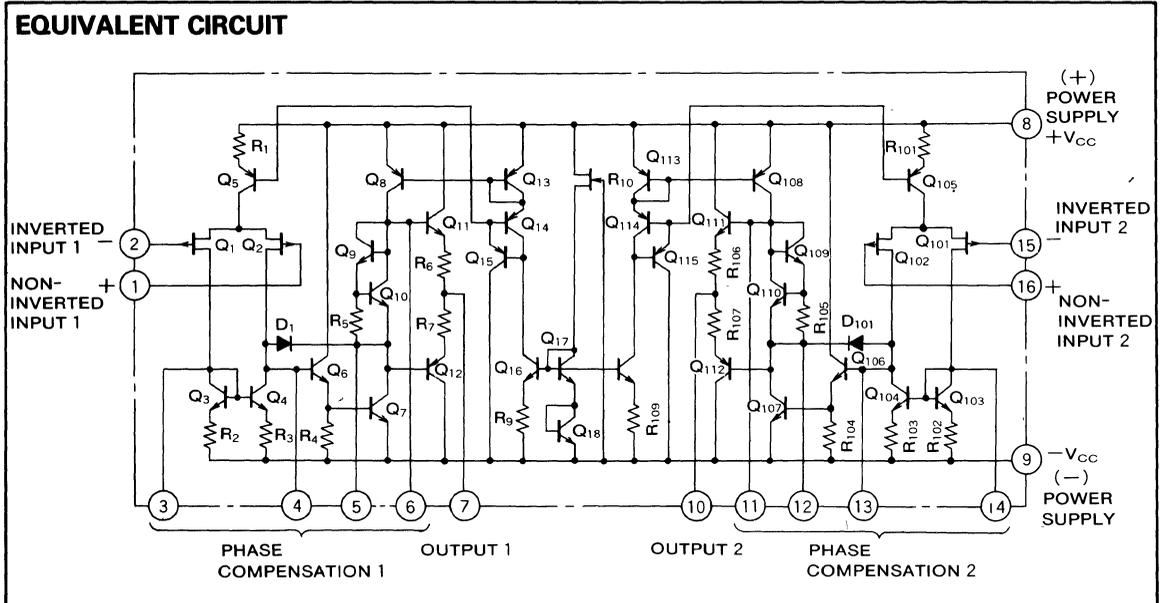
Amplifier in stereo equipment, tape decks and digital audio disc players, mixer.

RECOMMENDED OPERATING CONDITIONS

- Supply voltage range $\pm 5 \sim \pm 15V$
- Rated supply voltage $\pm 15V$



16-pin molded plastic DIP



DUAL LOW-NOISE J-FET INPUT OPERATIONAL AMPLIFIERS

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

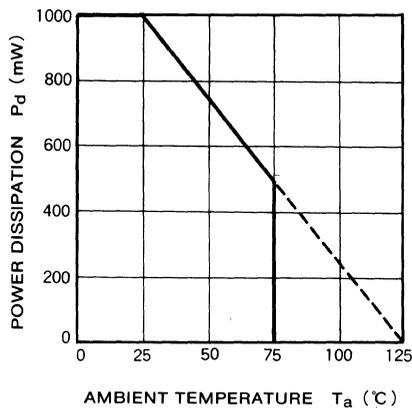
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		± 18	V
I_{LP}	Load current		± 50	mA
V_{id}	Differential input voltage		± 30	V
V_{ic}	Common input voltage		± 15	V
P_d	Power dissipation		1000	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	10	mW/ $^\circ\text{C}$
T_{opr}	Ambient temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-55 \sim +125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 15\text{V}$)

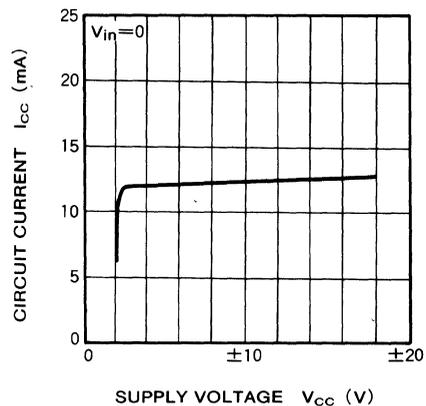
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{in}=0$		12.0	17.0	mA
V_{IO}	Input offset voltage	$R_s \leq 10\text{k}\Omega$		5.0	15.0	mV
I_{IO}	Input offset current			25	1000	pA
I_{IB}	Input bias current			120	2000	pA
R_{in}	Input resistance			10^3		M Ω
G_{VO}	Open loop voltage gain	$R_L \geq 2\text{k}\Omega$, $V_o = \pm 10\text{V}$		110		dB
V_{OM}	Maximum output voltage	$R_L \geq 10\text{k}\Omega$	± 12	± 14		V
		$R_L \geq 2\text{k}\Omega$	± 10	± 13		
V_{CM}	Common input voltage width		± 10	± 12		V
CMRR	Common mode rejection ratio	$R_s \leq 10\text{k}\Omega$	70	76		dB
SVRR	Supply voltage rejection ratio	$R_s \leq 10\text{k}\Omega$		30	150	$\mu\text{V}/\text{V}$
P_d	Power dissipation			360	510	mW
SR	Slew rate	$G_v = 16.5\text{dB}$		40		V/ μs
f_T	Gain bandwidth product			18		MHz
e_n	Equivalent input referred noise voltage	$R_s = 100\Omega$, BW=10Hz~30kHz		3.3		nV/ $\sqrt{\text{Hz}}$
V_{NI}		$R_g = 2.2\text{k}\Omega$, RIAA EQ		1.2	2.5	μV_{rms}
S/N	Signal-to-noise voltage ratio	$R_g = 47\Omega$, RIAA EQ, IHF-A		82		dB

TYPICAL CHARACTERISTICS

**THERMAL DERATING
(MAXIMUM RATING)**

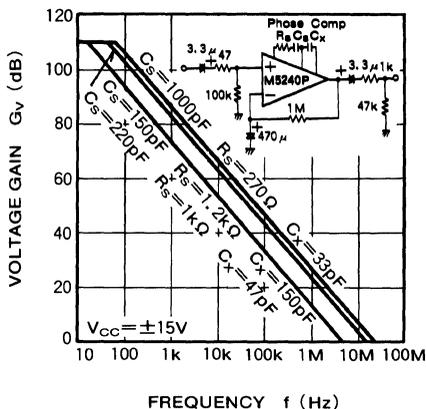


**CIRCUIT CURRENT VS.
SUPPLY VOLTAGE**

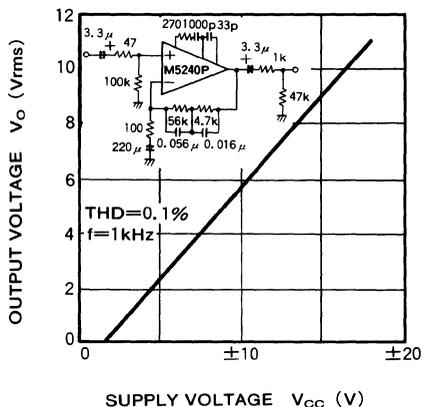


DUAL LOW-NOISE J-FET INPUT OPERATIONAL AMPLIFIERS

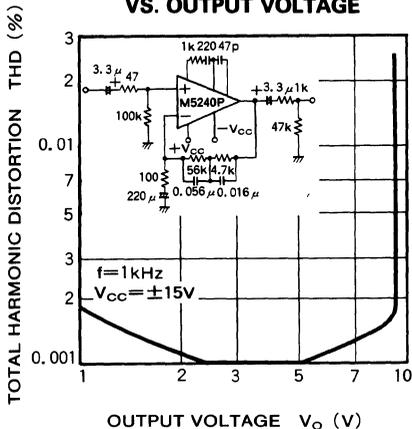
VOLTAGE GAIN VS. FREQUENCY RESPONSE



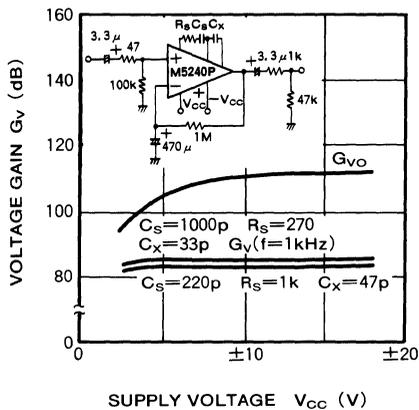
OUTPUT VOLTAGE VS. SUPPLY VOLTAGE



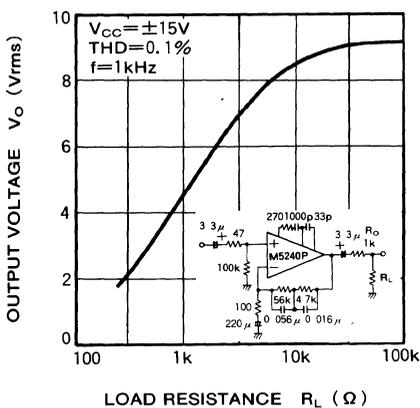
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



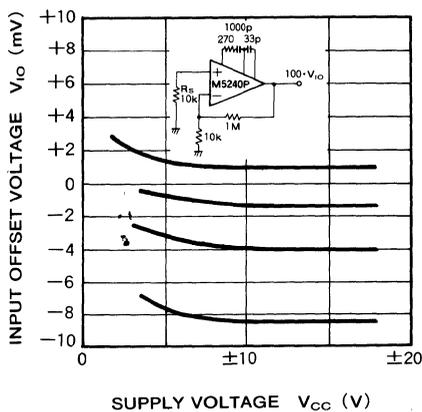
VOLTAGE GAIN VS. SUPPLY VOLTAGE



OUTPUT VOLTAGE VS. LOAD RESISTANCE



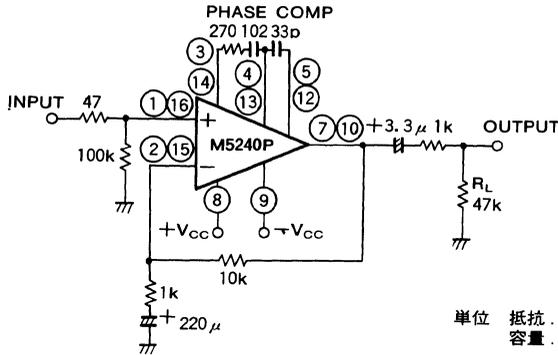
INPUT OFFSET VOLTAGE VS. SUPPLY VOLTAGE



DUAL LOW-NOISE J-FET INPUT OPERATIONAL AMPLIFIERS

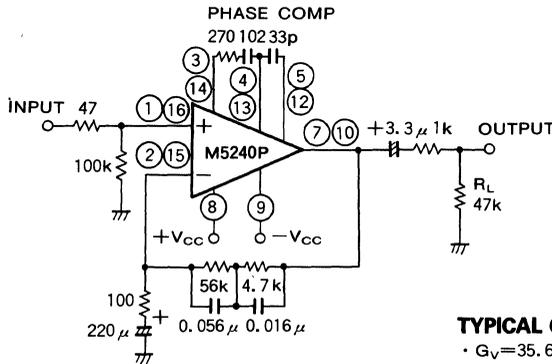
TYPICAL APPLICATION EXAMPLES

1. $G_V=20\text{dB}$ FLAT AMP



單位 抵抗 . Ω
 容量 . F

2. STEREO EQUALIZER AMP

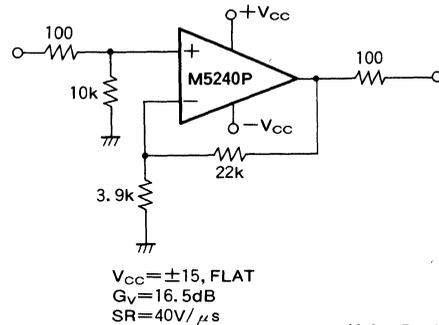


Unit Resistance : Ω
 Capacitance : F

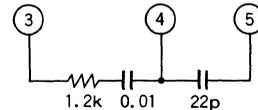
TYPICAL CHARACTERISTICS ($V_{CC}=\pm 15\text{V}$, RIAA)

- $G_V=35.6\text{dB}$ ($f=1\text{kHz}$)
- $V_{NI}=1.2\mu\text{Vrms}$ ($R_g=2.2\text{k}\Omega$, $\text{BW}=10\text{Hz}\sim 30\text{kHz}$)
- $\text{THD}=0.001\%$ ($f=1\text{kHz}$, $V_O=5\text{Vrms}$)
- $\text{S/N}=82\text{dB}$ (IHF-A network, shorted input, 2.5mrms input sensitivity)

3. HIGH SLEW RATE FLAT AMP



$V_{CC}=\pm 15\text{V}$, FLAT
 $G_V=16.5\text{dB}$
 $\text{SR}=40\text{V}/\mu\text{s}$



Phase compensation

SR +40V/μs
 -40V/μs

Unit Resistance : Ω
 Capacitance : F

M5201L, P, FP

GENERAL PURPOSE SWITCHING OPERATIONAL AMPLIFIER (DUAL INPUT, SINGLE OUTPUT TYPE)

DESCRIPTION

The M5201 is a semiconductor integrated circuit designed for an operational amplifier which adopts analog switch function, having dual inputs of A and B and a single output. The device comes in an 8-pin SIP, DIP or FP and contains input differential circuits of A and B type, single output circuit and a switching circuit of an operational amplifier, and can be used as a conventional operational amplifier, turning on A or B inputs by externally setting the control pin at high or low level. For a voltage follower condition where $G_v = 0\text{dB}$, the device functions merely as an analog switch, but, for an amplifier with a switching function, gain can be set independently for A and B inputs. The M5201 operational amplifier has basic characteristics similar to those of the M5218/M5R4558P and can be widely used as audio, video and musical instrument equipments.

FEATURES

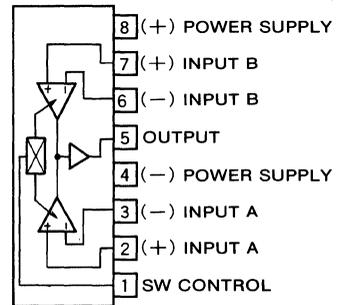
- Operational amplifier inputs of A and B type and gain can be set independently
- Applicable to both single and dual power supplies
- High gain, low distortion
..... $G_{V0} = 100\text{dB}$, $\text{THD} = 0.002\%$ (typ.)
- High slew rate, high f_T $\text{SR} = 2.2\text{V}/\mu\text{s}$, $f_T = 7\text{MHz}$ (typ.)
- Low noise ($R_S = 1\text{k}\Omega$) FLAT $V_{NI} = 2\mu\text{Vrms}$ (typ.)
- Small switching shock noise
- Large-current, large allowable dissipation
..... $I_{LP} = \pm 50\text{mA}$, $P_d = 800\text{mW}$ (SIP)
 $P_d = 625\text{mW}$ (DIP)
 $P_d = 440\text{mW}$ (FP)

RECOMMENDED OPERATING CONDITIONS

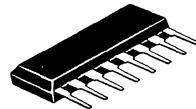
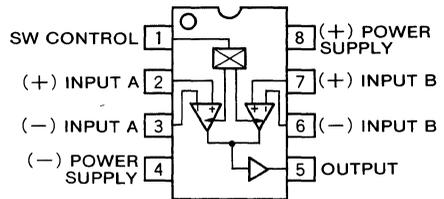
Supply voltage range $\pm 2.5 \sim \pm 16\text{V}$
Rated supply voltage $\pm 15\text{V}$

PIN CONFIGURATION (TOP VIEW)

SIP



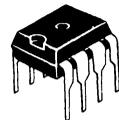
DIP, MINI FLAT



8-pin molded plastic SIP

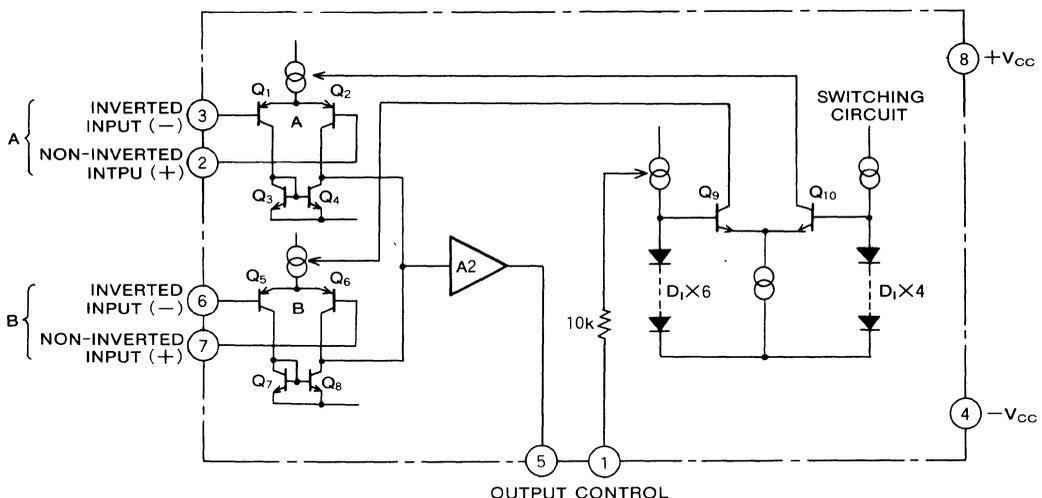


8-pin molded plastic FP
(MINI FLAT)



8-pin molded plastic DIP

EQUIVALENT CIRCUIT



COMPARATOR

QUICK REFERENCE TABLE OF COMPARATORS

Application	Type No.	Function	Features	Maximum ratings		Electrical characteristics (Typical characteristics)					Equivalent	Package outline	
				Operating supply voltage range V_{CC} (V)	Power dissipation P_d (mW)	Circuit current I_{CC} (mA)	Output sink current I_{sink} (mA)	Input offset voltage V_{IO} (mV)	Input bias current I_B (nA)	Response time (μs)			
Single comparator	M51201	Single (built-in reference voltage circuit for comparison)	<ul style="list-style-type: none"> Operates at low supply voltage Large output sink current 	1.7~6.5	TL:180 FP:300	2	60 (at $V_O=0.3V$)	—	8	0.2 (Rise of output) 50 (Fall of output)	—	TL: 5-pin SIP 	
	M51203		<ul style="list-style-type: none"> Wide supply voltage range Large output sink current 	3.0~28		2	60 (at $V_O=0.3V$)	—	20	1.0 (Rise of output) 10 (Fall of output)	—		
	M51205		<ul style="list-style-type: none"> Large output sink current 	Including zener diode for shunt regulator		1.9	60 (at $V_O=0.3V$)	—	20	1.0 (Rise of output) 10 (Fall of output)	—		
	M51202		Single	<ul style="list-style-type: none"> Operates at low supply voltage Large output sink current 		1.7~6.5	1.7	60 (at $V_O=0.3V$)	2	8	0.2 (Rise of output) 50 (Fall of output)	—	FP: 8-pin mini FLAT 
	M51204			<ul style="list-style-type: none"> Wide supply voltage range Large output sink current 		2.5~28	1.8	60 (at $V_O=0.3V$)	2	20	1.0 (Rise of output) 10 (Fall of output)	—	
	M51206			<ul style="list-style-type: none"> Large output sink current 		Including zener diode for shunt regulator	1.8	60 (at $V_O=0.3V$)	2	20	1.0 (Rise of output) 10 (Fall of output)	—	
Dual comparator	M51207L	Dual	<ul style="list-style-type: none"> Wide supply voltage range Large output sink current 	2.5~28	650	3.8	60 (at $V_O=0.2V$)	2	20	2.0 (Rise of output) 1.0 (Fall of output)	Functional equivalent of AN6916	L: 8-pin SIP 	
	M51922L		<ul style="list-style-type: none"> Wide supply voltage range Low power dissipation 	2.5~28	650	1.5 (All output ON) 0.4 (All output OFF)	20 (at $V_O=0.15V$)	2	25	2.0 (Rise of output) 0.2 (Fall of output)	—		
	M51923		<ul style="list-style-type: none"> Wide supply voltage range Low power dissipation 	2.5~28	650 (M51923P) 300 (M51923FP)	1.5 (All output ON) 0.4 (All output OFF)	20 (at $V_O=0.15V$)	2	25	2.0 (Rise of output) 0.2 (Fall of output)	2903	L: 8-pin SIP FP: 8-pin mini FLAT 	
	★ M5233		High output break down voltage dual	<ul style="list-style-type: none"> Wide supply voltage range Low power dissipation Large output sink current 	2~36	800 (M5233P) 625 (M5233L) 440 (M5233FP)	0.6	25	2	25	1.3	393 2903	L: 8-pin SIP P: 8-pin DIP FP: 8-pin mini FLAT 

★: New product

QUICK REFERENCE TABLE OF COMPARATORS (CONTINUED)

Application	Type No.	Function	Features	Maximum ratings		Electrical characteristics (Typical characteristics)					Equivalent	Package outline
				Operating supply voltage range V_{CC} (V)	Power dissipation P_d (mW)	Circuit current I_{CC} (mA)	Output sink current I_{sink} (mA)	Input offset voltage V_{IO} (mV)	Input bias current I_B (nA)	Response time (μs)		
Quad comparator	M51209P	Quad	<ul style="list-style-type: none"> Wide supply voltage range Large output sink current 	2.5~28	900	6.8	60 (at $V_O=0.2V$)	2	20	2.0 (Rise of output) 1.0 (Fall of output)	Functional equivalent of AN6918	14-pin DIP 
	M51924		<ul style="list-style-type: none"> Wide supply voltage range Low power dissipation current 	2.5~28	900 (M51924P) 350 (M51924FP)	3.0 (All output ON) 0.8 (All output OFF)	20 (at $V_O=0.15V$)	2	25	2.0 (Rise of output) 0.2 (Fall of output)	2901	P: 14-pin DIP  FP: 14-pin mini FLAT 
	★ M5234	High voltage quad	<ul style="list-style-type: none"> Wide supply voltage range Low power dissipation current Large output sink current 	2~36	700 (M5234P) 550 (M5234FP)	0.8	25	2	25	1.3	339 2901	P: 14-pin DIP  FP: 14-pin mini FLAT 

INDEX OF FUNCTIONAL EQUIVALENTS COMPARATORS

Application	Mitsubishi		Functional equivalents										
	Function	Type No.	JRC	NEC	Matsushita	Toshiba	Hitachi	Tokyo Sanyo	ROHM	TI	NSS	FC	MOTO
Comparator	High output break down voltage dual, general-purpose comparator	M5233L	NJM2903S(9L)		AN6913	TA75393S(9L)		LA69393S(9L)					
		M5233P	NJM2903D	μ PC393C	AN1393 AN6914	TA75393P	HA17903PS	LA6393D	BA6993	LM393 LM2903	LM393 LM2903	μ A393 μ A2903	
		★ M5233FP	NJM2903M	μ PC393G	AN1393S	TA75393F		LA6393M	BA6993F				
	High output break down voltage dual, general-purpose comparator	M5234P	NJM2901D	μ PC339C	AN1339 AN6912	TA75339P	HA17901P	LA6339	BA10339	LM339 LM2901	LM339 LM2901	μ A339 μ A2901	MLM339P MLM2901P
		M5234FP	NJM2901M	μ PC339G	AN1339NS	TA75339F		LA6339M	BA10339F				

L: SIP (8-pin), P: DIP (8-, 14-pin), (9L): SIP (9-pin), FP: mini FLAT (8-, 14-pin)

★: New product

MITSUBISHI LINEAR ICs
M51201TL,FP

VOLTAGE COMPARATOR

DESCRIPTION

The M51201TL,FP is a semiconductor integrated circuit for a voltage comparator that operates from a single power supply. Especially the M51201TL,FP has superiority as to characteristics of input current (high input resistance) and fits to wide ranged applications, for example, CR timer, etc. M51201TL,FP's package is a mini SIP and FLAT package, therefore can use very easily.

FEATURES

- Low input current8nA(typ.)
- Capable of directly driving a relay or a lamp
- Operates at low supply voltage1.7~6.5V
- Low power dissipation2mA typ. ($V_{CC}=2.65V$)
- Including reference voltage for comparison $V_{CC}/2$
- Including voltage surge absorbing zener diode for protection
- High output breakdown voltage 18V(max.)

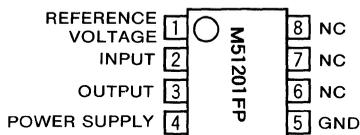
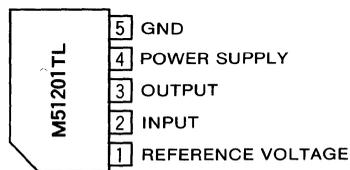
APPLICATION

Electric shutter, CR timer, comparator, time delay circuit, oscillator (square wave)

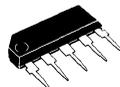
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range1.7~6.5V
- Rated supply voltage5.0V

PIN CONFIGURATION (TOP VIEW)



NC · NO CONNECTION

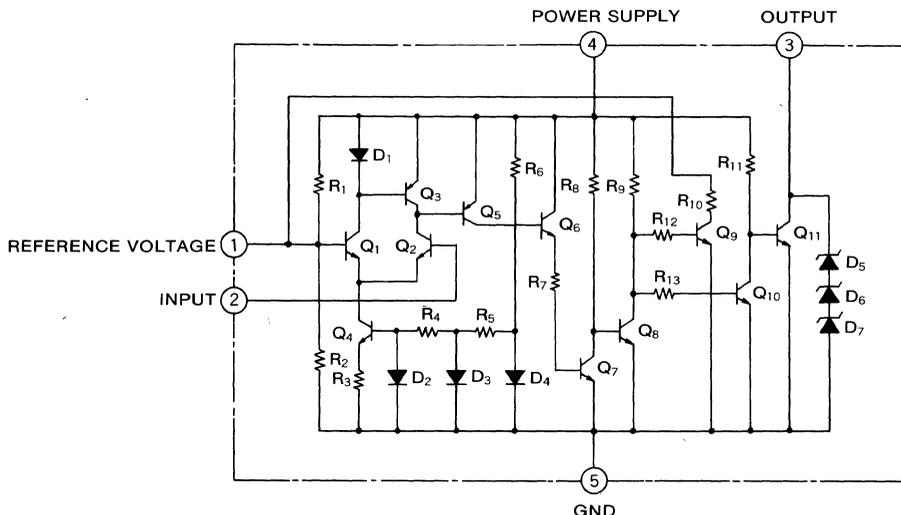


5-pin molded plastic SIP



8-pin molded plastic FP
(MINI FLAT)

EQUIVALENT CIRCUIT



VOLTAGE COMPARATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$, unless otherwise noted)

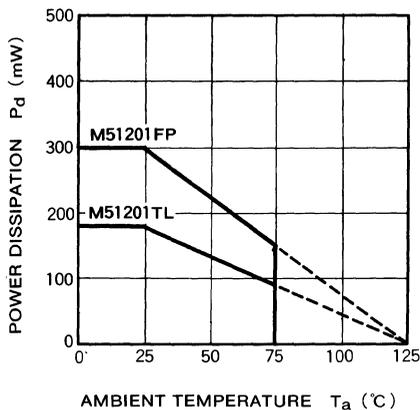
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		6.5	V
I_{sink}	Output sink current		200	mA
V_{IN}	Input voltage		V_{CC}	V
P_d	Power dissipation		180(M51201TL)	mW
			300(M51201FP)	
K_{θ}	Thermal derating	$T_a \geq 25^{\circ}\text{C}$	1.8(M51201TL) 3.0(M51201FP)	mW/ $^{\circ}\text{C}$
T_{opr}	Operating temperature		-20~+75	$^{\circ}\text{C}$
T_{stg}	Storage temperature		-40~+125	$^{\circ}\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$)

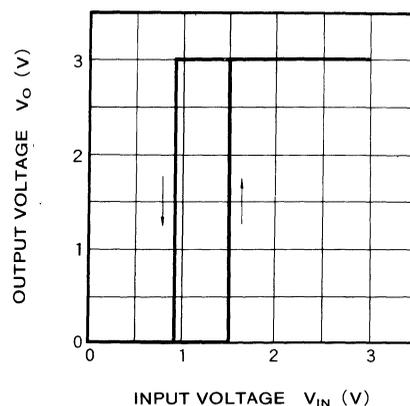
Symbol	Parameter	Test conditions	Limits			Unit	
			V_{CC} (V)	Min	Typ		Max
V_{CC}	Supply voltage range			1.7		6.5	V
I_{CC}	Circuit current		2.65		2.0	3.5	mA
			6.0		5.0	8.8	
I_{IN}	Input current		2.65		8	100	nA
V_{REF}	Reference voltage		6.0	2.55	3.0	3.45	V
V_{OL}	Output saturation voltage	$I_{sink}=60\text{mA}$ $I_{sink}=200\text{mA}$	6.0		0.2	0.6	V
						1	
V_Z	Zener voltage	$I_Z=5\text{mA}$		18	22	26	V
t_{PLH}	Output "L-H" propagation delay time		6.0		0.2		μs
t_{PHL}	Output "H-L" propagation delay time		6.0		50		μs
V_{IN}	Input voltage range			0.8		$V_{CC}-0.2$	V

TYPICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, $V_{CC}=3\text{V}$, unless otherwise noted)

THERMAL DERATING (MAXIMUM RATING)

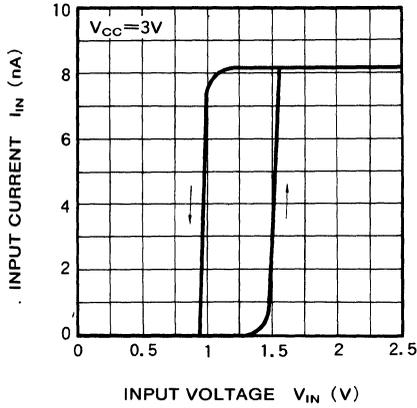


OUTPUT VOLTAGE VS. INPUT VOLTAGE

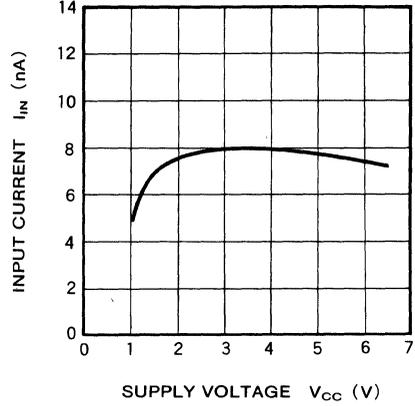


VOLTAGE COMPARATOR

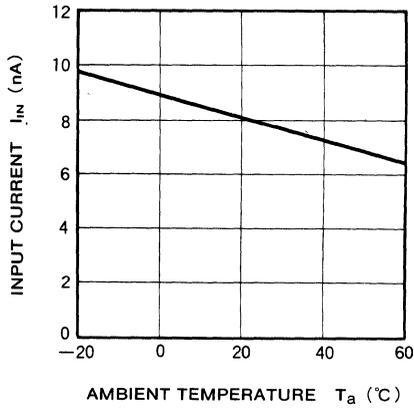
INPUT CURRENT VS. INPUT VOLTAGE



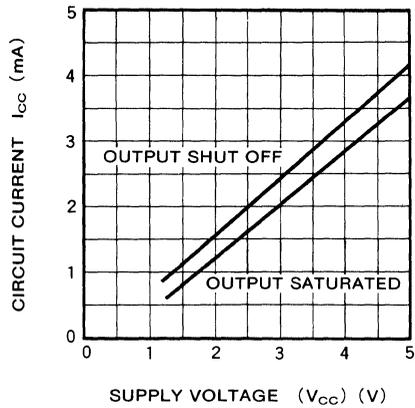
INPUT CURRENT VS. SUPPLY VOLTAGE



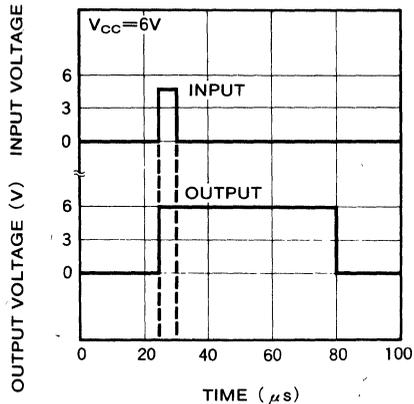
INPUT CURRENT VS. AMBIENT TEMPERATURE



CIRCUIT CURRENT VS. SUPPLY VOLTAGE

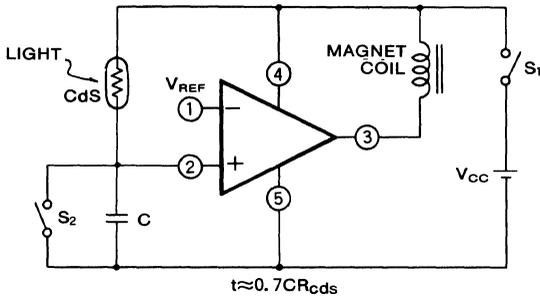


PULSE RESPONSE CHARACTERISTICS

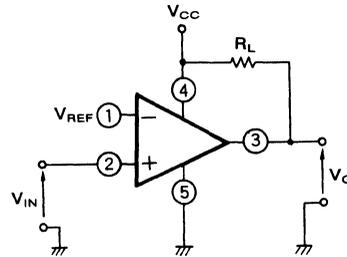


APPLICATION EXAMPLES

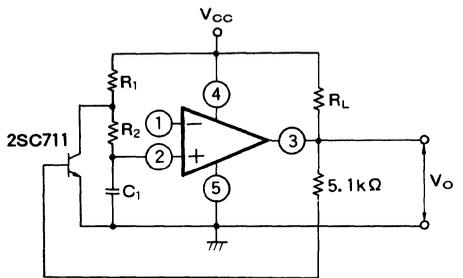
(1) Electric shutter



(2) Voltage comparator



(3) Oscillator



$$f = \frac{1}{C_1(0.337R_1 + 0.847R_2)} \text{ [Hz]}$$

PRECAUTIONS FOR USE

1. Paying much attention is necessary for fear that the M51201 TL,FP may flow large current and reach to destroy because of the structure when the terminals of V_{CC} and GND of the M51201 TL,FP is connected wrong position each other.
2. Reference voltage (V_{REF}) is adjustable for connecting external resistor, but adjustable voltage range is 0.8 to $(V_{CC}-0.2)(V)$.
3. Output is "open collector" and a loading resistor is not included. Connect a loading resistor to stabilize operation, in case of driving a next stage.

MITSUBISHI LINEAR ICs M51203TL,FP

VOLTAGE COMPARATOR

DESCRIPTION

The M51203TL,FP is a semiconductor integrated circuit for a voltage comparator that operates from a single power supply. Especially the M51203TL,FP has superiority as to characteristics of input current (high input resistance) and fits to wide ranged applications, for example CR timer, etc. M51203TL,FP's package is a mini SIP and FLAT package, therefore can use very easily.

FEATURES

- Low input current (high input resistance) 20nA (typ.)
- Operates at low supply voltage 3.0~28V
- Capable of directly driving a relay or a lamp
- Low power dissipation 2.5mA (max.)
- Including reference voltage for comparison
- Hysteresis characteristic between input and output
- High output breakdown voltage 30V (max.)

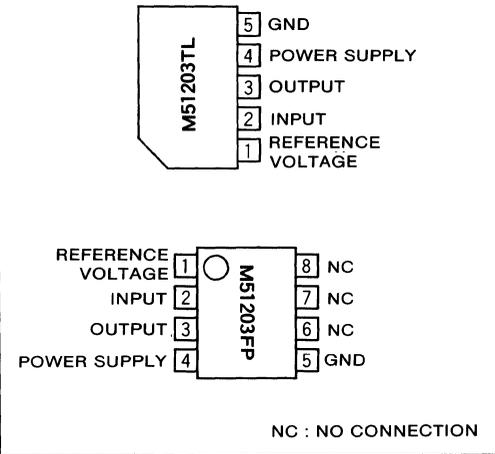
APPLICATION

Electric shutter, CR timer, voltage comparator, time delay circuit, oscillator (square wave)

RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 3.0~28V
- Rated supply voltage 12V

PIN CONFIGURATION (TOP VIEW)

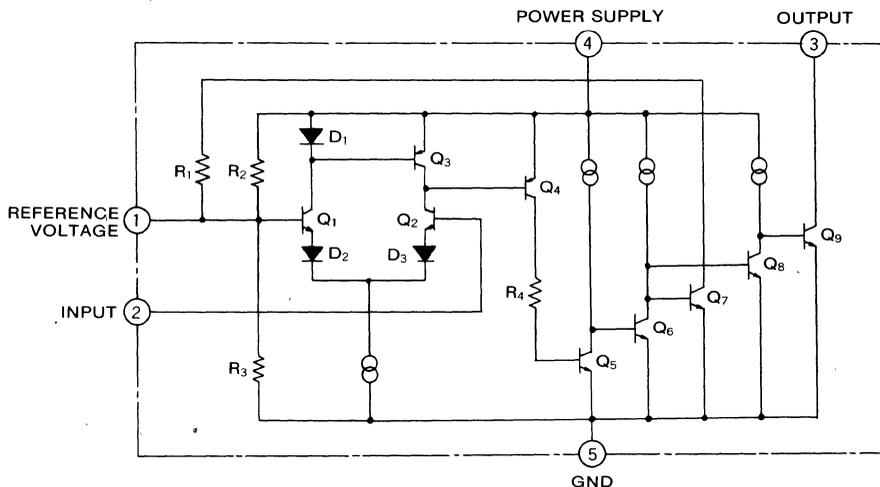


5-pin molded plastic SIP



8-pin molded plastic FLAT

EQUIVALENT CIRCUIT



VOLTAGE COMPARATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		28	V
V_{IN}	Input voltage		V_{CC}	V
I_{SINK}	Output sink current		200	mA
V_{OH}	Output drive voltage		30	V
P_d	Power dissipation		180(M51203TL)	mW
			300(M51203FP)	
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	1.8(M51203TL)	mW/ $^\circ\text{C}$
			3.0(M51203FP)	
T_{opr}	Operating temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^\circ\text{C}$

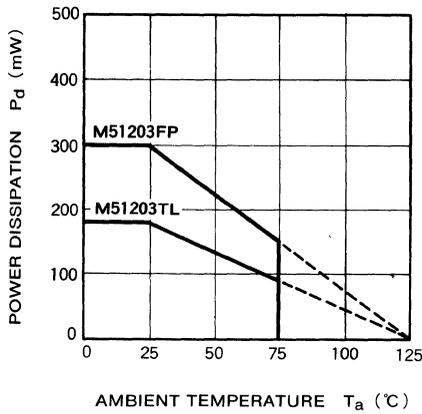
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$)

Symbol	Parameter	Test conditions	Limits			Unit	
			$V_{CC}(V)$	Min	Typ		Max
V_{CC}	Supply voltage range			3.0		28	V
I_{CC}	Circuit current		6.0		1.9	2.5	mA
			12.0		2.0		
			24.0		2.1		
V_{IN}	Input voltage		1.4		$V_{CC}-0.2$	V	
I_{IN}	Input current		6.0		20	75	nA
			12.0				
			24.0				
V_{REF}	Reference voltage	$V_{IN}=0V$	6.0	2.7	3.0	3.3	V
			12.0	5.4	6.0	6.6	
			24.0	10.8	12.0	13.2	
ΔV_{hys}	Input/output hysteresis width voltage		6.0	1.0	1.2	1.4	V
			12.0	1.9	2.4	2.9	
			24.0	3.8	4.8	5.8	
V_{OL}	Output saturation voltage	6.0	$I_{SINK}=60mA$ $I_{SINK}=200mA$		0.3	0.6	V
					1		
t_{PLH}	Output "L-H" propagation delay time	12.0			1		μs
t_{PHL}	Output "H-L" propagation delay time				10		μs

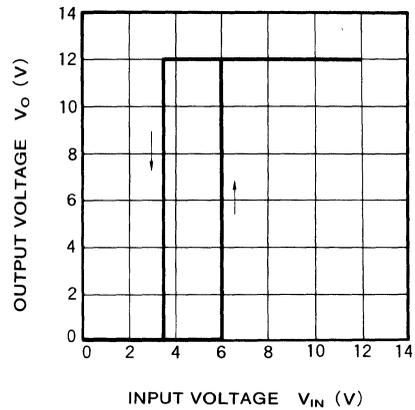
VOLTAGE COMPARATOR

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

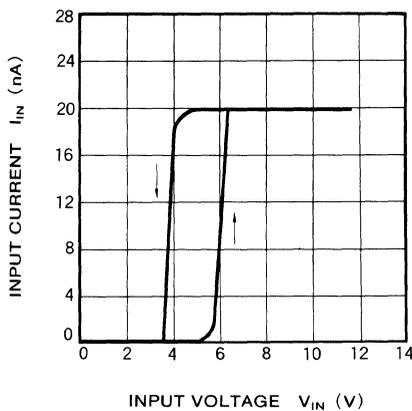
THERMAL DERATING (MAXIMUM RATING)



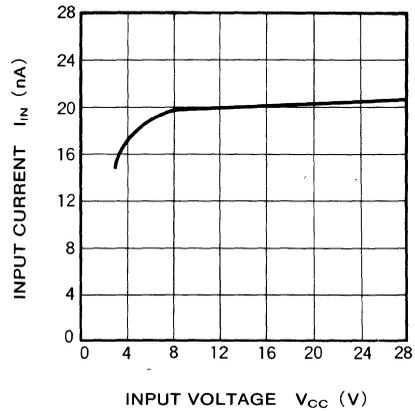
OUTPUT VOLTAGE VS. INPUT VOLTAGE



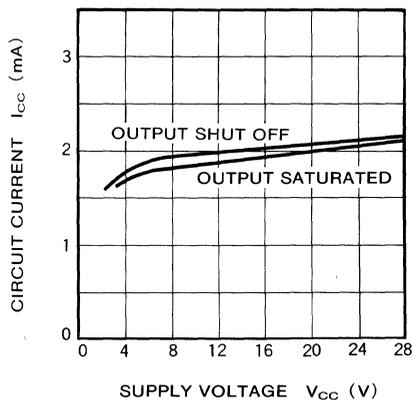
INPUT CURRENT VS. INPUT VOLTAGE



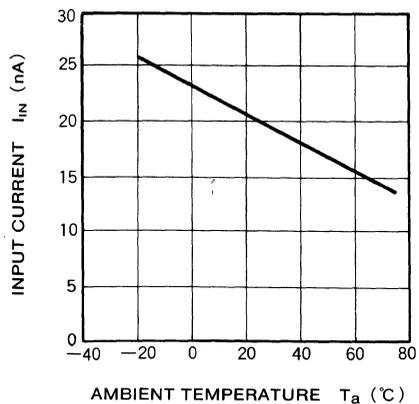
INPUT CURRENT VS. INPUT VOLTAGE



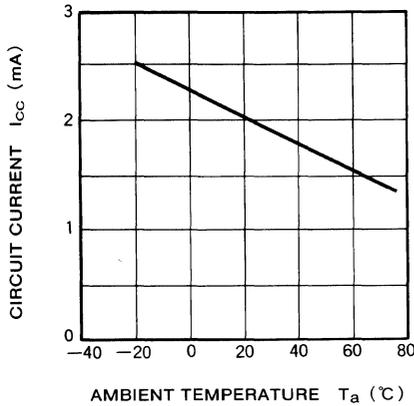
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



INPUT CURRENT VS. AMBIENT TEMPERATURE

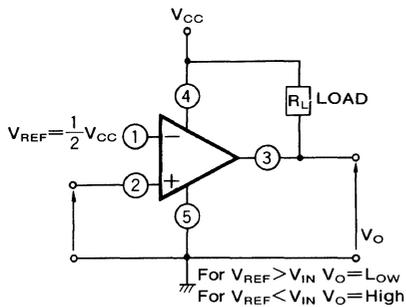


CIRCUIT CURRENT VS. AMBIENT TEMPERATURE

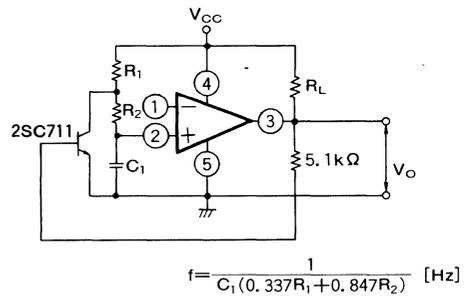


APPLICATION EXAMPLES

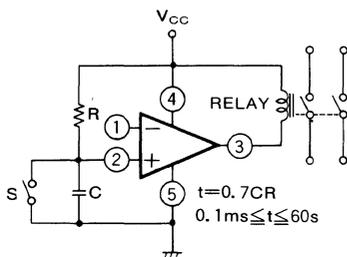
(1) Voltage comparator



(2) Oscillator



(3) CR Timer



PRECAUTIONS FOR USE

1. Paying much attention is necessary for fear that the M51203TL,FP may flow large current and reach to destroy because of the structure when the terminals of V_{CC} and GND of the M51203TL,FP is connected wrong position each other
2. Output is "open collector" and a loading resistor is not included. Connect a loading resistor to stabilize operation, in case of driving a next stage

MITSUBISHI LINEAR ICs
M51205TL,FP

VOLTAGE COMPARATOR

DESCRIPTION

The M51205TL,FP is a semiconductor integrated circuit for a voltage comparator that operates from a single power supply. Especially the M51205TL,FP has superiority as to characteristics of input current (high input resistance) and fits to wide ranged applications, for example CR timer, etc. M51205TL,FP's package is a mini SIP and FLAT package, therefore can use very easily.

FEATURES

- Low input current (high input resistance) 15nA(typ.)
- Built-in zener diode for stabilization of supply voltage between supply voltage terminal and GND terminal 5.6V(typ.)
- Capable of directly driving a relay or a lamp
- Including reference voltage for comparison
- Hysteresis between input and output
- High output breakdown voltage 30V(max.)

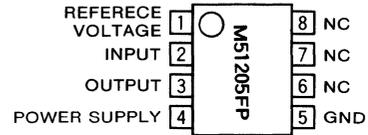
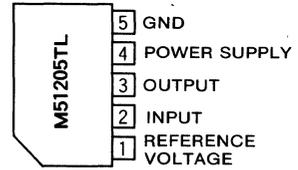
APPLICATION

Electric shutter, CR timer, voltage comparator, time delay circuit, oscillator (square wave)

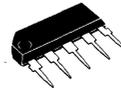
RECOMMENDED OPERATING CONDITIONS

Rated supply voltage 12V($R_d=1k\Omega$)

PIN CONFIGURATION (TOP VIEW)



NC : NO CONNECTION

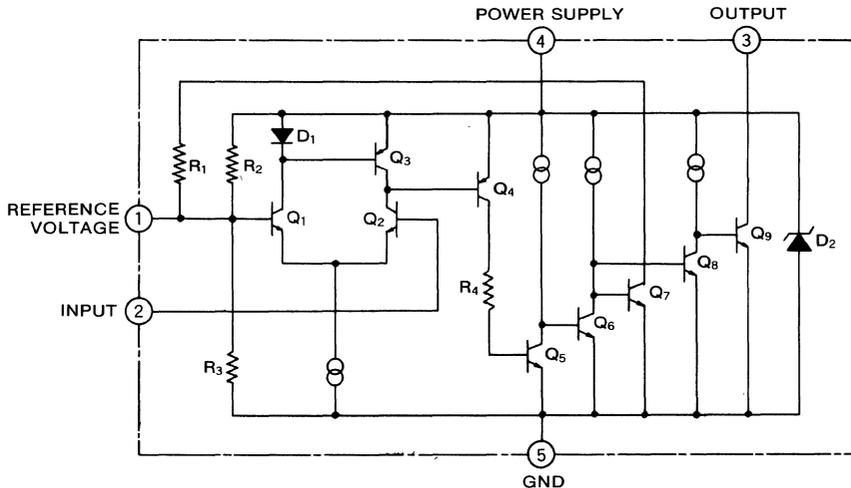


5-pin molded plastic SIP



8-pin molded plastic FLAT

EQUIVALENT CIRCUIT



VOLTAGE COMPARATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
I_{CC}	Circuit current		20	mA
V_{IN}	Input voltage		$V_{(4)}$ ※	V
I_{SINK}	Output sink current		200	mA
V_{OH}	Output drive voltage		30	V
P_d	Power dissipation		180(M51205TL)	mW
			300(M51205FP)	
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	1.8(M51205TL)	mW/°C
			3.0(M51205FP)	
T_{opr}	Operating temperature		-20~+75	°C
T_{stg}	Storage temperature		-40~+125	°C

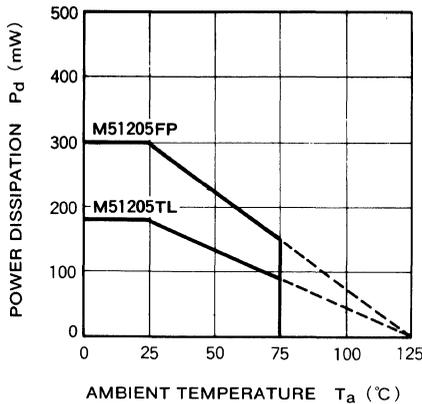
※ Voltage at pin ④

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, R_d : dropper resistor)

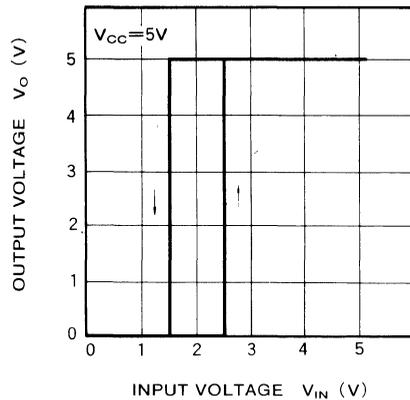
Symbol	Parameter	Test conditions			Limits			Unit
		$V_{CC}(V)$			Min	Typ	Max	
$V_{(4)}$	Zener voltage	12.0	$R_d=1k\Omega$		5.0	5.6	7.0	V
V_{IN}	Input voltage	12.0	$R_d=1k\Omega$		0.8		$V_{(4)}-0.2$	V
I_{IN}	Input current	12.0	$R_d=1k\Omega$			15	75	nA
V_{REF}	Reference voltage	5.00	$R_d=0\Omega, V_{IN}=0V$		2.25	2.50	2.75	V
ΔV_{hys}	Input/output hysteresis width voltage	5.00	$R_d=0\Omega, V_{IN}=3V$		0.80	1	1.20	V
V_{OL}	Output saturation voltage	6.0	$I_{SINK}=60mA$			0.2	0.6	V
			$I_{SINK}=200mA$			1		
t_{PLH}	Output "L-H" propagation delay time	12.0	$R_d=1k\Omega$			1		μs
t_{PHL}	Output "H-L" propagation delay time					10		μs
I_{CC}	Circuit current ($V_{CC} \leq V_{(4)}$)	5.00	$R_d=0\Omega$			1.9	2.4	mA

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

THERMAL DERATING (MAXIMUM RATING)

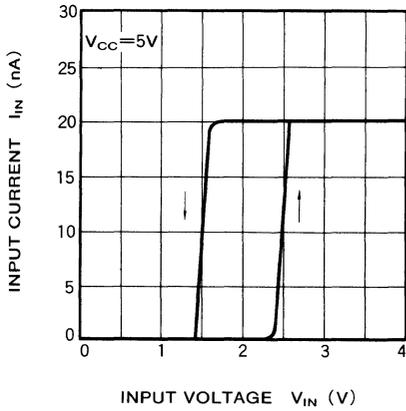


OUTPUT VOLTAGE VS. INPUT VOLTAGE

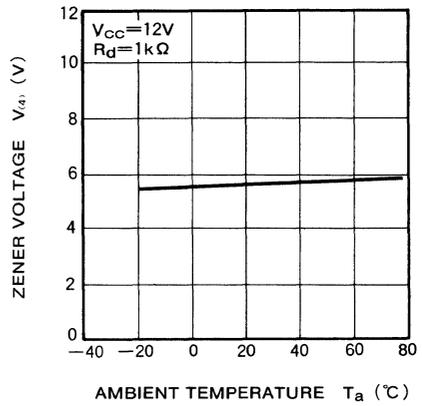


VOLTAGE COMPARATOR

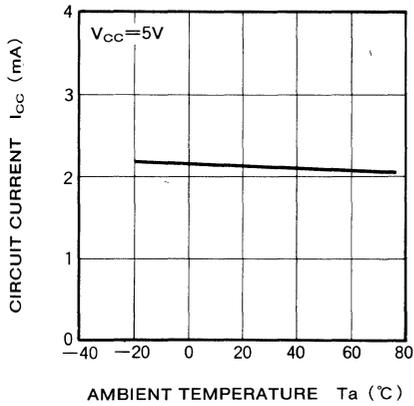
INPUT CURRENT VS. INPUT VOLTAGE



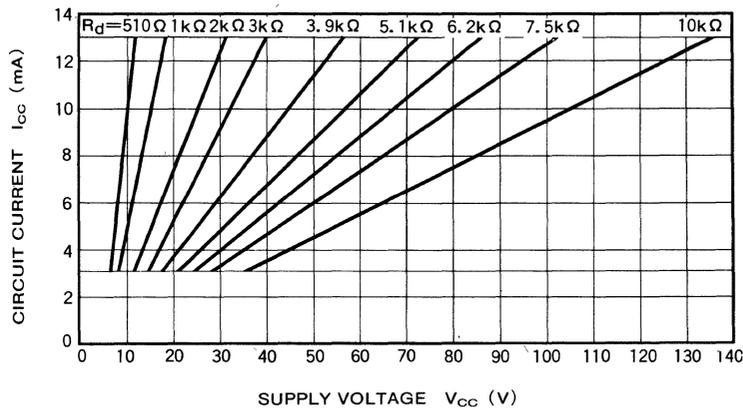
ZENER VOLTAGE VS. AMBIENT TEMPERATURE



CIRCUIT CURRENT VS. AMBIENT TEMPERATURE

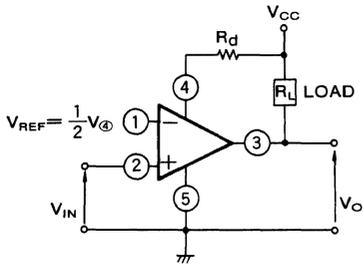


DROPPER RESISTOR (R_D) SELECTION GRAPH

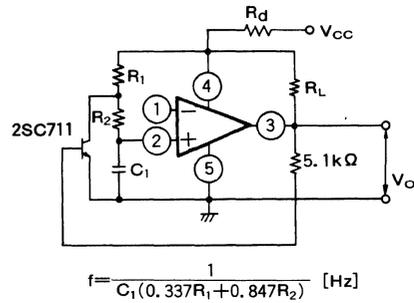


APPLICATION EXAMPLES

(1) Voltage comparator

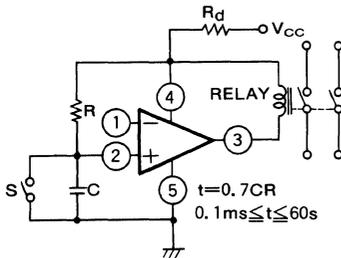


(2) Oscillator

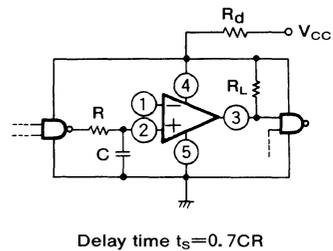


$$f = \frac{1}{C_1(0.337R_1 + 0.847R_2)} \text{ [Hz]}$$

(3) CR Timer



(4) Time delay circuit



PRECAUTIONS FOR USE

1. Paying much attention is necessary for fear that the M51205TL,FP may flow large current and reach to destroy because of the structure when the terminals of V_{CC} and GND of the M51205TL,FP is connected wrong position each other.
2. Output is "open collector" and a loading resistor is not included. Connect a loading resistor to stabilize operation, in case of driving a next stage.
3. Care should be taken not to apply over 5(V) directly to the terminals between Pin ④ and Pin ⑤. Connect a dropper resistor (R_d) in series to Pin ④, in case of applying over 5(V) between Pin ④ and Pin ⑤.

MITSUBISHI LINEAR ICs M51202TL,FP

VOLTAGE COMPARATOR

DESCRIPTION

The M51202TL,FP is a semiconductor integrated circuit for a voltage comparator that operates from a single power supply. Especially the M51202TL,FP has superiority as to characteristics of input current (high input resistance) and fits to wide ranged applications, for example CR timer, etc. M51202TL,FP's package is a mini SIP and FLAT package, therefore can use very easily.

FEATURES

- Low input current8nA(typ.)
- Operates at low supply voltage1.7~6.5V
- Capable of directly driving a relay or a lamp
- Including voltage surge absorbing zener diode for protection
- High output breakdown voltage18V(max.)

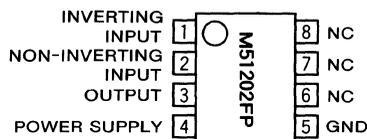
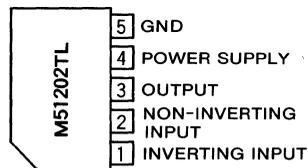
APPLICATION

Voltage comparator, CR timer, electric shutter, time delay circuit, oscillator (square wave)

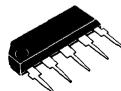
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range1.7~6.5V
- Rated supply voltage5.0V

PIN CONFIGURATION (TOP VIEW)



NC : NO CONNECTION

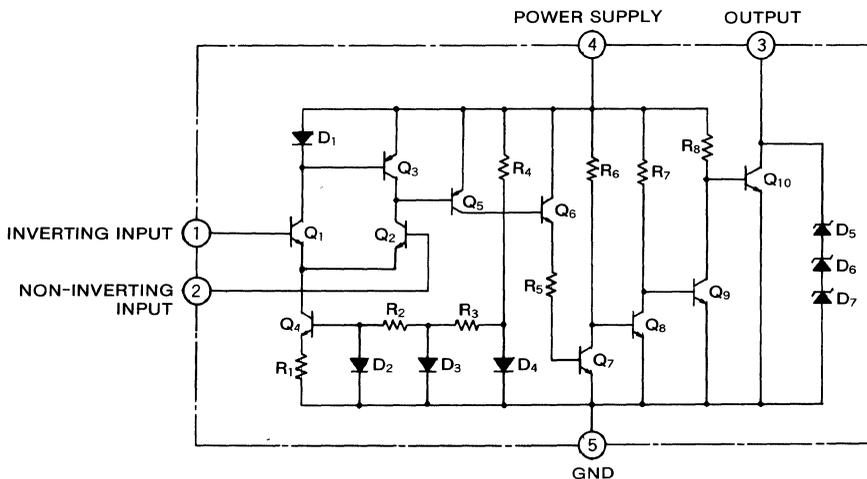


5-pin molded plastic SIP



8-pin molded plastic FLAT

EQUIVALENT CIRCUIT



VOLTAGE COMPARATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

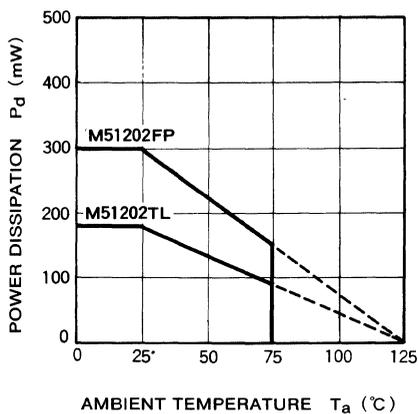
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		6.5	V
I_{SINK}	Output sink current		200	mA
V_{IN}	Input voltage		V_{CC}	V
P_d	Power dissipation		180(M51202TL)	mW
			300(M51202FP)	
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	1.8(M51202TL)	mW/ $^\circ\text{C}$
			3.0(M51202FP)	
T_{opr}	Operating temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$)

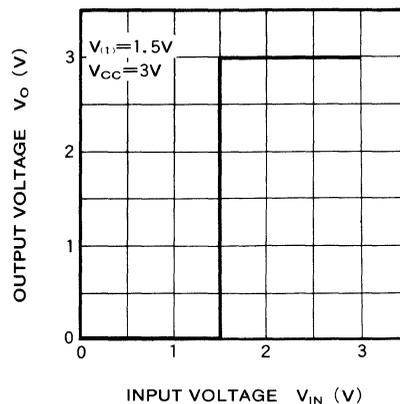
Symbol	Parameter	Test conditions		Limits			Unit
		$V_{CC}(V)$		Min	Typ	Max	
V_{CC}	Supply voltage range			1.7		6.5	V
I_{CC}	Circuit current	2.65			1.7	3.2	mA
		6.0			4.4	8.2	
I_{IN}	Input current	2.65			8	100	nA
V_{IO}	Input offset voltage	2.65			2	50	mV
V_{OL}	Output saturation voltage	6.0	$I_{SINK}=60\text{mA}$		0.2	0.6	V
			$I_{SINK}=200\text{mA}$		1		
V_Z	Zener voltage		$I_Z=5\text{mA}$	18	22	26	V
t_{PLH}	Output "L-H" propagation delay time	6.0	$V_{REF}=V_{CC}/2$		0.2		μs
t_{PHL}	Output "H-L" propagation delay time	6.0	$V_{REF}=V_{CC}/2$		50		μs
V_{IN}	Input voltage range			0.8		$V_{CC}-0.2$	V

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

THERMAL DERATING (MAXIMUM RATING)

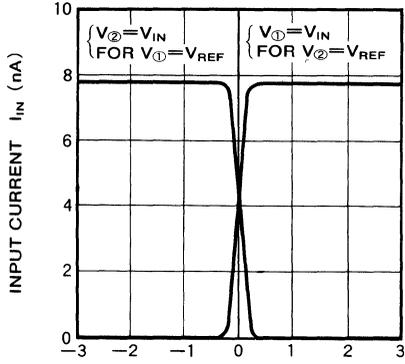


OUTPUT VOLTAGE VS. INPUT VOLTAGE



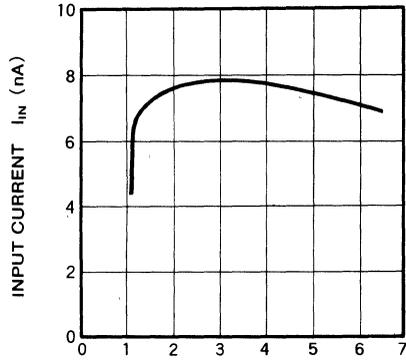
VOLTAGE COMPARATOR

INPUT CURRENT VS. INPUT VOLTAGE



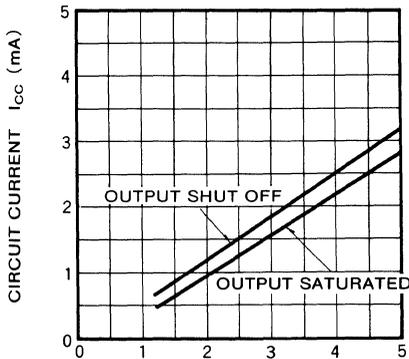
INPUT VOLTAGE AT PIN ① VS.
 INPUT VOLTAGE AT PIN ② $V_1 - V_2$ (V)

INPUT CURRENT VS. SUPPLY VOLTAGE



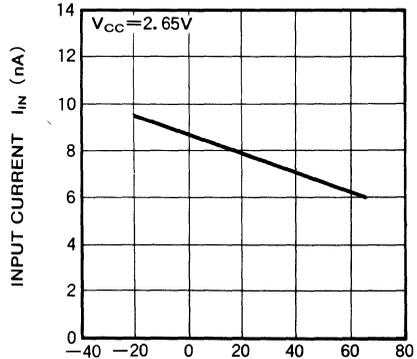
SUPPLY VOLTAGE V_{CC} (V)

CIRCUIT CURRENT VS. SUPPLY VOLTAGE



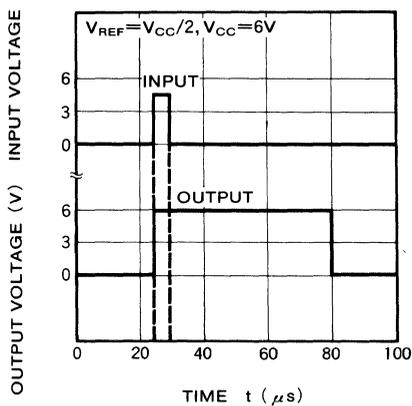
SUPPLY VOLTAGE V_{CC} (V)

INPUT CURRENT VS. AMBIENT TEMPERATURE



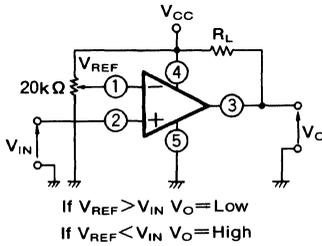
AMBIENT TEMPERATURE T_a (°C)

PULSE RESPONSE CHARACTERISTICS

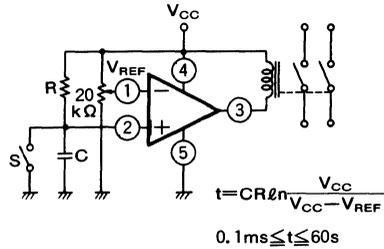


APPLICATION EXAMPLES

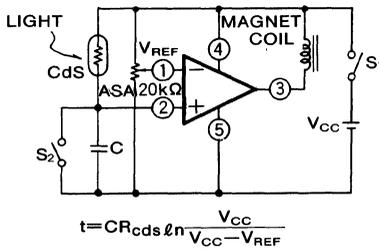
(1) Voltage comparator



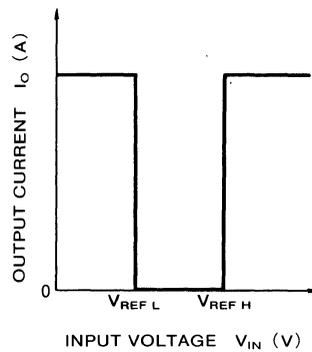
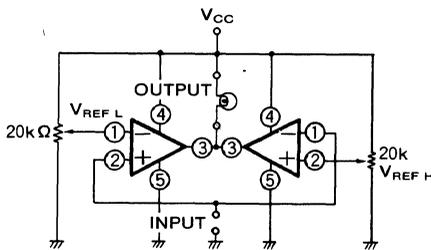
(2) CR Timer



(3) Electric shutter



(4) Window comparator (alarm circuit)



PRECAUTIONS FOR USE

1. Paying much attention is necessary for fear that the M51202TL,FP may flow large current and reach to destroy because of the structure when the terminals of V_{CC} and GND of the M51202TL,FP is connected wrong position each other.
2. Output is "open collector" and a loading resistor is not included. Connect a loading resistor to stabilize operation, in case of driving a next stage.

MITSUBISHI LINEAR ICs M51204TL, FP

VOLTAGE COMPARATOR

DESCRIPTION

The M51204TL,FP is a semiconductor integrated circuit for a voltage comparator that operates from a single power supply. Especially the M51204TL,FP has superior characteristics as to input current (high input resistance) and fits to wide ranged applications, for example CR timer, etc. M51204TL,FP's package is a mini SIP and FLAT package, therefore can use very easily.

FEATURES

- Low input current 20nA(typ.)
- Wide operating voltage range 2.5~28V
- Low power dissipation 2.5mA(max.)
- Capable of directly driving a relay or a lamp
- High output breakdown voltage 30V(max.)

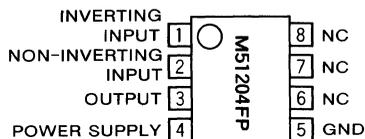
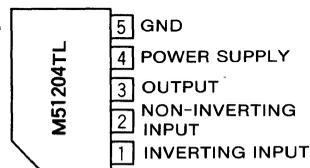
APPLICATION

Voltage comparator, electric shutter, CR timer, time delay circuit, oscillator (square wave)

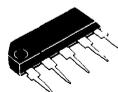
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 2.5~28V
- Rated supply voltage 12V

PIN CONFIGURATION (TOP VIEW)



NC NO CONNECTION

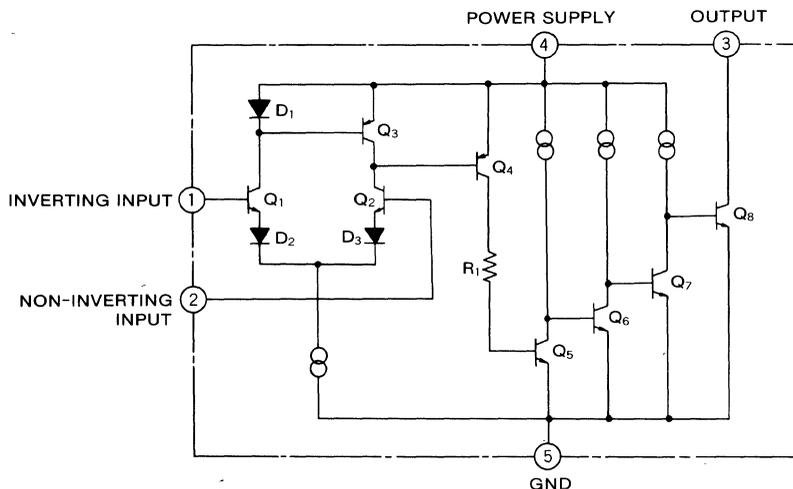


5-pin molded plastic SIP



8-pin molded plastic FLAT

EQUIVALENT CIRCUIT



VOLTAGE COMPARATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		28	V
V_{IN}	Input voltage		V_{CC}	V
I_{SINK}	Output sink current		200	mA
V_{OH}	Output drive voltage		30	V
P_d	Power dissipation		180(M51204TL)	mW
			300(M51204FP)	
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	1.8(M51204TL)	mW/°C
			3.0(M51204FP)	
T_{opr}	Operating temperature		-20~+75	°C
T_{stg}	Storage temperature		-40~+125	°C

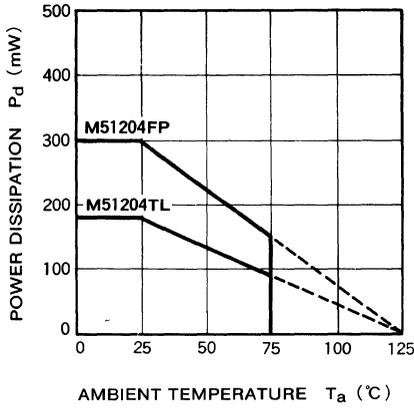
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$)

Symbol	Parameter	Test conditions		Limits			Unit
		V_{CC} (V)		Min	Typ	Max	
V_{CC}	Supply voltage range			2.5		28	V
I_{CC}	Circuit current	6.0			1.8	2.5	mA
		12.0					
		24.0					
$V_{IN①}$	Inverting input voltage	12.0		1.4		$V_{CC}-0.2$	V
$V_{IN②}$	Non-inverting input voltage	12.0			V		
$I_{IN①}$	Inverting input current	6.0			20	75	nA
		12.0					
		24.0					
$I_{IN②}$	Non-inverting input current	6.0			20	75	nA
		12.0					
		24.0					
V_{IO}	Input offset voltage	6.0	Reference voltage at Pin ①	-7	2	12	mV
		12.0					
		24.0					
V_{OL}	Output saturation voltage	6.0	$I_{SINK}=60\text{mA}$		0.2	0.6	V
			$I_{SINK}=200\text{mA}$		1		
t_{PLH}	Output "L-H" propagation delay time	12.0			1		μs
t_{PHL}	Output "H-L" propagation delay time				10		μs

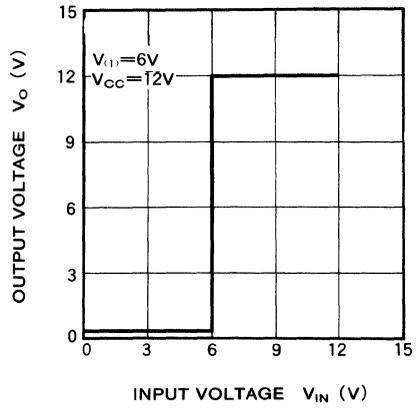
VOLTAGE COMPARATOR

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

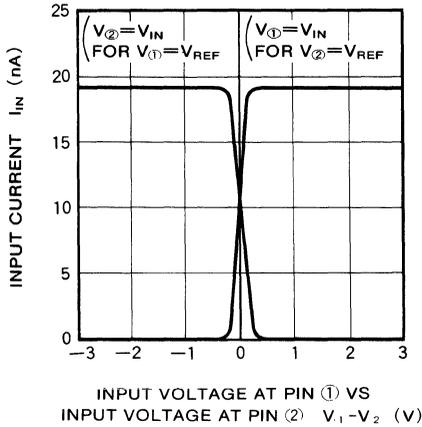
THERMAL DERATING (MAXIMUM RATING)



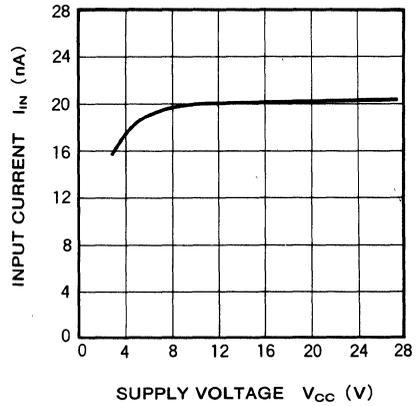
OUTPUT VOLTAGE VS. INPUT VOLTAGE



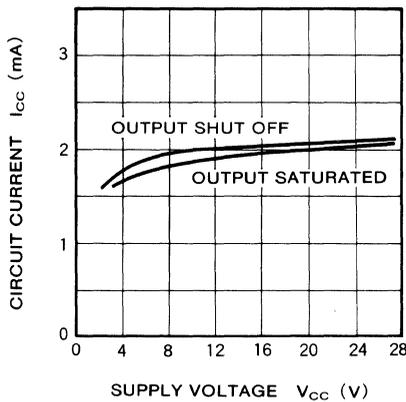
INPUT CURRENT VS. INPUT VOLTAGE



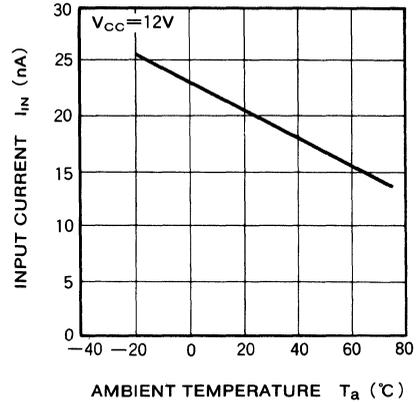
INPUT CURRENT VS. SUPPLY VOLTAGE



CIRCUIT CURRENT VS. SUPPLY VOLTAGE

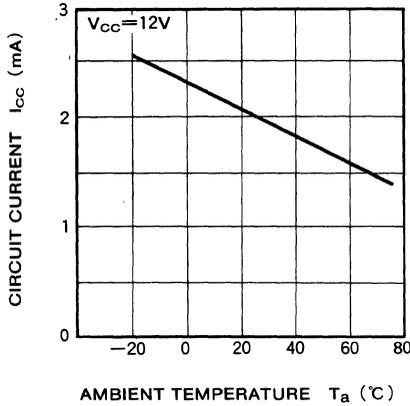


INPUT CURRENT VS. AMBIENT TEMPERATURE

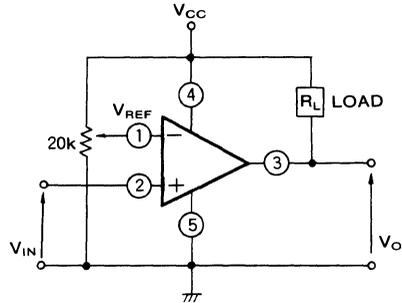


APPLICATION EXAMPLES

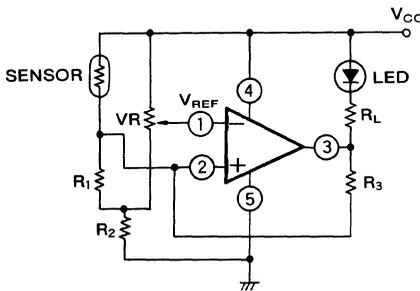
CIRCUIT CURRENT VS. AMBIENT TEMPERATURE



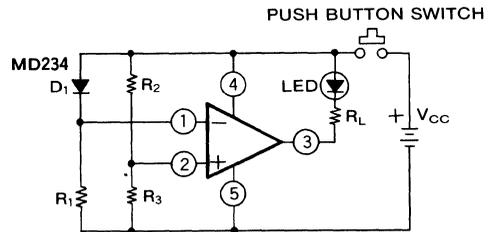
(1) Voltage comparator



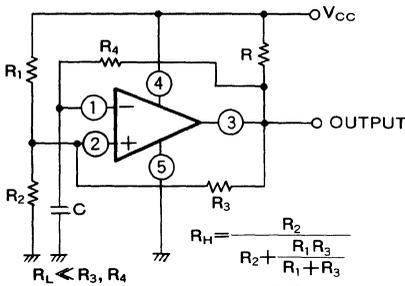
(2) Sensor detection circuit



(3) Battery check circuit



(4) Oscillator



$$f = \frac{1}{CR_4 \ln \frac{R_L}{R_H} \cdot \frac{1-R_L}{1-R_H}} \quad [\text{Hz}]$$

$$R_H = \frac{R_2}{R_2 + \frac{R_1 R_3}{R_1 + R_3}}$$

$$R_L = \frac{\frac{R_2 R_3}{R_2 + R_3}}{R_1 + \frac{R_2 R_3}{R_2 + R_3}}$$

$R_L < R_3, R_4$

PRECAUTIONS FOR USE

1. Paying much attention is necessary for fear that the M51204TL,FP may flow large current and reach to destroy because of the structure when the terminals of V_{CC} and GND of the M51204TL,FP is connected wrong position each other.
2. Output is "open collector" and a loading resistor is not included. Connect a loading resistor to stabilize operation, in case of driving a next stage.

M51206TL,FP

VOLTAGE COMPARATOR

DESCRIPTION

The M51206TL,FP is a semiconductor integrated circuit for a voltage comparator that operates at a single power supply. Especially the M51206TL,FP has superiority as to characteristics of input current (high input resistance) and fits to wide ranged applications, for example CR timer, etc. M51206TL,FP's package is a mini SIP and FLAT package, therefore can use very easily.

FEATURES

- Low input current 15nA(typ.)
- Built-in zener diode for stabilization of supply voltage between supply voltage terminal and GND terminal 5.6V(typ.)
- Capable of directly driving a relay or lamp
- High output breakdown voltage 30V(max.)

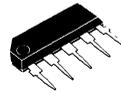
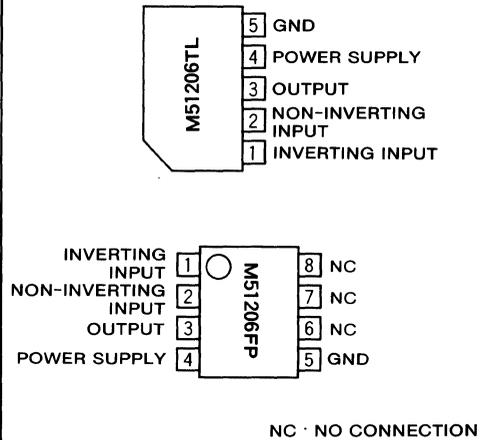
APPLICATION

Voltage comparator, CR timer, electric shutter, time delay circuit, oscillator (square wave).

RECOMMENDED OPERATING CONDITIONS

Rated supply voltage 12V($R_d=1k\Omega$)

PIN CONFIGURATION (TOP VIEW)

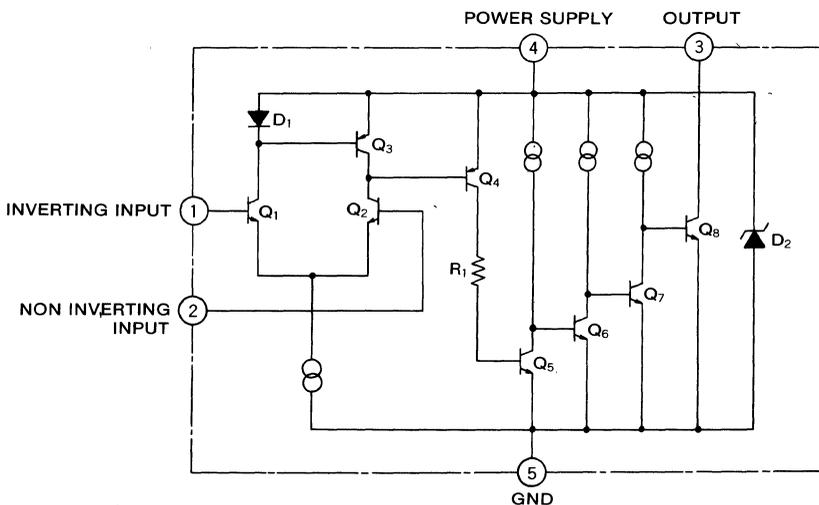


5-pin molded plastic SIP



8-pin molded plastic FLAT

EQUIVALENT CIRCUIT



VOLTAGE COMPARATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Rated	Unit
I_{CC}	Circuit current		20	mA
V_{IN}	Input voltage		$V_{\text{④}}^*$	V
I_{sink}	Output sink current		200	mA
V_{OH}	Output drive voltage		30	V
P_d	Power dissipation		180 (M51206TL)	mW
			300 (M51206FP)	
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	1.8 (M51206TL)	mW/°C
			3.0 (M51206FP)	
T_{opr}	Operating temperature		-20~+75	°C
T_{stg}	Storage temperature		-40~+125	°C

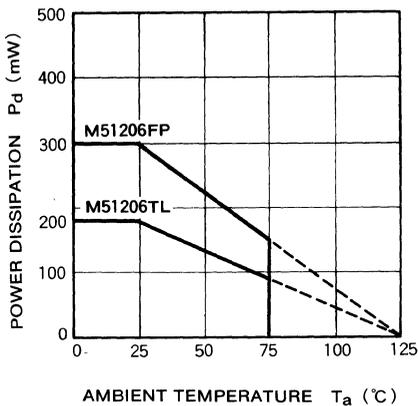
*Voltage at Pin ④

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, R_d dropper resistor)

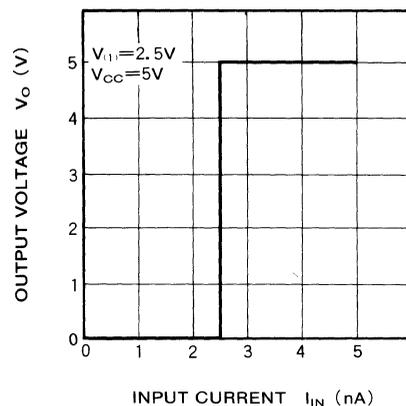
Symbol	Parameter	Test conditions		Limits			Unit
		V_{CC} (V)		Min	Typ	Max	
$V_{\text{④}}$	Zener voltage range	12.0	$R_d=1\text{k}\Omega$	5.0	5.6	7.0	V
V_{IN}	Input voltage range	12.0	$R_d=1\text{k}\Omega$	0.8		$V_{\text{④}}-0.2$	V
$I_{IN\text{①}}$	Input current at Pin ①	12.0	$R_d=1\text{k}\Omega$		15	75	nA
$I_{IN\text{②}}$	Input current at Pin ②	12.0	$R_d=1\text{k}\Omega$		15	75	nA
I_{CC}	Circuit current ($V_{CC} < V_{\text{④}}$)	5.0	$R_d=0\Omega$		1.8	2.4	mA
V_{IO}	Input offset voltage	12.0	$R_d=1\text{k}\Omega$	-20	-1.4	20	mV
V_{OL}	Output saturation voltage	6.0	$I_{\text{sink}}=60\text{mA}$		0.2	0.6	V
			$I_{\text{sink}}=200\text{mA}$		1		
t_{PLH}	Output "L-H" propagation delay time	12.0	$R_d=1\text{k}\Omega$		1		μs
t_{PHL}	Output "H-L" propagation delay time				10		μs

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

THERMAL DERATING (MAXIMUM RATING)

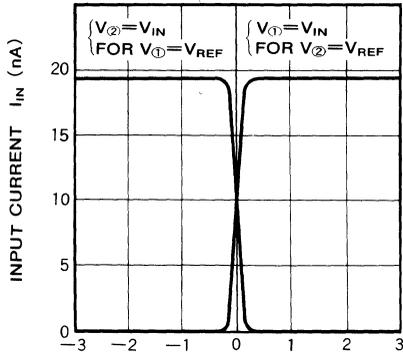


OUTPUT VOLTAGE VS. INPUT VOLTAGE



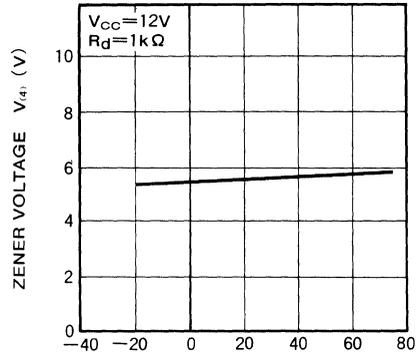
VOLTAGE COMPARATOR

INPUT CURRENT VS. INPUT VOLTAGE



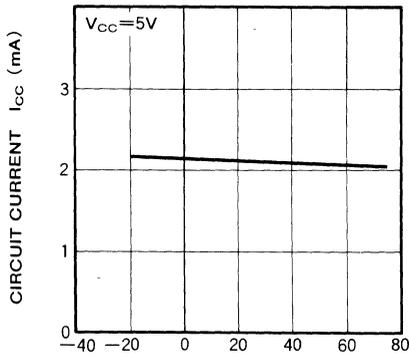
INPUT VOLTAGE AT PIN ① VS.
 INPUT VOLTAGE AT PIN ② V_{IN} (V)

ZENER VOLTAGE VS. AMBIENT TEMPERATURE



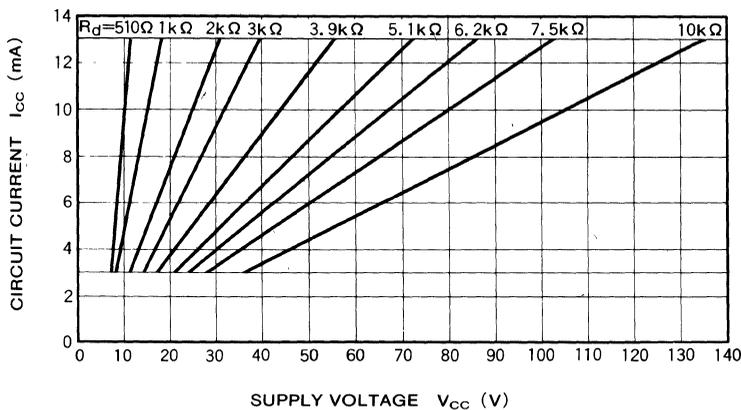
AMBIENT TEMPERATURE T_a (°C)

CIRCUIT CURRENT VS. AMBIENT TEMPERATURE



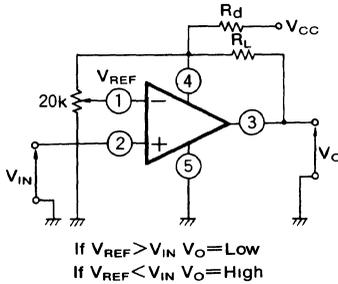
AMBIENT TEMPERATURE T_a (°C)

DROPPER RESISTOR (R_d) SELECTION GRAPH

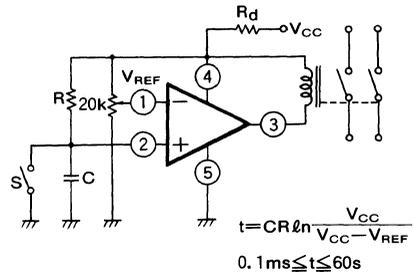


APPLICATION EXAMPLES

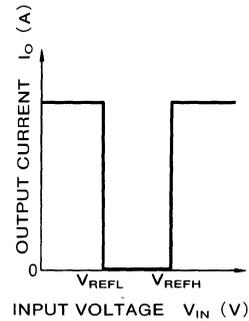
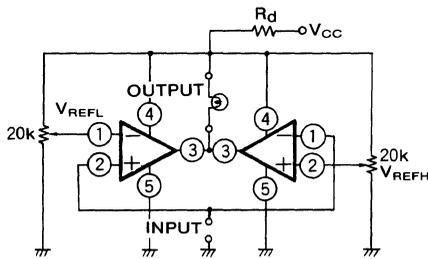
(1) Voltage comparator



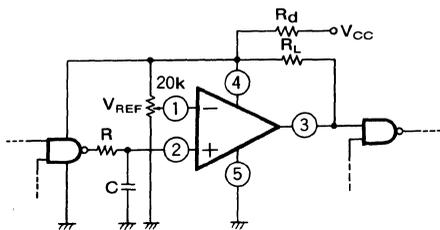
(2) CR Timer



(3) Window comparator (alarm circuit)



(4) Time delay circuit



$$t_s = CR \ln \frac{V_{CC}}{V_{CC} - V_{REF}}$$

Unit
 Resistance : Ω
 Capacitance : F

PRECAUTIONS FOR USE

1. Paying much attention is necessary for fear that the M51206TL,FP may flow large current and reach to destroy because of the structure when the terminals of V_{CC} and GND of the M51206TL,FP is connected wrong position each other.
2. Output is "open collector" and a loading resistor is not included. Connect a loading resistor to stabilize operation, in case of driving a next stage.
3. Care should be taken not to apply over 5(V) directly to the terminals between Pin ④ and Pin ⑤. Connect a dropper resistor (R_d) in series to Pin ④ in case of applying over 5(V) between Pin ④ and Pin ⑤.

M5123TL

VOLTAGE COMPARATOR

DESCRIPTION

The M5123TL is a semiconductor integrated circuit for a voltage comparator designed to operate from a single power supply. The comparator reference voltage is close to the supply voltage and, therefore, the M5123TL fits to the shaping of waveform of signal multiplied to power supply because the input voltage can be more than the supply voltage.

M5123TL's package is a mini SIP, making mounting of the component easy.

FEATURES

- Operates at low supply voltage2.5~5.5V
- Including comparator reference voltage
..... $V_{CC}-46\text{mV}(\text{typ.})$
- Low power dissipation in standby state $150\mu\text{A}(\text{typ.})$
- Built-in protection resistor and diode for input surge current

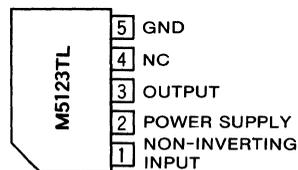
APPLICATION

Voltage comparator, general-purpose detection, alarm circuits.

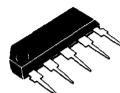
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range2.5~5.5V
- Rated supply voltage5.0V

PIN CONFIGURATION (TOP VIEW)

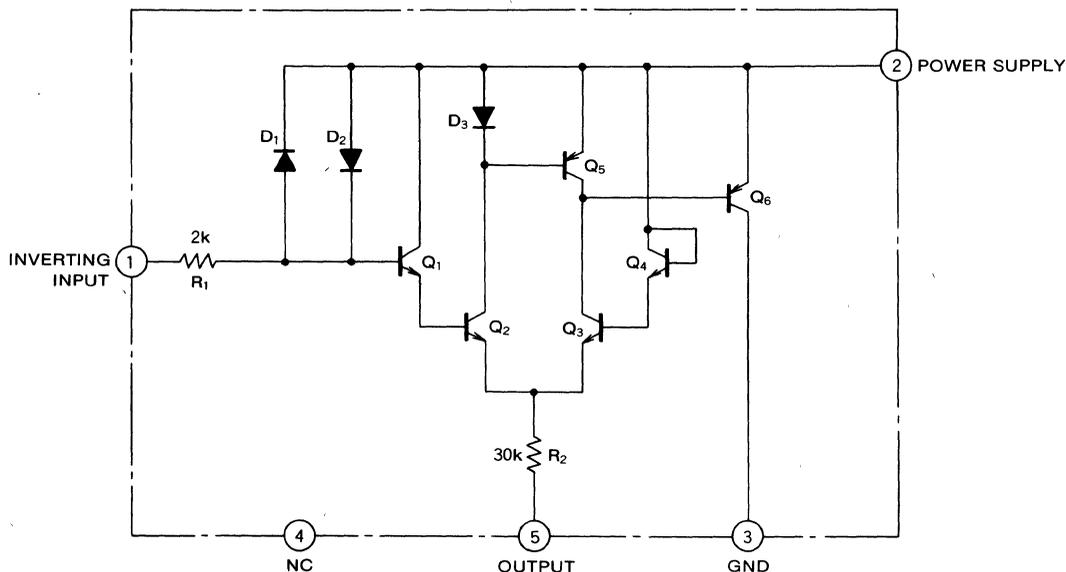


NC . NO CONNECTION



5-pin molded plastic SIP

EQUIVALENT CIRCUIT



VOLTAGE COMPARATOR

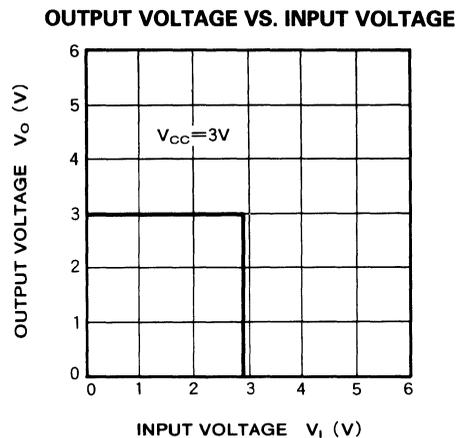
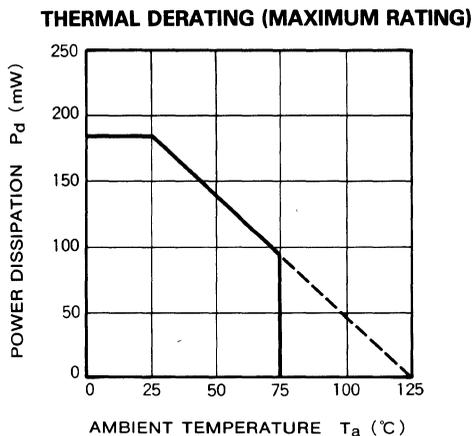
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		6	V
V_i	Input voltage		-0.2~10	V
V_o	Output voltage		V_{CC}	V
P_d	Power dissipation		180	mW
T_{opr}	Operating temperature		-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature		-40~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

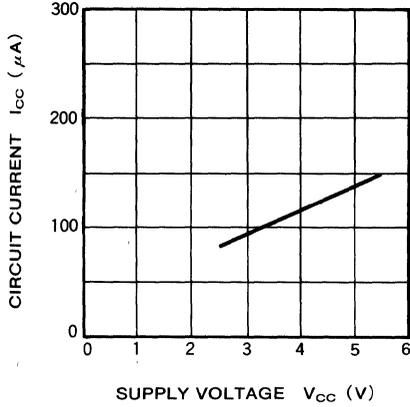
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage range		2.5		5.5	V
I_{CC1}	Circuit current	$V_{CC}=3.3\text{V}$		100	170	μA
I_{CC2}	Circuit current	$V_{CC}=5.5\text{V}$		150	270	μA
V_{IO}	Input offset voltage	$V_{CC}=2.5\sim 5.5\text{V}$, V_{CC} reference voltage	-100	-46	0	mV
V_{OS}	Output saturation voltage	$V_{CC}=5.5\text{V}$, $R_L=27\text{k}\Omega$		50	300	mV
I_{OL}	Output leak current	$V_{CC}=5.5\text{V}$	-1			μA

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=5.5\text{V}$, unless otherwise noted)

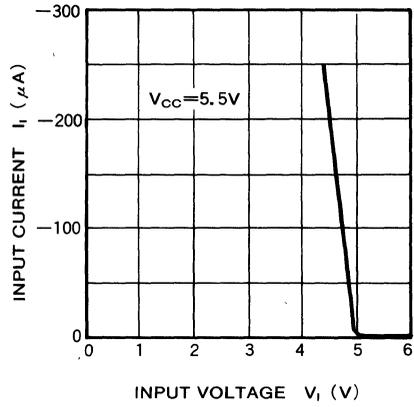


VOLTAGE COMPARATOR

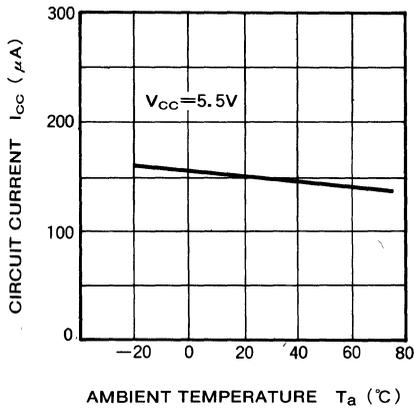
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



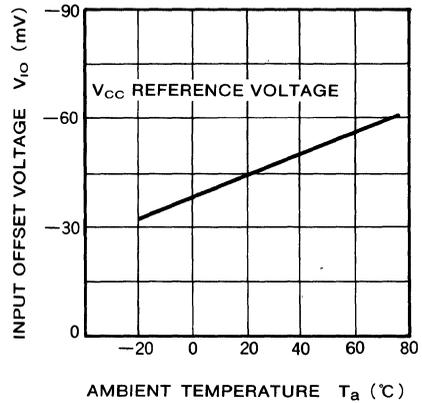
INPUT CURRENT VS. INPUT VOLTAGE



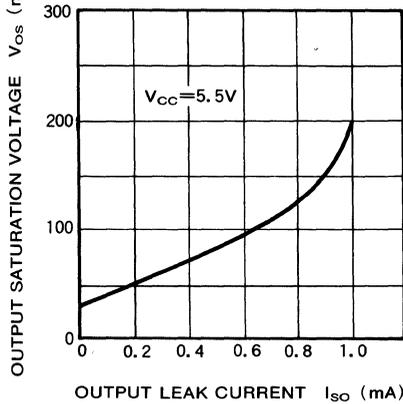
CIRCUIT CURRENT VS. AMBIENT TEMPERATURE



INPUT OFFSET VOLTAGE VS. AMBIENT TEMPERATURE



OUTPUT SATURATION VOLTAGE VS. OUTPUT LEAK CURRENT



M51207L

DUAL COMPARATOR

DESCRIPTION

The M51207L is a dual (two independent) comparator and operates over a wide voltage range from a single supply voltage. Especially the M51207L has superiority as to characteristics of input current (input resistance) and fits to wide ranged applications, for example CR Timer, oscillator, etc.

FEATURES

- Low input current (high input resistance)20nA(typ.)
- Wide supply voltage range 2.5V~28V
- Low dissipation current3.8mA(typ.)
- Capable of driving a relay or a lamp directly
60mA(max.)
- Includes voltage surge absorbing zener diodes
- High output breakdown voltage30V(max.)
- Low output voltage 0.2V(typ.)
- Low input offset voltage 2mV(typ.)

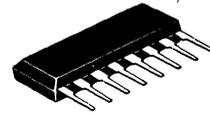
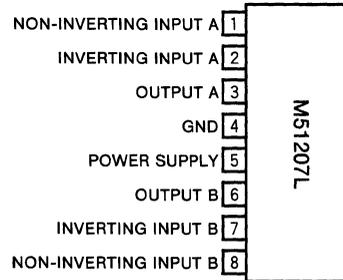
APPLICATION

Voltage comparator, window comparator, CR Timer, time delay circuit, oscillator, etc

RECOMMENDED OPERATING CONDITIONS

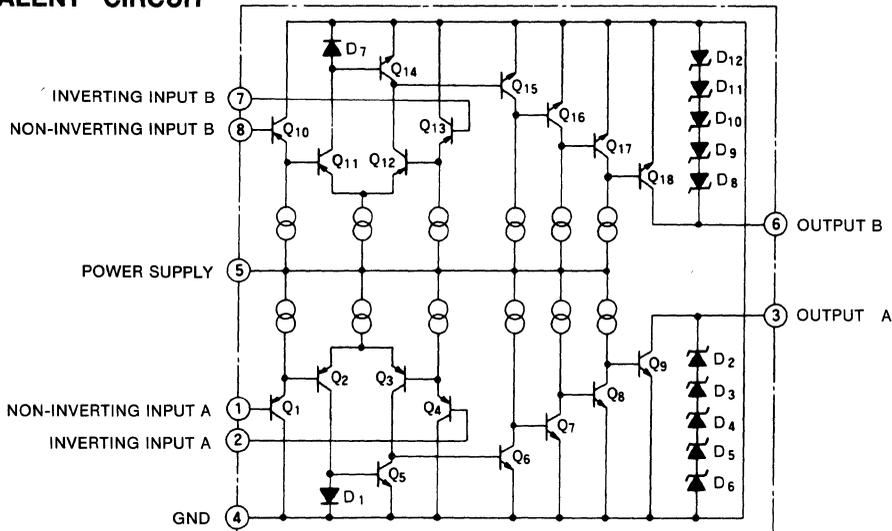
- Supply voltage range 2.5~28V
- Rated supply voltage 12V

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic SIP

EQUIVALENT CIRCUIT



DUAL COMPARATOR

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

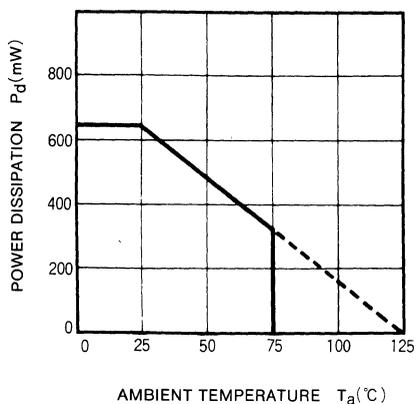
Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		28	V
V _{ID}	Differential input voltage		V _{CC}	V
V _{ICM}	Common mode input voltage range		-0.3~V _{CC}	V
I _{sink}	Output sink current		80	mA
V _{OH}	"H" output voltage		30	V
P _d	Power dissipation		650	mW
T _{opr}	Operating temperature		-20~+75	°C
T _{stg}	Storage temperature		-40~+125	°C

ELECTRICAL CHARACTERISTICS (T_a=25°C, V_{CC}=2.5~28V, unless otherwise noted)

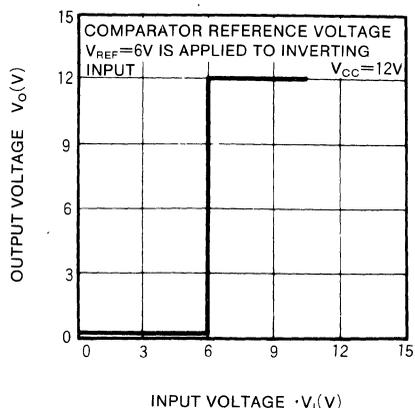
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage range		2.5		28	V
I _{CC}	Circuit current			3.8	5.3	mA
V _{I⊖}	Inverting input voltage range		0		V _{CC} -1.5	V
V _{I⊕}	Non-inverting input voltage range		0		V _{CC} -1.5	V
V _{IO}	Input offset voltage			2	7	mV
I _{I⊖}	Inverting input current			20	100	nA
I _{I⊕}	Non-inverting input current			20	100	nA
I _{IO}	Input offset current			5	50	nA
V _{OL}	"L" output voltage	I _{sink} =60mA I _{sink} =200mA		0.2 1	0.4	V
I _{LO}	Output leak current				0.1	μA
t _{PLH}	Output "L→H" propagation delay time			2		μs
t _{PHL}	Output "H→L" propagation delay time			1		μs

TYPICAL CHARACTERISTICS (T_a=25°C, unless otherwise noted)

THERMAL DERATING
(MAXIMUM RATING)

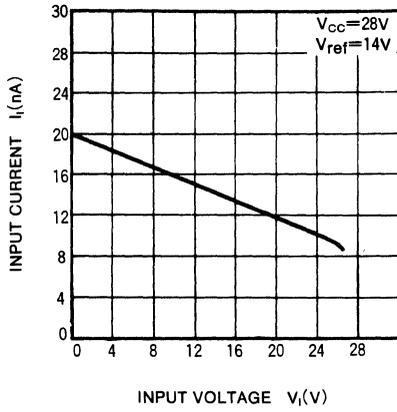


OUTPUT VOLTAGE VS
INPUT VOLTAGE

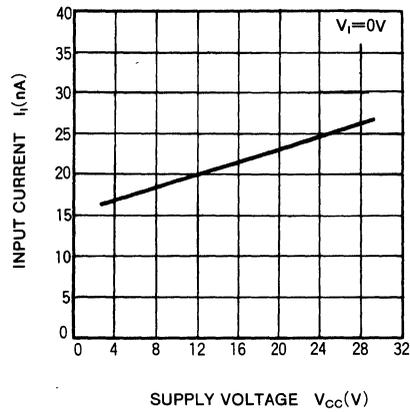


DUAL COMPARATOR

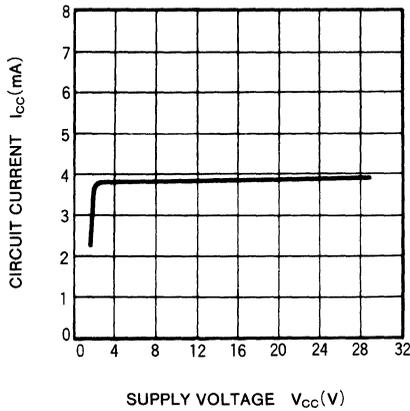
INPUT CURRENT VS INPUT VOLTAGE



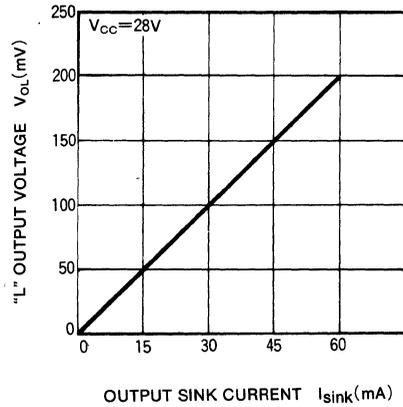
INPUT CURRENT VS SUPPLY VOLTAGE



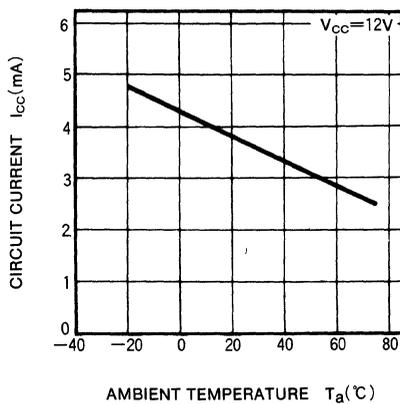
CIRCUIT CURRENT VS SUPPLY VOLTAGE



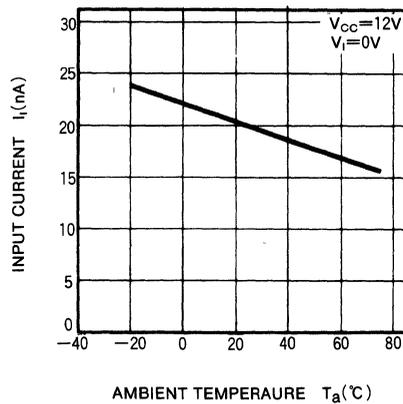
"L" OUTPUT VOLTAGE VS OUTPUT SINK CURRENT



CIRCUIT CURRENT VS AMBIENT TEMPERATURE



INPUT CURRENT VS AMBIENT TEMPERATURE



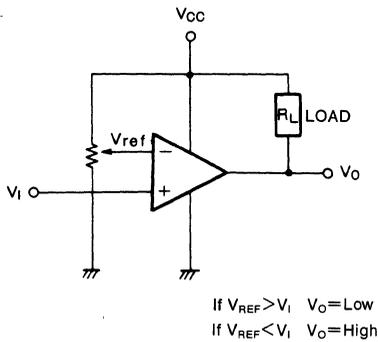
DUAL COMPARATOR

PRECAUTIONS FOR USE

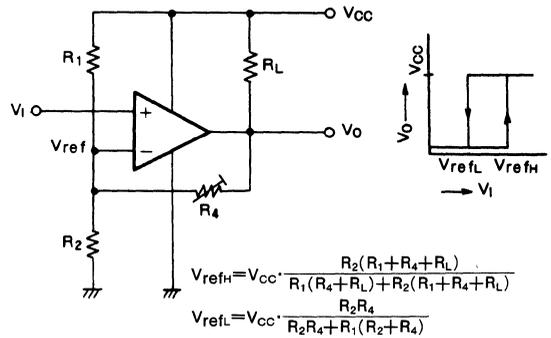
1. Special care must be taken to protect the M51209P from large surges in current, such as may result from the incorrect connection of the V_{CC} and GND terminals.
2. Output is "open collector" and a loading resistor is not included. Connect a loading resistor to stabilize operation, when driving another.

APPLICATION EXAMPLES

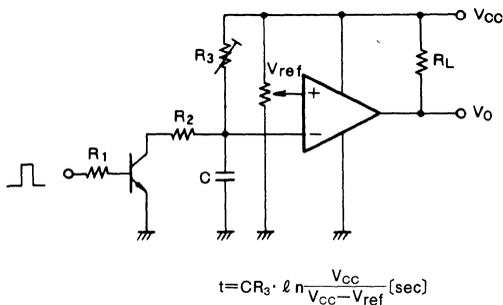
(1) Voltage comparator



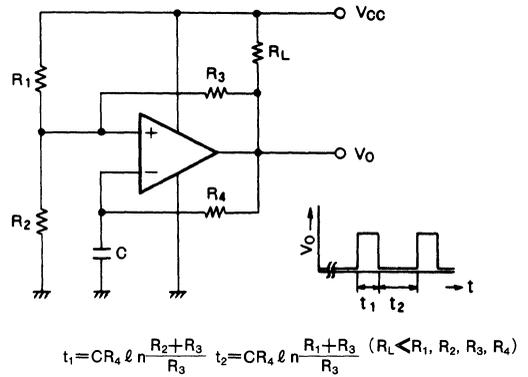
(2) Schmitt trigger circuit



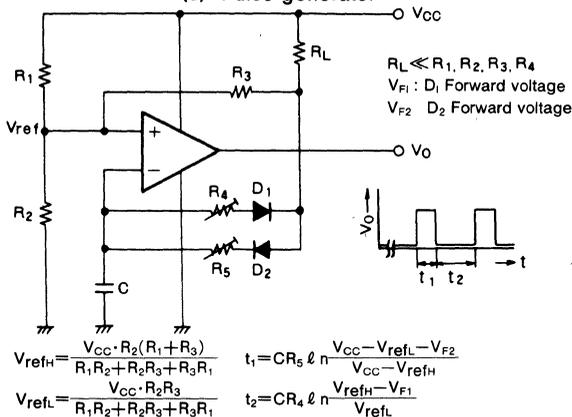
(3) Monostable multi-vibrator



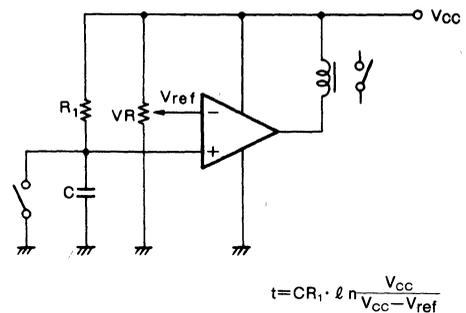
(4) Unstable multi-vibrator



(5) Pulse generator

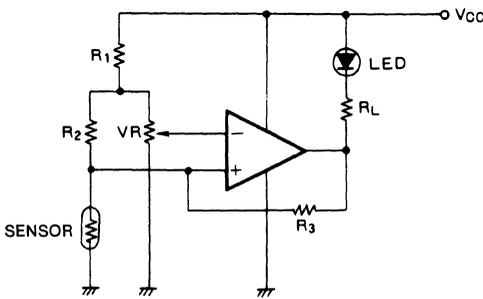


(6) CR Timer

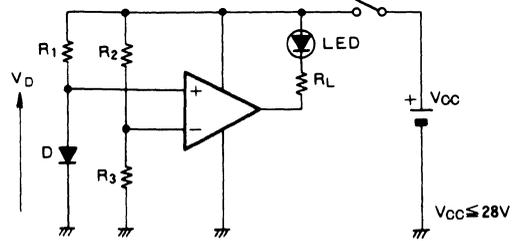


DUAL COMPARATOR

(7) Sensor detector



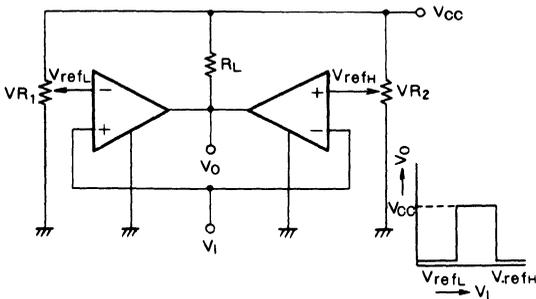
(8) Battery check circuit



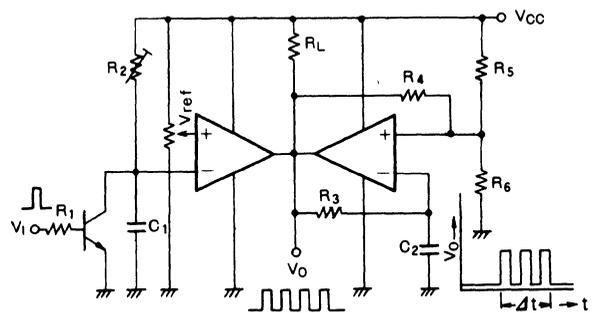
$$V_D < V_{CC} \cdot \frac{R_3}{R_2 + R_3}, \text{ LED} \rightarrow \text{ON}$$

$$V_D > V_{CC} \cdot \frac{R_3}{R_2 + R_3}, \text{ LED} \rightarrow \text{OFF}$$

(9) Window comparator

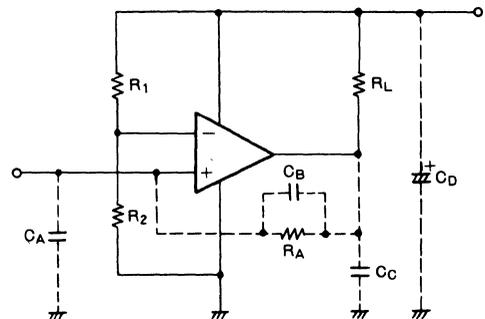


(10) Pulse train generator



$$\Delta t = C_1 R_2 \cdot \ln \frac{V_{CC}}{V_{CC} - V_{ref}}$$

(11) Countermeasure against oscillation



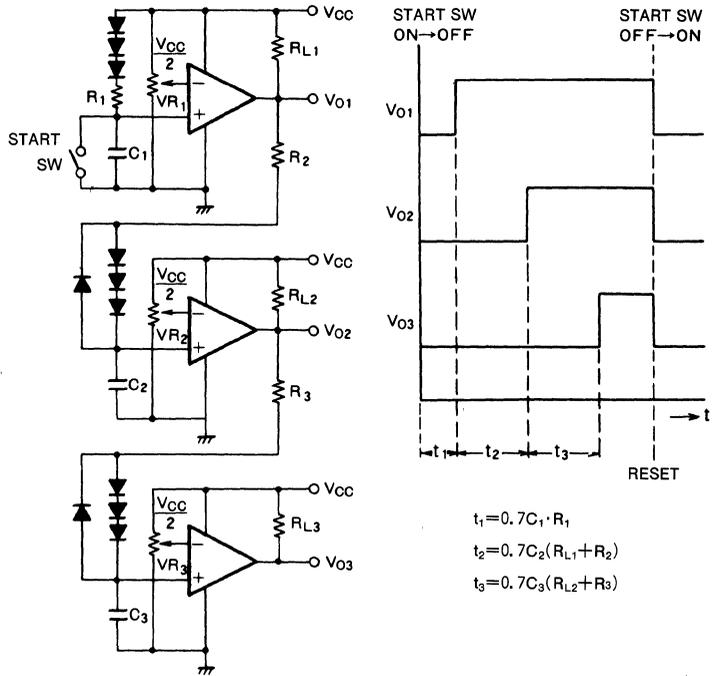
(Note) Taking steps against oscillation

The M51207L may oscillate according to input condition. If the M51207L should oscillate, the following countermeasures are applicable.

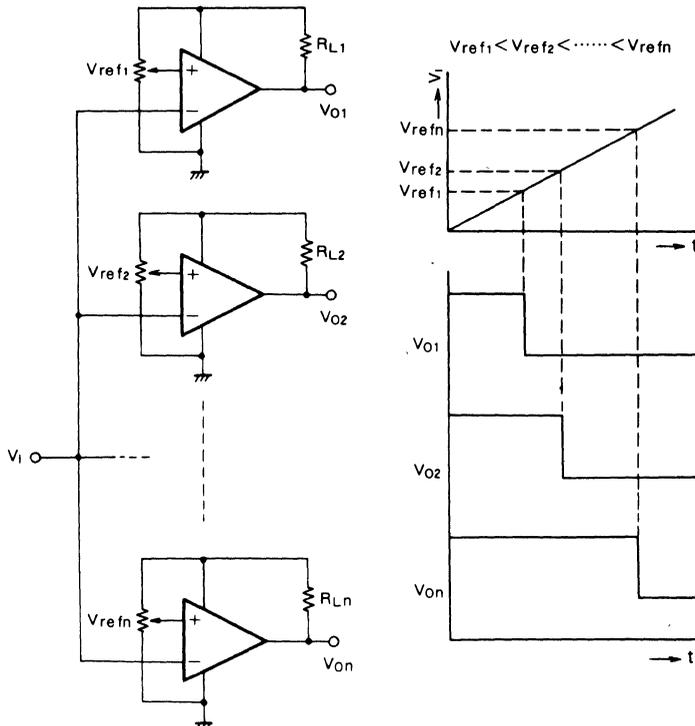
- In case of connecting input signal with chattering, connect a capacitor of small C_A value.
- In case of oscillation with ordinary input, employ positive feedback inserting R_A (large resistor), C_B (no polar) or connect C_C .
- When the supply voltage is not stabilized, connect C_D (a large electrolytic capacitor) to absorb the supply voltage change.

DUAL COMPARATOR

(12) Sequential timer



(13) Analog/Digital converter



M51922L

DUAL COMPARATOR

DESCRIPTION

The M51922L is a dual (two independent) comparator and operates over a wide voltage range from a single supply voltage. The M51922L has a low power dissipation characteristics but enables high output drive, and fits to wide ranged applications, for example CR timer, oscillator, etc.

FEATURES

- Low input current 25nA(typ.)
- Wide supply voltage range 2.5V~28V
- Low dissipation current 0.4mA(typ.) (All output OFF)
1.5mA(typ.) (All output ON)
- Enables high output drive $V_{OL}=0.15V$ (typ.)
(Output current 20mA)

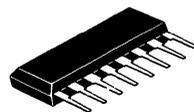
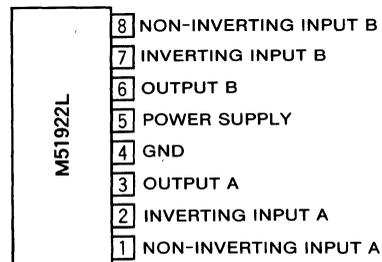
APPLICATION

Voltage comparator, window comparator, CR Timer, time delay circuit, oscillator, etc.

RECOMMENDED OPERATING CONDITIONS

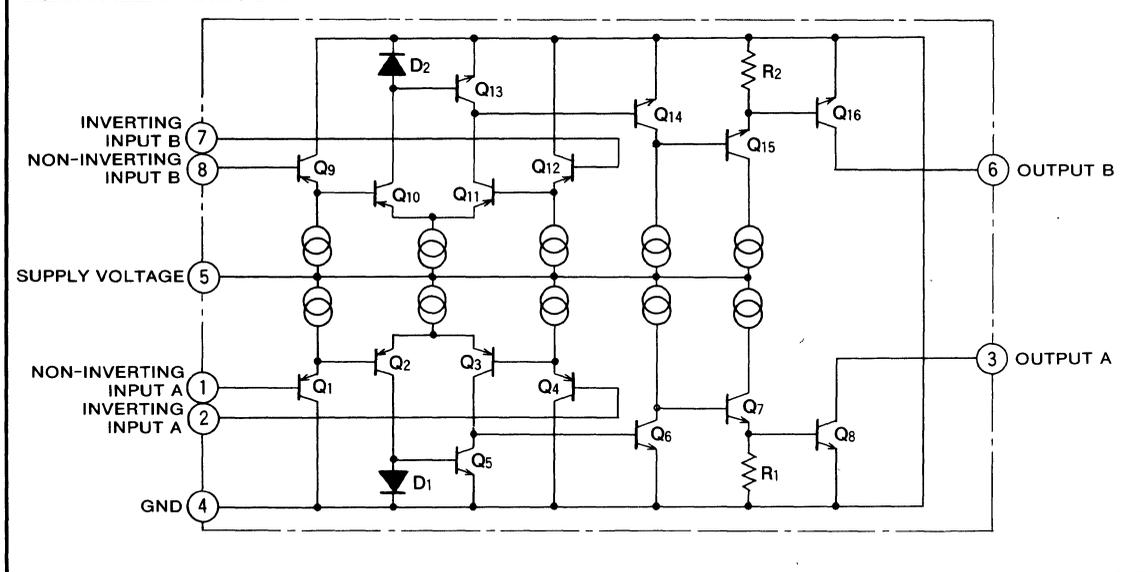
- Supply voltage range 2.5~28V
- Rated supply voltage 12V

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic SIP

EQUIVALENT CIRCUIT



DUAL COMPARATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		28	V
V_{ID}	Differential input voltage		V_{CC}	V
V_{ICM}	Common mode input voltage range		$-0.3 \sim V_{CC}$	V
I_{SINK}	Output sink current		80	mA
V_{OH}	"H" output voltage		30	V
P_d	Power dissipation		650	mW
T_{opr}	Operating temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^\circ\text{C}$

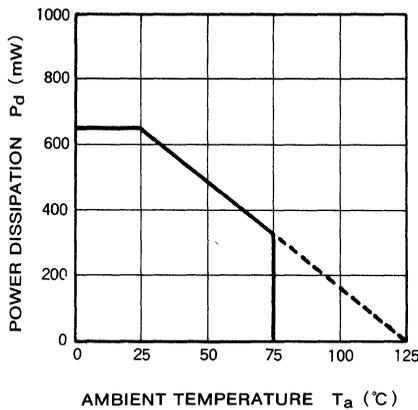
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=2.5 \sim 28\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage range		2.5		28	V
I_{CC1}	Circuit current 1	ALL OUTPUT ON		1.5	2.5	mA
I_{CC2}	Circuit current 2	ALL OUTPUT OFF		0.4	0.8	mA
$V_{I\ominus}$	Inverting input voltage range	NOTE	0		$V_{CC}-1.5$	V
$V_{I\oplus}$	Non-inverting input voltage range	NOTE	0		$V_{CC}-1.5$	V
V_{IO}	Input offset voltage			2	5	mV
$I_{I\ominus}$	Inverting input current			25	150	nA
$I_{I\oplus}$	Non-inverting input current			25	150	nA
I_{IO}	Input offset current			5	50	nA
V_{sat}	Output saturation voltage	$I_{OL}=20\text{mA}$		0.15	0.4	V
V_{OL}	"L" output voltage	$I_{SINK}=20\text{mA}$ $I_{SINK}=80\text{mA}$		0.15	0.4	V
I_{LO}	Output leak current				0.1	μA
t_{PLH}	Output "L→H" propagation delay time			2		μs
t_{PHL}	Output "H→L" propagation delay time			0.2		μs

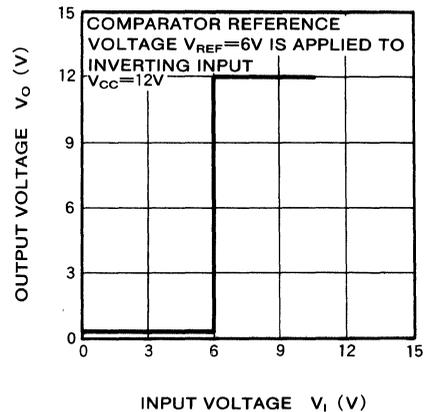
NOTE) Either inverting or non-inverting inputs (reference side) should be within this range. (Abnormal operation will not occur when the other is within the range of 0 to V_{CC} .)

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

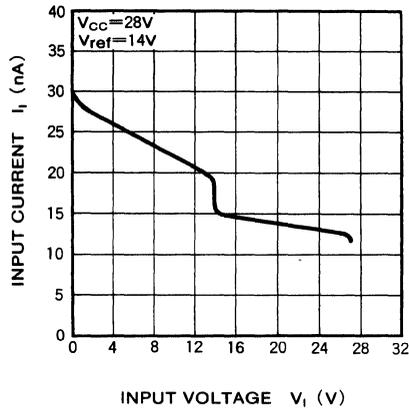
THERMAL DERATING (MAXIMUM RATING)



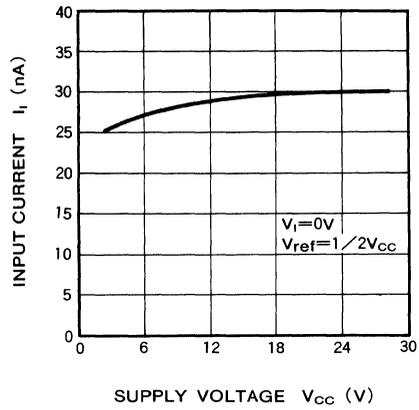
OUTPUT VOLTAGE VS. INPUT VOLTAGE



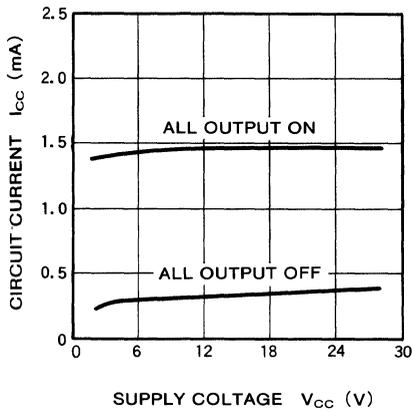
INPUT CURRENT VS. INPUT VOLTAGE



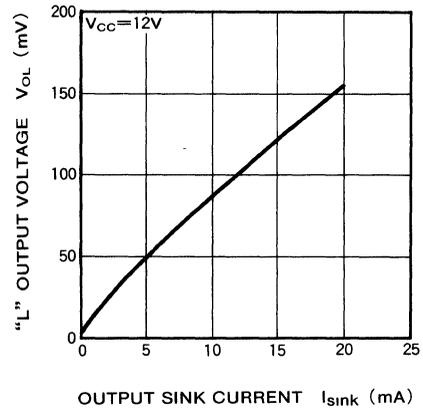
INPUT CURRENT VS. SUPPLY VOLTAGE



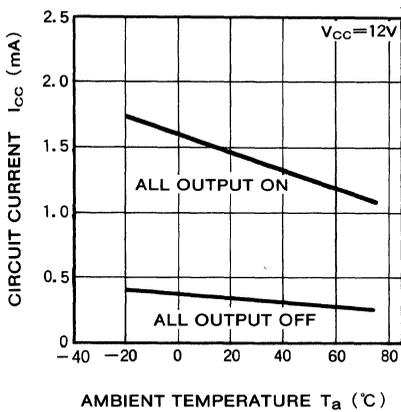
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



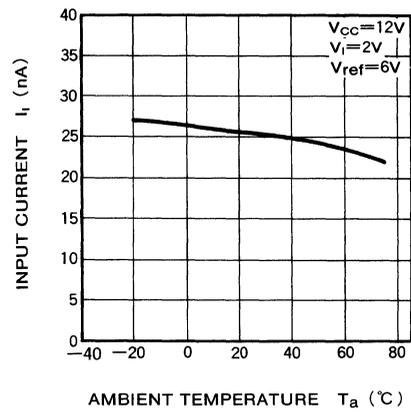
"L" OUTPUT VOLTAGE VS. OUTPUT SINK CURRENT



CIRCUIT CURRENT VS. AMBIENT TEMPERATURE

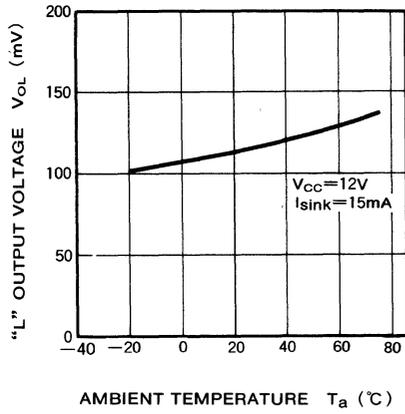


INPUT CURRENT VS. AMBIENT TEMPERATURE

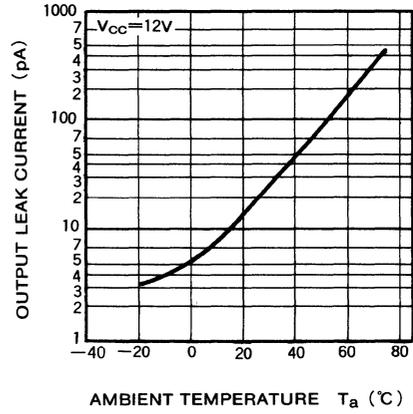


DUAL COMPARATOR

**"L" OUTPUT VOLTAGE
VS. AMBIENT TEMPERATURE**



**OUTPUT LEAK CURRENT VS.
AMBIENT TEMPERATURE**



MITSUBISHI LINEAR ICs

M51923P,FP

DUAL COMPARATOR

DESCRIPTION

The M51923P,FP is a dual (two independent) comparator and operates over a wide voltage range from a single supply voltage. The M51923P,FP has a characteristic of low power dissipation but enables high output drive, and fits to wide ranged applications, for example CR timer, oscillator, etc.

FEATURES

- Low input current 25nA(typ.)
- Wide supply voltage range 2.5V~28V
- Low dissipation current 0.4mA(typ.)(All output OFF)
1.5mA(typ.)(All output ON)
- Enables high output drive $V_{OL}=0.15V$ (typ.)
(Output current 20mA)

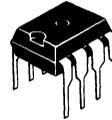
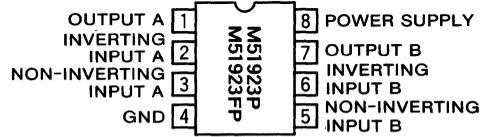
APPLICATION

Voltage comparator, window comparator, CR timer, time delay circuit, oscillator, etc.

RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 2.5~28V
- Rated supply voltage 12V

PIN CONFIGURATION (TOP VIEW)

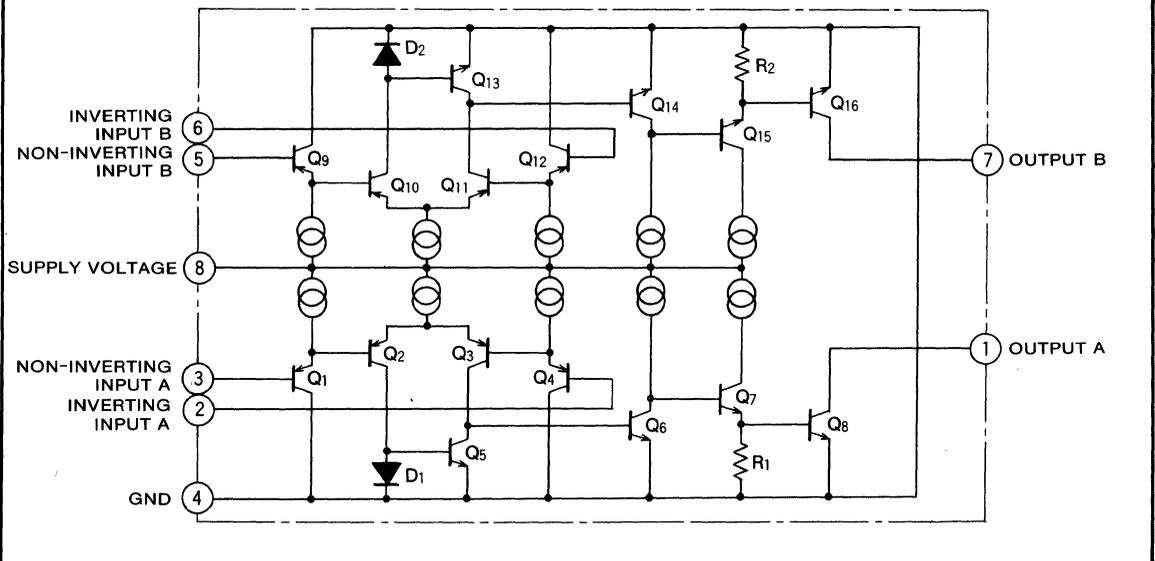


8-pin molded plastic DIP



8-pin molded plastic FLAT

EQUIVALENT CIRCUIT



DUAL COMPARATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		28	V
V_{ID}	Differential input voltage		V_{CC}	V
V_{ICM}	Common mode input voltage range		$-0.3 \sim V_{CC}$	V
I_{sink}	Output sink current		80	mA
V_{OH}	"H" output voltage		30	V
P_d	Power dissipation		650 (M51923P)	mW
			300 (M51923FP)	
T_{opr}	Operating temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^\circ\text{C}$

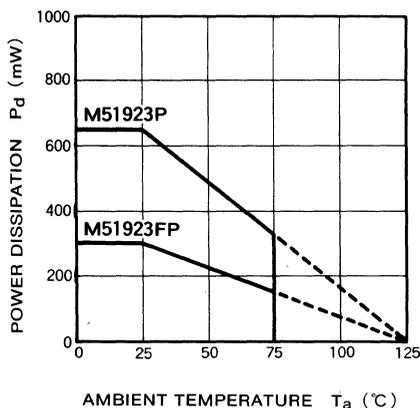
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=2.5 \sim 28\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage range		2.5		5.5	V
I_{CC1}	Circuit current 1	ALL OUTPUT ON		1.5	2.5	mA
I_{CC2}	Circuit current 2	ALL OUTPUT OFF		0.4	0.8	mA
$V_{I\ominus}$	Inverting input voltage range	NOTE	0		$V_{CC}-1.5$	V
$V_{I\oplus}$	Non-inverting input voltage range	NOTE	0		$V_{CC}-1.5$	V
V_{IO}	Input offset voltage			2	5	mV
$I_{I\ominus}$	Inverting input current			25	150	nA
$I_{I\oplus}$	Non-inverting input current			25	150	nA
I_{IO}	Input offset current			5	50	nA
V_{sat}	Output saturation voltage	$I_{OL}=20\text{mA}$		0.15	0.4	V
V_{OL}	"L" output voltage	$I_{sink}=20\text{mA}$		0.15	0.4	V
		$I_{sink}=80\text{mA}$		1		
I_{LO}	Output leak current			0.1	μA	
t_{PLH}	Output "L→H" propagation delay time			2	μs	
t_{PHL}	Output "H→L" propagation delay time			0.2	μs	

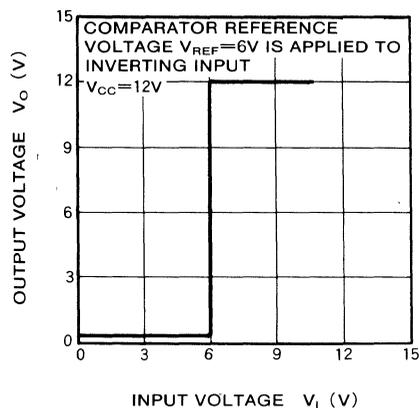
NOTE) Either inverting or non-inverting inputs (reference side) should be within this range. (Abnormal operation will not occur when the other is within the range of 0 to V_{CC} .)

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

THERMAL DERATING (MAXIMUM RATING)

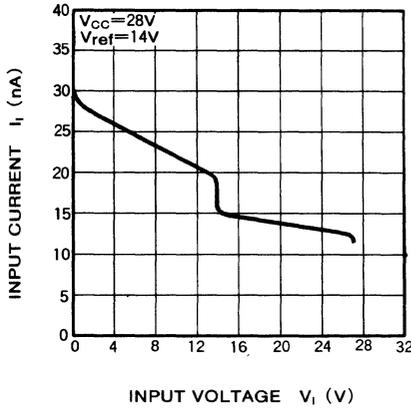


OUTPUT VOLTAGE VS. INPUT VOLTAGE

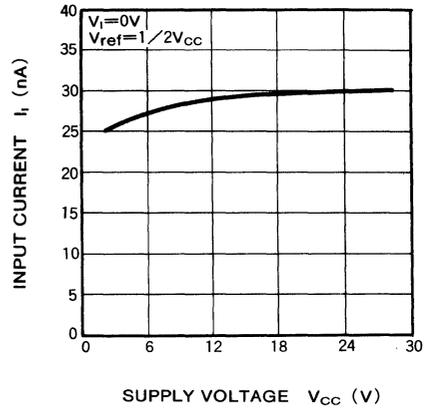


DUAL COMPARATOR

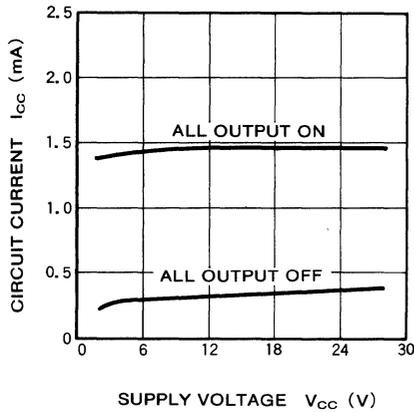
INPUT CURRENT VS. INPUT VOLTAGE



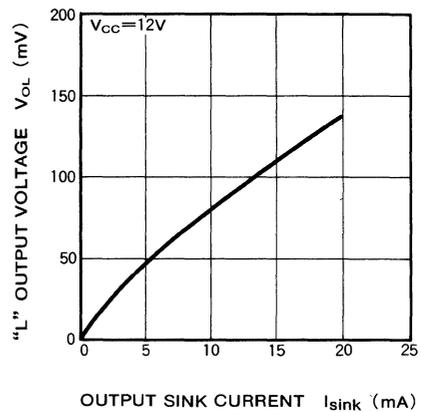
INPUT CURRENT VS. SUPPLY VOLTAGE



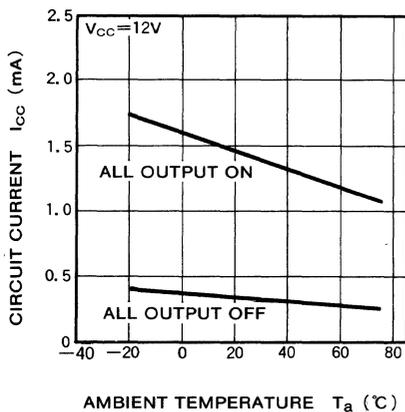
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



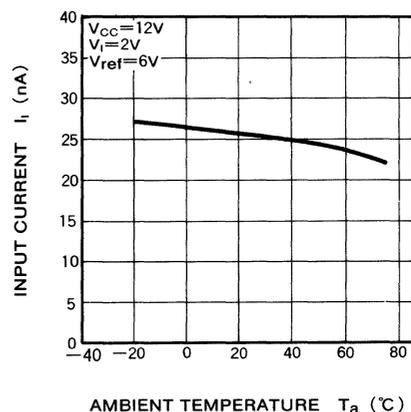
"L" OUTPUT VOLTAGE VS. OUTPUT SINK CURRENT



CIRCUIT CURRENT VS. AMBIENT TEMPERATURE

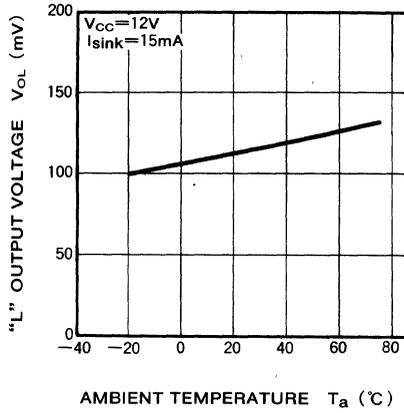


OUTPUT CURRENT VS. AMBIENT TEMPERATURE

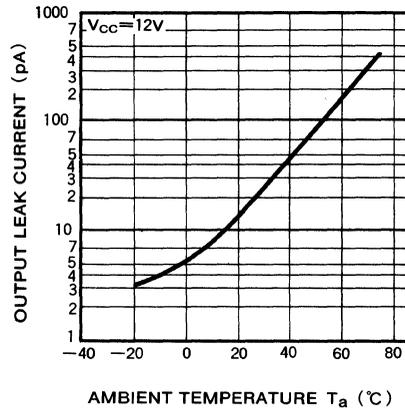


DUAL COMPARATOR

**"L" OUTPUT VOLTAGE
AMBIENT TEMPERATURE**



**OUTPUT LEAK CURRENT VS.
AMBIENT TEMPERATURE**



MITSUBISHI LINEAR ICs M5233L, P, FP

DUAL COMPARATOR

DESCRIPTION

The M5223L,P,FP is a semiconductor circuit for a comparator designed to operate over a wide supply voltage range from 2 to 36V from a single power supply, with two circuits in each 8-pin SIP and 8-pin DIP and 8-pin mini flat package. A differential circuit which is equivalent to a conventional single power supply operational amplifier is used to enable operation from GND level to improve input characteristics. Power dissipation (circuit current) is low and output voltage is large. It fits to a general-purpose comparator for a variety of electronic equipment.

FEATURES

- Wide operating supply voltage range 2V~36V
Dual power supplies: $\pm 1V \sim \pm 18V$
- Low circuit current 0.6mA (typ.)
- Wide common mode input voltage range
..... 0V~ $V_{CC}-1.5V$ (single power supply)
- Open collector output
- Output sink current 25mA
- Response time 13 μ sec
- Pin compatible with general-purpose comparators 393, 2930

APPLICATION

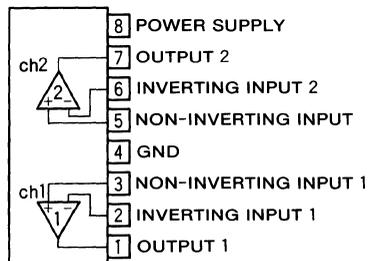
Voltage comparator, window comparator, CR timer, time delay circuit, oscillator, etc.

RECOMMENDED OPERATING CONDITIONS

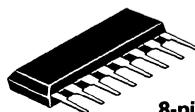
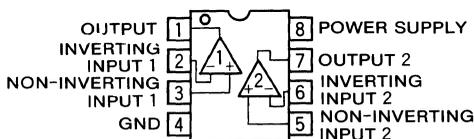
- Supply voltage range 2~36V
- Rated supply voltage 12V

PIN CONFIGURATION (TOP VIEW)

SIP



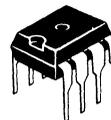
DIP, MINI FLAT



8-pin molded plastic SIP

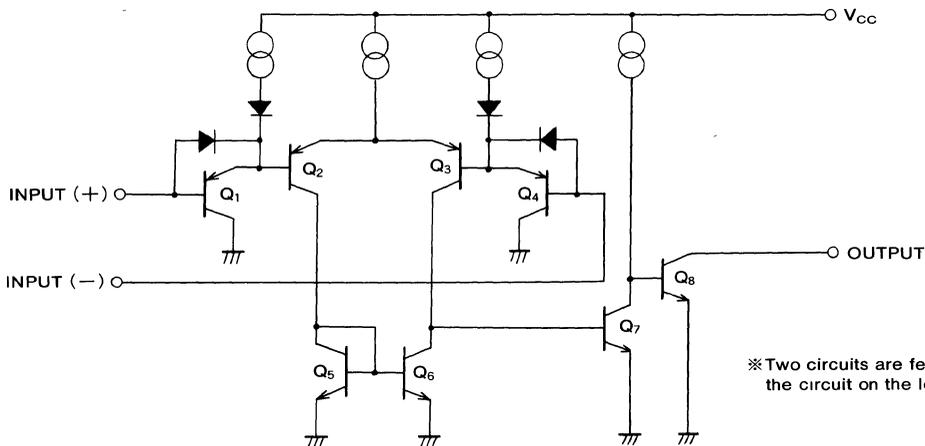


8-pin molded plastic FP
(MINI FLAT)



8-pin molded plastic DIP

EQUIVALENT CIRCUIT



※ Two circuits are featured in the circuit on the left

DUAL COMPARATOR

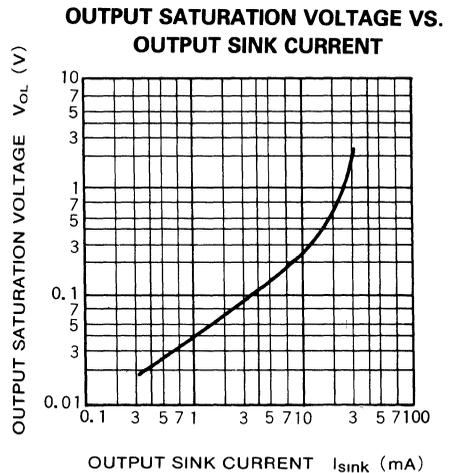
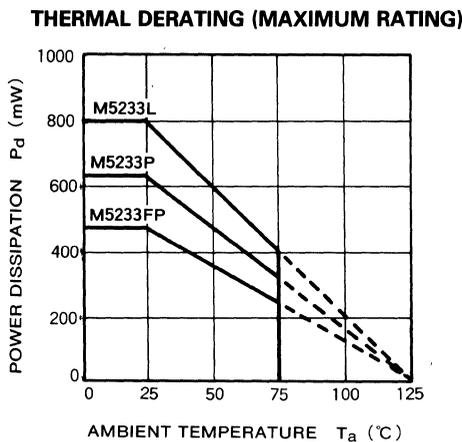
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Ratings	Unit
V_{CC}	Supply voltage	36(± 18)	V
V_{ID}	Differential input voltage	36	V
V_{ICM}	Common mode input voltage range	-0.3~+36	V
P_d	Power dissipation	800(SIP)/625(DIP)/440(FP)	mW
T_{opr}	Operating temperature	-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature	-55~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=5\text{V}$)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{IO}	Input offset voltage	$V_O=1.4\text{V}$, $V_{REF}=1.4\text{V}$, $R_S=0\Omega$		2	5	mV
I_{IO}	Input offset current			5	50	nA
I_B	Input bias current			25	250	nA
V_{ICM}	Common mode input voltage range		0		$V_{CC}-1.5$	V
G_V	Voltage gain	$R_L=15\text{k}\Omega$		200		V/mV
I_{CC}	Circuit current	$R_L=\infty$		0.6	1	mA
t_{PLH}	Response time	$R_L=5.1\text{k}\Omega$, $V_{RL}=5\text{V}$		1.3		μsec
I_{sink}	Output sink current	$V_{IN(-)}=1\text{V}$, $V_{IN(+)}=0\text{V}$, $V_O \leq 1.5\text{V}$	10	25		mA
V_{OL}	Output saturation voltage	$V_{IN(-)}=1\text{V}$, $V_{IN(+)}=0\text{V}$, $I_{sink}=8\text{mA}$		200	400	mV
I_{LO}	Output leak current	$V_{IN(+)}=1\text{V}$, $V_{IN(-)}=0\text{V}$, $V_O=5\text{V}$		0.1		nA

TYPICAL CHARACTERISTICS



M51200P

DUAL COMPARATOR

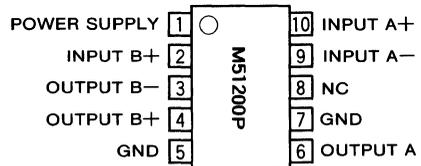
DESCRIPTION

The M51200P is a semiconductor integrated circuit for a dual voltage comparator designed to operate from a single power supply.

Comparator A has two inputs and one output while comparator B has one input and two outputs with half the supply voltage as its reference voltage. Especially the M51200P can operate at low supply voltage and has superiority as to input current (input resistance) and fits to wide ranged applications.

The M51200P's package is 10-pin FLAT package and makes mounting of components easy.

PIN CONFIGURATION (TOP VIEW)



NC : NO CONNECTION

FEATURES

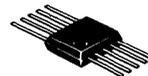
- Operates over low supply voltage range 1.4~6.0V
- Low input current 3nA(typ.)
- Capable of directly driving a relay or a lamp
..... 40mA(max.)
- High breakdown voltage 18V(max.)
- Includes voltage surge absorbing zener diode for protection

APPLICATION

Voltage comparator, electronic shutter, CR timer, time delay circuit, oscillator.

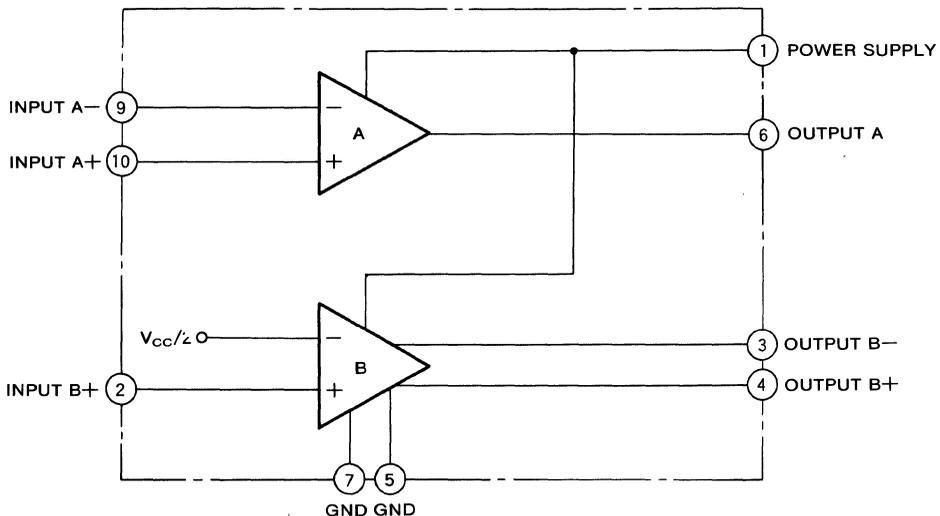
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 1.4~6.0V
- Rated supply voltage 3.0V



10-pin molded plastic FLAT

EQUIVALENT CIRCUIT



DUAL COMPARATOR

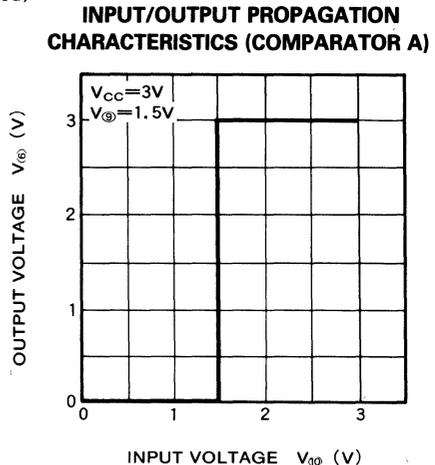
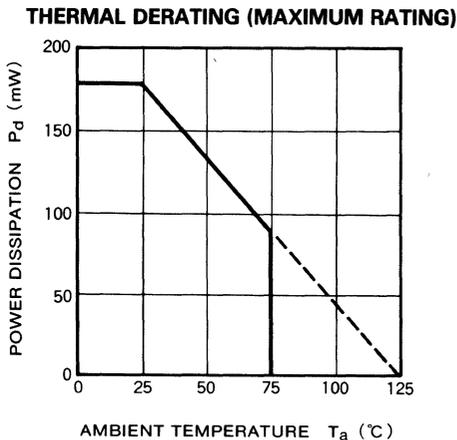
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Power supply		6	V
V_{IN}	Input voltage		V_{CC}	V
$I_{(6)}$	Output sink current		40	mA
$I_{(3)}$			2	mA
$I_{(4)}$			25	mA
$V_{(6)}$	Output drive voltage		18	V
$V_{(3)}$			10	V
$V_{(4)}$			18	V
P_d	Power dissipation		180	mW
T_{opr}	Operating temperature		$-20\sim+75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40\sim+125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=3.0\text{V}$, unless otherwise noted)

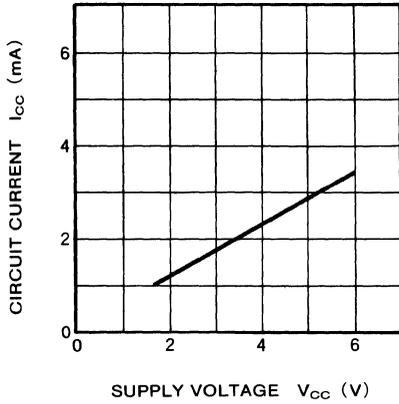
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage range		1.4		6.0	V
I_{CC}	Circuit current	$V_{CC}=3\text{V}$		1.8	2.8	mA
		$V_{CC}=6\text{V}$		3.4	4.9	
$I_{IN(1)}$	Input current	$V_{CC}=3\text{V}$		3	10	nA
$I_{IN(3)}$				3	10	nA
$I_{IN(2)}$				8	100	nA
V_{REF}	Reference voltage of comparator B	$V_{CC}=3\text{V}$	1.35	1.5	1.65	V
$V_{S(6)}$	Output saturation voltage	$I_{(6)}=20\text{mA}$		0.18	0.3	V
$V_{S(3)}$		$I_{(3)}=10\mu\text{A}$		38	60	mV
$V_{S(4)}$		$I_{(4)}=25\text{mA}$		0.2	0.5	V
$V_{Z(6)}$	Output protection zener voltage	$I_{(6)}=5\text{mA}$	18	21	26	V
$V_{Z(4)}$		$I_{(4)}=5\text{mA}$	18	21	26	V
V_{IN}	Input voltage range	$V_{CC}=3\text{V}$	0.8		$V_{CC}-0.2$	V
t_{PLH}	Output "L→H" propagation delay time	$V_{CC}=3\text{V}$		20		μs
t_{PHL}	Output "H→L" propagation delay time			20		μs

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

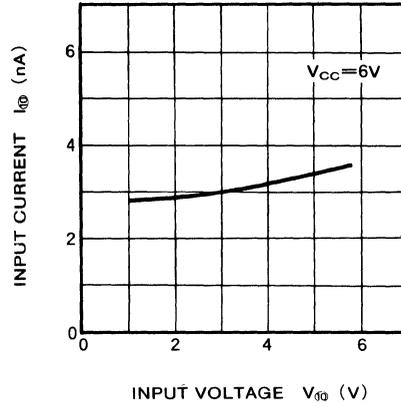


DUAL COMPARATOR

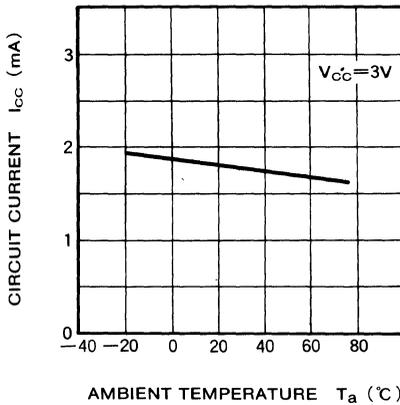
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



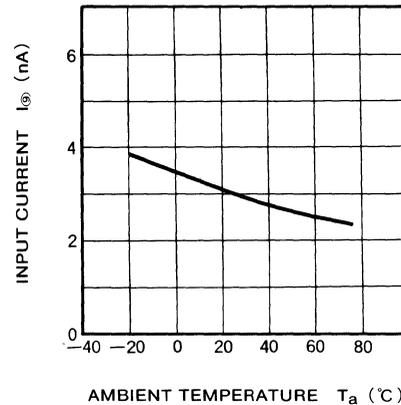
INPUT CURRENT AT PIN 10 VS. INPUT VOLTAGE



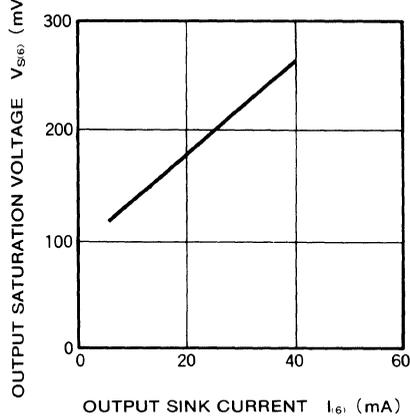
CIRCUIT CURRENT VS. AMBIENT TEMPERATURE



INPUT CURRENT AT PIN 9 VS. AMBIENT TEMPERATURE



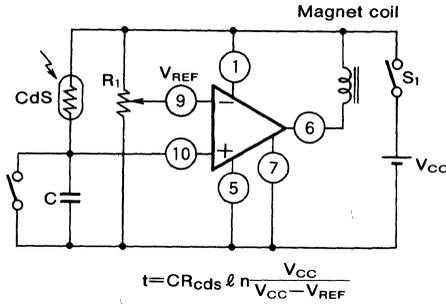
OUTPUT SATURATION VOLTAGE AT PIN 6 VS. OUTPUT SINK CURRENT



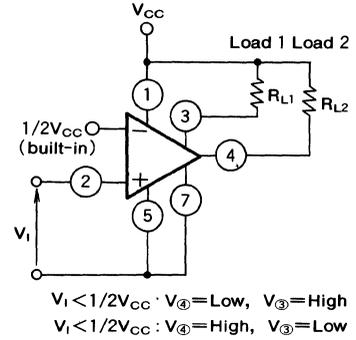
DUAL COMPARATOR

APPLICATION EXAMPLES

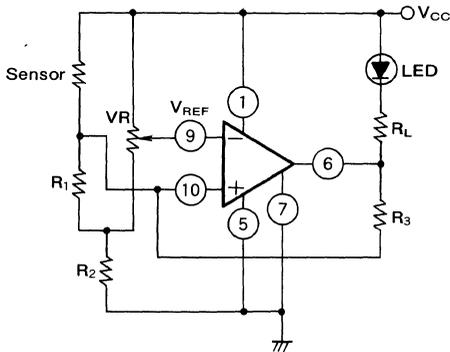
(1) Electronic shutter



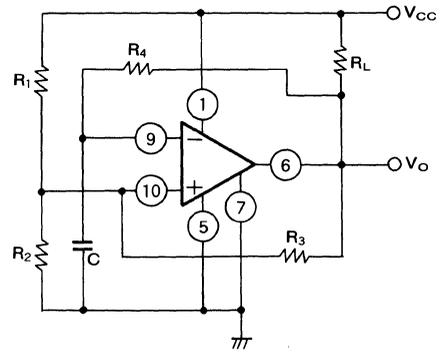
(2) Comparator



(3) Detection circuit



(4) Oscillator

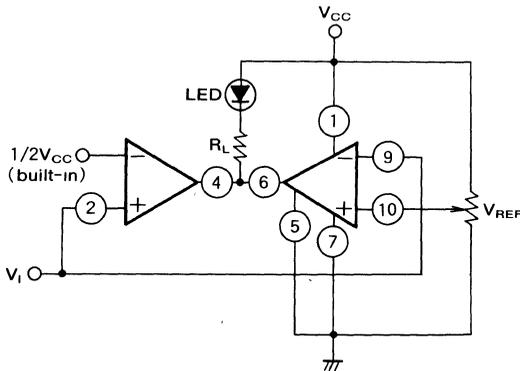


$R_L < R_3, R_4$

$f = \frac{1}{CR_4 \ln \frac{R_L}{R_h} \cdot \frac{1-R_L}{1-R_h}}$ (Hz)

$R_h = \frac{R_2}{R_2 + \frac{R_1 R_3}{R_1 + R_3}}$
 $R_L = \frac{\frac{R_2 R_3}{R_2 + R_3}}{R_1 + \frac{R_2 R_3}{R_2 + R_3}}$

(5) Window comparator



Note : Connect both pin ⑤ and pin ⑦ to GND

M51209P

QUAD COMPARATOR

DESCRIPTION

The M51209P is a quad (four independent) comparator and operates over a wide voltage range from a single supply voltage. Especially the M51209P has superiority as to characteristics of input current (input resistance) and fits to wide ranged applications, for example CR Timer, oscillator, etc.

FEATURES

- Low input current (high input resistance).....20nA(typ.)
- Wide supply voltage range.....2.5V~28V
- Low dissipation current.....6.8mA(typ.)
- Capable of driving a relay or a lamp directly
200mA(max.)
- Includes voltage surge absorbing zener diodes
- High output breakdown voltage.....30V(max.)
- Low output voltage ($I_{sink}=60mA$).....0.2V(typ.)
- Low input offset voltage.....2mV(typ.)

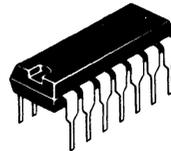
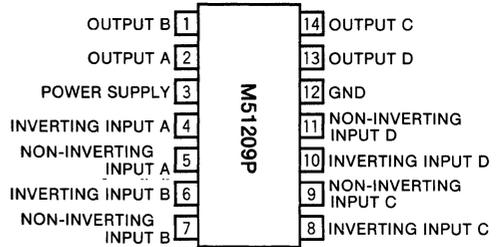
APPLICATION

Voltage comparator, sequential timer, pulse generator, analog / digital converter, time delay circuit

RECOMMENDED OPERATING CONDITIONS

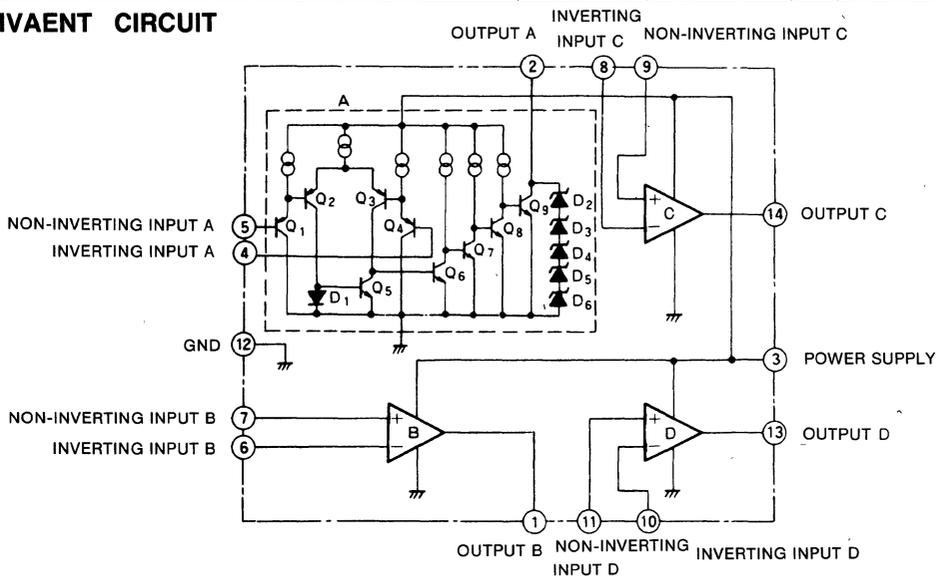
- Supply voltage range.....2.5~28V
- Rated supply voltage.....12V

PIN CONFIGURATION (TOP VIEW)



14-pin molded plastic DIP

EQUIVALENT CIRCUIT



QUAD COMPARATOR

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

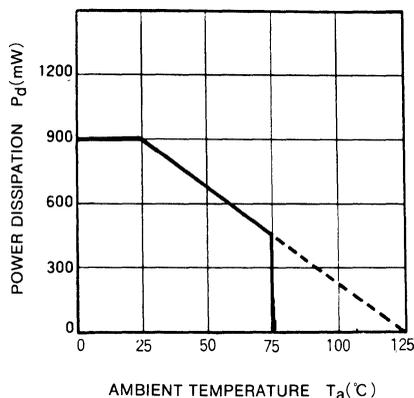
Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		28	V
V _{ID}	Differential input voltage		V _{CC}	V
V _{ICM}	Common mode input voltage range		-0.3~V _{CC}	V
I _{sink}	Output sink current		200	mA
V _{OH}	"H" output voltage		30	V
P _d	Power dissipation		900	mW
T _{opr}	Operating temperature		-20~+75	°C
T _{stg}	Storage temperature		-40~+125	°C

ELECTRICAL CHARACTERISTICS (T_a=25°C, V_{CC}=2.5~28V, unless otherwise noted)

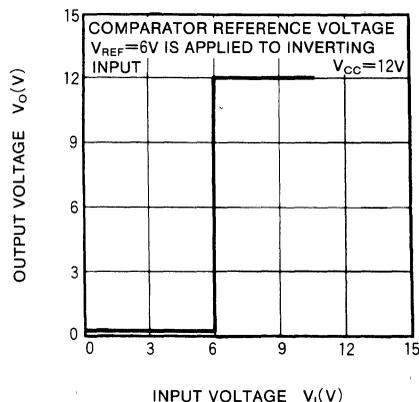
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage range		2.5		28	V
I _{CC}	Circuit current			6.8	9.5	mA
V _{I⊖}	Inverting input voltage range		0		V _{CC} -1.5	V
V _{I⊕}	Non-inverting input voltage range		0		V _{CC} -1.5	V
V _{IO}	Input offset voltage			2	7	mV
I _{I⊖}	Inverting input current			20	100	nA
I _{I⊕}	Non-inverting input current			20	100	nA
I _{IO}	Input offset current			5	50	nA
V _{OL}	"L" output voltage	I _{sink} =60mA I _{sink} =200mA		0.2	0.4	V
I _{LO}	Output leak current				0.1	μA
t _{PLH}	Output "L→H" propagation delay time			2		μs
t _{PHL}	Output "H→L" propagation delay time			1		μs

TYPICAL CHARACTERISTICS (T_a=25°C, unless otherwise noted)

THERMAL DERATING
(MAXIMUM RATING)

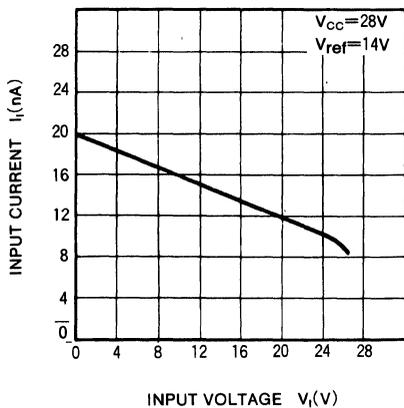


OUTPUT VOLTAGE VS
INPUT VOLTAGE

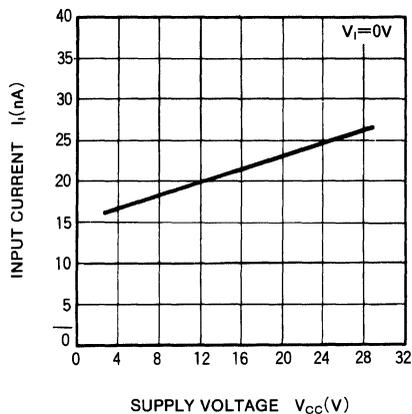


QUAD COMPARATOR

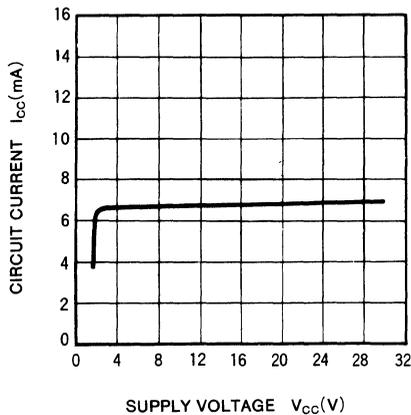
**INPUT CURRENT VS
 INPUT VOLTAGE**



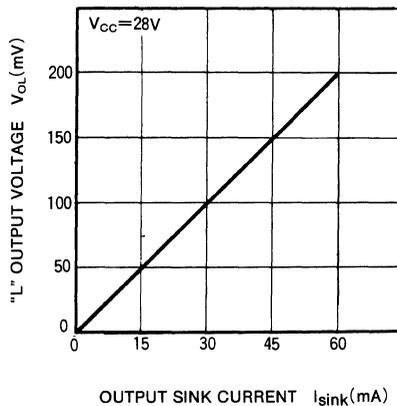
**INPUT CURRENT VS
 SUPPLY VOLTAGE**



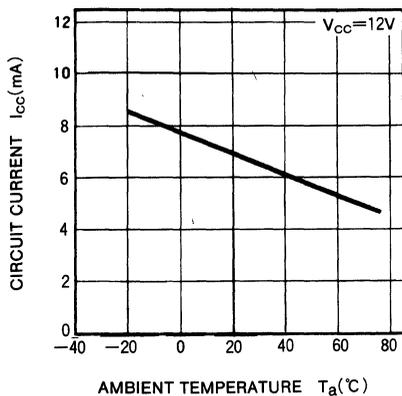
**CIRCUIT CURRENT VS
 SUPPLY VOLTAGE**



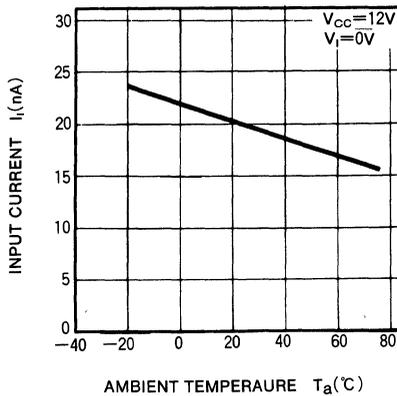
**"L" OUTPUT VOLTAGE VS
 OUTPUT SINK CURRENT**



**CIRCUIT CURRENT VS
 AMBIENT TEMPERATURE**



**INPUT CURRENT VS
 AMBIENT TEMPERATURE**



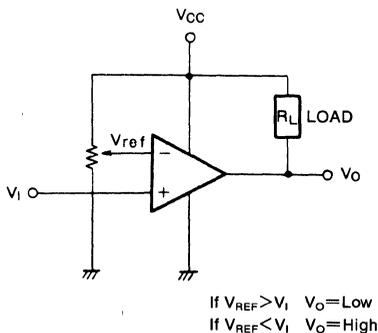
QUAD COMPARATOR

PRECAUTIONS FOU USE

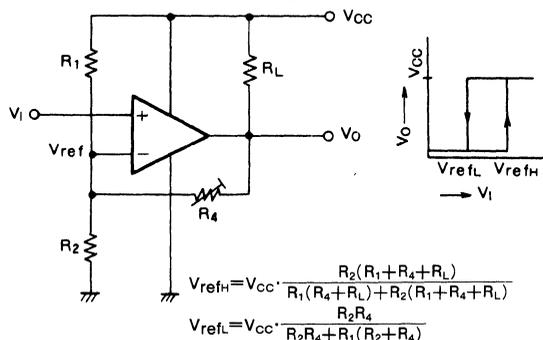
1. Special care must be taken to protect the M51209P from large surges in current, such as may result from the incorrect connection of the V_{CC} and GND terminals.
2. Output is "open collector" and a loading resistor is not included. Connect a loading resistor to stabilize operation, when driving another.

APPLICATION EXAMPLES

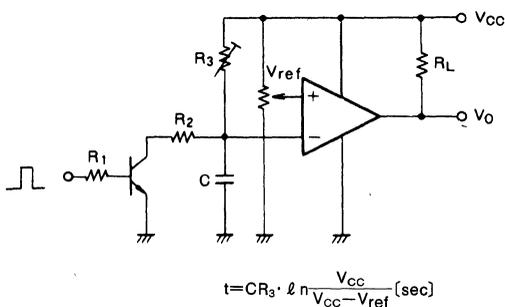
(1) Voltage comparator



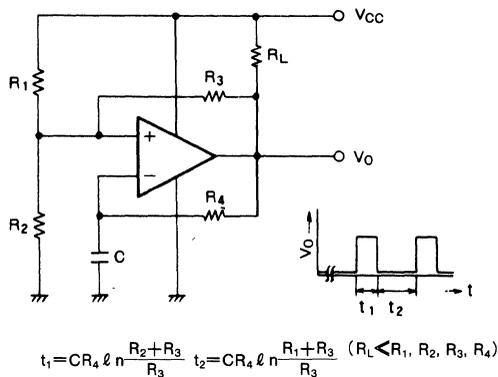
(2) Schmitt trigger circuit



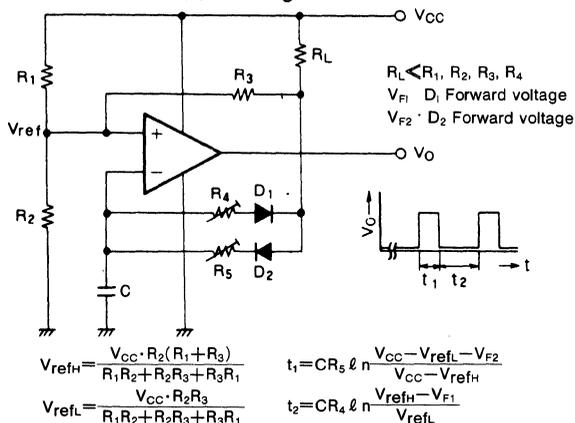
(3) Monostable multi-vibrator



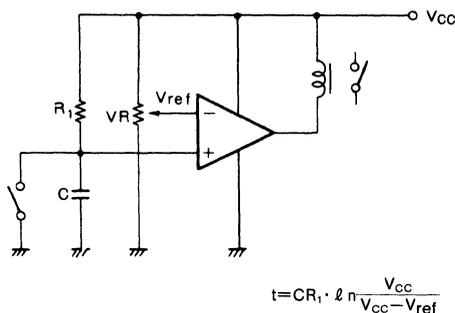
(4) Unstable multi-vibrator



(5) Pulse generator

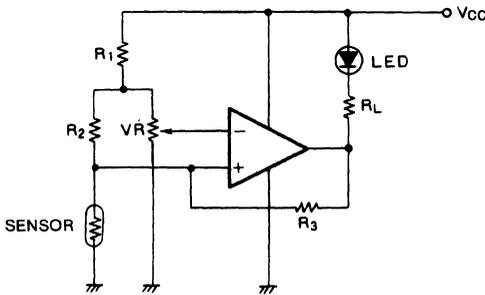


(6) CR Timer

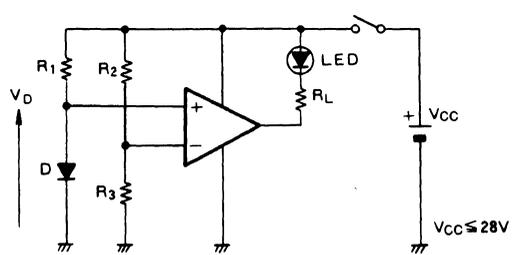


QUAD COMPARATOR

(7) Sensor detector



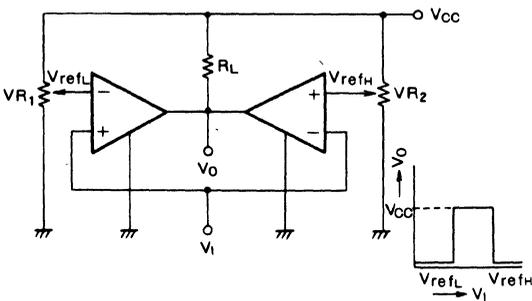
(8) Battery check circuit



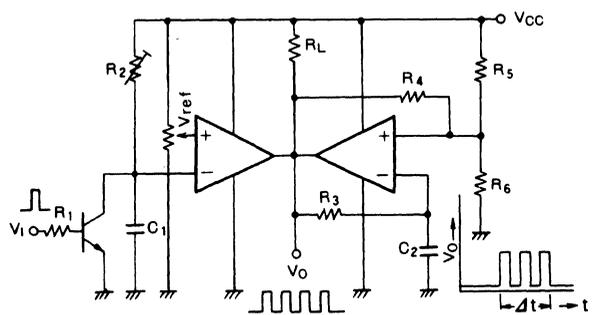
$$V_D < V_{CC} \cdot \frac{R_3}{R_2 + R_3}; \text{ LED} \rightarrow \text{ON}$$

$$V_D > V_{CC} \cdot \frac{R_3}{R_2 + R_3}; \text{ LED} \rightarrow \text{OFF}$$

(9) Window comparator

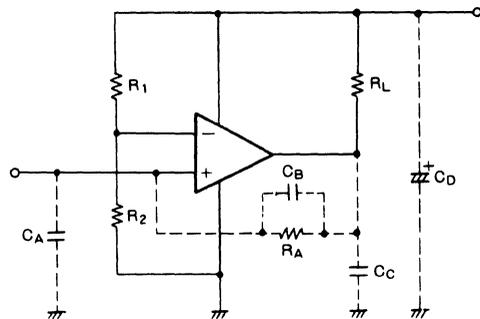


(10) Pulse train generator



$$\Delta t = C_1 R_2 \cdot \ln \frac{V_{CC}}{V_{CC} - V_{ref}}$$

(11) Countermeasure against oscillation



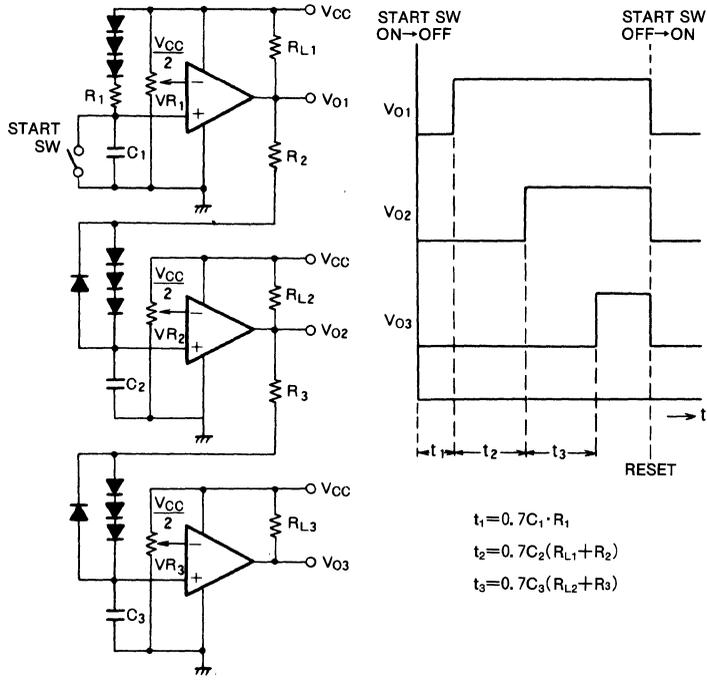
(Note) Taking steps against oscillation

The M51209P may oscillate according to input condition. If the M51209P should oscillate, the following countermeasures are applicable.

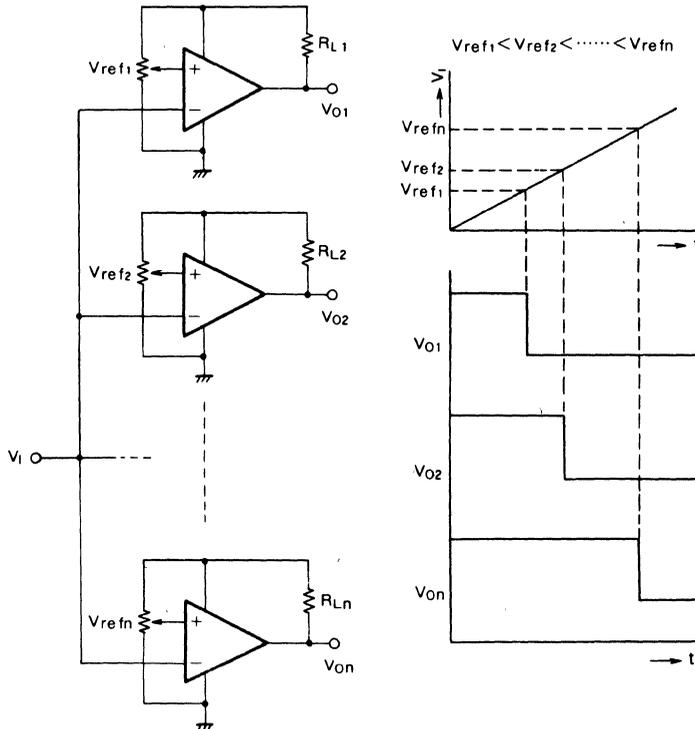
- In case of connecting input signal with chattering, connect a capacitor of small C_A value.
- In case of oscillation with ordinary input, employ positive feedback inserting R_A (large resistor), C_B (no polar) or connect C_C .
- When the supply voltage is not stabilized, connect C_D (a large electrolytic capacitor) to absorb the supply voltage change.

QUAD COMPARATOR

(12) Sequential timer



(13) Analog/Digital converter



MITSUBISHI LINEAR ICs

M51924P,FP

QUAD COMPARATOR

DESCRIPTION

The M51924P,FP is a quad (four independent) comparator and operates over a wide voltage range from a single supply voltage. Especially the M51924P,FP has superiority as to characteristics of input current (input resistance) and fits to wide ranged applications, for example CR timer, oscillator, etc.

FEATURES

- Low input current 25nA(typ.)
- Wide supply voltage range 2.5~28V
- Low dissipation current
..... 0.8mA(typ.)(All output OFF)
..... 3mA(typ.)(ALL output ON)
- Enable high output drive $V_{OL}=0.15V$ (typ.)
(Output current 20mA)

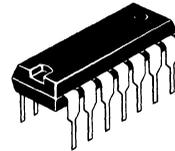
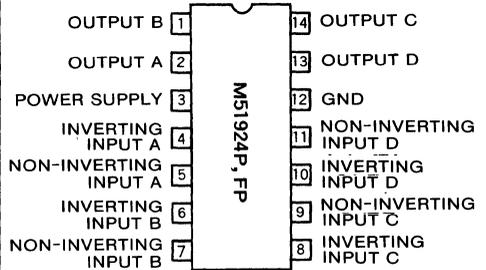
APPLICATION

Voltage comparator, window comparator, CR timer, time delay circuit, oscillator.

RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 2.5~28V
- Rated supply voltage 12V

PIN CONFIGURATION (TOP VIEW)

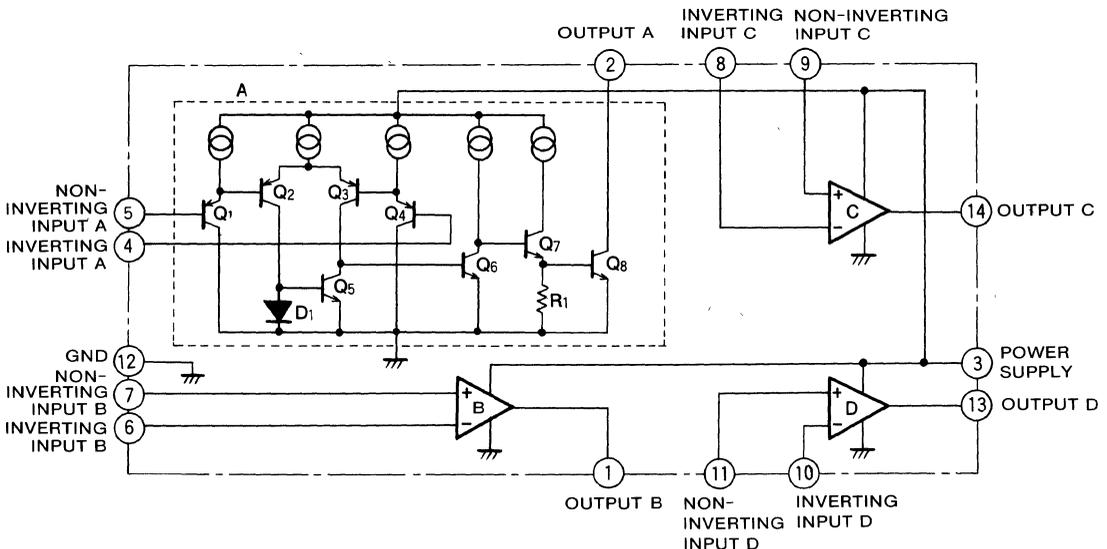


14-pin molded plastic DIP



14-pin molded plastic FLAT

EQUIVALENT CIRCUIT



QUAD COMPARATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

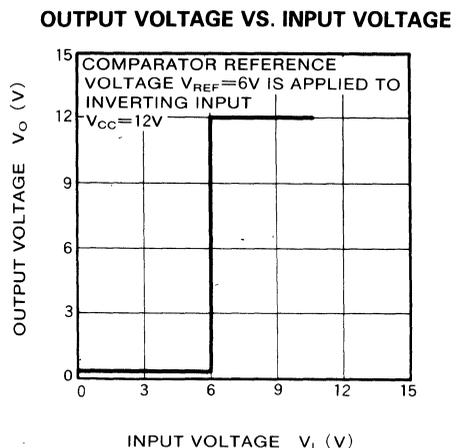
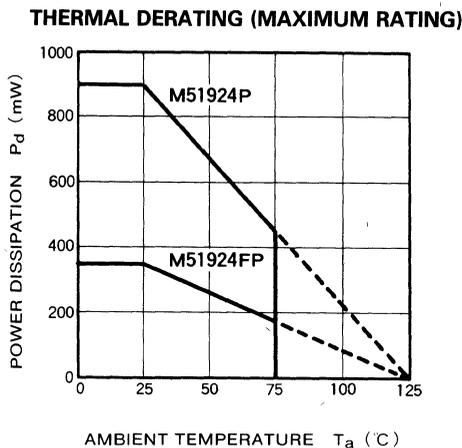
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		28	V
V_{ID}	Differential input voltage		V_{CC}	V
V_{ICM}	Common mode input voltage range		$-0.3 \sim V_{CC}$	V
I_{sink}	Output sink current		80	mA
V_{OH}	"H" output voltage		30	V
P_d	Power dissipation		900 (M51924P)	mW
			350 (M51924FP)	
T_{opr}	Operating temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=2.5 \sim 28\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage range		2.5		28	V
I_{CC1}	Circuit current 1	ALL OUTPUT ON		3.0	5.0	mA
I_{CC2}	Circuit current 2	ALL OUTPUT OFF		0.8	1.6	mA
$V_{I\ominus}$	Inverting input voltage range	NOTE	0		$V_{CC}-1.5$	V
$V_{I\oplus}$	Non-inverting input voltage range	NOTE	0		$V_{CC}-1.5$	V
V_{IO}	Input offset voltage			2	5	mV
$I_{I\ominus}$	Inverting input current			25	150	nA
$I_{I\oplus}$	Non-inverting input current			25	150	nA
I_{IO}	Input offset current			5	50	nA
V_{sat}	Output saturation voltage	$I_{OL}=20\text{mA}$		0.15	0.4	V
V_{OL}	"L" output voltage	$I_{sink}=20\text{mA}$		0.15	0.4	V
		$I_{sink}=80\text{mA}$		1		
I_{LO}	Output leak current			0.1	μA	
t_{PLH}	Output "L→H" propagation delay time			2	μs	
t_{PHL}	Output "H→L" propagation delay time			0.2	μs	

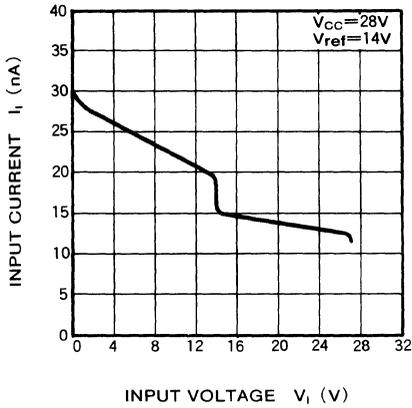
NOTE) Either inverting or non-inverting inputs (reference side) should be within this range (Abnormal operation will not occur when the other is within the range of 0 to V_{CC})

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

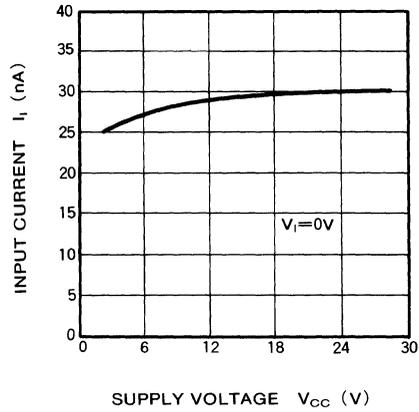


QUAD COMPARATOR

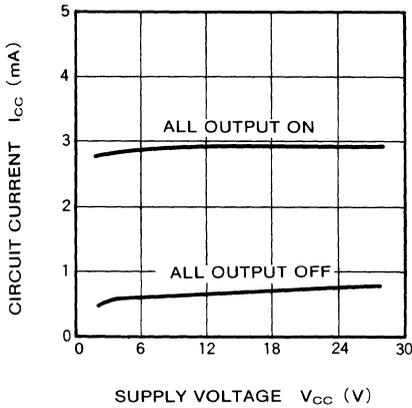
INPUT CURRENT VS. INPUT VOLTAGE



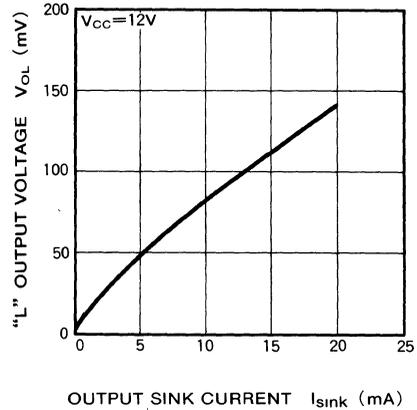
INPUT CURRENT VS. SUPPLY VOLTAGE



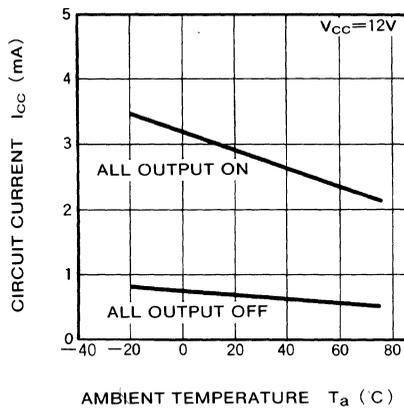
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



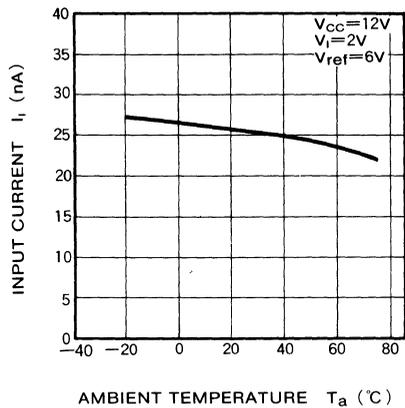
"L" OUTPUT VOLTAGE VS. OUTPUT SINK CURRENT



CIRCUIT CURRENT VS. AMBIENT TEMPERATURE

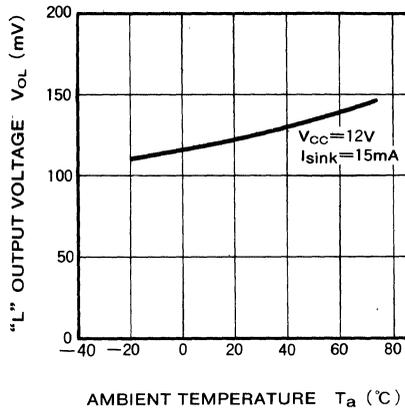


INPUT CURRENT VS. AMBIENT TEMPERATURE

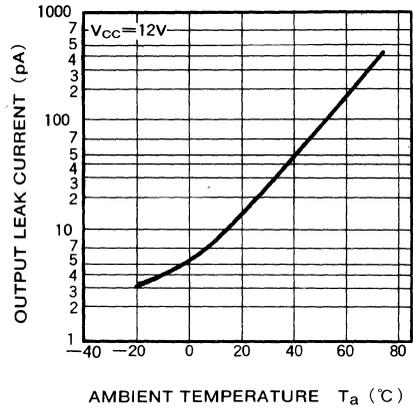


QUAD COMPARATOR

**"L" OUTPUT VOLTAGE VS.
AMBIENT TEMPERATURE**



**OUTPUT LEAK CURRENT VS.
AMBIENT TEMPERATURE**



MITSUBISHI LINEAR ICs M5234P, FP

QUAD COMPARATOR

DESCRIPTION

The M5234P,FP is a semiconductor circuit for a comparator designed to operate over a wide supply voltage range from 2 to 36V from a single power supply, with four circuits in each 14-pin DIP and 14-pin mini flat package. A differential circuit which is equivalent to a conventional single power supply operational amplifier is used to enable operation from GND level to improve input characteristics. Power dissipation (circuit current) is low and output voltage is large. It is a general-purpose comparator suitable for a variety of electronic equipment.

FEATURES

- Wide operating supply voltage range 2V~36V
Dual power supplies: $\pm 1V \sim \pm 18V$
- Low circuit current 0.8mA (typ.)
- Wide common mode input voltage range
..... 0V~ $V_{CC}-1.5V$ (single power supply)
- Open collector output
- Output sink current 25mA
- Response time 1.3 μ sec
- Pin compatible with general-purpose comparators 339, 2901

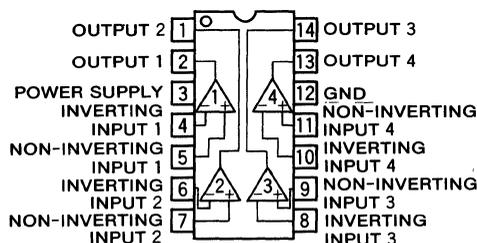
APPLICATION

Voltage comparator, window comparator, CR timer, time delay circuit, oscillator, etc.

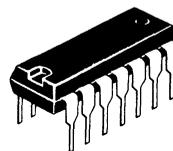
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 2~36V
- Rated supply voltage 12V

PIN CONFIGURATION (TOP VIEW) DIP, MINI FLAT

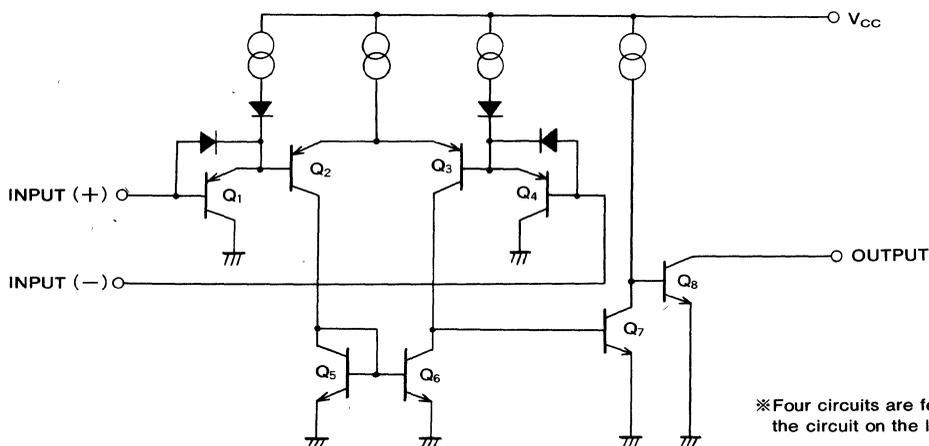


14-pin molded plastic FP
(MINI FLAT)



14-pin molded plastic DIP

EQUIVALENT CIRCUIT



※ Four circuits are featured in the circuit on the left.

QUAD COMPARATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

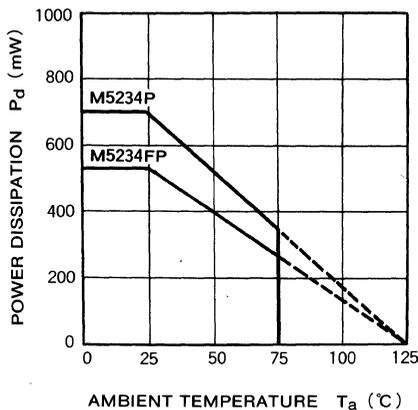
Symbol	Parameter	Ratings	Unit
V_{CC}	Supply voltage	36(± 18)	V
V_{ID}	Differential input voltage	36	V
V_{ICM}	Common mode input voltage range	-0.3~+36	V
P_d	Power dissipation	700(DIP)/550(FP)	mW
T_{opr}	Operating temperature	-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature	-55~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 15\text{V}$)

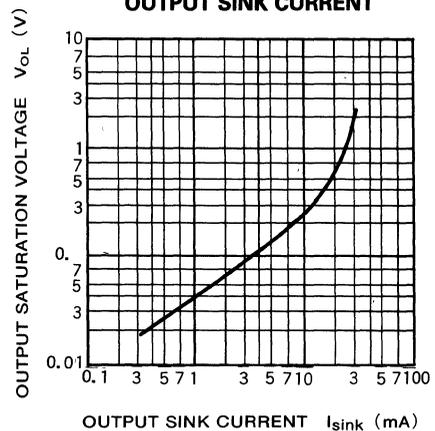
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{IO}	Input offset voltage	$V_O=1.4\text{V}$, $V_{REF}=1.4\text{V}$, $R_S=0\Omega$		2	5	mV
I_{IO}	Input offset current			5	50	nA
I_B	Input bias current			25	250	nA
V_{ICM}	Common mode input voltage range		0		$V_{CC}-1.5$	V
G_V	Voltage gain	$R_L=15\text{k}\Omega$		200		V/mV
I_{CC}	Circuit current	$R_L=\infty$		0.8	2	mA
t_{PLH}	Response time	$R_L=5.1\text{k}\Omega$, $V_{RL}=5\text{V}$		1.3		μsec
I_{sink}	Output sink current	$V_{IN(-)}=1\text{V}$, $V_{IN(+)}=0\text{V}$, $V_O \leq 1.5\text{V}$	10	25		mA
V_{OL}	Output saturation voltage	$V_{IN(-)}=1\text{V}$, $V_{IN(+)}=0\text{V}$, $I_{sink}=8\text{mA}$		200	400	mV
I_{LO}	Output leakage current	$V_{IN(+)}=1\text{V}$, $V_{IN(-)}=0\text{V}$, $V_O=5\text{V}$		0.1		nA

TYPICAL CHARACTERISTICS

THERMAL DERATING (MAXIMUM RATING)



OUTPUT SATURATION VOLTAGE VS. OUTPUT SINK CURRENT

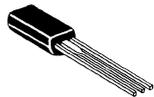


VOLTAGE REGULATOR

QUICK REFERENCE TABLE OF VOLTAGE REGULATORS

Application	Type No.	Function	Features	Maximum ratings			Electrical characteristics (typ.)					Equivalent	Package outline			
				Input voltage range V_i (V)	load current I_L (mA)	Power dissipation P_d (mW)	Input bias current I_B (mA)	Ripple rejection ratio RR (dB)	Reference voltage V_{REF} (V)	Minimum I/O voltage differential V_{IO} (V)	Output noise voltage V_{NO} (μ Vrms)					
Variable output	M5230L	Positive/negative variable voltage regulator (tracking type)	<ul style="list-style-type: none"> Withstands high voltage, and has high ripple rejection ratio Capable of ON/OFF control and variable rising time constant of output voltage 	$\pm 8 \sim \pm 35$	30	800	1.3	68	1.8	2.5 (0.1 possible)	12	TA7179P	8-pin SIP			
	M5231TL	Variable positive single power supply	<ul style="list-style-type: none"> Withstands high voltage, and has high ripple rejection ratio Constant variable and ON/OFF control at the time of output voltage rise 	8~70	30	300	1.2	62	1.8	2.5 (0.1 possible)	6	μ PC141 μ PC305 TA7089P	5-pin SIP			
Low voltage differential operation	M5235L	3-Terminal low voltage variable output voltage regulator (drive type)	<ul style="list-style-type: none"> Very low input-output voltage differential operation Low operating voltage ($V_{INmin}=1.3V$) Small outline 	$V_{IN}1.3\sim 7$	20	300	1.6	56	0.79	0.1	18	—	Compack 3-pin SIP			
	M5236	General-purpose 3-terminal variable output voltage regulator (driver type)	<ul style="list-style-type: none"> Low input-output voltage differential operation High withstand voltage High ripple rejection ratio 	$V_{IN}3.5\sim 36$	30	900	1.7	68	1.26	0.2	33	—	L: 3-pin SIP ML: 3-pin mini FLAT			
Fixed positive output	*M5278L05	5V 3-terminal regulator	<ul style="list-style-type: none"> Internal fold-back protection circuit for load terminal short circuit High ripple rejection ratio High output-voltage accuracy 	8~36	100	900	4.8	73	Output voltage V_O 5 ± 0.25	2	49	78L	3-pin SIP (TO-92L)			
	*M5278L56	5.6V 3-terminal regulator		8.6~36										73	5.6 ± 0.28	55
	*M5278L08	8V 3-terminal regulator		11~36										59	8 ± 0.4	80
	*M5278L09	9V 3-terminal regulator		12~36										58	9 ± 0.45	90
	*M5278L10	10V 3-terminal regulator		13~36										57	10 ± 0.5	100
	*M5278L12	12V 3-terminal regulator		15~36										55	12 ± 0.6	110
	*M5278L15	15V 3-terminal regulator		18~36										52	15 ± 0.75	140
	Switching	M5T494		Switching regulator control										<ul style="list-style-type: none"> Internal under voltage lock-out circuit Output voltage of reference voltage circuit.....$5V\pm 5\%$ Applicable to single-end, push-pull 	7~40V	200

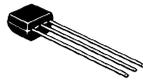
★: New product



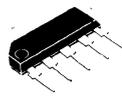
3-pin SIP (TO-92L)



3-pin mini FLAT



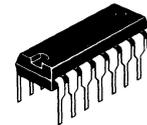
Small 3-pin SIP



5-pin SIP



8-pin SIP



16-pin DIP



16-pin FLAT

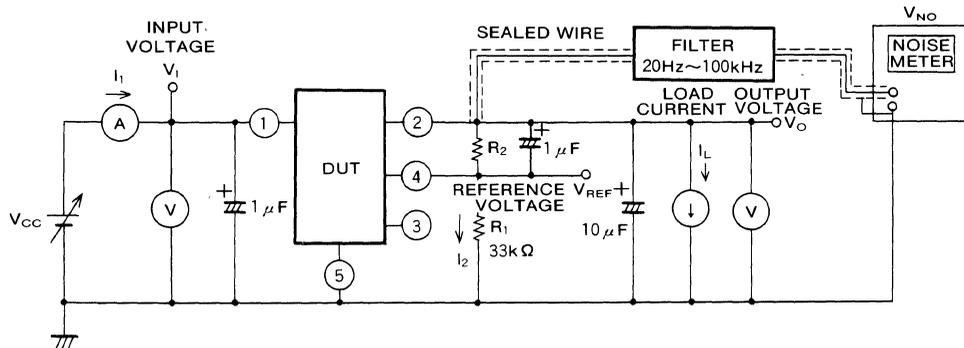
TEST METHOD, APPLICATION EXAMPLE PRECAUTIONS FOR USE OF VOLTAGE REGULATOR

TEST METHOD, APPLICATION EXAMPLE AND PRECAUTIONS FOR USE OF VOLTAGE REGULATOR

Methods of measuring various electric characteristics, applications and general precautions in the use of power-supply ICs are explained, with M5231, a general purpose variable-voltage positive power-supply IC as an example.

1. Test method

1-1. Test circuit 1



V_{REF} : Reference voltage

Reference voltage for variable output voltage regulator. Voltage between pins ④—⑤ and is 1.8V typ.

V_O : Output voltage

Voltage between pins ②—⑤. V_O is determined by the following equation.

$$V_O = V_{REF} \times \left(1 + \frac{R_2}{R_1}\right) (V)$$

V_{I-O} : Minimum input-output voltage differential

Voltage difference of input voltage (V_I) and output voltage (V_O) required for stable operation of the device.

Difference of input voltage and output voltage is measured when input voltage regulation (Reg-in) is reduced by 6dB.

Reg-in: Input regulation

Indicates ratio of output voltage regulation against input voltage regulation. When input voltage changes from V_{I1} to V_{I2} , and output voltage changes from V_{O1} to V_{O2} , the Reg-in is obtained in the following equation.

$$\text{Reg-in} = \frac{V_{O2} - V_{O1}}{V_{O1}} \times \frac{1}{V_{I2} - V_{I1}} \times 100 (\%/V)$$

Reg-L: Load regulation

Ratio of output voltage regulation against load current regulation. When load current (I_L) changes from I_{L1} to I_{L2} , Reg-L is obtained in the following equation.

$$\text{Reg-L} = \frac{V_{O1} - V_{O2}}{V_{O1}} \times 100 (\%)$$

I_B : Bias current

Circuit current (power dissipation current) required for stable operation of the device. If load current (I_L) is 0 and input current is I_1 , the I_B is expressed by the following equation. Current in reference voltage pin ④ is minimum and can be disregarded.

$$I_B = I_1 - I_2 = I_1 - \frac{V_{REF}}{R_1} \\ \approx I_1 - \frac{1.8V}{3.3k\Omega} \approx I_1 - 0.55mA$$

V_{NO} : Output noise voltage

Noise at output pin is measured with a noise meter using band pass filter of 20Hz~100kHz.

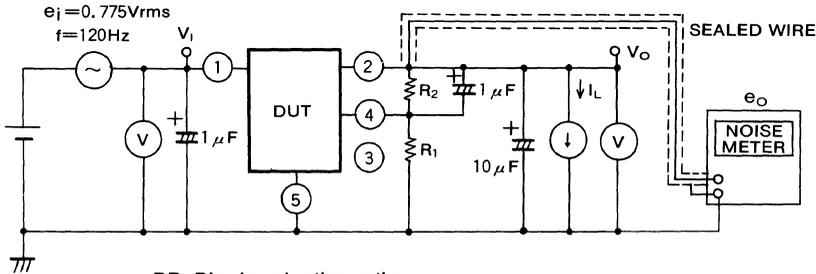
TCV_O: Temperature coefficient of output voltage

Output voltage regulation owing to ambient temperature change. When the device is placed in isothermal bath, and ambient temperature changes from T_{A1} to T_{A2} , and output voltage changes from V_{O1} to V_{O2} , TCV_O is expressed by the following equation.

$$\text{TCV}_O = \frac{V_{O2} - V_{O1}}{V_O (@T_A = 25^\circ C)} \times \frac{1}{T_{A1} - T_{A2}} \times 100 (\%/^\circ C)$$

TEST METHOD, APPLICATION EXAMPLE PRECAUTIONS FOR USE OF VOLTAGE REGULATOR

1-2. Test circuit 2



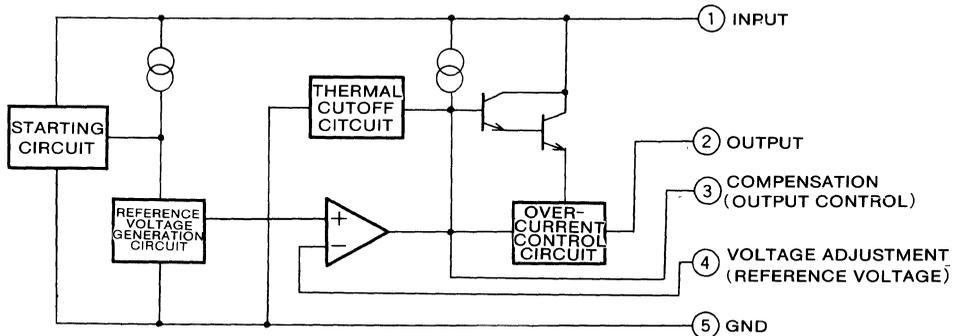
RR: Ripple rejection ratio

Ratio of output ripple (e_o) against input ripple (e_i) is expressed by the following equation.

$$RR = \log \frac{e_i}{e_o} \text{ (dB)} \quad \text{where, } V_1 = V_o + 5V$$

2. Application example

2-1. Block diagram (M5231TL)

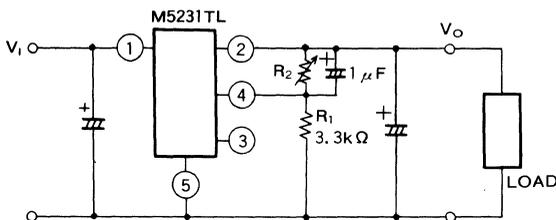


2-2. Input voltage range

Minimum input voltage required for stable operation and maximum input voltage (maximum rating) to prevent destruction of the device.

For the M5231TL, the minimum input voltage is 8V for generating internal reference voltage, and maximum input voltage is 70V.

2-3. Output voltage setting



Output voltage (V_o) can be set freely by externally connected resistors R_1 , R_2 from the following equation.

$$V_o = V_{REF} \times \left(1 + \frac{R_2}{R_1}\right)$$

The R_2 is obtained from the following equation.

$$R_2 = R_1 \times \left(\frac{V_o}{V_{REF}} - 1\right)$$

For the M5231TL, $V_{REF} = 1.8V$ typ.

$R_1 = 3.3k\Omega$ (recommended value)
output voltage setting range $V_o = 3V \sim 50V$

2-4. Input-output voltage difference

Input voltage should be higher than output voltage for stable operation of the device. The minimum voltage difference is shown by V_{i-o} .

If the input voltage contains ripple element, the input voltage must be determined considering the ripple voltage.

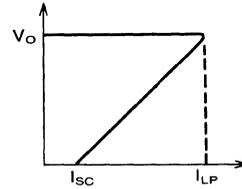
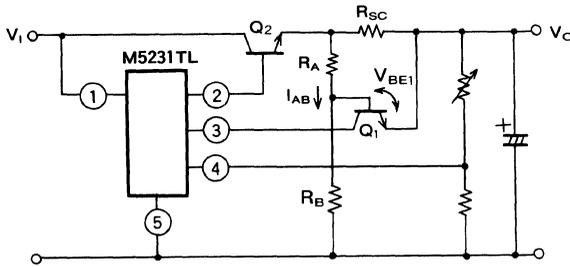
$$\text{Input voltage } V_i > V_o + (V_{i-o}) + \frac{V_{RR \text{ P-P}}}{2}$$

(V_o : Output voltage)

(V_{i-o} : Minimum input-output voltage difference)

$\frac{V_{RR \text{ P-P}}}{2}$: 1/2 of ripple voltage (peak to peak)

TEST METHOD, APPLICATION EXAMPLE PRECAUTIONS FOR USE OF VOLTAGE REGULATOR



$$= 0.417 + 1.68 = 2.09(\text{A})$$

(Designed value 2A)

The design procedure is shown in the following. ($V_i = 18\text{V}$, $V_o = 12\text{V}$, $I_{LP} = 2\text{A}$, $I_{SC} \approx 0.4\text{A}$)

Control circuit consists of Q_1 , R_A , R_B and R_{SC} .

To determine R_{SC} ,

when base-emitter voltage (V_{BE1}) of Q_1 is 0.6V,

$$R_{SC} > \frac{V_{BE1}}{I_{SC}} = \frac{0.6}{0.4} = 1.5 \rightarrow 1.8\Omega.$$

To determine R_A , R_B ,

$$R_{SC} \cdot I_{LP} - \left(\frac{V_o + R_{SC} \cdot I_{LP}}{R_A + R_B} \right) \cdot R_A = V_{BE1} \quad (1)$$

when () shows breeder current (I_{AB}) in resistors R_A , R_B and is 5mA,

$$R_A = \frac{R_{SC} \cdot I_{LP} - V_{BE1}}{I_{AB}} = \frac{R_{SC} \cdot I_{LP} - V_{BE1}}{5 \times 10^{-3}}$$

$$= 600 \rightarrow 680\Omega.$$

$$\text{From } \frac{V_o + R_{SC} \cdot I_{LP}}{R_A + R_B} = I_{AB}$$

$$R_B = \frac{V_o + R_{SC} \cdot I_{LP}}{I_{AB}} - R_A = \frac{12 + 1.8 \times 2}{5 \times 10^{-3}}$$

$$= 2440 \rightarrow 2.7\text{k}\Omega$$

With $R_{SC} = 1.5\Omega$, $R_A = 480\Omega$, $R_B = 5.1\text{k}\Omega$, I_{LP} and I_{SC} are calculated from equation 1 below.

$$I_{LP}(@I_{Lmax}) = \frac{V_{BE1}(R_A + R_B)}{R_{SC} \cdot R_B} + \frac{V_o \cdot R_A}{R_{SC} \cdot R_B}$$

$$= \frac{0.6(680 + 2.7 \times 10^3)}{1.8 \times 2.7 \times 10^3} + \frac{12 \times 680}{1.8 \times 2.7 \times 10^3}$$

$$I_{SC}(@V_o=0) = \frac{V_{BE1}(R_A + R_B)}{R_{SC} \cdot R_B} = 0.417$$

(Designed value 0.4A)

Power dissipation of each resistor

$$R_A: P_d = (I_{AB})^2 \times R_A = 0.017(\text{W})$$

$$R_B: P_d = (I_{AB})^2 \times R_B = 0.067(\text{W})$$

$$R_{SC}: P_d(\text{max}) = I_{LP}^2 \times R_{SC} = 7.86(\text{W})$$

(determined by load current)

To determine Q_1 ,

Maximum collector dissipation,

$$P_{C(\text{max})} = (V_{3-5} + I_{SC} \cdot R_{SC}) \times I_{O(\text{max})}$$

$$= 2.8\text{V} \times 60\text{mA} = 168(\text{mW})$$

$$V_{CEO} > V_{3-5} + I_{LP} + R_{SC} = 2.1 + 3.76 = 5.86(\text{V})$$

$$I_C > I_{O(\text{max})} = 60\text{mA}$$

To determine Q_2 ,

Maximum collector dissipation,

$$P_{C(\text{max})} = V_{CE} \cdot I_{LP}$$

$$= [V_i - (V_o + R_{SC} \cdot I_{LP})] \cdot I_{LP} = 4.67(\text{W})$$

In shorted load condition,

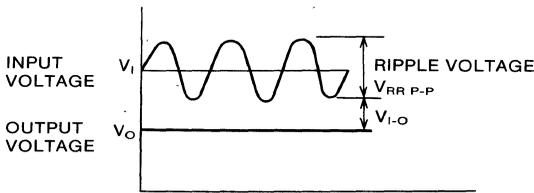
$$P_C = (V_i - I_{SC} \cdot R_{SC}) \cdot I_{SC} = 7.19(\text{W})$$

$$V_{CEO} > V_i = 18\text{V} \quad I_C > I_{LP} = 2.09\text{A}$$

If the output voltage of M5231TL $I_O = 10\text{mA}$,

$$h_{FE(\text{min})} > \frac{I_{LP}}{I_O} = \frac{2.09\text{A}}{10\text{mA}} = 209$$

TEST METHOD, APPLICATION EXAMPLE PRECAUTIONS FOR USE OF VOLTAGE REGULATOR



But the input voltage merely increases power dissipation if increased more than necessary, and should be set at the most suitable value.

2-5. Power dissipation

Power dissipation of the device in normal operating condition and in shorted output condition is shown below.

In normal operating condition,

$$P_d = V_i \times I_B + (V_i - V_o) \times I_L$$

In shorted output condition,

$$P_d = V_i \times (I_{SC} + I_B) \dots\dots\dots \text{Note 1.}$$

- (V_i : Input voltage)
- (I_B : Bias current)
- (V_o : Output voltage)
- (I_L : Load current)
- (I_{SC} : Load current in shorted output condition)

The maximum power dissipation P_{dMAX} should be

$$P_{dMAX} < \frac{T_{jMAX} - T_a}{R_{th}}$$

- (T_{jMAX} : Maximum junction temperature)
- (T_a : Ambient temperature)
- (R_{th} : Thermal resistance)

For the M5231TL, $V_{i-o} = 3V$, $I_{SC} \approx 50mA$, $T_{jMAX} = 125^\circ C$, $R_{th} = 333^\circ C/W$.

Note 1) The device contains an over-current control (protection) circuit to prevent large current when output is shorted. An application example to minimize the current is shown in section (2-9).

2-6. Output current range

I_L is the maximum rated output current of the device for itself. The M5231TL has the I_{MAX} of 30mA and can be used from 0 to 30mA. The output current over 30mA can be obtained, if required, by externally connected transistors at the value of h_{FE} times of the transistor. (Refer to application example.)

2-7. Phase compensation

If oscillation occurs owing to external wiring condition, a capacitor of 1000pF to 0.1 μF must be connected between pins ③ and ⑤.

2-8. Input/output capacitors

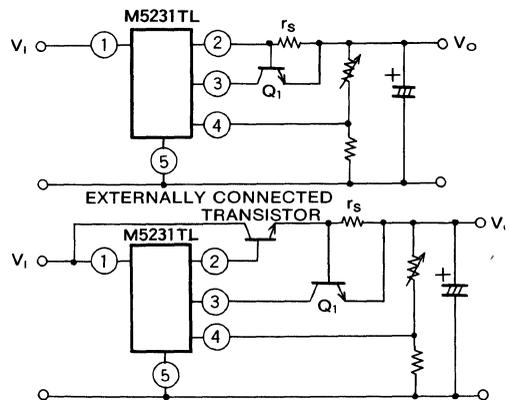
A capacitor of 0.33 μF or over must be connected on the input side for stable operation of the device. An electrolytic capacitor of some tens of μF or over and film capacitor of 0.1 μF must be connected on the output side to prevent oscillation and reduce output impedance in high-frequency range. In both cases, capacitors should be mounted as close to the IC as possible.

2-9. Output current control

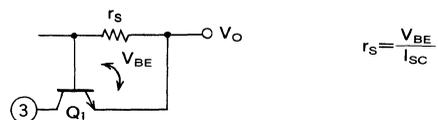
The device contains an output current control circuit to prevent accidental destruction due to output short circuit.

The M5231TL controls current over approximately 50~60mA. To minimize the control current value, the following method can be adopted. The over-current control circuit is recommended when externally connected transistors are used as shorted current is multiplied by h_{FE} times of transistors in shorted output condition.

2-9-1. Over-current control circuit (drooping type)

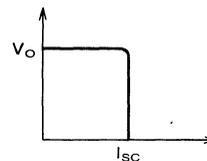


The control circuit consists of Q_1 and r_s . If the control current is I_{SC} , r_s is obtained from the equation below.



$$r_s = \frac{V_{BE}}{I_{SC}}$$

Care should be taken as V_{BE} is derated by $-2.3mV/^\circ C$.



2-9-2. Over-current control circuit (fold-back protection circuit)

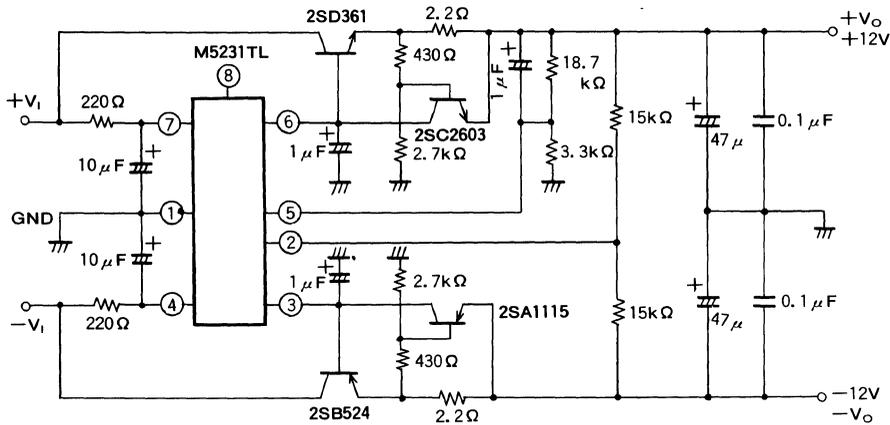
The fold-back protection circuit instantly reduces over-current generated by load short circuit as shown below and the name is derived from the fact that the curve in typical characteristic graph of output voltage V_o and load current I_L is folded back.

TEST METHOD, APPLICATION EXAMPLE PRECAUTIONS FOR USE OF VOLTAGE REGULATOR

2-9-3. Dual variable output voltage regulator M5231TL with fold-back protection circuit

+12V/1A power supply with a fold-back protection circuit

Specification $I_{SC}=0.31A$ $I_{LP}=1.18A$

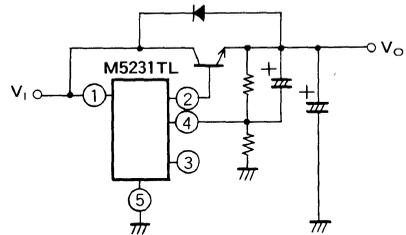
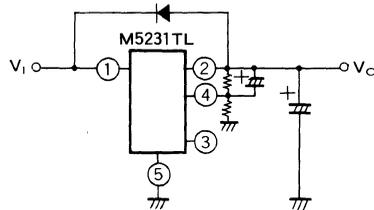


3. Precautions for use

- Application of supply voltage with the device mounted the wrong way or the pins not securely inserted may lead to deterioration of the characteristics or destruction of the device and should be avoided at all costs.

- Application of voltage to an output pin higher than the voltage in an input pin may destroy the element. If an input pin is shorted to GND under normal operating condition, the output pin voltage becomes higher than the input pin voltage and the electric load of an electrolytic capacitor connected to the output pin is reversed from the device or from externally connected transistors to the input side, causing destruction of the device. To prevent this, an ordinary silicon diode can be connected as shown below.

- Parasitic oscillation may be generated due to wiring condition. Proper capacitors should be connected between input pin and GND or output pin and GND. Capacitors with small capacitive change due to changes in ambient temperature should be used.



M5230L

VARIABLE OUTPUT VOLTAGE REGULATOR (DUAL TRACKING TYPE)

DESCRIPTION

The M5230L is a semiconductor integrated circuit which is designed for variable output voltage regulator of dual tracking type.

It is housed in an 8-pin SIP. The output voltage can be adjusted over a wide range from $\pm 3 \sim \pm 30V$ by adjusting the value of the voltage setting external resistors. By adjusting the resistance of the external balance setting resistors the positive/negative output voltage ratio can also be set freely. Again by attaching power transistors high current gains can be achieved, making the device suitable for use in the power supplies of a wide variety of equipment.

FEATURES

- High input voltage $V_I = \pm 35V$
- Wide range of output voltage $V_O = \pm 3 \sim \pm 30V$
- Low output noise voltage $V_{NO} = 12 \mu V_{rms}(typ.)$
- Built-in current limiting and thermal shutdown circuit
- The output voltage rise time constant of the coefficients can be adjusted by the value of the external capacitor.
- Capability of operation control by the external control signal (Pin ⑧).

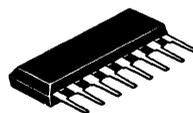
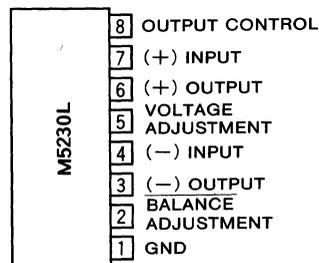
APPLICATION

Dual voltage power supplies for stereo preamplifiers, for the power supplies of other equipment, including operational amplifiers.

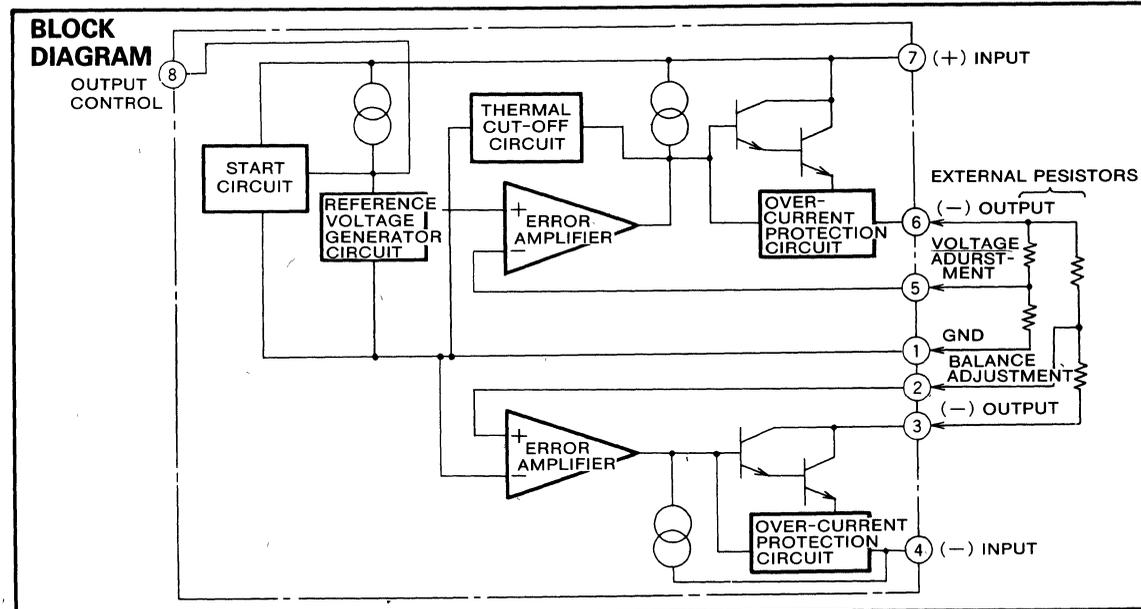
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range $\pm 8 \sim \pm 35V$
- Rated supply voltage $\pm 20V$

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic SIP



VARIABLE OUTPUT VOLTAGE REGULATOR(DUAL TRACKING TYPE)

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$)

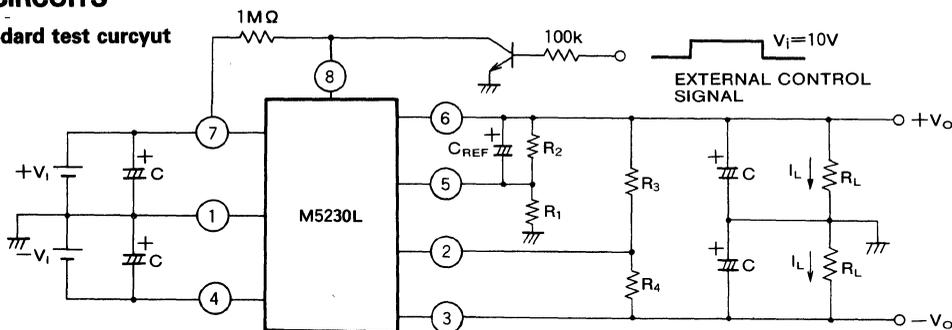
Symbol	Parameter	Ratings	Unit
V_i	Input voltage	± 35	V
I_L	Load current	± 30	mA
$V_i - V_o$	Input-output voltage difference	± 32	V
P_d	Power dissipation	800	mW
T_{opr}	Ambient temperature	$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature	$-55 \sim +125$	$^\circ\text{C}$

ELECTICAL CHARACTERISTICS (measurement circuit (a) is used with, $T_a=25^\circ\text{C}$, $V_i=\pm 20\text{V}$, $V_o=\pm 15\text{V}$, $I_L=10\text{mA}$, $C=10\mu\text{F}$, $C_{REF}=1\mu\text{F}$, $R_1=3.3\text{k}\Omega$)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_i	Input voltage		± 8		± 35	V
V_o	Output voltage	$R_2 \approx 1.5 \sim 55\text{k}\Omega$	± 3		± 30	V
V_{REF}	Reference voltage	(between pin ⑤ and pin ①)	(1.66)	1.8	(1.95)	V
$V_i - V_o$	Minimum input-output voltage difference			2.5	3	V
$\Delta V_o \pm$	Dual voltage tracking				1	%
Reg_{in}	Input regulation	$V_i = \pm 18 \sim \pm 30\text{V}$		0.02	0.1	%/V
Reg_L	Load regulation	$I_L = 0 \sim 20\text{mA}$		0.02	0.1	%
I_B	Bias current	$I_L = 0$ (disregarding the current in resistors R_1, R_2, R_3, R_4)		1.3	3.0	mA
TC_{V_o}	Temperature coefficient of output voltage	$T_a = 0 \sim 75^\circ\text{C}$, $V_o = \pm 3 \sim \pm 30\text{V}$		0.01		%/ $^\circ\text{C}$
RR	Ripple rejection	$f = 120\text{Hz}$ (measured with circuit (b))		68		dB
V_{NO}	Output noise voltage	$f = 20\text{Hz} \sim 100\text{kHz}$ (between the output terminal and ground)		12		μVrms
$V_{O(OFF)}$	Output cut-off voltage	$V_i = 10\text{V}$			± 0.1	V

TEST CIRCUITS

(a) Standard test circuit



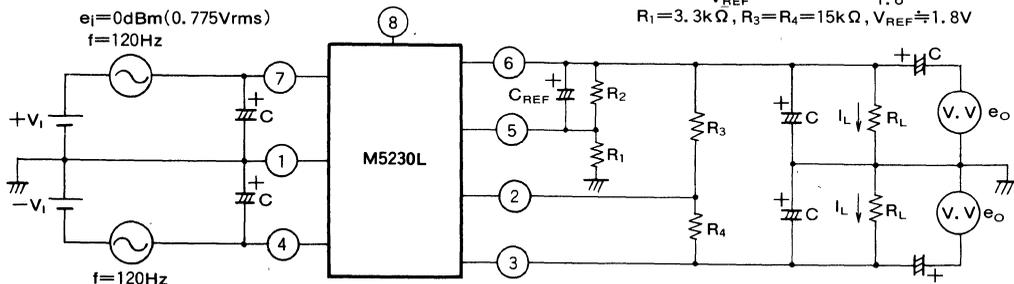
$$+V_o = V_{REF} \left(1 + \frac{R_2}{R_1}\right) \approx 1.8 \times \left(1 + \frac{R_2}{3.3}\right) \text{ (V)}$$

$$-V_o = +V_o \cdot \frac{R_4}{R_3} \text{ (V)}$$

$$R_2 = R_1 \left(\frac{+V_o}{V_{REF}} - 1\right) \approx 3.3 \times \left(\frac{+V_o}{1.8} - 1\right) \text{ (k}\Omega\text{)}$$

$$R_1 = 3.3\text{k}\Omega, R_3 = R_4 = 15\text{k}\Omega, V_{REF} \approx 1.8\text{V}$$

(b) Ripple rejection test circuit



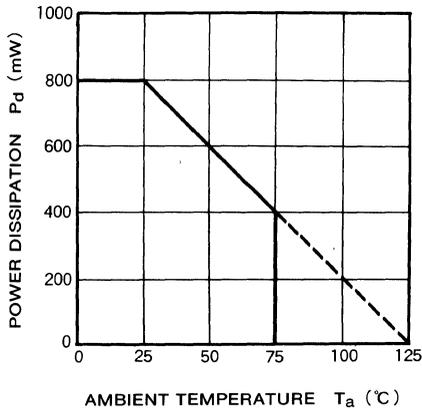
$e_i = 0\text{dBm}$ (0.775Vrms)
 $f = 120\text{Hz}$

$$RR = 20 \log_{10} \left(\frac{e_i}{e_o}\right) \text{ (dB)}$$

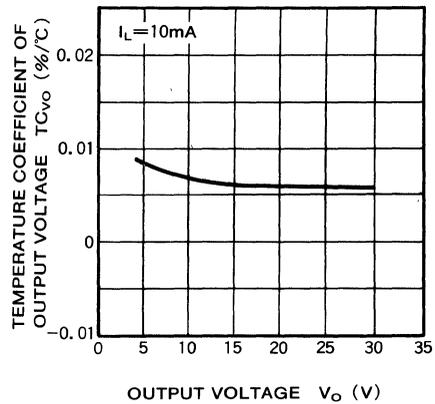
VARIABLE OUTPUT VOLTAGE REGULATOR (DUAL TRACKING TYPE)

TYPICAL CHARACTERISTICS

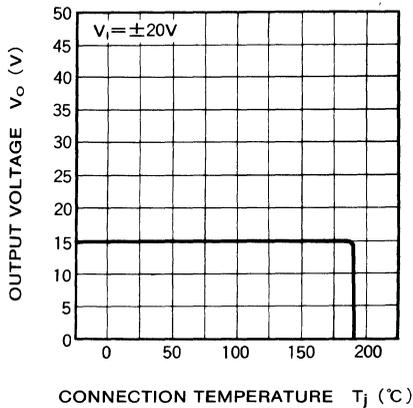
THERMAL DERATING (MAXIMUM RATING)



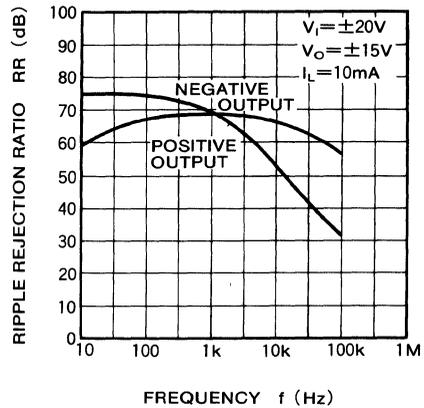
TEMPERATURE COEFFICIENT OF OUTPUT VOLTAGE VS. OUTPUT VOLTAGE CHARACTERISTICS



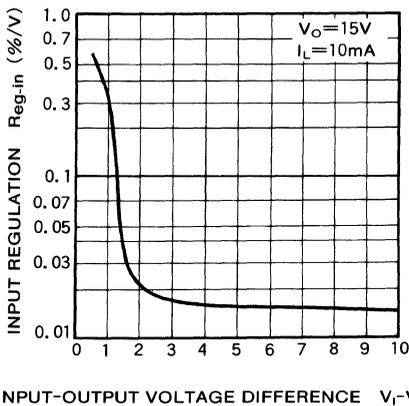
THERMAL CUTOFF



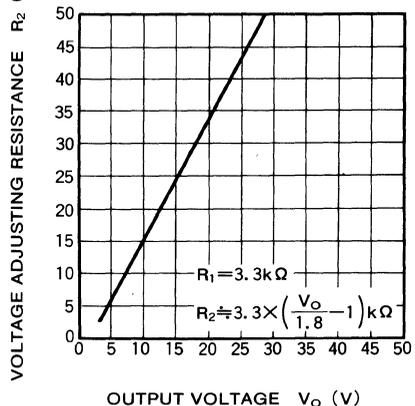
RIPPLE REJECTION



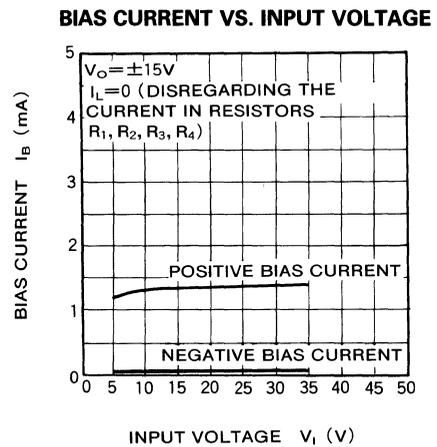
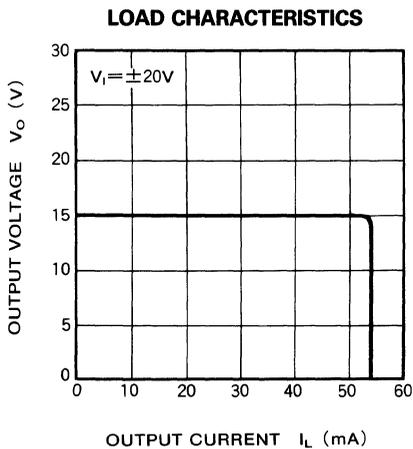
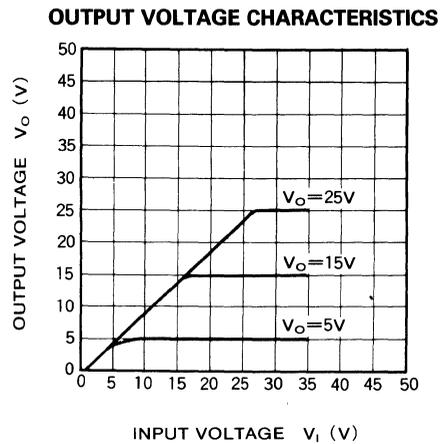
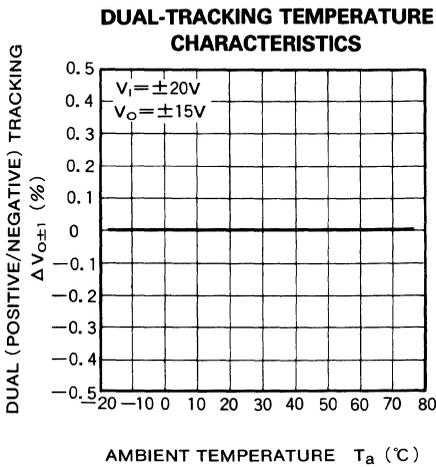
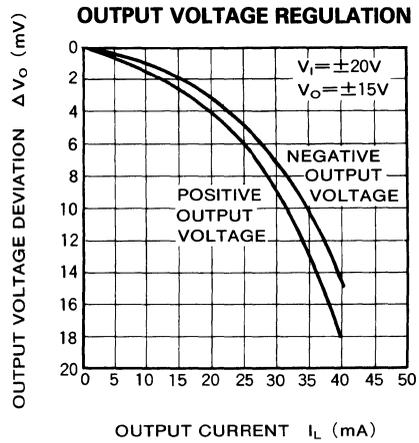
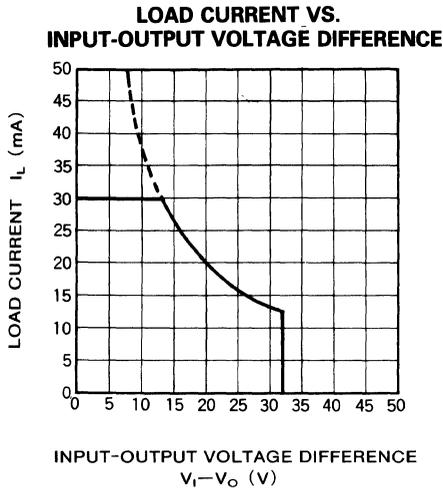
INPUT REGULATION VS. INPUT-OUTPUT VOLTAGE DIFFERENCE



VOLTAGE ADJUSTING RESISTANCE VS. OUTPUT VOLTAGE

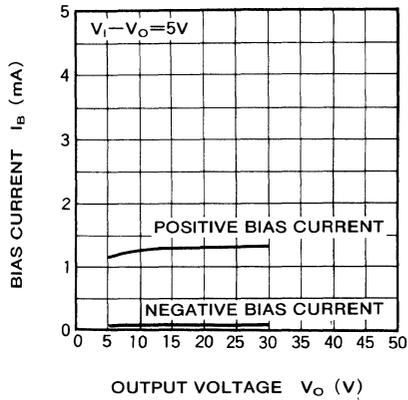


VARIABLE OUTPUT VOLTAGE REGULATOR (DUAL TRACKING TYPE)

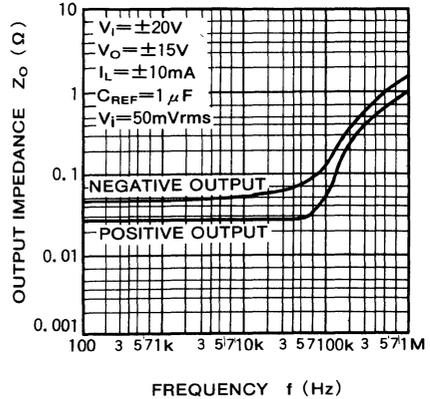


VARIABLE OUTPUT VOLTAGE REGULATOR (DUAL TRACKING TYPE)

BIAS CURRENT VS. OUTPUT VOLTAGE

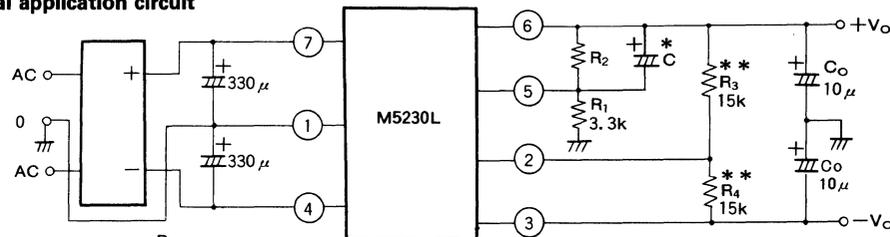


OUTPUT IMPEDANCE VS. FREQUENCY



APPLICATION EXAMPLES

(1) Typical application circuit



$$+V_o \approx 1.8 \times \left(1 + \frac{R_2}{3.3}\right) \text{ (V)}$$

$$R_2 \approx 3.3 \times \left(\frac{V_o}{1.8} - 1\right) \text{ (k}\Omega\text{)} \quad -V_o = (+V_o) \cdot \frac{R_4}{R_3} \text{ (V)}$$

Note When the input power supply lines become long, a 0.1 μF capacitor should be connected between input power supply pins ⑦ and ④ and ground.

* C

Connection of this capacitor gives the following characteristics.

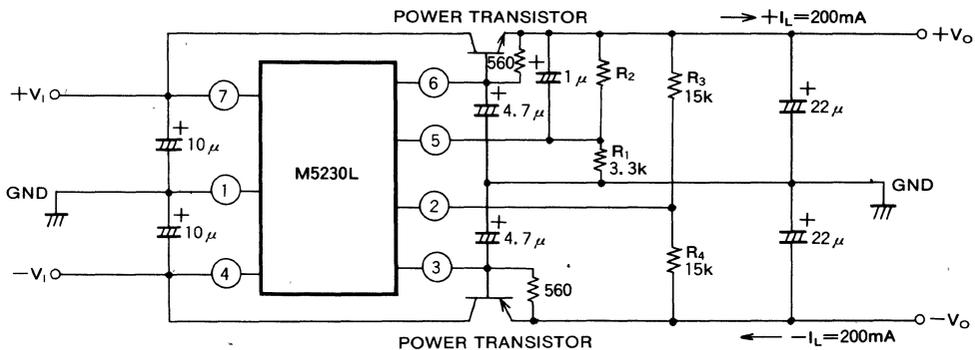
- 1) The rise time constant of the output voltage can be adjusted (slowed) (See Fig. 1)
- 2) The ripple rejection ratio is improved.
- 3) Noise output voltage is reduced.

** R₃, R₄

By changing the ratio of these two resistances the positive/negative voltage ratio can also be set freely. (See Fig. 2)

Unit Resistance Ω
 Capacitance F

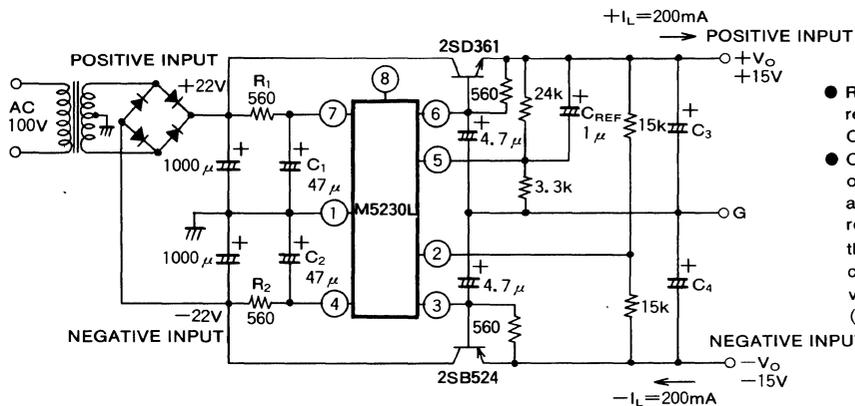
(2) Typical application circuit with power transistors connected



Unit Resistance : Ω
 Capacitance : F

VARIABLE OUTPUT VOLTAGE REGULATOR(DUAL TRACKING TYPE)

(3) High ripple rejection circuit (80dB)

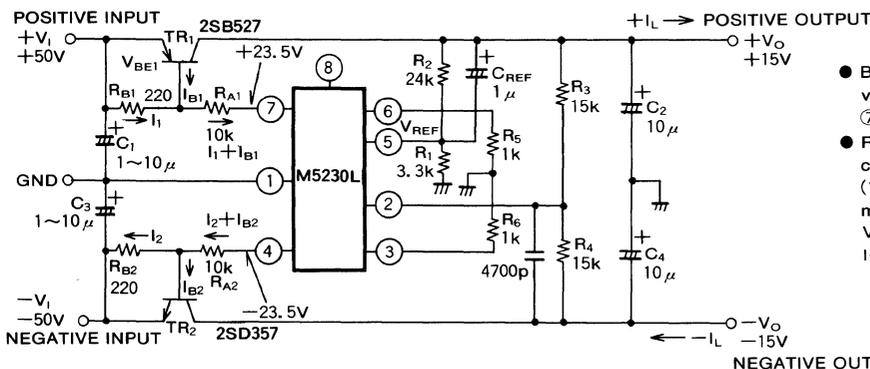


- Ripple regained by the input resistance R_1 , (R_2), and capacitor C_1 , (C_2).
- C_{REF} can reduce noise to the 1/10 of that of 3 Terminal regulator IC, and also can improve the ripple rejection. In addition, by increasing the capacitance of the C_{REF} , the constant at set-up of the output voltage V_O can be adjusted (Soft set-up enable).

Unit Resistance : Ω
 Capacitance : F

(4) High input voltage ($V_I = \pm 50V$)

RIPLLE REJECTION 65dB

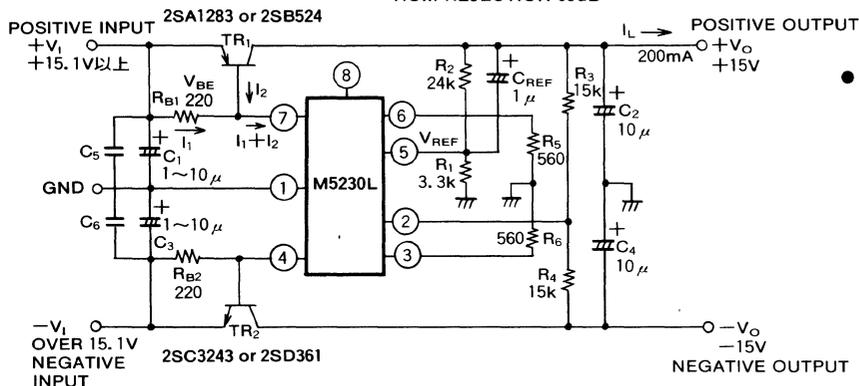


- By the resistance R_{A1} , (R_{A2}), the voltage that will be supplied to pins ⑦, ④ can be lower
- Resistances R_5 and R_6 are for load current limit. Keep the I/O voltage (V_{I0}) difference between ⑦-⑥ more than 6V
 $V_I - V_{BE} - (I_1 + I_{B1} - I_B) R_5 > 3V$
 $I_1 = V_{BE} / R_{B1}$, $I_{B1} = I_L / h_{FE}$

Unit Resistance : Ω
 Capacitance : F

(5) Low dropout regulator circuit ($V_{IO} = 100mV$)

HUM REJECTION 65dB



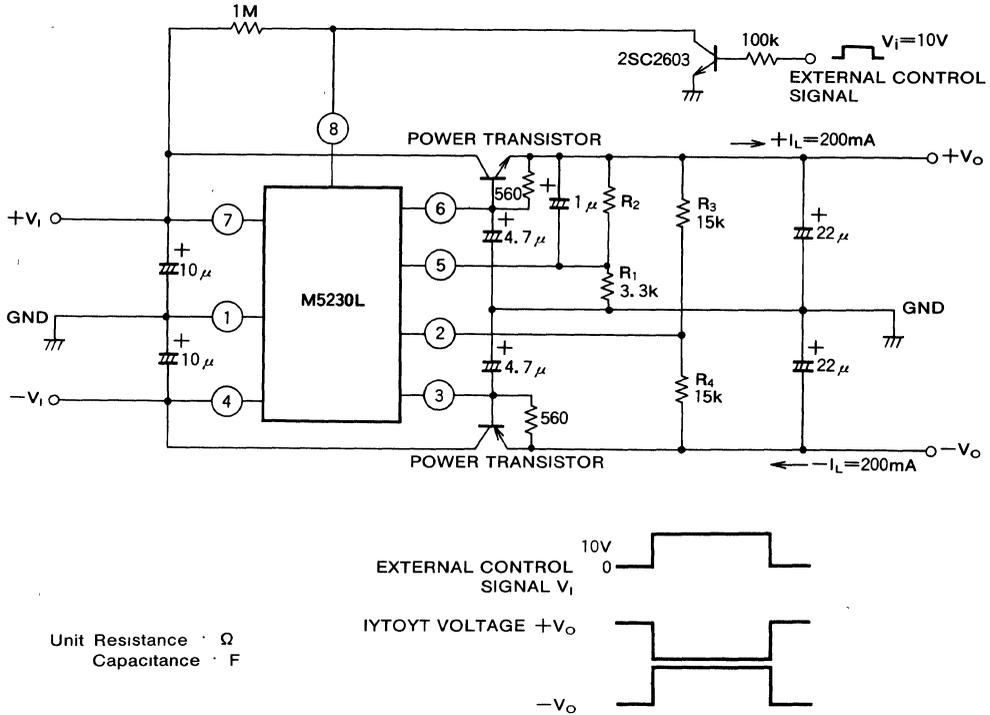
- Stable operations are expected even if the input-output voltage differences are quite low as 0.1V. The heat sink of power TR can become small in size owing to the low dissipation

Unit Resistance : Ω
 Capacitance : F

(Note) The load current can be over 1A by connecting the external power TR

VARIABLE OUTPUT VOLTAGE REGULATOR (DUAL TRACKING TYPE)

(6) ON/OFF control of output voltage circuit



EXAMPLES OF THE CHARACTERISTICS ACHIEVED

Fig. 1 OUPUT VOLTAGE CHARACTERISTICS FOR EXTERNAL CAPACITORS (*C)

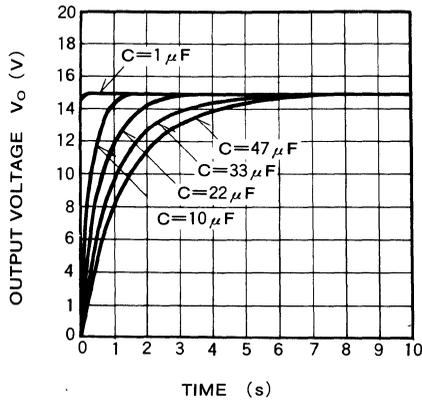
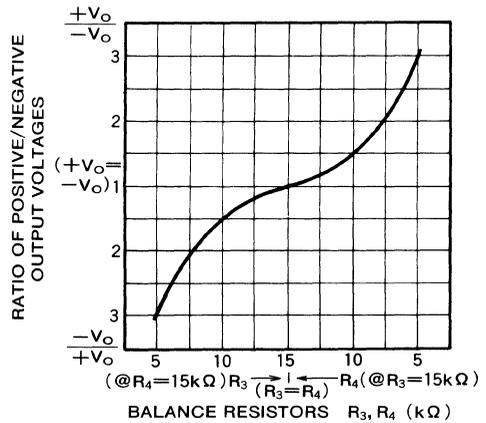


Fig. 2 OUTPUT VOLTAGE RATIO VS. BALNCE VOLTAGE ADJUSTING RESISTANCE CHARACTERISTICS



M5231TL

VARIABLE OUTPUT VOLTAGE REGULATOR

DESCRIPTION

The M5231TL is a semiconductor integrated circuit which is designed for variable output voltage regulator and is housed in a small 5-pin SIP.

The input range 8~70V, and the output voltage range 3~50V can be optionally adjusted by the external resistors. In addition, by attaching power transistors, high current gains can be achieved, making the device suitable for used in the power supplies of a wide variety of equipment.

FEATURES

- High input voltage $V_I=70V$
- Wide range of output voltages $V_O=3V\sim 50V$
- Low output noise voltage $V_{NO}=6\mu V_{rms}(typ.)$
- Built-in current limiting and thermal shutdown circuits
- Capability of adjusting the output voltage rise time constant of the coefficients by the external capacitor
- Capability of the operating control by the external signal

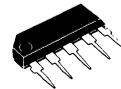
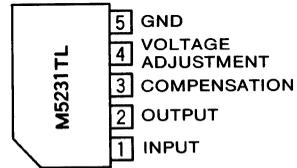
APPLICATION

Audio, VTR equipment, and a variety of electronic musical instruments.

RECOMMENDED OPERATING CONDITIONS

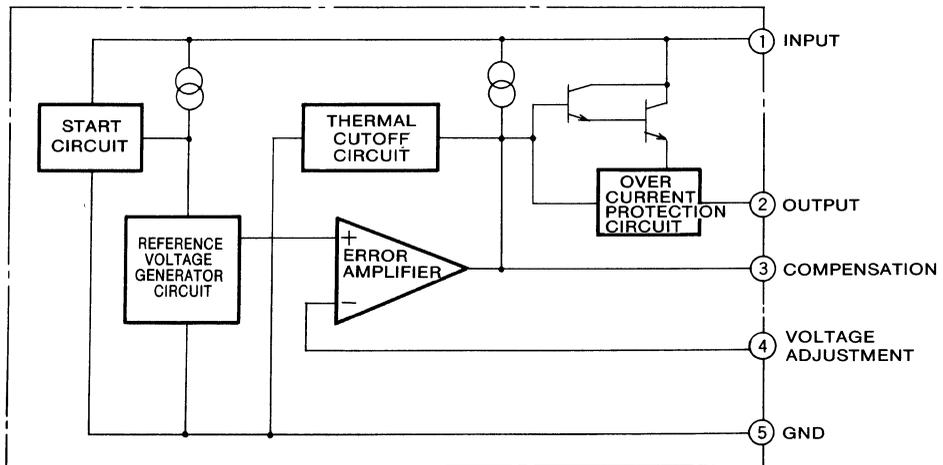
- Supply voltage range 8~70V
- Rated supply voltage 35V

PIN CONFIGURATION (TOP VIEW)



5-pin molded plastic SIP

BLOCK DIAGRAM



VARIABLE OUTPUT VOLTAGE REGULATOR

ABSOLUTE MAXIMUM RATINGS (T_a=25°C)

Symbol	Parameter	Ratings	Unit
V _I	Input voltage	70	V
I _L	Load current	30	mA
V _I -V _O	Input-output voltage difference	67	V
P _d	Power dissipation	300(L)/450(TL)	mW
T _{opr}	Operating temperature	-20~+75	°C
T _{stg}	Storage temperature	-55~+125	°C

ELECTRICAL CHARACTERISTICS (measurement circuit (a) is used with, T_a=25°C, V_I=40V, V_O=35V, I_L=10mA, C=10μF, C_{REF}=1μF, R₁=3.3kΩ, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _I	Input voltage		8		70	V
V _O	Output voltage	R ₂ ≐1.5~88kΩ	3		50	V
V _I -V _O	Minimum input-output voltage differential			2.5	3.0	V
V _{REF}	Reference voltage	(between pin ④ and pin ⑤)	(1.66)	1.8	(1.90)	V
R _{reg-in}	Input regulation	V _I =38~60V		0.04	0.1	%/V
R _{ge-L}	Load regulation	I _L =0~20mA		0.03	0.1	%
I _B	Bias current	I _L =0 (disregarding the current in resistors R ₁ , R ₂)		1.2	2.5	mA
TC _{VO}	Temperature coefficient of output voltage	T _a =0~75°C, V _O =3~50V		0.02		%/°C
RR	Ripple rejection	f=120Hz(measured with circuit (b))		62		dB
V _{NO}	Output noise voltage	f=20Hz~100kHz		6		μVrms

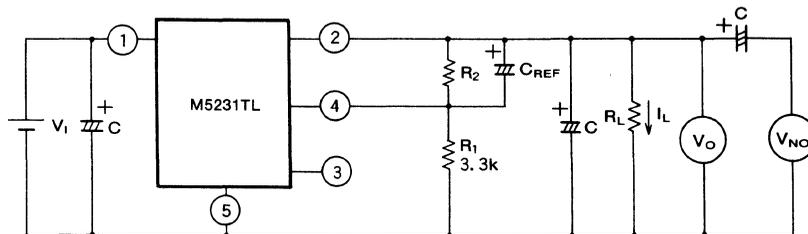
TEST CIRCUITS

(a) Standard test circuit

$$V_O = V_{REF} \left(1 + \frac{R_2}{R_1}\right) \doteq 1.8 \times \left(1 + \frac{R_2}{3.3}\right) \text{ (V)}$$

$$R_2 = R_1 \left(\frac{V_O}{V_{REF}} - 1\right) \doteq 3.3 \times \left(\frac{V_O}{1.8} - 1\right) \text{ (k}\Omega\text{)}$$

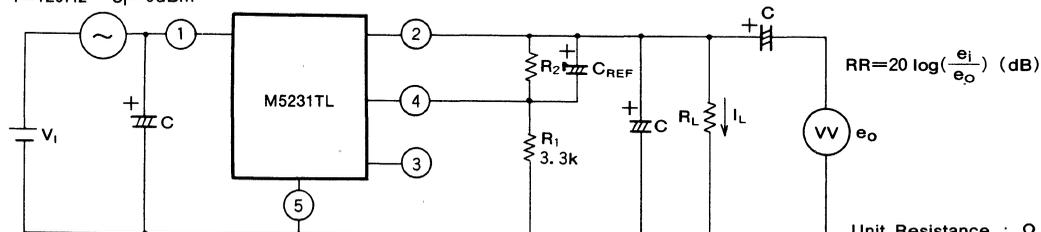
$$R_1 = 3.3 \text{ k}\Omega, V_{REF} \doteq 1.8 \text{ V}$$



Unit Resistance : Ω
 Capacitance : F

(b) Ripple rejection test circuit

f=120Hz e_i=0dBm

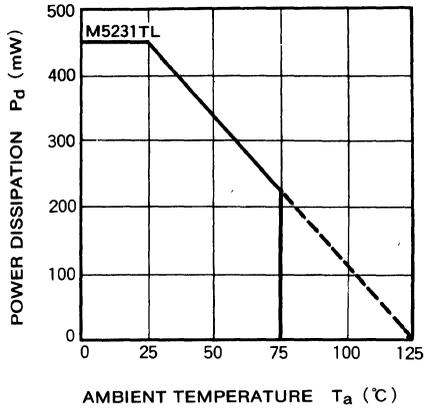


Unit Resistance : Ω
 Capacitance : F

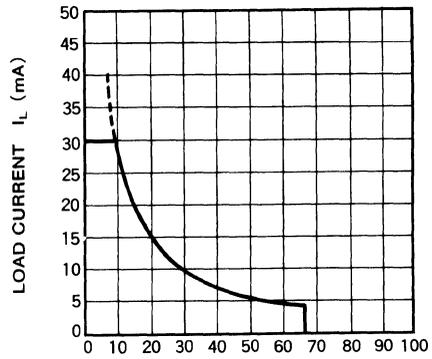
VARIABLE OUTPUT VOLTAGE REGULATOR

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

THERMAL DERATING (MAXIMUM RATINGS)

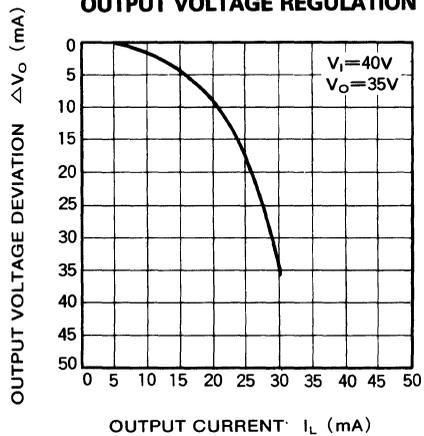


LOAD CURRENT VS. INPUT-OUTPUT VOLTAGE DIFFERENCE



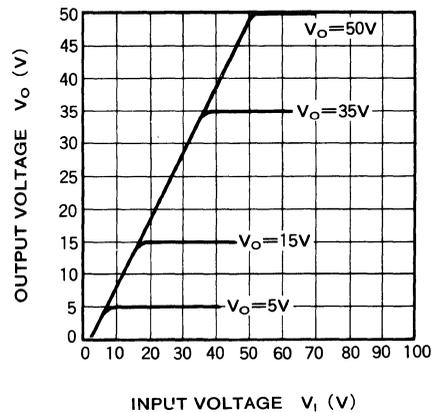
INPUT-OUTPUT VOLTAGE DIFFERENCE $V_i - V_o$ (V)

OUTPUT VOLTAGE REGULATION



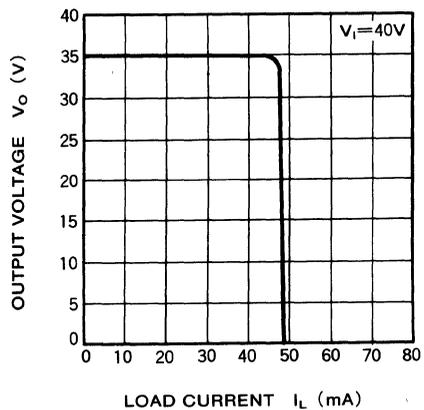
OUTPUT CURRENT I_L (mA)

OUTPUT VOLTAGE CHARACTERISTICS



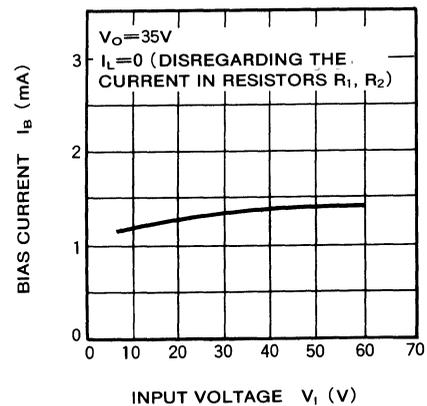
INPUT VOLTAGE V_i (V)

LOAD CHARACTERISTICS



LOAD CURRENT I_L (mA)

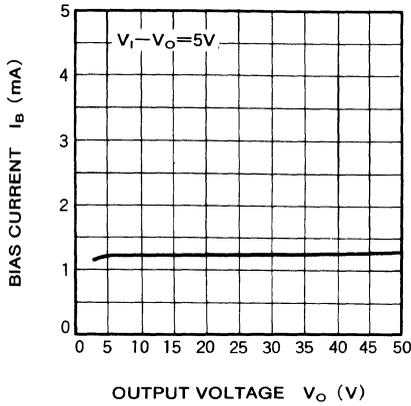
BIAS CURRENT VS. INPUT VOLTAGE



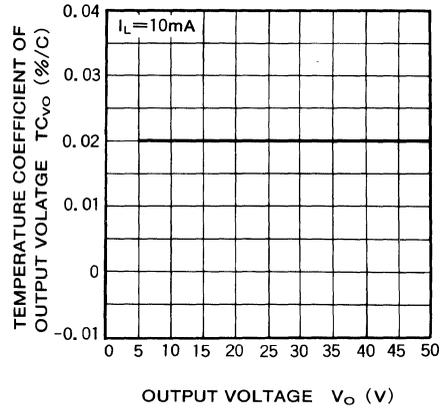
INPUT VOLTAGE V_i (V)

VARIABLE OUTPUT VOLTAGE REGULATOR

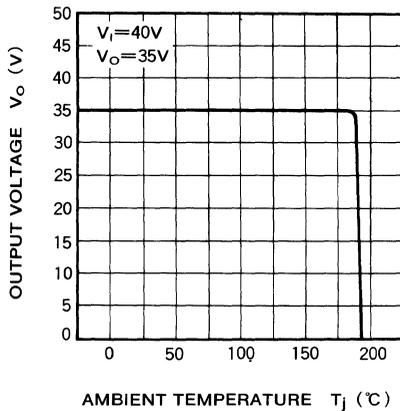
BIAS CURRENT VS. OUTPUT VOLTAGE



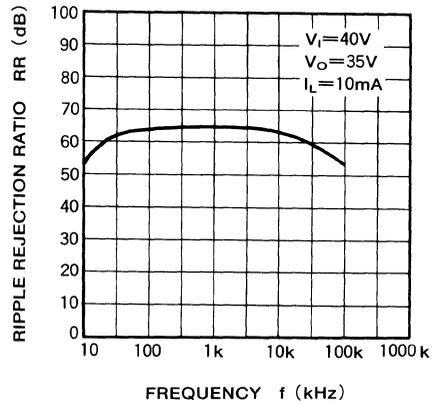
TEMPERATURE COEFFICIENT OF OUTPUT VOLTAGE VS. OUTPUT VOLTAGE CHARACTERISTICS



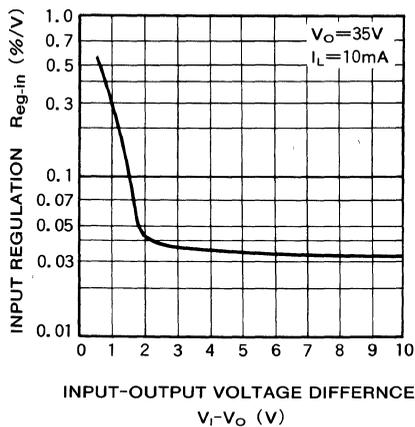
THERMAL CUTOFF



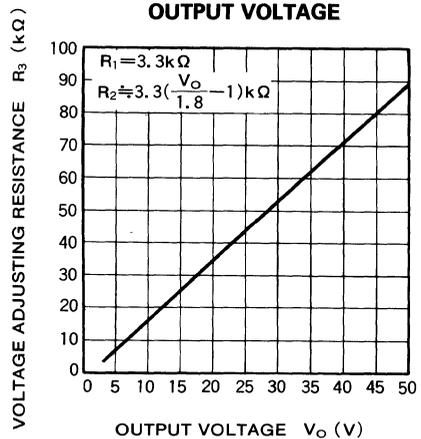
RIPPLE REJECTION



INPUT REGULATION VS. INPUT-OUTPUT VOLTAGE DIFFERENCE



VOLTAGE ADJUSTING RESISTANCE VS. OUTPUT VOLTAGE



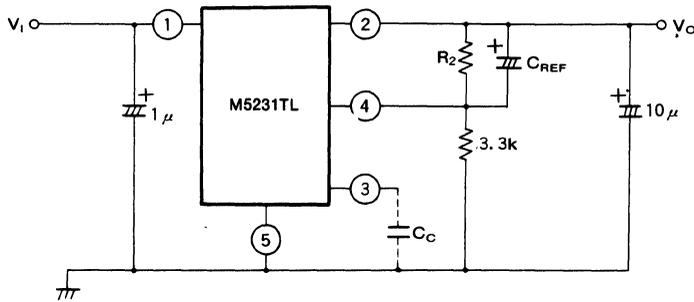
VARIABLE OUTPUT VOLTAGE REGULATOR

APPLICATION EXAMPLES

(1) Standard application example

$$V_o \approx 1.8 \times \left(1 + \frac{R_2}{3.3}\right) \text{ (V)}$$

$$R_2 \approx 3.3 \times \left(\frac{V_o}{1.8} - 1\right) \text{ (k}\Omega\text{)}$$



C_{REF}

Connection of this capacitor gives the following characteristics.

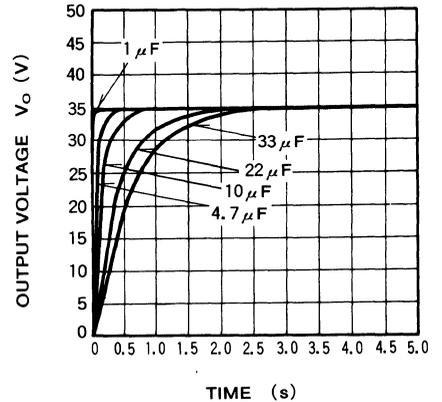
- 1) The rise time of the output voltage can be adjusted (slowed).
- 2) The ripple rejection ratio is improved.
- 3) Output noise voltage is reduced down to 1/10 of three terminal regulator IC.

C_C

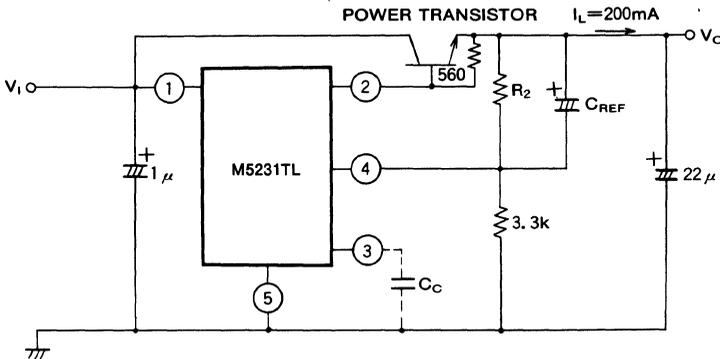
Compensation capability by connection of a capacitor.

EXAMPLE 1 OF CHARACTERISTIC ACHIEVED

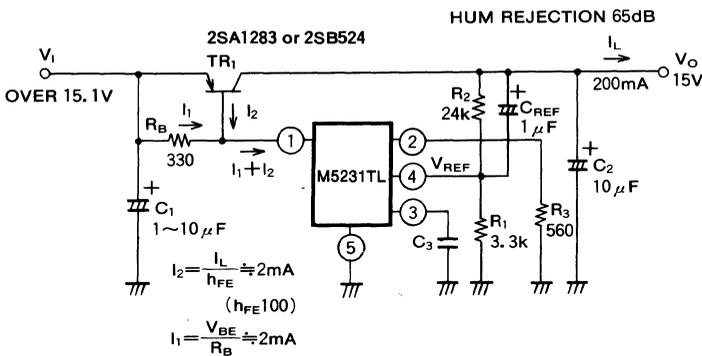
OUTPUT VOLTAGE CHARACTERISTICS FOR EXTERNAL CAPACITORS (C_{REF})



(2) Current boost circuit with NPN external power TR



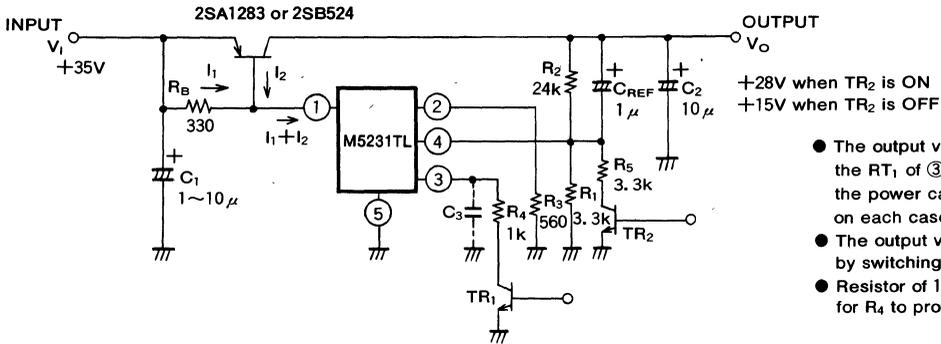
(3) Low dropout regulator circuits (V₁₀ = 100mV)



- Stable operation is expected even if the input-output voltage differences are as low as 0.1V. The heat sink of power TR can become small in size owing to low dissipation. The R₃ is a load current limit resistor and the input-output voltage differences between ① and ② pins must be over 3V. $V_1 - V_{RE} - (I_1 + I_2 - I_B)R_3 > 3\text{V}$

VARIABLE OUTPUT VOLTAGE REGULATOR

(4) Output voltage ON/OFF controller, Step UP/DOWN controller



- The output voltage V_o IS 0V when the TR_1 of ③ pin is ON. Therefore, the power can be on/off depending on each case.
- The output voltage is changeable by switching the TR_2 .
- Resistor of $1k\Omega$ must be connected for R_4 to protect the integrated circuit

(Note) The load current can be over 1A by connection the external power TR.

M5235L

3V,3-TERMINAL VARIABLE VOLTAGE OUTPUT REGULATOR (FOR DRIVER)

DESCRIPTION

The M5235L is a semiconductor integrated circuit designed for 3V system output voltage regulator.

A high-performance low-voltage, 3-terminal output voltage regulator with small input-output voltage differences can be made in combination with externally connected PNP transistors. It is housed in a small 3-pin package, including a reference voltage circuit, error amplifier, and driver, and the output voltage can be set freely by externally connected resistors, and a small, compact power supply circuit can be achieved, making the device suitable for use in small electronic equipment, such as headphone stereo, VTR camera, and radio cassette recorder.

FEATURES

- Operates at low input voltage $V_{IN(min)}=1.3V$
- Capable of operating with low input-output voltage for driver by externally connected power transistors
 $[V_{CE(sat)} \text{ state of } T_R]$
 $V_{I-O(min)} \cong 0.1V$
- Output voltage can be set freely by externally connected resistors
- A compact device is possible due to small package
- Capable of taping (automated insertion) and lead forming

APPLICATION

For two dry-battery (3V) type headphone stereo (especially to stabilize recording bias circuit), portable VTR, small radio cassette recorder, radio-controlled and electronic toys, or electronic games, and for power supplies of other general battery-driven electronic equipment.

RECOMMENDED OPERATING CONDITIONS

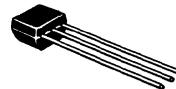
- Supply voltage range $V_{IN}=1.5V \sim 6V$
- Rated supply voltage $V_O=1.0V \sim 5.5V$

PIN CONFIGURATION



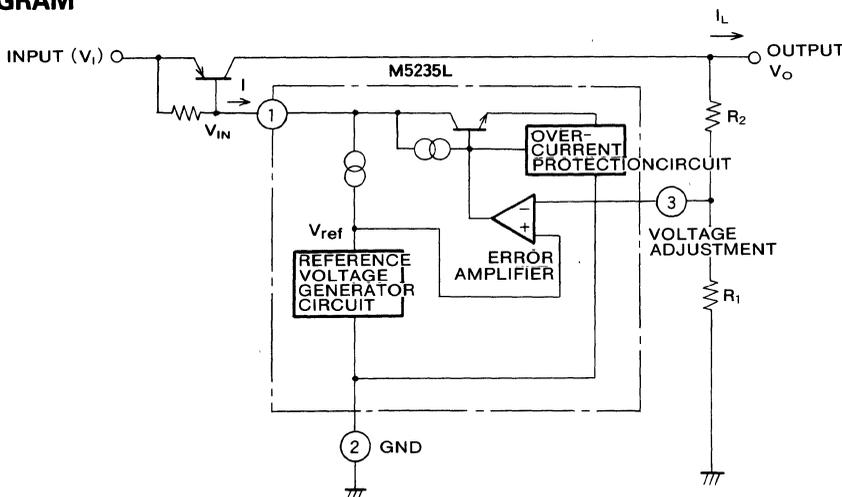
PIN CONNECTION

- ① INPUT
- ② GND
- ③ VOLTAGE ADJUSTMENT



Small 3-pin package

BLOCK DIAGRAM



3V,3-TERMINAL VARIABLE VOLTAGE OUTPUT REGULATOR (FOR DRIVER)

ABSOLUTE MAXIMUM RATINGS (T_a=25°C)

Symbol	Parameter	Ratings	Unit
V _{IN}	Input voltage	7(10) ^{Note 1}	V
I _D	Drive current	20	mA
V _I -V _O	Input-output voltage difference	6	V
P _d	Power dissipation	300	mW
T _{opr}	Operating temperature	-20~+75	°C
T _{stg}	Storage temperature	-55~+125	°C

Note 1. Refer to APPLICATION EXAMPLES

ELECTRICAL CHARACTERISTICS (measurement circuit (a) is used with, T_a=25°C, V_I=4.5V, V_O=3V, I_L=50mA, C=10μF, C_{REF}=1μF, R₁=5.1kΩ, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{IN}	Input voltage	(between pin ① and pin ②)	1.3		7	V
V _O	Output voltage	R ₂ ≅1.35kΩ~36.8kΩ	1.0		6.5	V
V _I -V _O	Minimum input-output voltage difference			0.1	0.5	V
V _{REF}	Reference voltage	(between pin ③ and pin ②)	0.75	0.79	0.83	V
R _{reg-in}	Input regulation	V _I =4~5V		0.3	0.8	%/V
R _{reg-L}	Load regulation	I _L =10~50mA		0.03	0.2	%
I _B	Bias current	I _L =0(disregarding the current in resistors R ₁ , R ₂)		1.6	3.0	mA
TC _{VO}	Temperature coefficient of output voltage	T _a =0~75°C		-0.1		%/°C
RR	Ripple rejection ratio	f=120Hz(measured with circuit (b))		50		dB
V _{NO}	Output noise voltage	f=20Hz~100kHz		24		μVrms

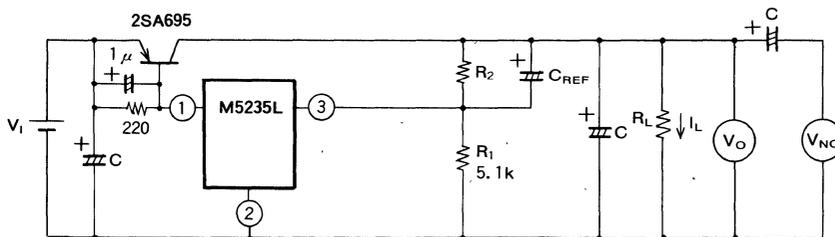
TEST CIRCUITS

(a) Standard test circuit

$$V_O = V_{REF} \left(1 + \frac{R_2}{R_1}\right) \cong 0.79 \times \left(1 + \frac{R_2}{5.1}\right) \text{ (V)}$$

$$R_2 = R_1 \left(\frac{V_O}{V_{REF}} - 1\right) \cong 5.1 \times \left(\frac{V_O}{0.79} - 1\right) \text{ (k}\Omega\text{)}$$

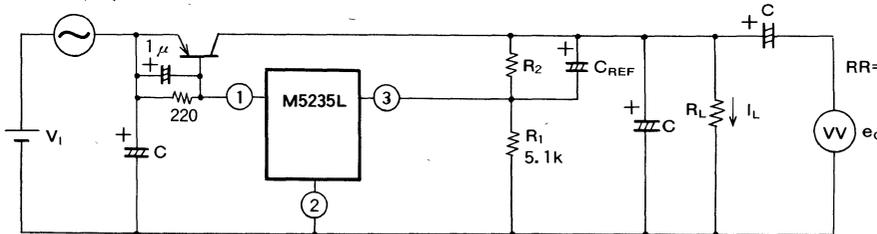
(R₁=5.1kΩ, V_{REF}≅0.79V)



Unit Resistance : Ω
 Capacitance : F

(d) Ripple rejection test circuit

f=120Hz, e_i=0.1Vrms



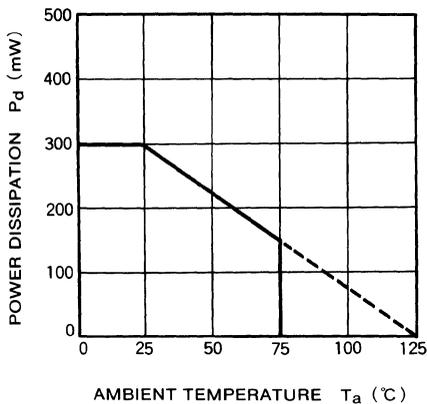
$$RR = 20 \log \left(\frac{e_i}{e_o}\right) \text{ (dB)}$$

Unit Resistance : Ω
 Capacitance : F

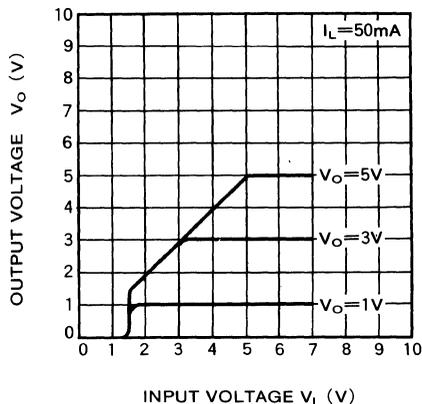
3V,3-TERMINAL VARIABLE VOLTAGE OUTPUT REGULATOR (FOR DRIVER)

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

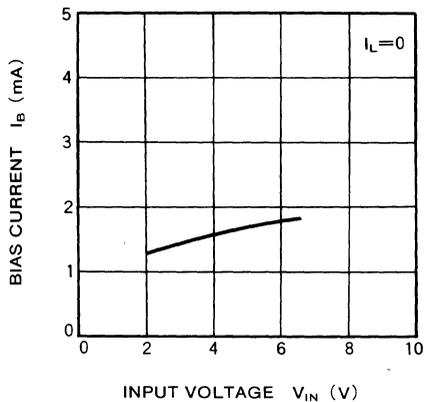
THERMAL DERATING (MAXIMUM RATINGS)



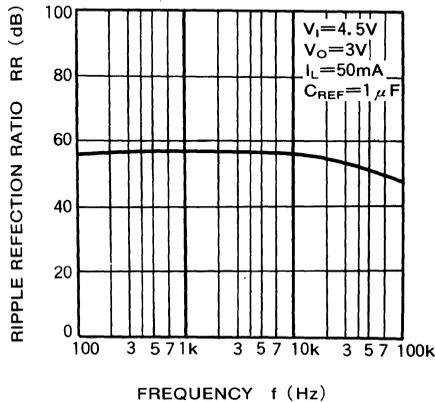
OUTPUT VOLTAGE CHARACTERISTICS



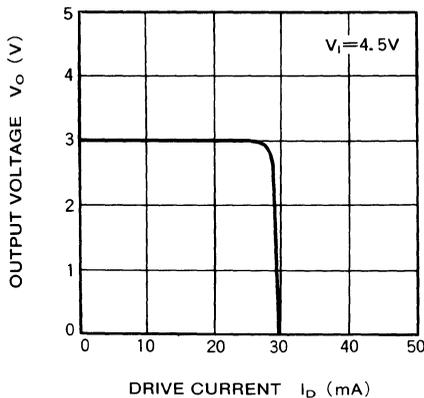
BIAS CURRENT VS. INPUT VOLTAGE



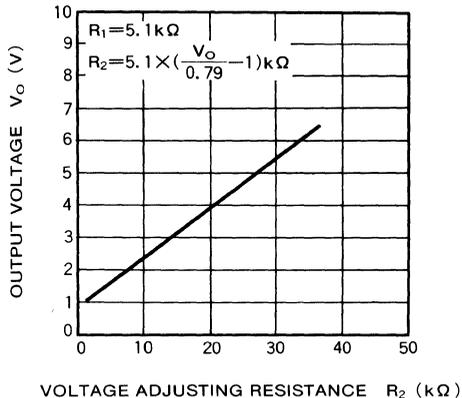
RIPPLE REJECTION



OUTPUT VOLTAGE VS. DRIVE CURRENT



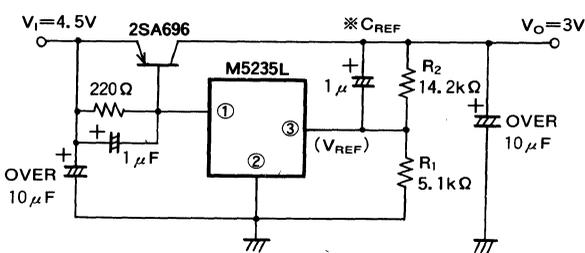
OUTPUT VOLTAGE VS. VOLTAGE ADJUSTING RESISTANCE



3V,3-TERMINAL VARIABLE VOLTAGE OUTPUT REGULATOR (FOR DRIVER)

APPLICATION EXAMPLES

1. Standard application circuit



$$V_O = V_{REF} \times \left(1 + \frac{R_2}{R_1}\right) V$$

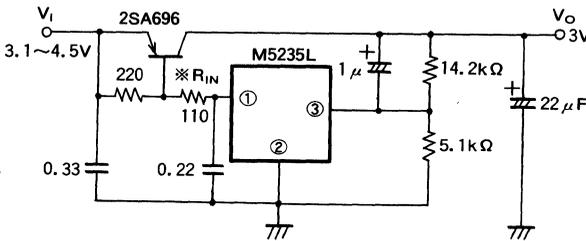
$$V_{REF} = 0.79V (\text{typ.})$$

※C_{REF}

Connection of this capacitor improves ripple rejection ratio, improves output noise voltage and adjusts the rise time constants of the output voltage. (Capacitor over 1000pF must be connected.)

Note. Capacitors displaying small capacity change with temperature should be used.

2. Control circuit for maximum drive current (I_p)



When the input voltage (V_i) is lower than the set output voltage (V_o) or when the output pin is shorted, drive current of approximately 20mA TO 35mA runs in Pin ① of the integrated circuit.

In this case, the current does no harm to the device, but the current can be limited by connecting a resistor R_{IN} if required due to input supply voltage (dry battery). (Fig. 1 shows relationship between control current vs. input voltage. In the application example, V_i controls the drive current approximately down to 10mA, when it reaches 3V.

Note1. Resistors ranging from 180Ω to 200Ω should be inserted between the emitter and base of externally connected PNP transistors.
Resistors from 100Ω to 120Ω should be used when the input voltage is 7V to 10V.

Fig.1 MAXIMUM DRIVE CURRENT CONTROL CHARACTERISTICS (I_{DM})

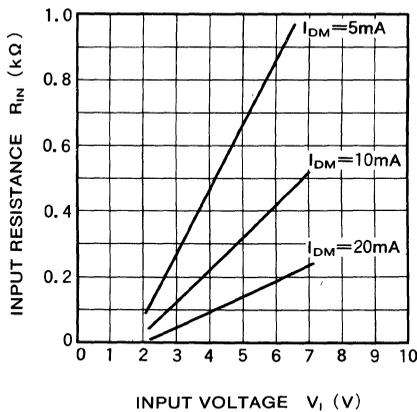
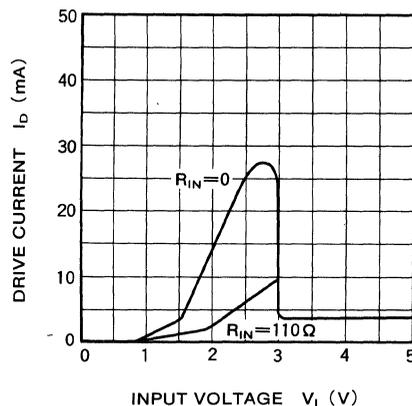
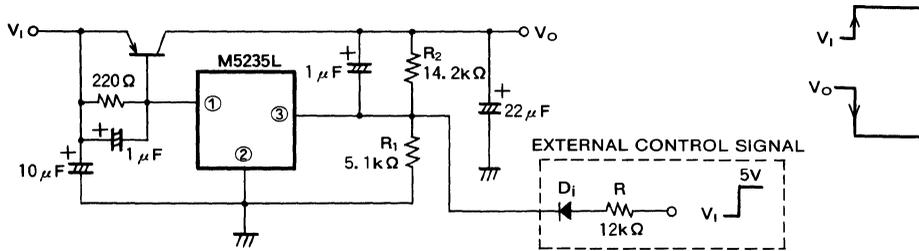


Fig.2 I_D - V_{IN} CHARACTERISTICS IN APPLICATION EXAMPLE 2



3V, 3-TERMINAL VARIABLE VOLTAGE OUTPUT REGULATOR (FOR DRIVER)

3. ON/OFF control of output voltage circuit



Resistor R in the control circuit is determined by the following equation.

$$R \doteq \frac{V_i - V_F - V_{REF'}}{V_{REF'} + \frac{V_{REF'} - V_{O'}}{R_1 + R_2}}$$

where, V_i : External control voltage

V_F : Forward voltage of diode (D_i)

$V_{REF'}$: 0.9V Pin ③ voltage when $V_{REF'}$ is

$V_{O(OFF)}$

$V_{O'}$: 0V output voltage when $V_{O'}$ is $V_{O(OFF)}$

MITSUBISHI LINEAR ICs
M5236L, ML

**GENERAL PURPOSE 3-TERMINAL VARIABLE VOLTAGE OUTPUT REGULATOR
 (FOR DRIVER)**

DESCRIPTION

The M5236 is a semiconductor integrated circuit designed for general-purpose output voltage regulator.

A high-performance variable output voltage regulator with small input-output voltage differences can be made in combination with externally connected PNP transistors.

It is housed in a small 3-pin package, including a reference voltage circuit, error amplifier, and driver, and the output voltage can be set freely by externally connected resistors, and a small, compact power supply circuit can be achieved, making the device suitable for use in small electronic equipment, such as car stereo, radio cassette recorder and portable stereo equipment.

FEATURES

- Wide operating voltage range
 $V_{IN}=3.5V\sim 36V$, $V_O=1.5V\sim 33V$
- Capable of operating at low input-output voltage for driver by externally connected power transistors
 [$V_{CE(sat)}$ state of TR] $V_{I-O(min)}\cong 0.2V$
- Output voltage can be set freely by externally connected resistors
- Built-in ASO protection and thermal cutoff circuits
- Capable of taping (automated insertion) and lead forming

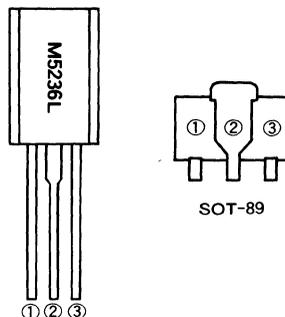
APPLICATION

For car stereo equipment, radio cassette recorder, portable stereo and other general electronic equipment.

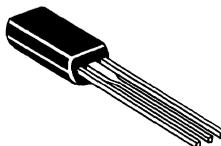
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range $V_{IN}=3.5V\sim 30V$
- Rated supply voltage $V_O=1.5V\sim 25V$

PIN CONFIGURATION



- PIN CONNECTION
- ① INPUT
 - ② GND
 - ③ VOLTAGE ADJUSTMENT

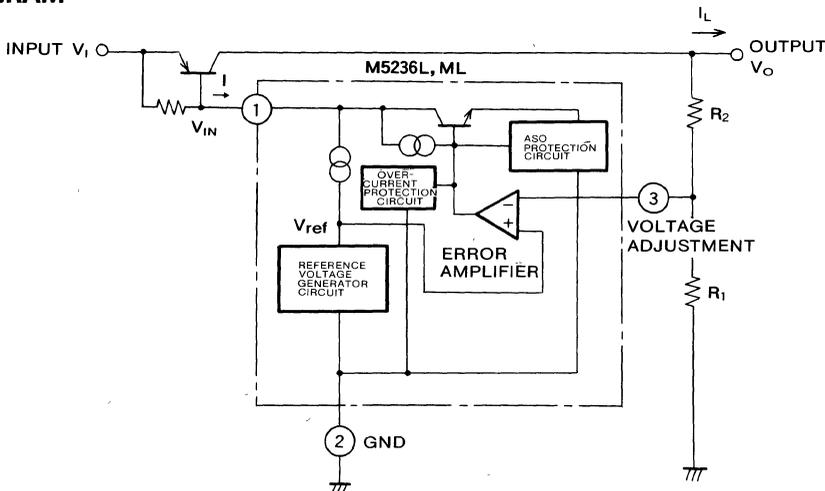


TO-92L package



SOT-89 package

BLOCK DIAGRAM



**GENERAL PURPOSE 3-TERMINAL VARIABLE VOLTAGE OUTPUT REGULATOR
 (FOR DRIVER)**

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$)

Symbol	Parameter	Ratings	Unit
V_{IN}	Input voltage	36	V
I_D	Drive current	30	mA
V_I-V_O	Input-output voltage difference	30	V
P_d	Power dissipation	900(SIP)/500(ML)	mW
T_{opr}	Operating temperature	-20~+75	°C
T_{stg}	Storage temperature	-55~+150	°C

ELECTRICAL CHARACTERISTICS (measurement circuit (a) is used with, $T_a=25^\circ\text{C}$, $V_I=15\text{V}$, $V_O=12\text{V}$, $I_L=200\text{mA}$, $C_{REF}=1\mu\text{F}$, $R_1=4.3\text{k}\Omega$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{IN}	Input voltage	(between pin ① and pin ②)	3.5		36	V
V_O	Output voltage	$R_2 \approx 0.82\text{k}\Omega \sim 108\text{k}\Omega$	1.5		33	V
V_I-V_O	Minimum input-output voltage difference			0.2		V
V_{REF}	Reference voltage	(between pin ③ and pin ②)	1.20	1.26	1.32	V
Reg_{in}	Input regulation	$V_I=15\sim 20\text{V}$		0.02	0.1	%/V
Reg_L	Load regulation	$I_L=10\sim 200\text{mA}$		0.02	0.1	%
I_B	Bias current	$I_L=0$ (disregarding the current in resistors R_1, R_2)		1.7	3.0	mA
TC_{VO}	Temperature coefficient of output voltage	$T_a=0\sim 75^\circ\text{C}$		0.02		%/°C
RR	Ripple rejection ratio	$f=120\text{Hz}$ (measured with circuit (b))		68		dB
V_{NO}	Output noise voltage	$f=20\text{Hz}\sim 100\text{kHz}$		33		μVrms

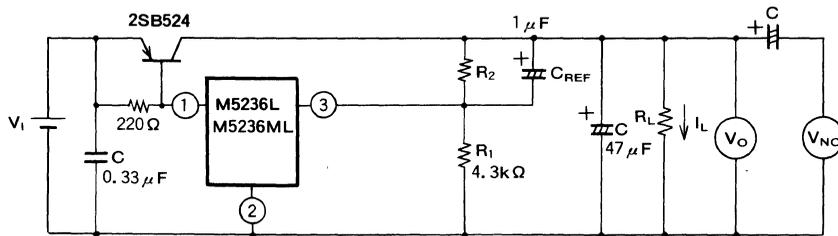
TEST CIRCUITS

(a) Standard test circuit

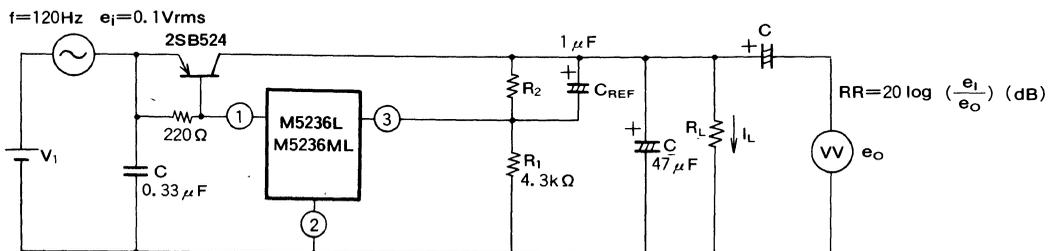
$$V_O = V_{REF} \left(1 + \frac{R_2}{R_1}\right) \approx 1.26 \times \left(1 + \frac{R_2}{4.3}\right) \text{ (V)}$$

$$R_2 = R_1 \left(\frac{V_O}{V_{REF}} - 1\right) \approx 4.3 \times \left(\frac{V_O}{1.26} - 1\right) \text{ (k}\Omega\text{)}$$

$$(R_1 = 4.3\text{k}\Omega, V_{REF} \approx 1.26\text{V})$$



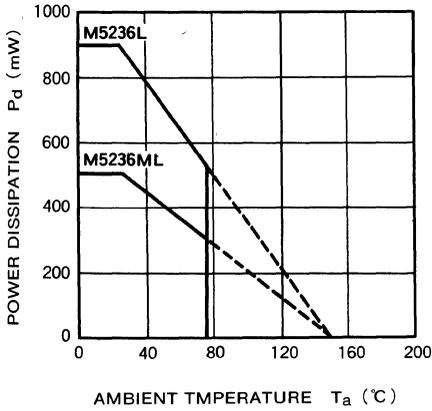
(b) Ripple rejection test circuit



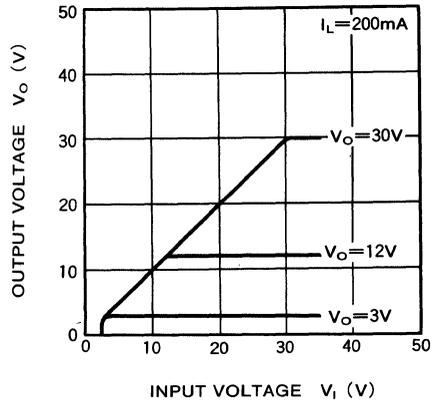
**GENERAL PURPOSE 3-TERMINAL VARIABLE VOLTAGE OUTPUT REGULATOR
 (FOR DRIVER)**

TYPICAL CHARACTERISTICS

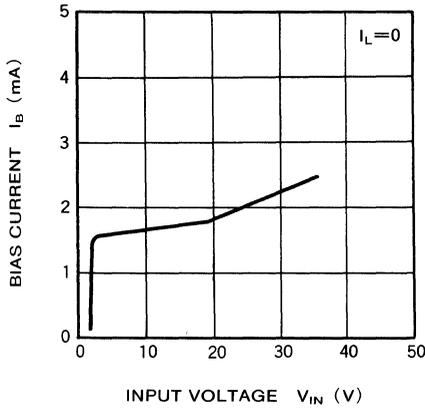
THERMAL DERATING (MAXIMUM RATINGS)



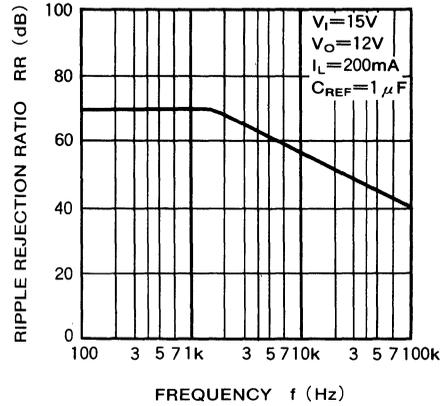
OUTPUT VOLTAGE CHARACTERISTICS



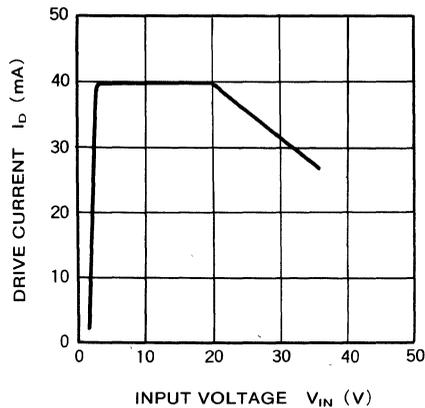
BIAS CURRENT VS. INPUT VOLTAGE



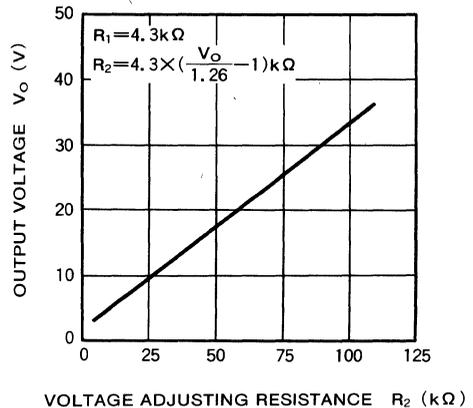
RIPPLE REJECTION



DRIVE CURRENT VS. INPUT VOLTAGE



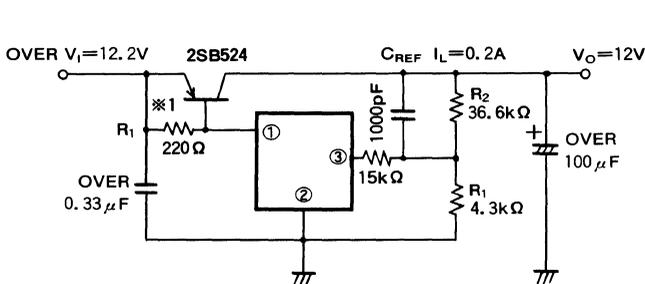
OUTPUT VOLTAGE VS. VOLTAGE ADJUSTING RESISTANCE



**GENERAL PURPOSE 3-TERMINAL VARIABLE VOLTAGE OUTPUT REGULATOR
 (FOR DRIVER)**

APPLICATION EXAMPLES

1. Standard application circuit



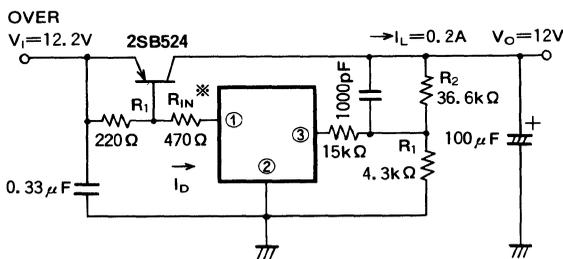
$$V_O = V_{REF} \times \left(1 + \frac{R_2}{R_1}\right) V$$

$$V_{REF} = 1.26V (\text{typ.})$$

※1. R₁ of 180~220Ω should be used.

Note) Capacitors displaying small capacity change with temperature should be used.

2. Control circuit for maximum drive current (I_{DM})



When the input voltage (V_i) is lower than the set output voltage (V_o), drive current of approximately 30mA to 45mA runs in Pin ① of the integrated circuit. (Refer to TYPICAL CHARACTERISTICS DRIVE CURRENT VS INPUT VOLTAGE. For example, when the input voltage V_i of 20V is higher than the fixed output voltage of 20V or above, and input and output are inverted, power dissipation in the circuit is P_d = 20V × 45mA = 900mW, and reaches the maximum rating, making it necessary to control the drive current.) When the input power supply is supplied by batteries and the current needs to be controlled, a resistor R_{IN} can be inserted to control the drive current. (Fig. 1 shows input voltage dependency of the control current and input resistor R_{IN}.)

When the input voltage reaches 12V (=V_o), the current at Pin ① is limited to approximately 20mA.

Fig. 2 shows I_D-V_i characteristics of the circuit.

Fig. 1 MAXIMUM DRIVE CURRENT CONTROL CHARACTERISTICS (I_{DM})

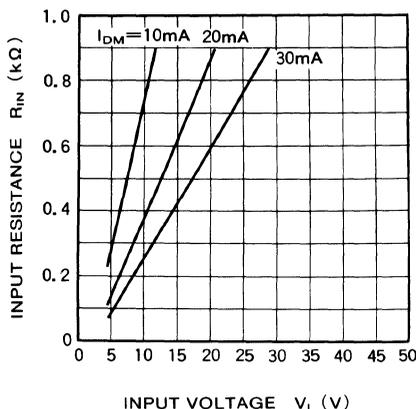
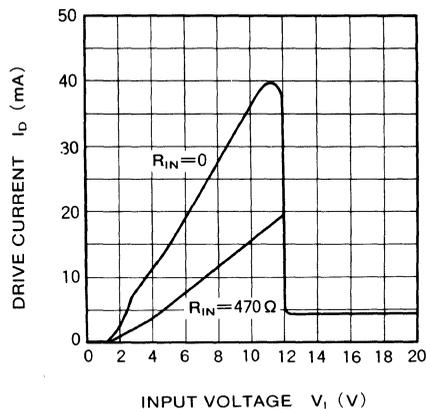
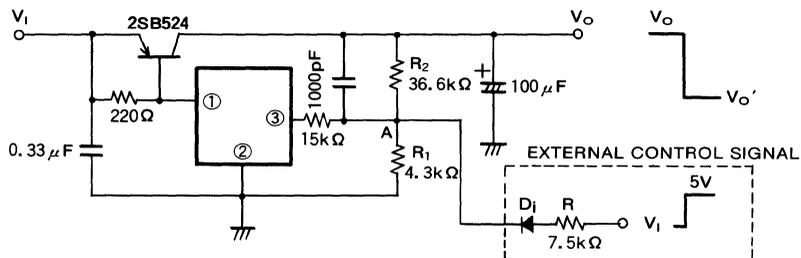


Fig. 2 I_D-V_i CHARACTERISTICS IN APPLICATION EXAMPLE 2



**GENERAL PURPOSE 3-TERMINAL VARIABLE VOLTAGE OUTPUT REGULATOR
 (FOR DRIVER)**

3. ON/OFF control of output voltage circuit



Resistor R in the control circuit is determined by the following equation.

$$R = \frac{V_i - V_F - V_{REF'}}{V_{REF'} + \frac{V_{REF'} - V_{O'}}{R_1 + R_2}}$$

where, V_i : External control voltage

V_F : Forward voltage of diode (D_i)

$V_{REF'}$: 1.4V Pin ③ voltage when $V_{REF'}$ is $V_{O(OFF)}$

$V_{O'}$: 0V output voltage when $V_{O'}$ is $V_{O(OFF)}$

MITSUBISHI LINEAR ICs M5278LXX

(5V, 5.6V, 8V, 9V, 10V, 12V, 15V) 3-TERMINAL FIXED POSITIVE OUTPUT VOLTAGE REGULATOR (WITH FOLD-BACK PROTECTION CIRCUIT)

DESCRIPTION

The M5278L series consists of four monolithic integrated circuits, each a three-terminal regulator with maximum load capabilities of 100mA and featuring output voltages of 5, 5.6, 8, 9, 10, 12, or 15V. The high-performance, fixed-positive output power supply ICs are packaged in 3-pin TO-92L packages featuring fold-back protective circuits for limiting current when load short. They are especially appropriate for use in personal computers and general power supplies of electrical appliances.

FEATURES

- Interchangeable with other brand 78L series
..... $I_{Lmax}=100\text{mA}$
- Internal fold-back protection circuit limits current due to shorted loads.
M5278L05 16.5mA (typ.) (1/10 that of other brands)
- High ripple division factor
M5278L05 73dB (typ.)
- Low output voltage tolerance $\pm 5\%$ (max.)
- High level of permissible internal heat dissipation
..... 900mW (max.)

APPLICATION

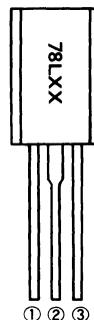
Power supply for microcomputer in VTR equipment, general power supply for electronic equipment of tape decks, car stereo equipment and radio cassette recorder.

FUNCTION CODE

M5278LXX
└── OUTPUT VOLTAGE VALUE

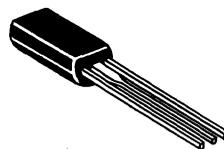
Type	Marking	Output voltage	Type	Marking	Output voltage
M5278L05	78L05	5V	M5278L10	78L10	10V
M5278L56	78L56	5.6V	M5278L12	78L12	12V
M5278L08	78L08	8V	M5278L15	78L15	15V
M5278L09	78L09	9V			

PIN CONFIGURATION



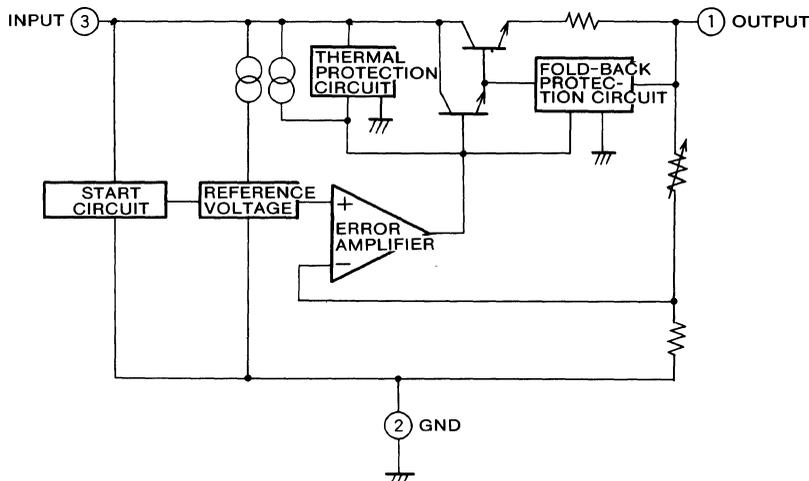
ELECTRODE CONNECTION

- ① OUTPUT EIAJ : TO-92L
- ② : GND
- ③ : INPUT



TO-92L package

BLOCK DIAGRAM

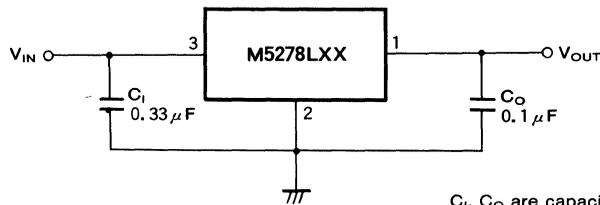


**(5V, 5.6V, 8V, 9V, 10V, 12V, 15V) 3-TERMINAL FIXED POSITIVE
 OUTPUT VOLTAGE REGULATOR (WITH FOLD-BACK PROTECTION CIRCUIT)**

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$)

Symbol	Parameter	Conditions	Ratings	Unit
V_i	Input voltage		36	V
I_L	Load current		100	mA
P_d	Power dissipation		900	mW
T_{opr}	Operating temperature		-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature		-55~+150	$^\circ\text{C}$

STANDARD CONNECTION



C_1 , C_0 are capacitors to prevent oscillation.
 Make connections as close to the IC as possible.

ELECTRICAL CHARACTERISTICS

M5278L05 ($V_i=10\text{V}$, $I_L=40\text{mA}$, $T_a=25^\circ\text{C}$, $C_1=0.33\ \mu\text{F}$, $C_0=0.1\ \mu\text{F}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		4.75	5.0	5.25	V
Reg-in	Input regulation	$7.5\text{V} \leq V_i \leq 20\text{V}$		50	200	mV
		$8\text{V} \leq V_i \leq 20\text{V}$		30	150	
Reg-L	Load regulation	$1\text{mA} \leq I_L \leq 100\text{mA}$		10	60	mV
		$1\text{mA} \leq I_L \leq 40\text{mA}$		5	30	
V_O	Output voltage	$7.5\text{V} \leq V_i \leq 20\text{V}$, $1\text{mA} \leq I_L \leq 40\text{mA}$	4.7	5.0	5.3	V
		$V_i=10\text{V}$, $1\text{mA} \leq I_L \leq 70\text{mA}$	4.7	5.0	5.3	
I_B	Bias current	$I_L=0$		4.8	6.7	mA
ΔI_B	Bias current variability	$8\text{V} \leq V_i \leq 20\text{V}$, $I_L=40\text{mA}$		0.1	1.5	mA
		$V_i=10\text{V}$, $1\text{mA} \leq I_L \leq 40\text{mA}$			0.2	
V_{NO}	Output noise voltage	BW: 10Hz~100kHz		49		μVrms
RR	Ripple rejection ratio	$f=120\text{Hz}$, $V_{in}=0\text{dBm}$	63	73		dB
V_i-V_O	Minimum input-output voltage difference			2.0		V
I_{LP}	Peak load current			130		mA
I_{OS}	Output short holding current			16.5		mA
TC_{VO}	Temperature coefficient of output voltage	$I_L=5\text{mA}$		-0.6		mV/ $^\circ\text{C}$

**(5V, 5.6V, 8V, 9V, 10V, 12V, 15V) 3-TERMINAL FIXED POSITIVE
OUTPUT VOLTAGE REGULATOR (WITH FOLD-BACK PROTECTION CIRCUIT)****M5278L56** ($V_i=11V, I_L=40mA, T_a=25^\circ C, C_1=0.33\mu F, C_O=0.1\mu F$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		5.32	5.6	5.88	V
R_{reg-in}	Input regulation	$8.5V \leq V_i \leq 21V$		50	200	mV
		$9V \leq V_i \leq 21V$		30	150	
R_{reg-L}	Load regulation	$1mA \leq I_L \leq 100mA$		10	60	mV
		$1mA \leq I_L \leq 40mA$		5	30	
V_O	Output voltage	$8.5V \leq V_i \leq 21V, 1mA \leq I_L \leq 40mA$	5.27	5.6	5.93	V
		$V_i=11V, 1mA \leq I_L \leq 70mA$	5.27	5.6	5.93	
I_B	Bias current	$I_L=0$		4.8	6.7	mA
ΔI_B	Bias current differential	$9V \leq V_i \leq 21V, I_L=40mA$		0.1	1.5	mA
		$V_i=11V, 1mA \leq I_L \leq 40mA$			0.2	
V_{NO}	Output noise voltage	BW: 10Hz~100kHz		55		μV_{rms}
RR	Ripple rejection ratio	$f=120Hz, V_{in}=0dBm$	63	73		dB
V_i-V_O	Minimum input-output voltage difference			2.0		V
I_{LP}	Peak load current			150		mA
I_{OS}	Output short circuit sustain current			16.5		mA
TC_{VO}	Temperature coefficient of output voltage	$I_L=5mA$		-0.65		mV/ $^\circ C$

M5278L08 ($V_i=14V, I_L=40mA, T_a=25^\circ C, C_1=0.33\mu F, C_O=0.1\mu F$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		7.6	8.0	8.4	V
R_{reg-in}	Input regulation	$10.5V \leq V_i \leq 23V$		60	200	mV
		$11V \leq V_i \leq 23V$		40	150	
R_{reg-L}	Load regulation	$1mA \leq I_L \leq 100mA$		10	80	mV
		$1mA \leq I_L \leq 40mA$		5	40	
V_O	Output voltage	$10.5V \leq V_i \leq 23V, 1mA \leq I_L \leq 40mA$	7.52	8.0	8.48	V
		$V_i=14V, 1mA \leq I_L \leq 70mA$	7.52	8.0	8.48	
I_B	Bias current	$I_L=0$		4.8	6.7	mA
ΔI_B	Bias current differential	$10.5V \leq V_i \leq 23V, I_L=40mA$		0.1	1.5	mA
		$V_i=14V, 1mA \leq I_L \leq 40mA$			0.2	
V_{NO}	Output noise voltage	BW: 10Hz~100kHz		80		μV_{rms}
RR	Ripple rejection ratio	$f=120Hz, V_{in}=0dBm$		59		dB
V_i-V_O	Minimum input-output voltage difference			2.0		V
I_{LP}	Peak load current			150		mA
I_{OS}	Output short circuit sustain current			12.5		mA
TC_{VO}	Temperature coefficient of output voltage	$I_L=5mA$		-0.7		mV/ $^\circ C$

**(5V, 5.6V, 8V, 9V, 10V, 12V, 15V) 3-TERMINAL FIXED POSITIVE
OUTPUT VOLTAGE REGULATOR (WITH FOLD-BACK PROTECTION CIRCUIT)**

M5278L09 ($V_I=16V, I_L=40mA, T_a=25^\circ C, C_I=0.33\mu F, C_O=0.1\mu F$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		8.55	9.0	9.45	V
R_{reg-in}	Input regulation	$12V \leq V_I \leq 24V$ $13V \leq V_I \leq 24V$		70 40	225 170	mV
R_{reg-L}	Load regulation	$1mA \leq I_L \leq 100mA$ $1mA \leq I_L \leq 40mA$		10 5	90 40	mV
V_O	Output voltage	$12V \leq V_I \leq 24V, 1mA \leq I_L \leq 40mA$ $V_I=16V, 1mA \leq I_L \leq 70mA$	8.46 8.46	9.0	9.54 9.54	V
I_B	Bias current	$I_L=0$		4.8	6.7	mA
ΔI_B	Bias current differential	$13V \leq V_I \leq 24V, I_L=40mA$ $V_I=16V, 1mA \leq I_L \leq 40mA$		0.1	1.5 0.2	mA
V_{NO}	Output noise voltage	BW: 10Hz~100kHz		90		μV_{rms}
RR	Ripple rejection ratio	$f=120Hz, V_{in}=0dBm$		58		dB
$V_I - V_O$	Minimum input-output voltage difference			2.0		V
I_{LP}	Peak load current			150		mA
I_{OS}	Output short circuit sustain current			12.5		mA
TC_{VO}	Temperature coefficient of output voltage	$I_L=5mA$		-0.8		mV/°C

M5278L10 ($V_I=17V, I_L=40mA, T_a=25^\circ C, C_I=0.33\mu F, C_O=0.1\mu F$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		9.5	10.0	10.5	V
R_{reg-in}	Input regulation	$12.5V \leq V_I \leq 25V$ $13V \leq V_I \leq 25V$		70 50	230 170	mV
R_{reg-L}	Load regulation	$1mA \leq I_L \leq 100mA$ $1mA \leq I_L \leq 40mA$		10 5	90 45	mV
V_O	Output voltage	$12.5V \leq V_I \leq 25V, 1mA \leq I_L \leq 40mA$ $V_I=17V, 1mA \leq I_L \leq 70mA$	9.4 9.4	10.0	10.6 10.6	V
I_B	Bias current	$I_L=0$		4.8	6.7	mA
ΔI_B	Bias current differential	$13V \leq V_I \leq 25V, I_L=40mA$ $V_I=17V, 1mA \leq I_L \leq 40mA$		0.1	1.5 0.2	mA
V_{NO}	Output noise voltage	BW: 10Hz~100kHz		100		μV_{rms}
RR	Ripple rejection ratio	$f=120Hz, V_{in}=0dBm$		57		dB
$V_I - V_O$	Minimum input-output voltage difference			2.0		V
I_{LP}	Peak load current			130		mA
I_{OS}	Output short circuit sustain current			12.5		mA
TC_{VO}	Temperature coefficient of output voltage	$I_L=5mA$		-0.9		mV/°C

**(5V, 5.6V, 8V, 9V, 10V, 12V, 15V) 3-TERMINAL FIXED POSITIVE
OUTPUT VOLTAGE REGULATOR (WITH FOLD-BACK PROTECTION CIRCUIT)**

M5278L12 ($V_i=19V$, $I_L=40mA$, $T_a=25^\circ C$, $C_1=0.33\mu F$, $C_o=0.1\mu F$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_o	Output voltage		11.4	12.0	12.6	V
Reg_{-in}	Input regulation	$14.5V \leq V_i \leq 27V$		70	250	mV
		$16V \leq V_i \leq 27V$		50	200	
Reg_{-L}	Load regulation	$1mA \leq I_L \leq 100mA$		10	100	mV
		$1mA \leq I_L \leq 40mA$		5	50	
V_o	Output voltage	$14.5V \leq V_i \leq 27V$, $1mA \leq I_L \leq 40mA$	11.3	12.0	12.7	V
		$V_i=19V$, $1mA \leq I_L \leq 70mA$	11.3	12.0	12.7	
I_B	Bias current	$I_L=0$		4.8	6.7	mA
ΔI_B	Bias current differential	$16V \leq V_i \leq 27V$, $I_L=40mA$		0.1	1.5	mA
		$V_i=19V$, $1mA \leq I_L \leq 40mA$			0.2	
V_{NO}	Output noise voltage	BW: 10Hz~100kHz		110		μV_{rms}
RR	Ripple rejection ratio	$f=120Hz$, $V_{in}=0dBm$		55		dB
V_i-V_o	Minimum input-output voltage difference			2.0		V
I_{LP}	Peak load current			150		mA
I_{OS}	Output short circuit sustain current			12.5		mA
TC_{V_o}	Temperature coefficient of output voltage	$I_L=5mA$		-1.0		mV/ $^\circ C$

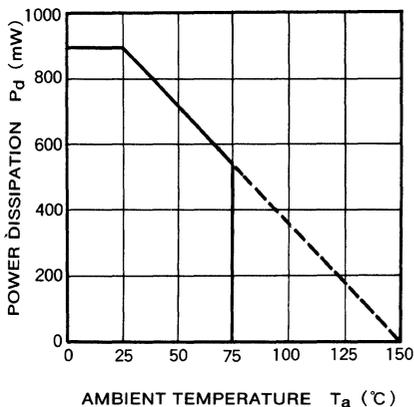
M5278L15 ($V_i=23V$, $I_L=40mA$, $T_a=25^\circ C$, $C_1=0.33\mu F$, $C_o=0.1\mu F$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_o	Output voltage		14.25	15.0	15.75	V
Reg_{-in}	Input regulation	$17.5V \leq V_i \leq 30V$		80	300	mV
		$20V \leq V_i \leq 30V$		60	250	
Reg_{-L}	Load regulation	$1mA \leq I_L \leq 100mA$		10	150	mV
		$1mA \leq I_L \leq 40mA$		5	75	
V_o	Output voltage	$17.5V \leq V_i \leq 30V$, $1mA \leq I_L \leq 40mA$	14.1	15.0	15.9	V
		$V_i=23V$, $1mA \leq I_L \leq 70mA$	14.1	15.0	15.9	
I_B	Bias current	$I_L=0$		4.8	6.7	mA
ΔI_B	Bias current differential	$23V \leq V_i \leq 30V$, $I_L=40mA$		0.1	1.5	mA
		$V_i=23V$, $1mA \leq I_L \leq 40mA$			0.2	
V_{NO}	Output noise voltage	BW: 10Hz~100kHz		140		μV_{rms}
RR	Ripple rejection ratio	$f=120Hz$, $V_{in}=0dBm$		52		dB
V_i-V_o	Minimum input-output voltage difference			2.0		V
I_{LP}	Peak load current			150		mA
I_{OS}	Output short circuit sustain current			12.5		mA
TC_{V_o}	Temperature coefficient of output voltage	$I_L=5mA$		-1.2		mV/ $^\circ C$

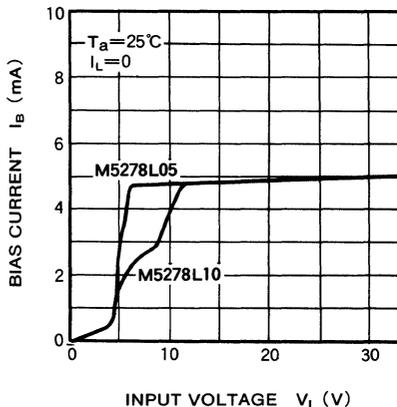
(5V, 5.6V, 8V, 9V, 10V, 12V, 15V) 3-TERMINAL FIXED POSITIVE OUTPUT VOLTAGE REGULATOR (WITH FOLD-BACK PROTECTION CIRCUIT)

TYPICAL CHARACTERISTICS

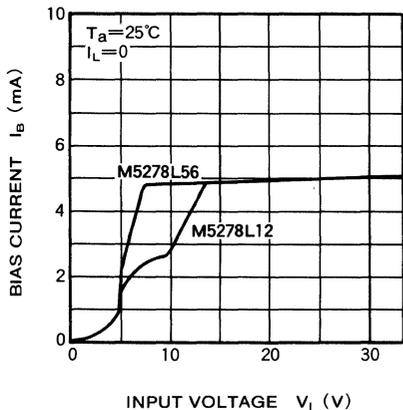
THERMAL DERATING (MAXIMUM RATING)



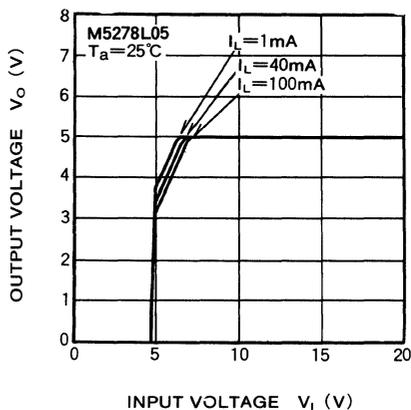
BIAS CURRENT VS. INPUT VOLTAGE



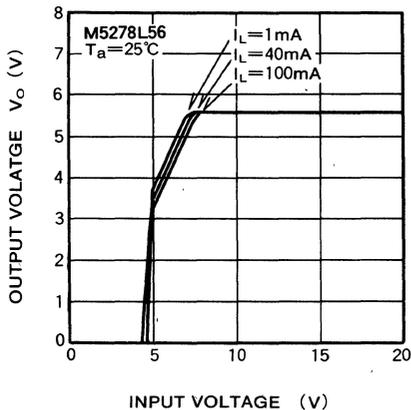
BIAS CURRENT VS. INPUT VOLTAGE



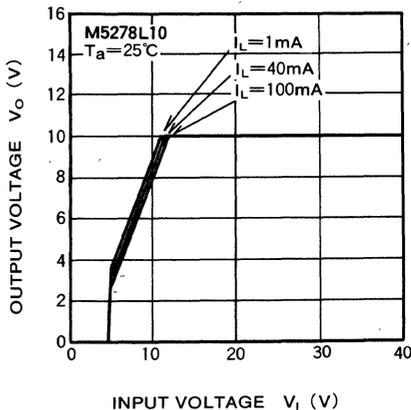
OUTPUT VOLTAGE VS. INPUT VOLTAGE



OUTPUT VOLTAGE VS. INPUT VOLTAGE

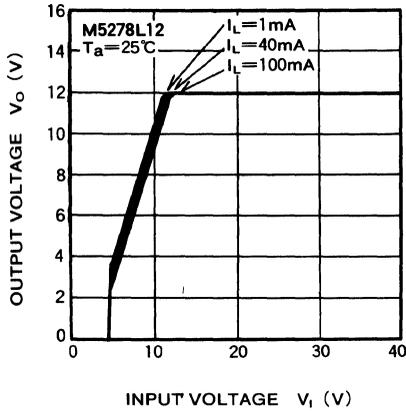


OUTPUT VOLTAGE VS. INPUT VOLTAGE

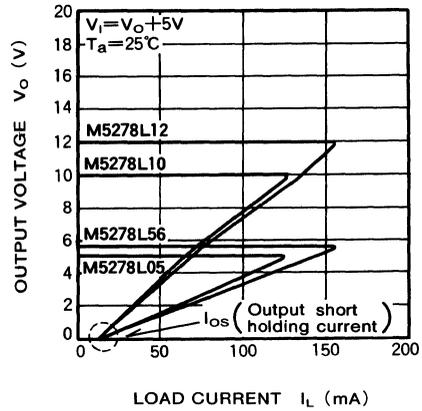


**(5V, 5.6V, 8V, 9V, 10V, 12V, 15V) 3-TERMINAL FIXED POSITIVE
 OUTPUT VOLTAGE REGULATOR (WITH FOLD-BACK PROTECTION CIRCUIT)**

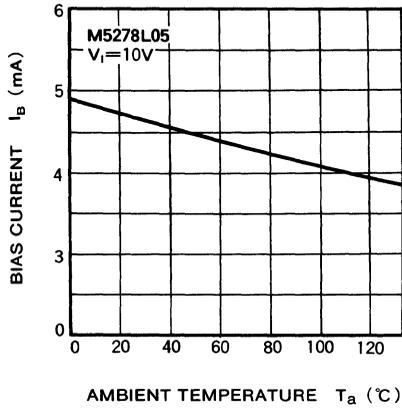
OUTPUT VOLTAGE VS. INPUT VOLTAGE



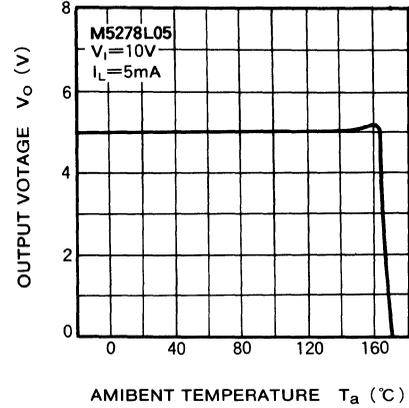
OUTPUT VOLTAGE VS. LOAD CURRENT



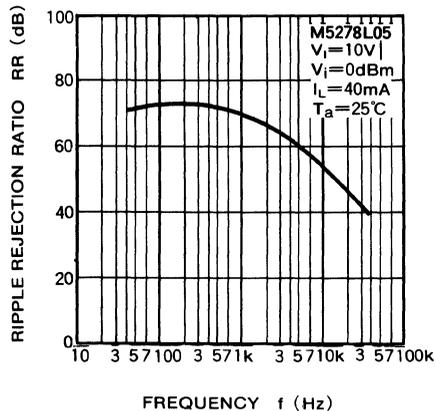
**BIAS CURRENT VS.
 AMBIENT TEMPERATURE**



**OUTPUT VOLTAGE VS.
 AMBIENT TEMPERATURE**

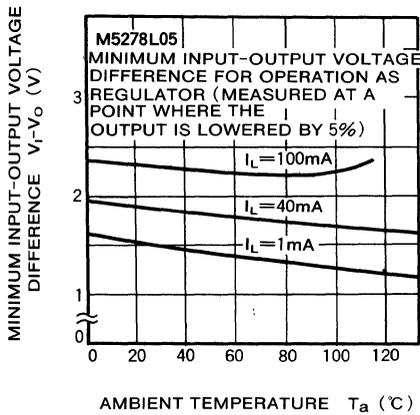


RIPPLE REJECTION VS. FREQUENCY



**(5V, 5.6V, 8V, 9V, 10V, 12V, 15V) 3-TERMINAL FIXED POSITIVE
 OUTPUT VOLTAGE REGULATOR (WITH FOLD-BACK PROTECTION CIRCUIT)**

**MINIMUM INPUT-OUTPUT
 VOLTAGE DIFFERENCE
 VS. AMBIENT TEMPERATURE**



FOLD-BACK PROTECTION CIRCUIT

The M5278L series has been designed to be complete with three-pin, 78L type regulators manufactured by other companies for applications with loading currents in the 100mA class. They additionally have an internalized fold-back protection circuit for protection against shorted loads.

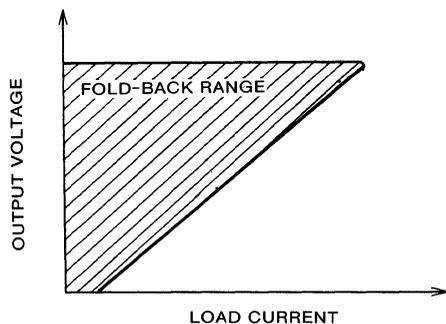
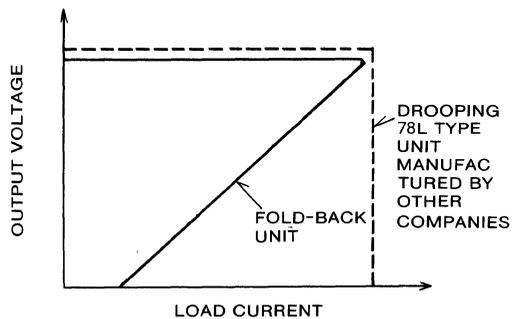
Other 78L units do have an internalized protection circuit known as a drooping type circuit that are rether simple, merely limiting maximum loading current. When large current begins to flow in these units due to a short circuit, abnormal temperatures are generated leading to breakdown and effective setting reliability.

As shown in the diagram, the fold-back protection circuit employed in the M5278L decreases immediately excessive current caused by a short in the load. This not only improves set reliability but permits the elimination of such protection circuits as fuses in the protection circuit.

PRECAUTIONS FOR USE

The current-control circuit requires that, as an IC power supply, this device be operated within the fold-back operating range shown in the accompanying chart.

OUTPUT VOLTAGE VS. LOAD CURRENT



MITSUBISHI LINEAR ICs

M5T494P, FP

SWITCHING REGULATOR CONTROL

DESCRIPTION

The M5T494P,FP is a monolithic IC designed for a pulse-width-modulation control circuit.

It contains all functions necessary to control a switching power supplies. It employs an on-chip reference voltage circuit, error amplifier, saw-teeth wave oscillator, dead-time control comparator, a flip-flop buffer capable of both sink and source.

It also contains an undervoltage-lockout (UVLO) function, which turns the output completely even in transit state due to the rising and falling of power supply.

FEATURES

- Built-in undervoltage lock-out circuit
- Built-in reference regulator $5V \pm 5\%$
- Applicable to single-end and push-pull circuits
- Uncommitted outputs for 200mA sink or source outputs

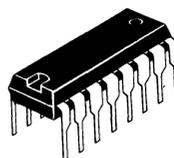
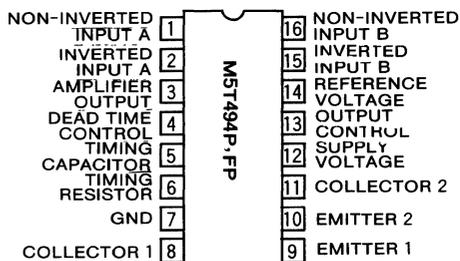
APPLICATION

Switching regulator, step-up regulator, step-down regulator, voltage inversion regulator.

RECOMMENDED OPERATING CONDITIONS

Supply voltage range	7~40V
Pin ③ current	less than 0.3mA
Timing capacitor C_T	470pF~3.3 μ F
Timing resistor R_T	1.8~500k Ω
Oscillator frequency	lower than 300kHz

PIN CONFIGURATION (TOP VIEW)

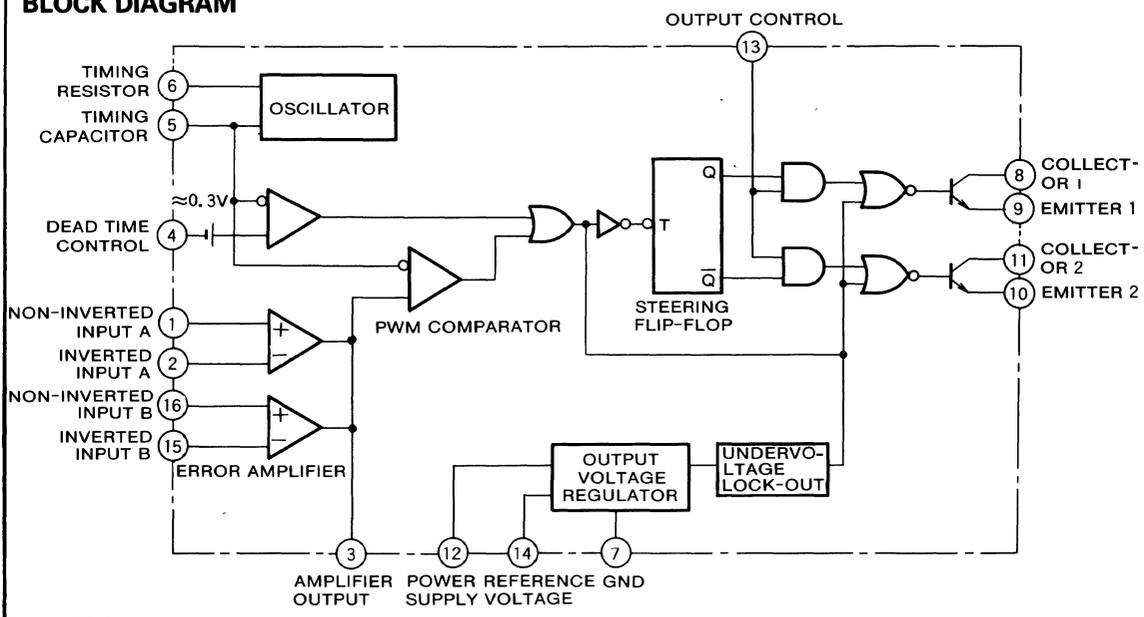


16-pin molded plastic DIP



16-pin molded plastic FLAT

BLOCK DIAGRAM



SWITCHING REGULATOR CONTROL

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

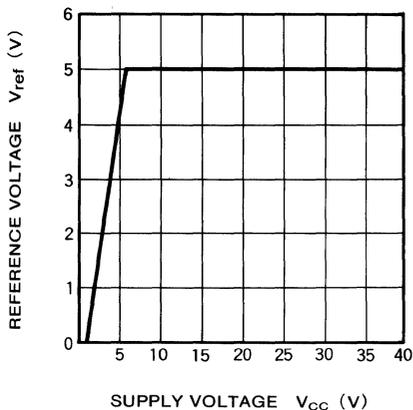
Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		41	V
V _{ICM}	Common phase input voltage range of error amplifier		-0.3~V _{CC}	V
V _{ID}	Differential input voltage of error amplifier		V _{CC}	V
V _O	Output voltage		41	V
I _O	Output current		200	mA
V _③	Pin ③ applied voltage range		-0.3~V _③ +0.3	V
P _d	Power dissipation		1000(M5T494P)	mW
			800(M5T494FP)	
K _θ	Thermal derating	T _a ≥25°C	8(M5T494P)	mW/°C
			6.4(M5T494FP)	
T _{opr}	Operating temperature		-20~+85	°C
T _{stg}	Storage temperature		-40~+125	°C

ELECTRICAL CHARACTERISTICS (V_{CC}=15V, f_{OSC}=40kHz, T_a=-20~+70°C, unless otherwise noted)

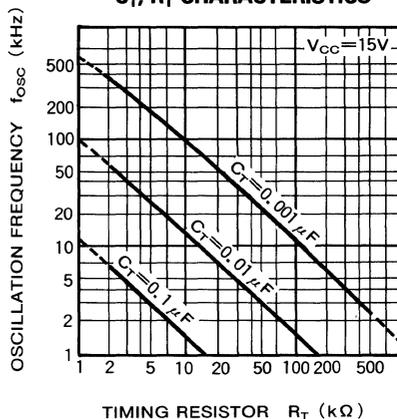
Symbol	Block	Parameter	Test conditions	Limits			Unit
				Min	Typ	Max	
V _{ref}	REFERENCE	Output voltage	I _{ref} =-1mA, T _j =25°C	4.75	5	5.25	V
ΔV _{ref IN}		Input stability	V _{CC} =7~40V I _{ref} =-1mA, T _j =25°C		1	10	mV
ΔV _{ref L}		Load stability	I _{ref} =-1~ -10mA, T _j =25°C		2	20	mV
ΔV _{ref} /ΔT _a		Output voltage temperature regulation	T _a =-20~+85°C, I _{ref} =-1mA		0.01	0.03	%/°C
I _s		Output short current	V _{ref} =0	-50	-30	-15	mA
f _{OSC}	OSCILLATOR	Oscillation frequency setting value	C _T =4700pF, R _T =6.2kΩ	37	41	45	kHz
Δf/f _s		Oscillation frequency setting accuracy	V _{CC} =7~40V, T _a =25°C C _T , R _T , within recommended ratings		10		%
Δf/f _{IN}		Frequency input stability	V _{CC} =7~40V, T _a =25°C C _T =4700pF, R _T =6.2kΩ		0.5	1.5	%
Δf/f _{Ta}		Frequency thermal regulation	T _a =0~70°C C _T =4700pF, R _T =6.2kΩ		1	2	%
I _③	DEAD-TIME CONTROL	Input bias current	V _③ =0~5.25V	-7	-0.7		μA
D _{max}		Maximum duty (each output stage)	V _③ =0V	42	45	48	%
V _{③TH1}	CONTROL	Input threshold voltage 1	Output pulse 0% duty		2.45	2.80	V
V _{③TH2}		Input threshold voltage 2	Output pulse maximum duty		0		V
V _{AMPIO}	ERROR AMPLIFIER	Input offset voltage	V _③ =2.5V		1	7	mV
I _{AMPIO}		Input offset current	V _③ =2.5V		5	200	nA
I _{AMPIB}		Input bias current	V _③ =2.5V	-700	-100		nA
V _{AMPICM}		Common phase input voltage range	V _{CC} =7~40V	-0.3		V _{CC} -2	V
A _v		Open loop voltage gain	V _③ =0.5~3.5V, T _a =25°C	70	110		dB
f _T		Unity gain frequency	T _a =25°C	500	900		kHz
CMRR		Common mode signal rejection ratio	V _{CC} =40V, T _a =25°C	65	85		dB
I _{③sink}		Output sink current	V _③ =0.7V	0.3	0.7		mA
I _{③source}		Output source current	V _③ =3.5V		-10	-2	mA
V _{③RANGE}		Output voltage range	Low level			0.1	0.3
	High level		I _③ =0	4.2	4.9		V
V _{③TH}	PWM COMPARATOR	Input threshold voltage	Output pulse 0% duty		3.4	3.8	V
I _{③sink}		Input sink current	V _③ =0.7V	0.3	0.7		mA
I _{CL}	OUTPUT	Collector leak current	V _{CE} =40V, V _{CC} =40V Common emitter		0.01	100	μA
I _{EL}		Emitter leak current	V _{CC} =V _C =40V, V _E =0V Emitter follower	-100	-0.01		μA
V _{CESAT}		Collector saturation voltage (emitter grounded)	I _C =200mA, V _E =0V		0.95	1.3	V
V _{CEON}		Collector saturation voltage (emitter follower)	I _E =-200mA, V _C =15V		1.6	2.5	V

SWITCHING REGULATOR CONTROL

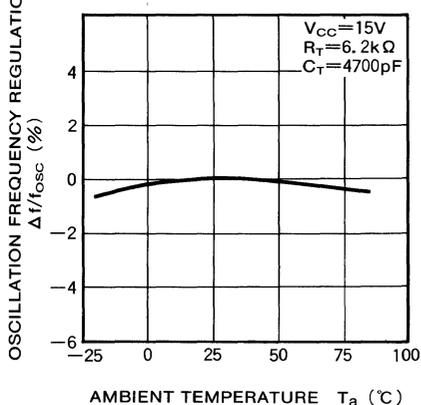
REFERENCE VOLTAGE VS. SUPPLY VOLTAGE



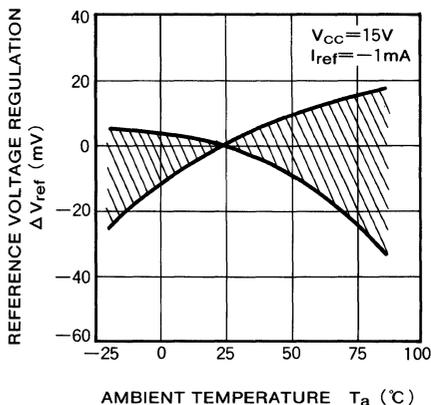
OSCILLATION FREQUENCY VS. C_T, R_T CHARACTERISTICS



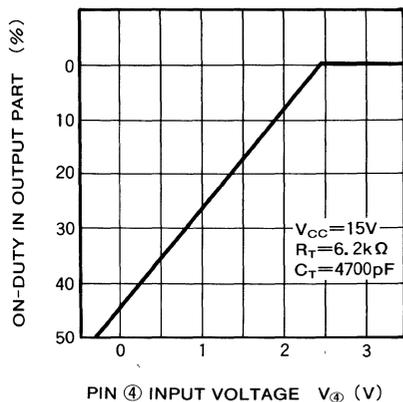
OSCILLATION FREQUENCY REGULATION VS. AMBIENT TEMPERATURE



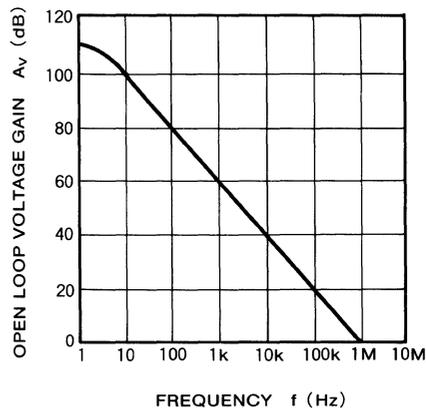
REFERENCE VOLTAGE REGULATION VS. AMBIENT TEMPERATURE



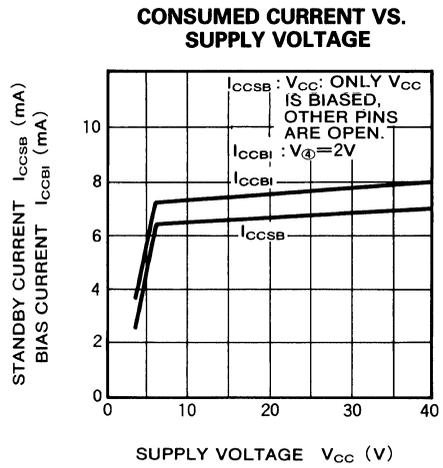
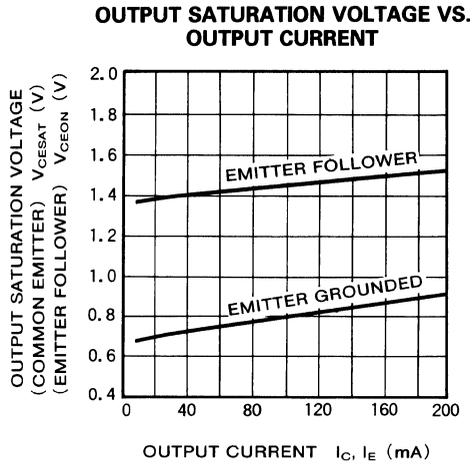
INPUT VOLTAGE AT ON-DUTY PIN ④ OF OUTPUT PART



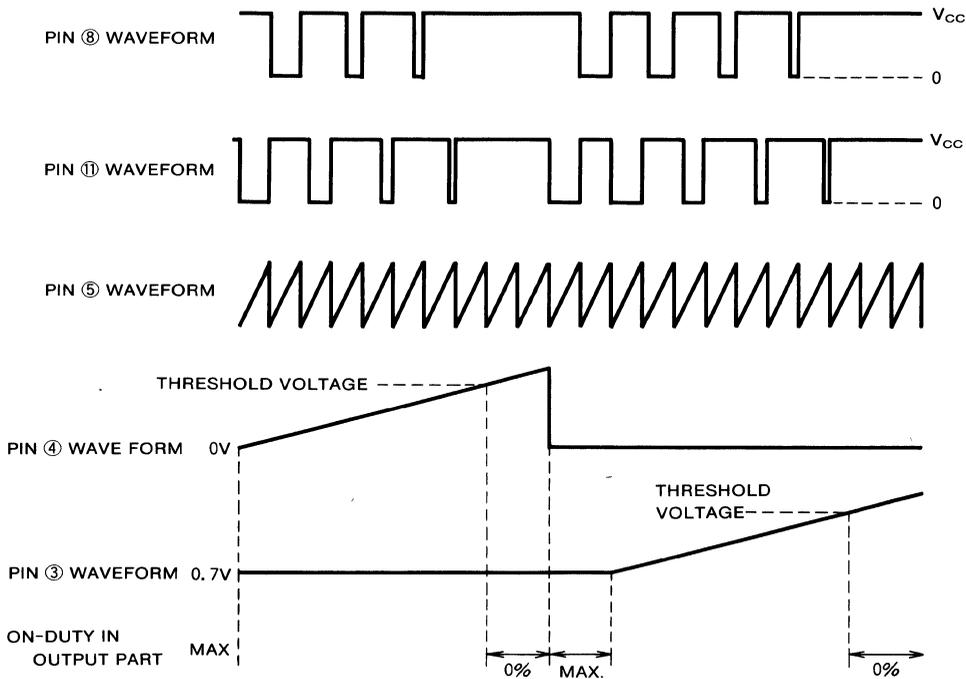
OPEN LOOP VOLTAGE GAIN OF ERROR AMPLIFIER VS. FREQUENCY



SWITCHING REGULATOR CONTROL



VOLTAGE WAVEFORMS



PIN ⑬ VOLTAGE AND OPERATION MODE

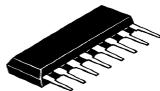
Pin ⑬ Voltage	Operational mode
V_{ref}	Push-pull operation
GND	Single-end or parallel operation

TIMER

5

QUICK REFERENCE TABLE OF TIMERS

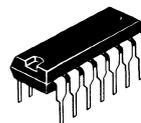
Application	Type No	Processing time	Features	Electrical characteristics			Package outline
				Supply voltage range	Output current (max.)	Circuit current (typ.)	
CR timer	M51841P	100 μ s~180sec	High output current 200mA	4.5~16V	200mA	10mA	8-pin DIP
	M51843P	10ms~180sec	Internal open collector and emitter outputs Built-in ON/OFF control and discharge circuit	4.5~16V	200mA	13mA	14-pin DIP
	M51848	10 μ s~180sec	Maximum operating frequency 100kHz Built-in supply voltage reset circuit	4~17V	200mA	7mA	8-pin DIP 8-pin SIP
Dual CR timer	M51847P	10 μ s~180sec	Maximum operating frequency 100kHz Built-in supply voltage reset circuit	4~17V	100mA	10mA	14-pin DIP
High-speed CR timer	M52051P	1 μ s~1sec	Maximum oscillator frequency 1MHz Built-in reset circuit	4.5~5.5V	5mA	11mA	8-pin DIP
III counter timer	M51849L	100ms~50hr	Built-in supply voltage reset circuit and stabilization zener Built-in 11 stage divider	5~V _Z	30mA	3.5mA	8-pin SIP
CMOS counter timer	M58479P	50ms~4800hr	Low power dissipation Selection of division ratio	7.4~9V	1.6mA	0.25mA	14-pin DIP
	M58482P	50ms~4800hr	Low power dissipation Selection of division ratio	3~9V	1.6mA	25 μ A	14-pin DIP



8-pin DIP



8-pin SIP



14-pin DIP

M51841P

SINGLE TIMER

DESCRIPTION

The M51841P monolithic timing circuit is highly stable controller capable of producing accurate time delays, or oscillation. Additional terminals are provided for two voltage comparators, triggering or resetting, if desired, applicable for a wide range of usage as monostable or astable multivibrators. The output can be directly connected to TTL or DTL circuits.

The M51841P is interchangeable with the Signetics NE555.

FEATURES

- Capable of forming a monostable multivibrator with one resistor and capacitor
- Capable of forming an astable multivibrator with two resistors and a capacitor
- Supply voltage rejection ratio.....0.01%V(typ.)
- Temperature coefficient.....50ppm/°C(typ.)
- High output current.....200mA(max.)

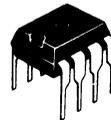
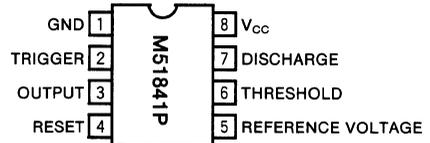
APPLICATION

Time-delay generator (monostable multivibrator), pulse oscillator (astable multivibrator), pulsewidth modulation, pulse position modulation, sequential timer, DC-DC converter

RECOMMENDED OPERATING CONDITIONS

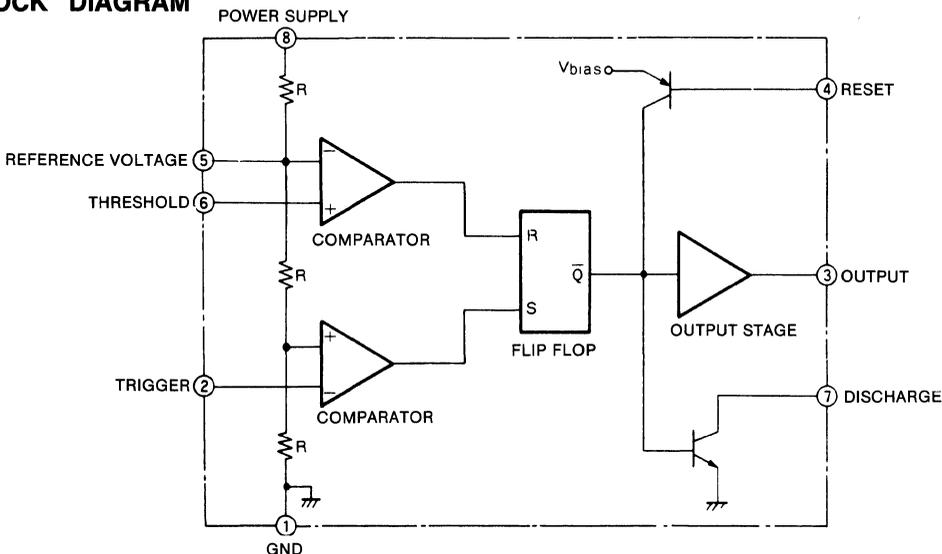
- Supply voltage range.....4.5~16V
- Rated supply voltage.....5V, 12V, 15V

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic DIP

BLOCK DIAGRAM



ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

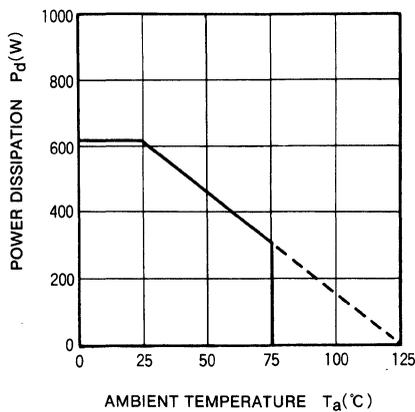
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
$I_{O(\text{peak})}$	Output current		200	mA
P_d	Power dissipation		625	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	6.25	mW/ $^\circ\text{C}$
T_{opr}	Operating ambient temperature		-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature		-40~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=5\sim 15\text{V}$, unless otherwise noted)

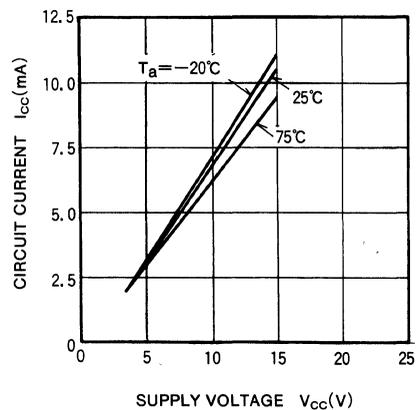
Symbol	Parameter	$V_{CC}(\text{V})$	Test conditions	Limits			Unit
				Min	Typ	Max	
V_{CC}	Supply voltage			4.5		16	V
I_{CC}	Circuit current	5	$R_L = \infty$		3	6	mA
		15	$R_L = \infty$		10	15	
V_{REF}	Reference voltage	5		2.6	3.33	4.0	V
		15		9	10	11	
V_{TH}	Threshold voltage				$2/3V_{CC}$		V
I_{TH}	Threshold current				0.1	0.25	μA
V_T	Trigger voltage				$1/3V_{CC}$		V
I_T	Trigger current				0.5	1.0	μA
V_R	Reset voltage				0.7	1.0	V
I_R	Reset current					0.1	mA
V_{OL}	Low output voltage	5	$I_{SINK}=5\text{mA}$		0.25	0.35	V
		15	$I_{SINK}=10\text{mA}$		0.1	0.25	
		15	$I_{SINK}=50\text{mA}$		0.4	0.75	
		15	$I_{SINK}=100\text{mA}$		2.0	2.5	
V_{OH}	High output voltage	5	$I_{SOURCE}=100\text{mA}$	2.75	3.3		V
		15	$I_{SOURCE}=100\text{mA}$	12.75	13.3		
—	Timing error				0.5		%
—	Temperature coefficient				50		ppm/ $^\circ\text{C}$
—	Supply voltage rejection ratio				0.01		%/V

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

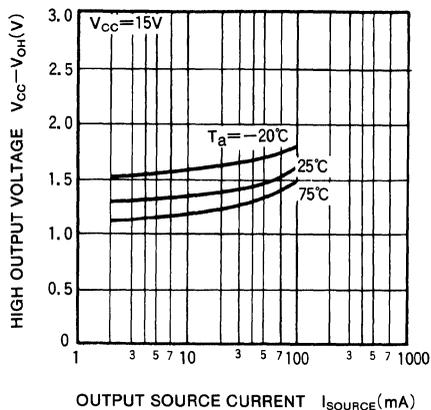
**THERMAL DERATING
(MAXIMUM RATING)**



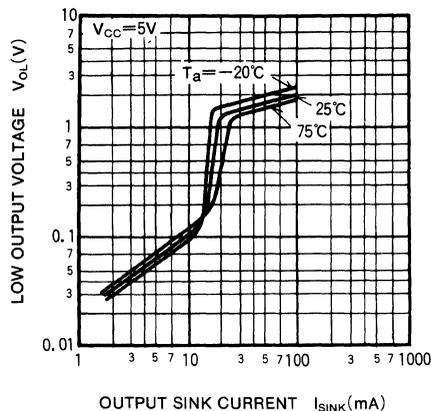
**SUPPLY VOLTAGE
VS. CIRCUIT CURRENT**



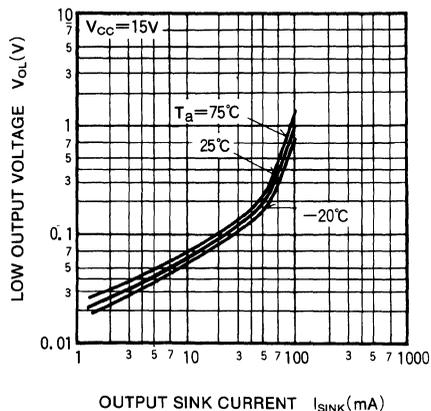
HIGH OUTPUT VOLTAGE VS. OUTPUT SOURCE CURRENT



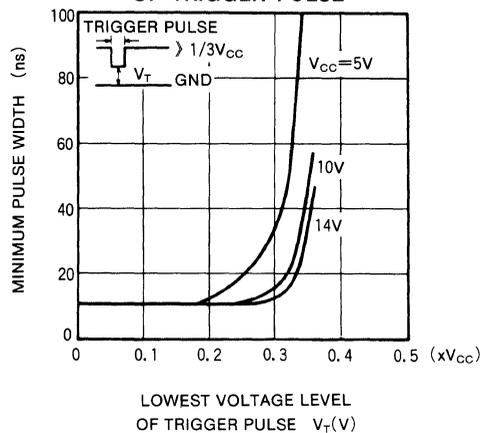
LOW OUTPUT VOLTAGE VS. OUTPUT SINK CURRENT



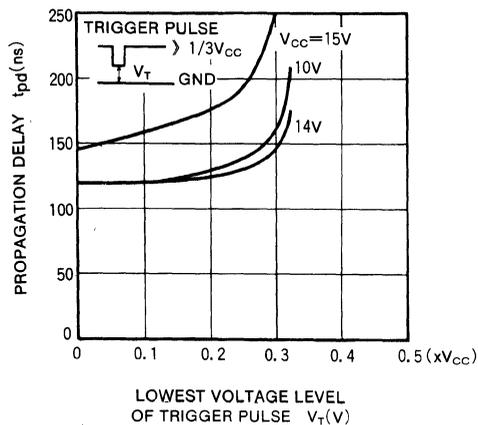
LOW OUTPUT VOLTAGE VS. OUTPUT SINK CURRENT



MINIMUM PULSE WIDTH VS. LOWEST VOLTAGE LEVEL OF TRIGGER PULSE



PROPAGATION DELAY VS. LOWEST VOLTAGE LEVEL OF TRIGGER PULSE



PIN DESCRIPTION

1. Trigger pin (pin ②)

When the voltage at the trigger pin is reduced to lower than $1/3V_{CC}$, timing operation is started. Time of applying the trigger voltage "L" should be shorter than the time set by CR.

2. Output pin (pin ③)

Logic output level is normally in the low state but is in the high state during timing operation. The output circuit is shaped like a totem-pole and withstands maximum load current of 200mA. The circuit can be directly connected to TTL or DTL circuits.

3. Reset pin (pin ④)

Timing operation can be interrupted by applying the reset signal to a reset pin. (The voltage at this pin must be less than 1V.)

If the reset signal is applied, the output is in the low state and the reset condition is maintained as long as the signal is applied. The output stays in the low state until the reset signal is replaced with the set signal (trigger signal at a trigger pin). Connect this pin to V_{CC} if not used.

4. Reference voltage pin (pin ⑤)

The voltage at this pin is normally set at $2/3V_{CC}$. By applying reference voltage, delay-time or oscillation frequency can be changed. The control signal for pulse-width modulation is applied at this pin. Connect a capacitor of $0.01\mu F$ between this pin and GND as a noise filter, if the pin is not used.

5. Threshold pin (pin ⑥)

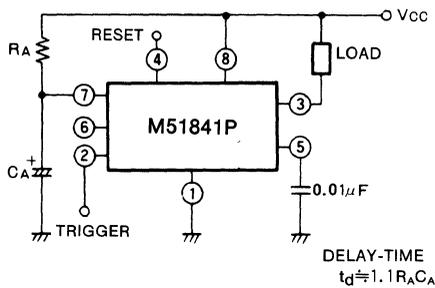
The delay-time is determined by CR time constants connected to this pin. The delay-time of a monostable multivibrator t_d is expressed by the equation $t_d \approx 1.1R_A C_A$.

6. Discharge pin (pin ⑦)

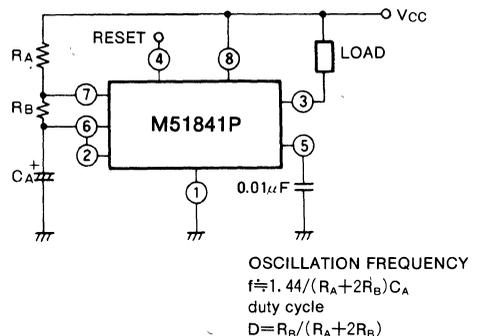
The timed capacitor discharges at this pin. The discharge is enabled when a flip-flop in the device is ON and the reset signal is applied. In a monostable multivibrator, the capacitor is connected to pin ⑥

APPLICATION EXAMPLES

(1) Delay-timer
(monostable multivibrator)



(2) Pulse oscillator
(astable multivibrator)



Precautions for use

- Care must be taken not to connect the discharge pin (pin ⑦) directly to V_{CC} . Such connection causes a short circuit of V_{CC} and GND through a discharge transistor (built in to the device) and may destroy the integrated circuit.
- Connect the reset pin (pin ④) to V_{CC} , if not used.
- A capacitor connected to a reference voltage pin (pin ⑤) functions as a noise filter. Choose appropriate capacitance according to the noise induced. The M51843P is recommended for use in an environment where noise may cause misoperation.

M51843P

SINGLE TIMER

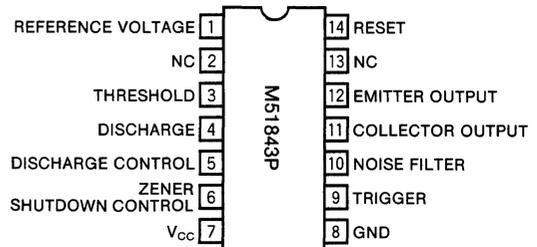
DESCRIPTION

The M51843P monolithic timing circuit is highly stable controller capable of producing accurate time delays, or oscillation. Additional terminals are provided for two voltage comparators, triggering or resetting, if desired, applicable for a wide range of usage as monostable or astable multivibrators. The circuit consists of noise filter pin, shutdown control circuit, and zener diode for supply voltage regulation, providing excellent anti-noise characteristics.

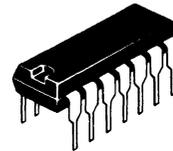
FEATURES

- Capable of forming a monostable multivibrator with a resistor and a capacitor
- Capable of forming an astable multivibrator with two resistors and a capacitor
- Supply voltage rejection ratio..... 0.01%V (typ.)
- Temperature coefficient..... 50ppm/°C (typ.)
- High output current..... 100mA (max.)
- Noise filter pin provided
- Discharge control, and shutdown control pins
- Built-in zener diode for supply voltage regulation
- Emitter and collector outputs

PIN CONFIGURATION (TOP VIEW)



NC : NO CONNECTION



14-pin molded plastic DIP

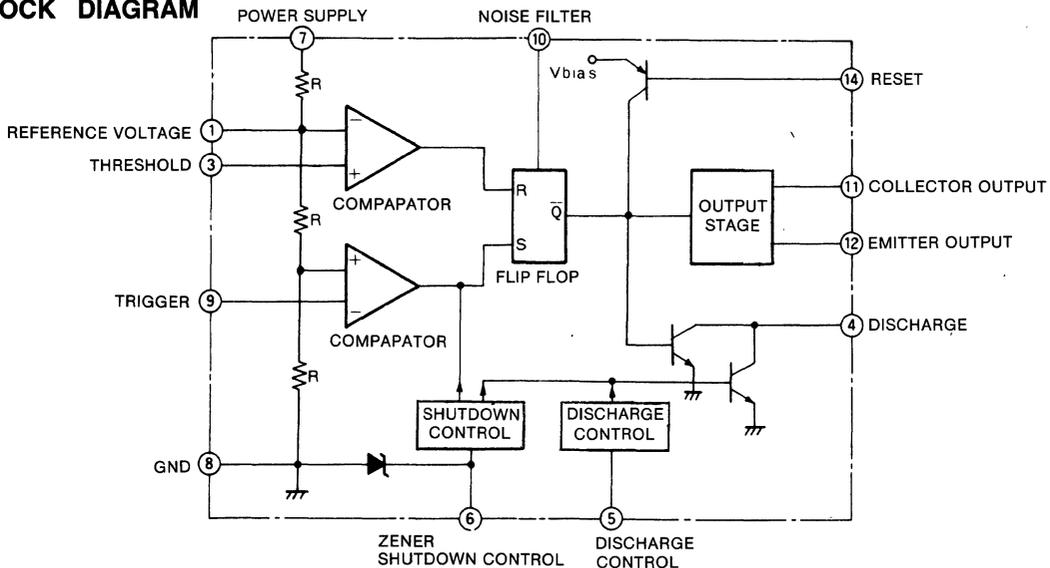
APPLICATION

Time delay generator (monostable multivibrator), pulse oscillator, pulsewidth modulation, pulse position modulation, sequential timer

RECOMMENDED OPERATING CONDITIONS

Supply voltage range..... 4.5~16V
 Rated supply voltage..... 5V, 12V, 15V

BLOCK DIAGRAM



ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

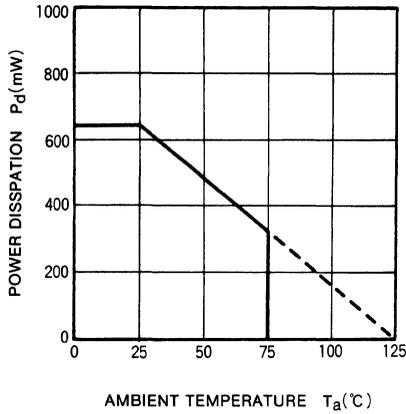
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage (pin②)		18	V
I_Z	Zener Current (pin⑥)		10	mA
I_{OC}	Collector output current	Saturation	200	mA
$BV_{(1)}$	Breakdown voltage (pin①)		27	V
P_d	Power dissipation		650	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	6.5	mW/ $^\circ\text{C}$
Topg	Operating temperature		-20~+75	$^\circ\text{C}$
Tstg	Storage temperature		-40~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$ $V_{CC}=5\sim 15\text{V}$)

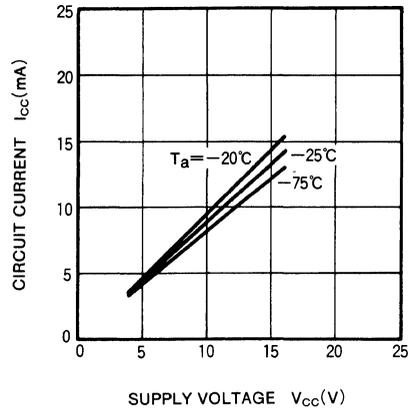
Symbol	Parameter	$V_{CC}(\text{V})$	Test Conditions	Limits			Unit
				Min	Typ	Max	
V_{CC}	Supply voltage		Excluding pin $R_L = \infty$	4.5		16	V
I_{CC}	Circuit current	5				4	10
		15			13	22	
V_{REF}	Reference voltage	5		2.6	3.33	4.0	V
		15		9	10	11	
V_{TH}	Threshold voltage				$2/3V_{CC}$		V
I_{TH}	Threshold current				0.1	0.25	μA
V_T	Trigger voltage				$1/3V_{CC}$		V
I_T	Trigger current				0.5	1.0	μA
V_R	Reset voltage				0.7	1.0	V
I_R	Reset current				0.1		mA
V_Z	Zener voltage		$I_Z=2\text{mA}$	6.5	7.5	8.5	V
V_{SC}	Voltage range of shutdown control at pin⑥				2.4	4.0	V
V_{DCC}	Voltage range of discharge control				$0.7+V_{CC}$	$1.0+V_{CC}$	V
V_{OC}	Collector output voltage	5	$I_{OC}=30\text{mA}$		0.15	0.3	V
		15	$I_{OC}=10\text{mA}$		0.05	0.1	
		15	$I_{OC}=100\text{mA}$		0.3	1.0	
I_{OE}	Emitter output current	15		1	2		mA
—	Timing accuracy				0.5		%
—	Temperature coefficient				50		ppm/ $^\circ\text{C}$
—	Supply voltage rejection ratio				0.01		%/V

TYPICAL CHARACTERISTICS (unless otherwise noted, $T_a=25^\circ\text{C}$)

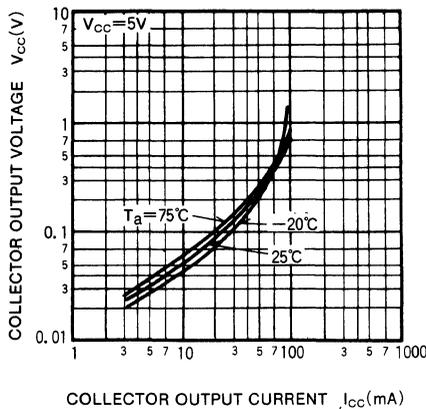
**THERMAL DERATING
(MAXIMUM RATING)**



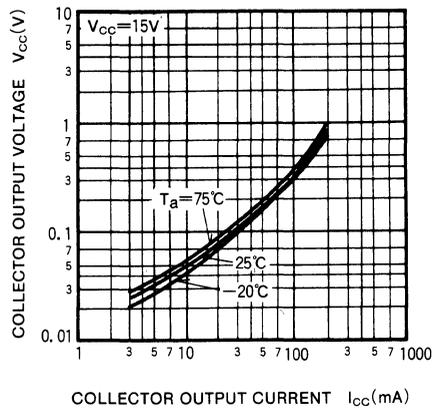
**CIRCUIT CURRENT
VS. SUPPLY VOLTAGE**



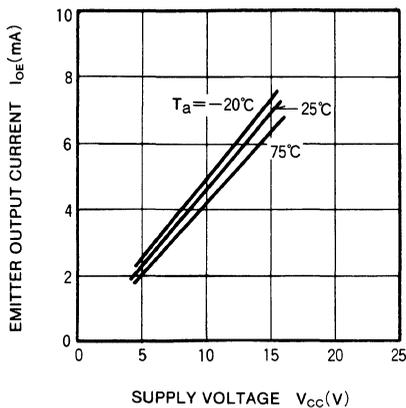
**COLLECTOR OUTPUT VOLTAGE
VS. OUTPUT CURRENT**



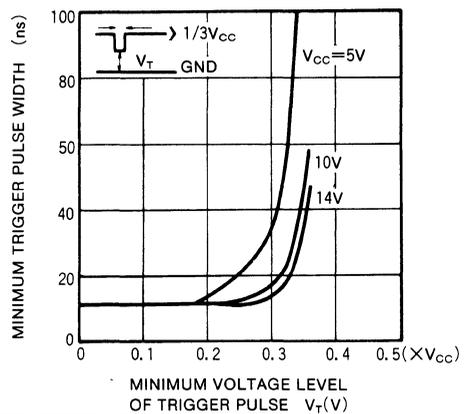
**COLLECTOR OUTPUT VOLTAGE
VS. OUTPUT CURRENT**

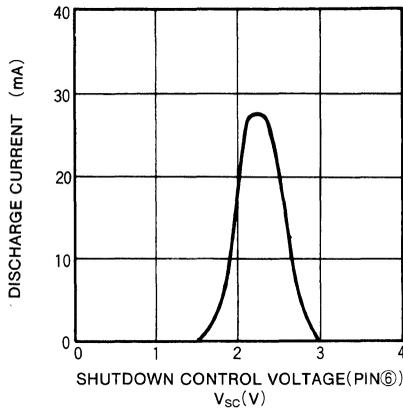
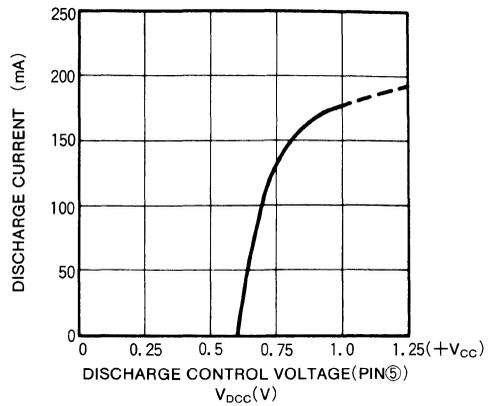


**EMITTER OUTPUT CURRENT
VS. SUPPLY VOLTAGE**



**MINIMUM PULSE WIDTH
VS. LOWEST VOLTAGE LEVEL
OF TRIGGER**



SHUTDOWN CONTROL
CHARACTERISTICSDISCHARGE CONTROL
CHARACTERISTICS

PIN DESCRIPTION

1. Reference voltage pin (pin ①)

The voltage at this pin is normally set at $2/3V_{CC}$. By applying reference voltage, delay time or oscillation frequency can be changed. The control signal for pulse-width modulation is applied at this pin. Connect a capacitor of $0.01\mu\text{F}$ between this pin and GND as the noise filter, if the pin is not used.

2. Threshold pin (pin ③)

The delay time is determined by CR time constants connected to this pin. The delay time of a monostable multivibrator t_d is expressed by the equation $t_d \approx 1.1R_A C_A$.

3. Discharge pin (pin ④)

The timed capacitor discharges at this pin. The discharge is enabled when a flip flop in the device is ON and the reset signal, discharge control signal or shutdown signal are applied. In a monostable multivibrator, the capacitor is connected to pin ③.

4. Discharge control pin (pin ⑤)

If a capacitor is connected between this pin and GND, the discharge pin (discharge transistor) is enabled by applying discharge control signal (approximately $V_{CC} + 0.7\text{V}$) to this pin. The timing capacitor can be automatically discharged when power supply drops. The capacitance should be $1/10$ - $1/30$ of that of the timing capacitor.

5. Zener/shutdown control pin (pin ⑥)

Supply voltage can be regulated (at approximately 8V) by connecting this pin to V_{CC} (pin ⑦). By controlling the voltage at this pin as the shutdown control voltage (approximately 2.4V), discharge of timing capacitor and OFF of the output stage can be controlled.

6. Trigger pin (pin ⑨)

When the voltage level at trigger pin is reduced to lower than $1/3V_{CC}$, timing operating is started.

7. Noise filter pin (pin ⑩)

Connect a capacitor between this pin and GND as the noise filter. Appropriate capacitance must be chosen according to the noise induced. (Normally less than $0.1\mu\text{F}$)

8. Collector output pin (pin ⑪)

The maximum load current of 200mA can be applied. Logic output level is in the low state from the high state when the output stage is ON.

9. Emitter output pin (pin ⑫)

This pin drives externally connected transistors or thyristors. The load current is 2mA typ.

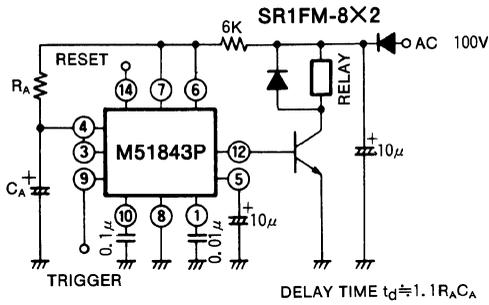
10. Reset pin (pin ⑭)

Timing operation can be interrupted by applying the reset signal to a reset pin. (The voltage at this pin must be less than 1V.)

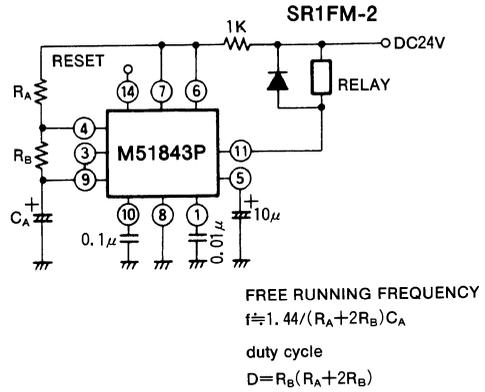
If the reset signal is applied, the output is in the low state and the reset condition is maintained as long as the signal is applied. The output stays in the low state until the reset signal is replaced with the set signal (trigger signal at a trigger pin). Connect this pin to V_{CC} if not used.

APPLICATION EXAMPLES

(1) DELAY TIMER



(2) ASTABLE OPERATION



M51848L,P

SINGLE TIMER

DESCRIPTION

The M51848L, P are semiconductor integrated circuits designed for producing accurate timing pulse and time delay. The delay time can be set from microseconds through minutes by externally connected resistor and capacitor. In an astable multivibrator, the maximum frequency is 100kHz. The reset voltage is 1.4V typ. ($T_a=25^\circ\text{C}$), and compatible with TTL level.

FEATURES

- Timing from microseconds through minutes
- Wide range of unage as monostable or astable multivibrators
- Maximum oscillation frequency 100kHz
- Reset voltage is compatible with TTL level
- Built-in supply voltage reset circuit for setting the output in the low state, when power is supplied

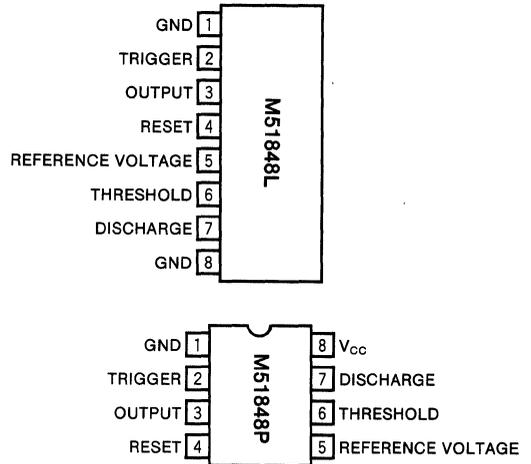
APPLICATION

Monostable multivibrator, astable multivibrator, pulsewidth modulation

RECOMMENDED OPERATING CONDITIONS

Supply voltage range 4.5~17V

PIN CONFIGURATION (TOP VIEW)

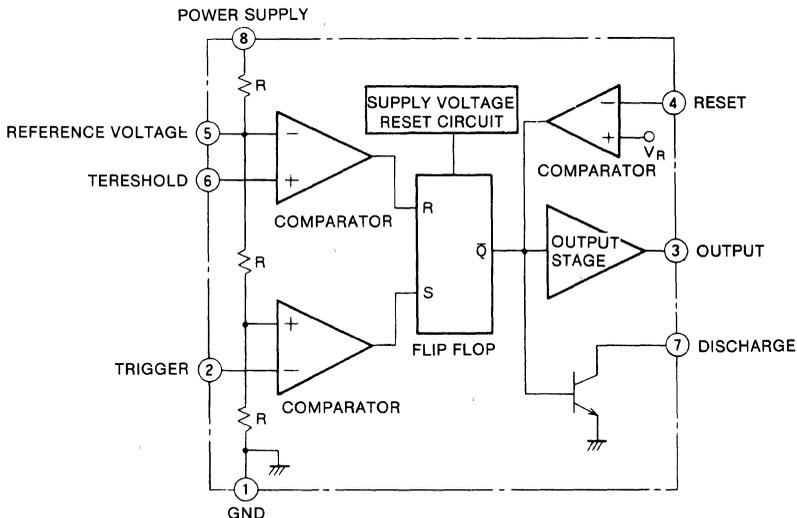


8-pin molded plastic DIP



8-pin molded plastic SIP

BLOCK DIAGRAM



SINGLE TIMER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

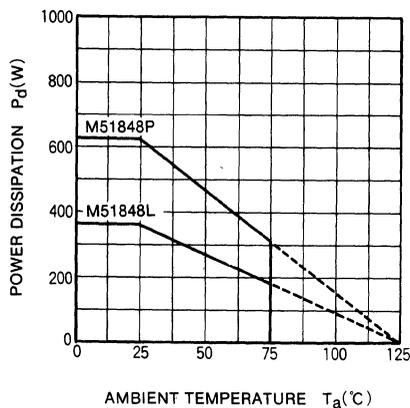
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
$I_{O(\text{peak})}$	Output current		200	mA
P_d	Power dissipation		360 (M51848L)	mW
			625 (M51848P)	
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	3.6 (M51848L)	mW/ $^\circ\text{C}$
			6.25 (M51848P)	
T_{opr}	Operating ambient temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-48 \sim +125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=5\sim 15\text{V}$, unless otherwise noted)

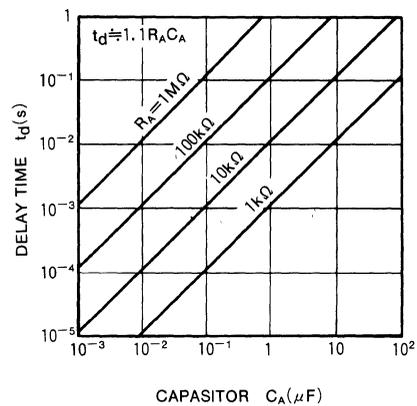
Symbol	Parameter	Test conditions	Limits			Units
			Min	Typ	Max	
V_{CC}	Supply voltage		4		17	V
I_{CC}	Circuit current	$V_{CC}=5\text{V}, R_L=\infty$		3	5.5	mA
		$V_{CC}=15\text{V}, R_L=\infty$		7	10	
V_{REF}	Reference voltage	$V_{CC}=5\text{V}$	2.6	3.33	4.0	V
		$V_{CC}=15\text{V}$	9	10	11	
V_{TH}	Threshold voltage			$2/3V_{CC}$	V	
I_{TH}	Threshold current			0.05	0.3	μA
V_T	Trigger voltage			$1/3V_{CC}$	V	
I_T	Trigger current			0.1	0.5	μA
V_R	Reset voltage		1.0	1.4	2.0	V
I_R	Reset current			0.05	0.2	μA
V_{OL}	Low output voltage	$V_{CC}=5\text{V}, I_{SINK}=5\text{mA}$		0.05	0.2	V
		$V_{CC}=15\text{V}, I_{SINK}=10\text{mA}$		0.05	0.2	
		$V_{CC}=15\text{V}, I_{SINK}=50\text{mA}$		0.2	0.5	
		$V_{CC}=15\text{V}, I_{SINK}=100\text{mA}$		0.5	2.0	
V_{OH}	High output voltage	$V_{CC}=5\text{V}, I_{SOURCE}=100\text{mA}$	2.8	3.3		V
		$V_{CC}=15\text{V}, I_{SOURCE}=100\text{mA}$	12.8	13.3		
—	Supply voltage rejection ratio			0.01		%
f_{max}	Maximum operating frequency	$R_A=R_B=2\text{k}\Omega, C_A=200\text{pF}$	100			KHz

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

THERMAL DERATING (MAXIMUM RATING)

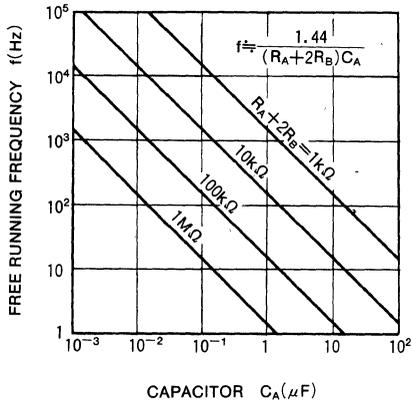


DELAY TIME VS. TIMING RESISTOR, CAPACITANCE

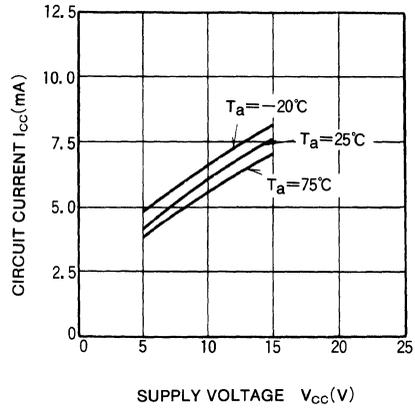


SINGLE TIMER

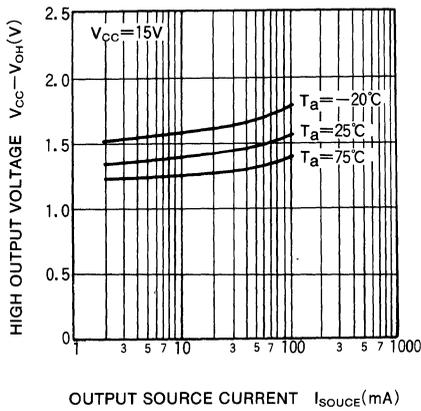
FREE RUNNING FREQUENCY VS. TIMING RESISTOR, CAPACITOR



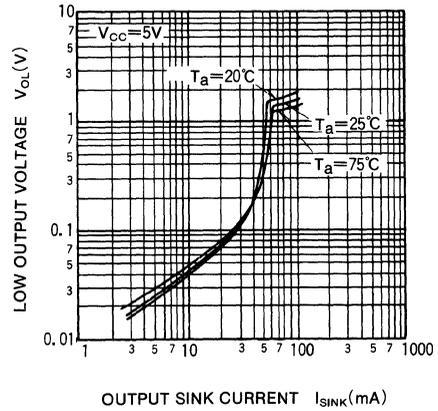
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



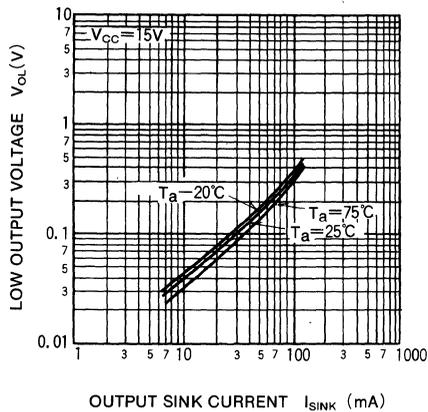
HIGH OUTPUT VOLTAGE VS. OUTPUT SOURCE CURRENT



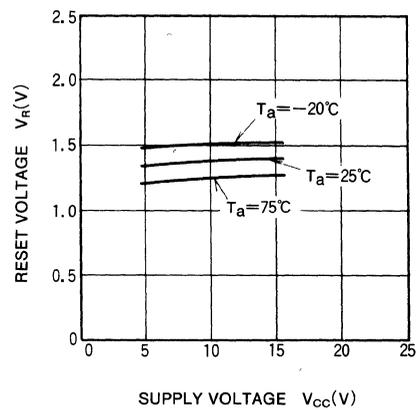
LOW OUTPUT VOLTAGE VS. OUTPUT SINK CURRENT



LOW OUTPUT VOLTAGE VS. OUTPUT SINK CURRENT



RESET VOLTAGE VS. SUPPLY VOLTAGE



PIN DESCRIPTION

1. Trigger pin (pin ②)

When the voltage at the trigger pin is reduced to lower than $1/3V_{CC}$, timing operation is started. Once triggered, the voltage is not affected during operation by the trigger voltage.

2. Output pin (pin ③)

Logic output level is normally in the low state but is in the high state during timing operation. The output circuit is shaped like a totem-pole and withstands maximum load current of 200mA. The circuit can directly drive the TTL or DTL circuits.

3. Reset pin (pin ④)

Timing operation can be interrupted by applying the reset signal to a reset pin. (The voltage at this pin must be less than 1.4V.)

If the reset signal is applied, the output is in the low state and the reset condition is maintained as long as the signal is applied. The output stays in the low state until the reset signal is removed and the set signal (trigger signal at a trigger pin) is applied. Connect this pin to V_{CC} if not used.

4. Reference voltage pin (pin ⑤)

The voltage at this pin is normally set at $2/3V_{CC}$. By applying reference voltage, delay time or oscillation frequency can be changed. The control signal for pulse-width modulation is applied through this pin. Connect a capacitor of $0.01 \mu F$ between this pin and GND as noise filter, if the pin not used.

5. Threshold pin (pin ⑥)

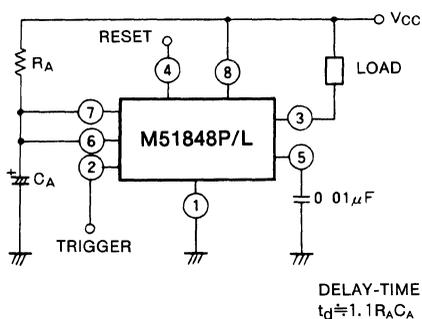
The delay time is determined by CR time constants connected to this pin. The delay time of a monostable multivibrator t_d is expressed by the equation $t_d \approx 1.1R_A C_A$.

6. Discharge pin (pin ⑦)

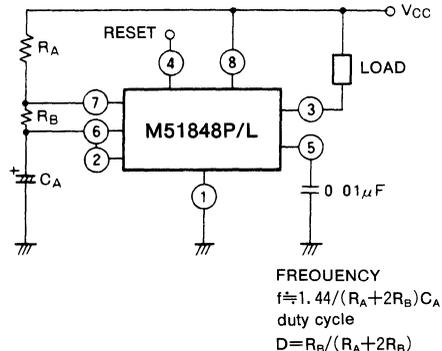
The timing capacitor discharges at this pin. The discharge is enabled when a flip flop in the device is ON and the reset signal is applied. In a monostable multivibrator, the capacitor is connected to pin ⑥.

APPLICATION EXAMPLES

(1) Delay-timer (monostable multivibrator)



(2) Pulse oscillator (astable multivibrator)



Precautions for us

1. Care must be taken not to connect the discharge pin (pin ⑦) directly to V_{CC} . Such connection causes short circuit of V_{CC} and GND through a discharge transistor (built into the device) and may destroy the integrated circuit.
2. Connect the reset pin (pin ④) to V_{CC} , if not used.

3. A capacitor connected to a reference voltage pin (pin ⑤) functions as the noise filter. Choose appropriate capacitance according to the noise induced. The M51843P is recommended for use in any environment where noise causes misoperation.

M51847P

DUAL TIMER

DESCRIPTION

The M51847P is a semiconductor integrated circuit consisting of circuits of the M51848P timer. The two circuits operate independently.

In an astable multivibrator, the maximum frequency is 100kHz.

The reset voltage is 1.4V typ. ($T_a=25^\circ\text{C}$), and compatible with TTL level.

FEATURES

- Timing from microseconds through minutes
- Wide range of usage as monostable or astable multivibrators
- Maximum oscillation frequency 100kHz
- Reset voltage is compatible with TTL level
- Built-in supply voltage reset circuit for setting the output in the low state, when power is supplied

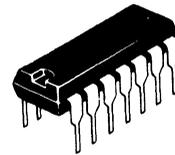
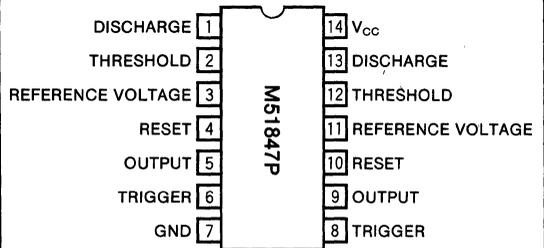
APPLICATION

Monostable multivibrator, astable multivibrator, pulsewidth modulation

RECOMMENDED OPERATING CONDITIONS

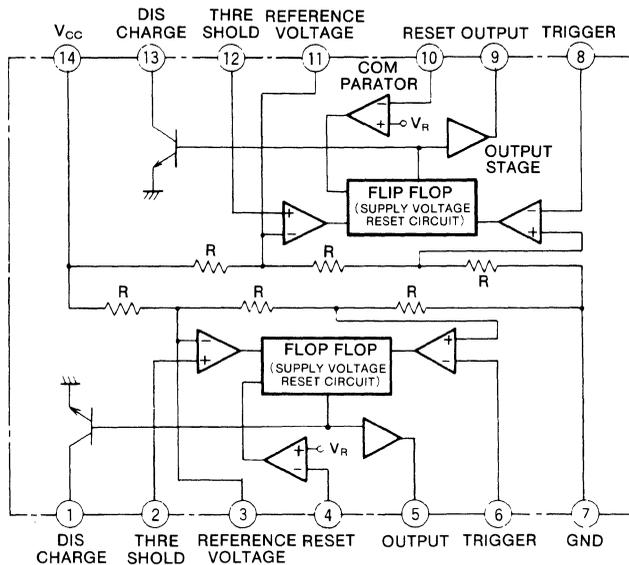
Supply voltage range 4.5~17V

PIN CONFIGURATION (TOP VIEW)



14-pin molded plastic DIP

BLOCK DIAGRAM



ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

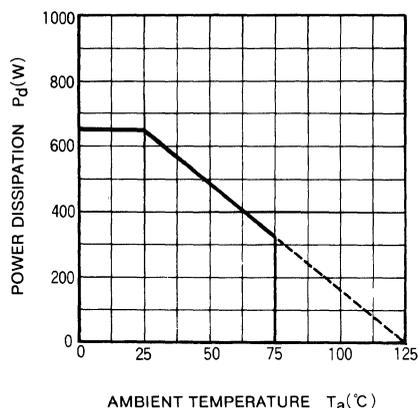
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
$I_{OC(peak)}$	Collector output current		100	mA
P_d	Power dissipation		650	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	6.5	mW/°C
T_{opr}	Operating ambient temperature		-20~+75	°C
T_{stg}	Storage temperature		-40~+125	°C

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=5\sim 15\text{V}$, unless otherwise noted)

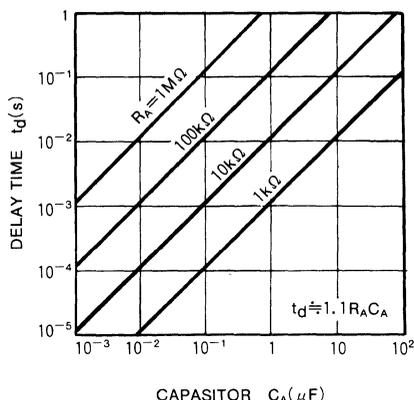
Symbol	Parameter	Test conditions	Limits			Units
			Min	Typ	Max	
V_{CC}	Supply voltage		4		17	V
I_{CC}	Circuit current	$V_{CC}=5\text{V}, R_L=\infty$		5	9	mA
		$V_{CC}=15\text{V}, R_L=\infty$		10	19	
V_{REF}	Reference voltage	$V_{CC}=5\text{V}$	2.6	3.33	4.0	V
		$V_{CC}=15\text{V}$	9	10	11	
V_{TH}	Threshold voltage			$2/3V_{CC}$		V
I_{TH}	Threshold current			0.03	0.2	μA
V_T	Trigger voltage			$1/3V_{CC}$		V
I_T	Trigger current			0.05	0.4	μA
V_R	Reset voltage		1.0	1.4	2.0	V
I_R	Reset current			0.05	0.2	μA
V_{OL}	Low output voltage	$V_{CC}=5\text{V}, I_{SINK}=5\text{mA}$		0.05	0.2	V
		$V_{CC}=15\text{V}, I_{SINK}=10\text{mA}$		0.05	0.2	
		$V_{CC}=15\text{V}, I_{SINK}=50\text{mA}$		0.2	0.5	
		$V_{CC}=15\text{V}, I_{SINK}=100\text{mA}$		1.0	2.0	
V_{OH}	High output voltage	$V_{CC}=5\text{V}, I_{SOURCE}=100\text{mA}$	2.8	3.3		V
		$V_{CC}=15\text{V}, I_{SOURCE}=100\text{mA}$	12.8	13.3		
f_{max}	Maximum operating frequency	$R_A=R_B=2\text{k}\Omega, C_A=200\text{pF}$	100			kHz

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

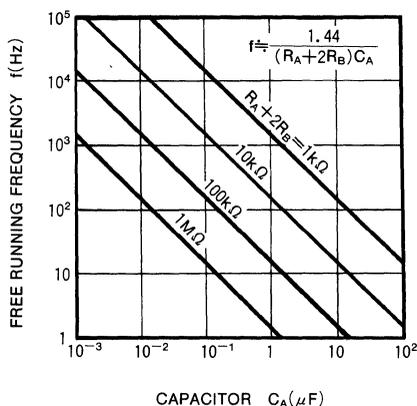
THERMAL DERATING (MAXIMUM RATING)



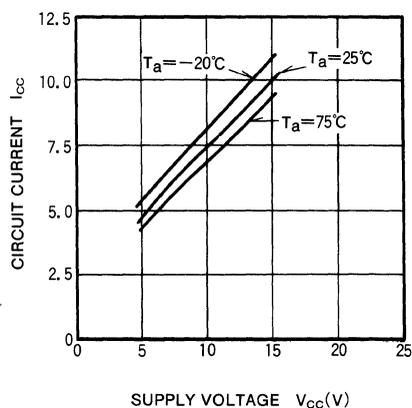
DELAY TIME VS. TIMING RESISTOR, CAPACITANCE



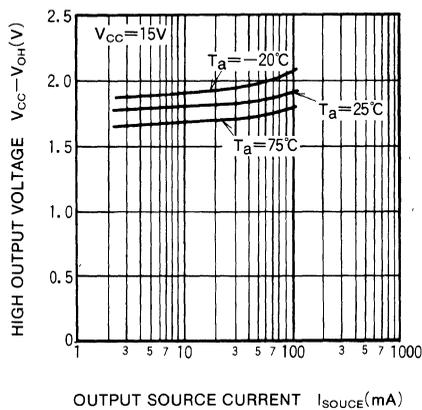
FREE RUNNING FREQUENCY VS. TIMING RESISTOR, CAPACITOR



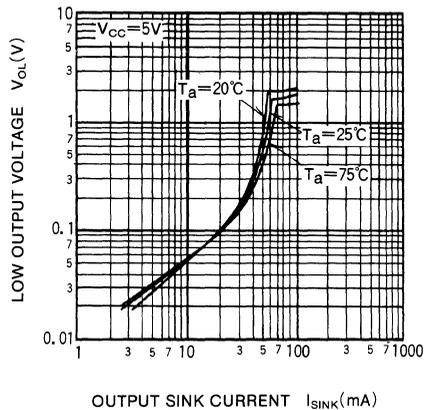
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



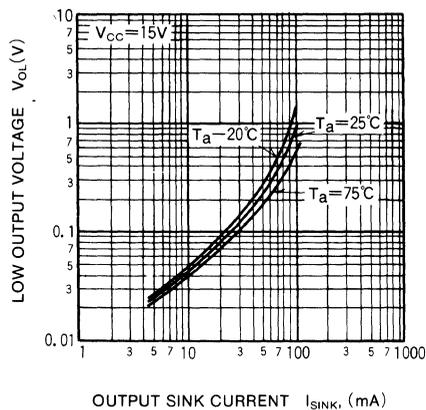
HIGH OUTPUT VOLTAGE VS. OUTPUT SOURCE CURRENT



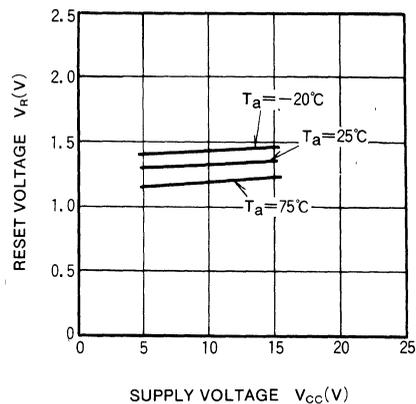
LOW OUTPUT VOLTAGE VS. OUTPUT SINK CURRENT



LOW OUTPUT VOLTAGE VS. OUTPUT SINK CURRENT



RESET VOLTAGE VS. SUPPLY VOLTAGE



MITSUBISHI LINEAR ICs
M52051P

SINGLE TIMER

DESCRIPTION

The M52051P is a semiconductor integrated circuit designed to produce an accurate timing pulse and time delay. The delay time can be set from microseconds through minutes by an externally connected resistor and capacitor.

In an astable multivibrator, the maximum frequency is 1MHz.

The reset voltage is 1.4V typ. ($T_a=25^\circ\text{C}$), and compatible with TTL level.

FEATURES

- Timing from microseconds through minutes
- Wide range of usage as monostable or astable multivibrators
- Maximum oscillation frequency1MHz
- Reset voltage is compatible with TTL level
- Built-in supply voltage reset circuit for setting the output in the low state, when power is supplied

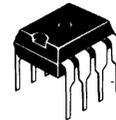
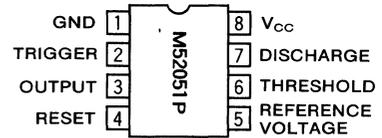
APPLICATION

Monostable multivibrator, astable multivibrator, pulsewidth modulation.

RECOMMENDED OPERATING CONDITIONS

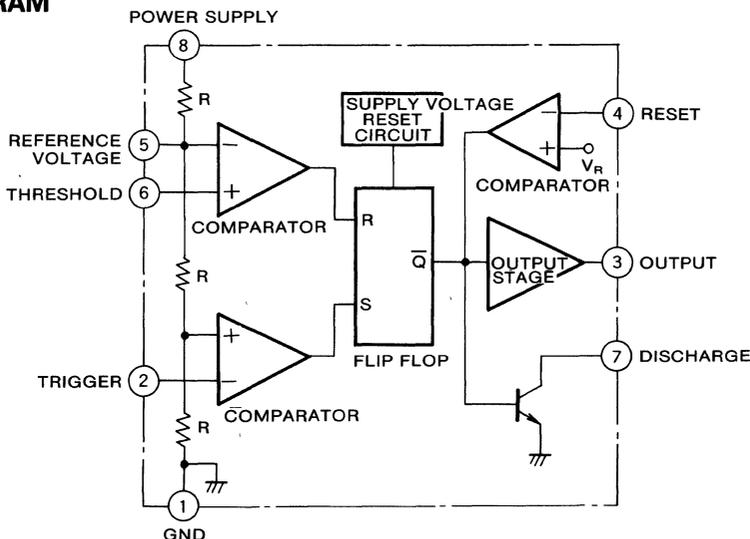
Supply voltage range4.5~5.5V

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic DIP

BLOCK DIAGRAM



ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

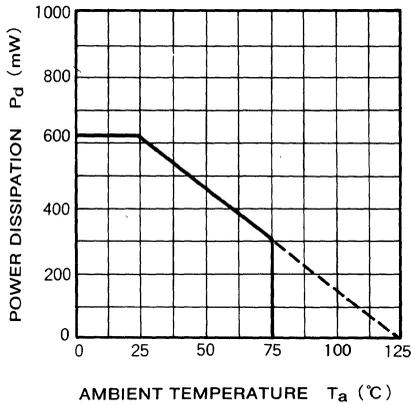
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		5.5	V
I_O	Output current		10	mA
P_d	Power dissipation		625	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	6.25	mW/ $^\circ\text{C}$
T_{opr}	Operating temperature		-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature		-40~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

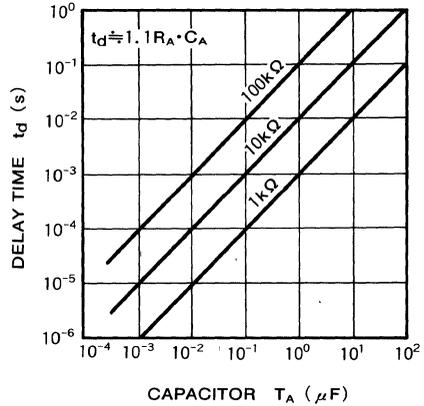
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4.5		5.5	V
I_{CC}	Circuit current	$V_{CC}=5\text{V}, R_L=\infty$	6.0	11.0	16.0	mA
V_{REF}	Reference voltage	$V_{CC}=5\text{V}$	3.0	3.33	3.7	V
V_{TH}	Threshold voltage	$V_{CC}=5\text{V}$	3.0	3.33	3.7	V
I_{TH}	Threshold current			2.0	20	μA
V_T	Trigger voltage	$V_{CC}=5\text{V}$	1.4	1.67	2.0	V
I_T	Trigger current			4.0	40	μA
V_R	Reset voltage		1.0	1.4	2.0	V
I_R	Reset current			4.0	40	μA
V_{OL}	Low output voltage	$V_{CC}=5\text{V}, I_{sink}=5\text{mA}$		0.3	0.5	V
V_{OH}	High output voltage	$V_{CC}=5\text{V}, I_{source}=10\text{mA}$	2.2	2.7		V
f_{max}	Maximum operating frequency		1			MHz
R_A	Resistance range for timing setting	Monostable multivibrator	0.5		100	k Ω
R_A+2R_B		Astable multivibrator	1		100	k Ω
C_A	Capacitance range for timing setting		3.3×10^{-4}		100	μF

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

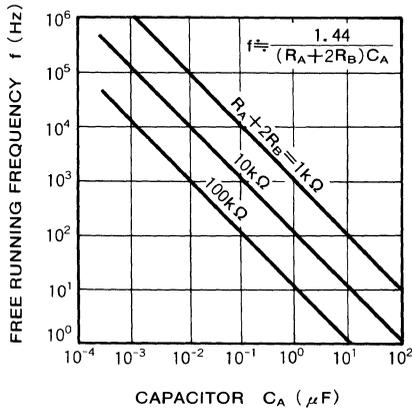
THERMAL DERATING (MAXIMUM RATING)



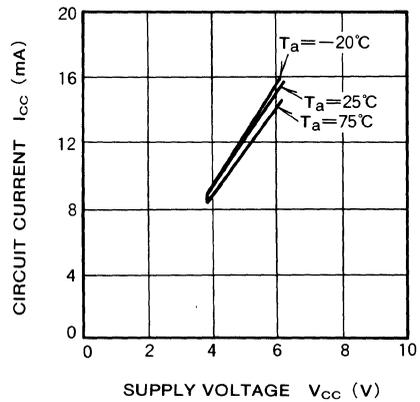
DELAY TIME VS. TIMING RESISTOR, CAPACITANCE



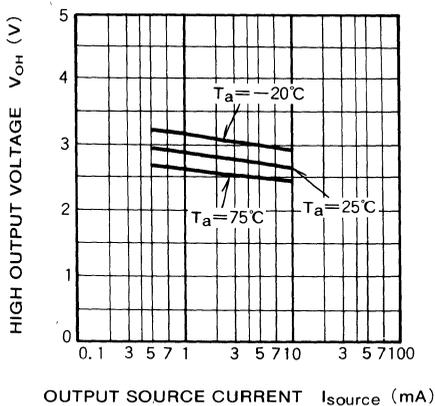
FREE RUNNING FREQUENCY VS. TIMING RESISTOR, CAPACITOR



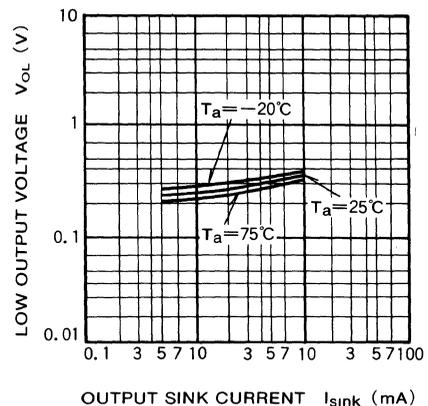
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



HIGH OUTPUT VOLTAGE VS. OUTPUT SOURCE CURRENT



LOW OUTPUT VOLTAGE VS. OUTPUT SINK CURRENT



PIN DESCRIPTION

1. Trigger pin (Pin ②)

When the voltage at the trigger pin is reduced to lower than $1/3V_{CC}$, timing operation is started. Once triggered, the voltage is not affected during operation by the trigger voltage.

2. Output pin (Pin ③)

Logic output level is normally in the low state but is in the high state during timing operation. The output circuit is shaped like a totem-pole and withstands maximum load current of 10mA. The circuit can directly drive the TTL or DTL circuits.

3. Reset pin (Pin ④)

Timing operation can be interrupted by applying the reset signal to a reset pin. (The voltage at this pin must be less than 1.4V).

If the reset signal is applied, the output is in the low state and the reset condition is maintained as long as the signal is applied. The output stays in the low state until the reset signal is removed and the set signal (trigger signal at a trigger pin) is applied. Connect this pin to V_{CC} if not used.

4. Reference voltage pin (pin ⑤)

The voltage at this pin is normally set at $2/3V_{CC}$. By applying reference voltage, delay time or oscillation frequency can be changed. The control signal for pulsewidth modulation is applied through this pin. Connect a capacitor of $0.01 \mu F$ between this pin and GND as the noise filter, if the pin is not used.

5. Threshold pin (Pin ⑥)

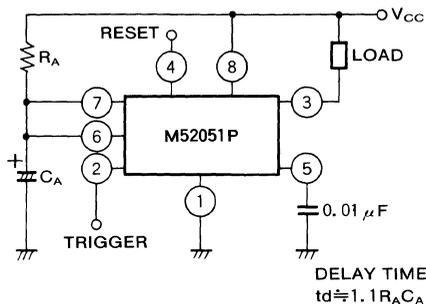
The delay time is determined by CR time constants connected to this pin. The delay time of a monostable multivibrator t_d is expressed by the equation $t_d \approx 1.1R_A C_A$.

6. Discharge pin (Pin ⑦)

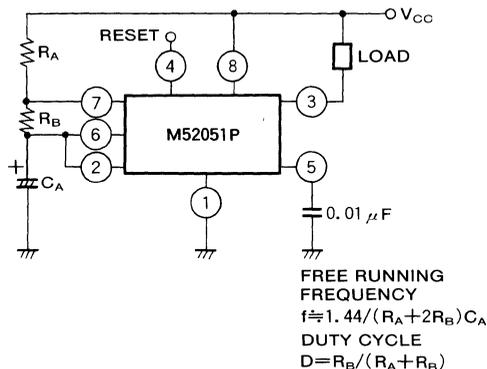
The timing capacitor discharges at this pin. The discharge is enabled when a flip flop in the device is ON and the reset signal is applied. In a monostable multivibrator, the capacitor is connected to Pin ⑥.

APPLICATION EXAMPLE

**(1) Delay timer
(monostable multivibrator)**



**(2) Pulse oscillator
(astable multivibrator)**



PRECAUTIONS FOR USE

1. Care must be taken not to connect the discharge pin (Pin ⑦) directly to V_{CC} . Such connection causes short circuit of V_{CC} and GND through a discharge transistor (built into the device) and may destroy the integrated circuit.

2. Connect the reset pin (Pin ④) to V_{CC} , if not used.
3. A capacitor connected to a reference voltage pin (Pin ⑤) functions as the noise filter. Choose appropriate capacitance according to the noise induced. The M51843P is recommended for use in an environment where the noise causes misoperation.

M51849L

COUNTER TIMER

DESCRIPTION

The M51849L is a semiconductor integrated circuit designed for controller of long time delay, consisting of 11 stage divider by ILL. The time base period of oscillator is extended 1024 times, determining by the 11 stage divider by ILL, and the maximum output period is 50 hours.

FEATURES

- Timing from 100ms through 50 hr
- High current output can sink 30mA(peak)
- Built-in stabilization zener
- Built-in power on reset

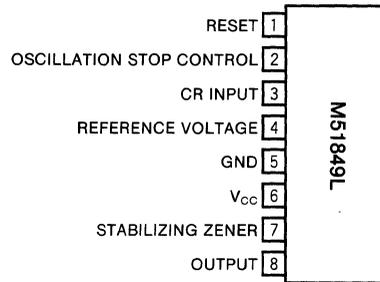
APPLICATION

- Precision timing, time delay generation, Ultra-low-frequency oscillator

RECOMMENDED OPERATING CONDITIONS

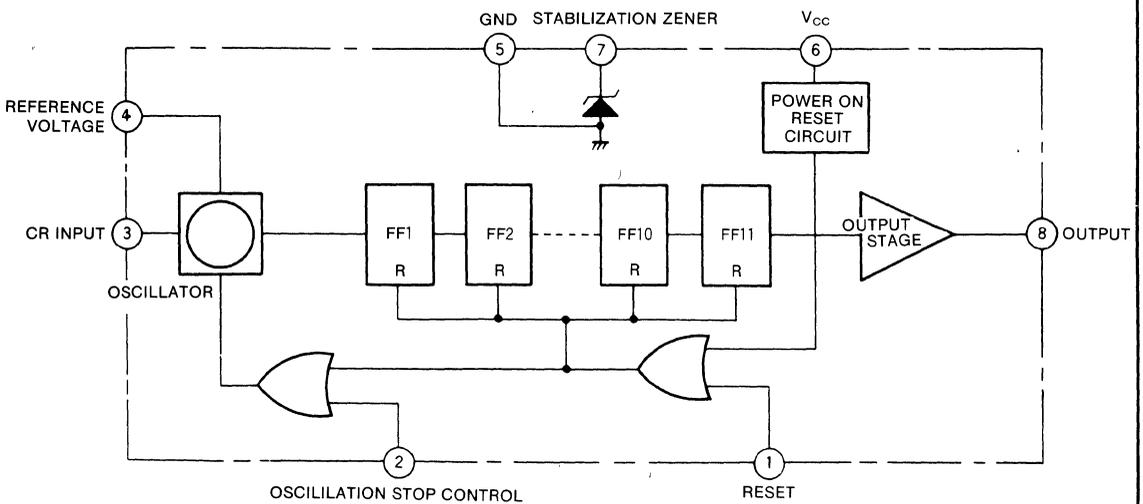
Supply voltage.....5V~V_Z(V_Z=pin ⑦ Zener voltage)
 Rated supply voltage.....6V±10%

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic SIP

BLOCK DIAGRAM



COUNTER TIMER**ABSOLUTE MAXIMUM RATINGS** ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		7.0	V
I_Z	Zener current		20	mA
I_O	Output sink current		30	mA
P_d	Power dissipation		360	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	3.6	mW/ $^\circ\text{C}$
T_{opr}	Operating temperature		0~+60	$^\circ\text{C}$
T_{stg}	Storage temperature		-40~+125	$^\circ\text{C}$

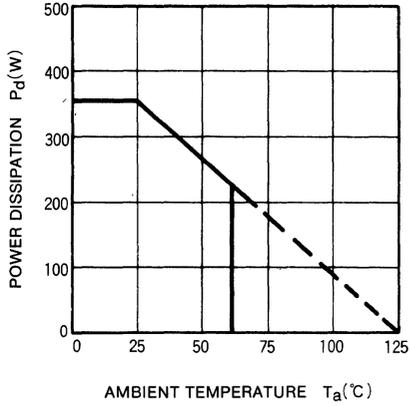
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=6\text{V}$ unless otherwise noted)

Symbol	Parameter	Test condition	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{CC}=6.0\text{V}$		3.5	6.0	mA
V_Z	Zener voltage	$I_Z=1\text{mA}$	6.0	6.4	6.8	V
V_R	Reset voltage			1.5	1.8	V
I_R	Reset current			0.1		mA
V_{OS}	Oscillation stop voltage			1.6	1.9	V
I_{OS}	Oscillation stop input current			0.1		mA
V_{OH}	High output voltage	$I_{SOURCE}=10\text{mA}$	4.0	4.5		V
V_{OL}	Low output voltage	$I_{SINK}=10\text{mA}$		0.2	0.6	V

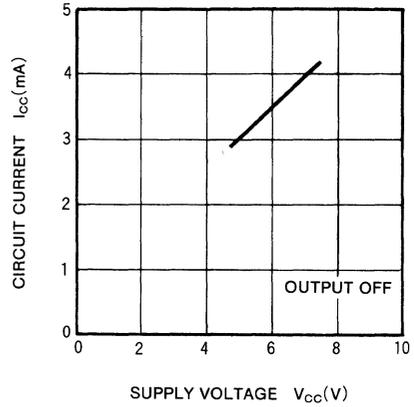
COUNTER TIMER

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

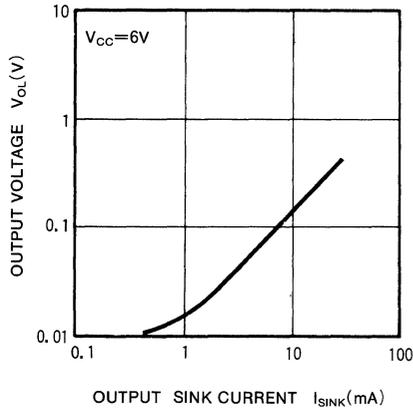
**THERMAL DERATING
(MAXIMUM RATING)**



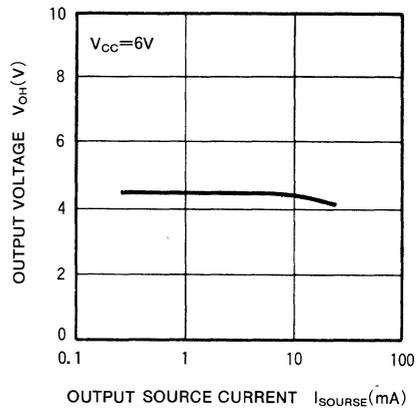
**CIRCUIT CURRENT VS.
SUPPLY VOLTAGE**



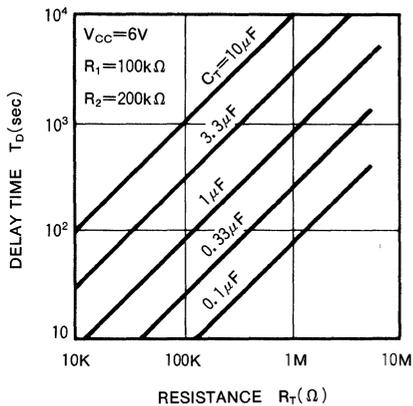
**OUTPUT VOLTAGE VS.
OUTPUT SINK CURRENT**



**OUTPUT VOLTAGE VS.
OUTPUT SOURCE CURRENT**



RESISTANCE VS. DELAY TIME



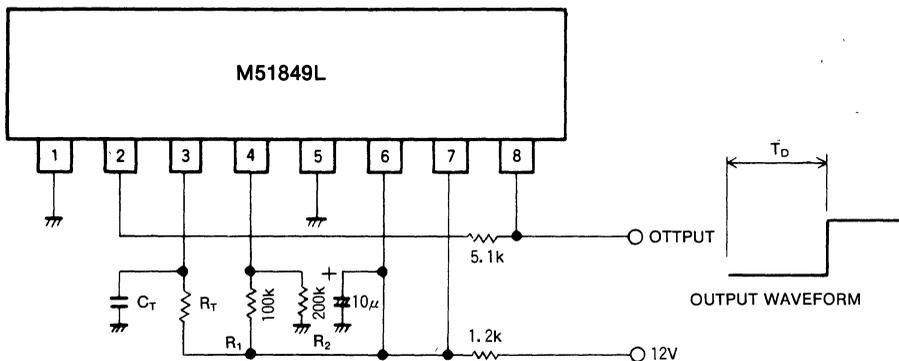
PIN DESCRIPTION

1. **Reset pin (pin ①)**
This pin is used to stop counter operation. If the pin is in the high state, the counter is cleared and oscillation is stopped. Connect this pin to GND, if not used.
2. **Oscillation stop control pin (pin ②)**
If the pin is in the high state, the oscillation stops. But the status of counter is not maintained, and therefore, if the pin is reset to the low state, timing operation is resumed. Connect the pin to GND, if not used.
3. **CR input pin (pin ③)**
The capacitor C_T and resistor R_T are connected to this pin. The oscillation period is given in the following equation. The resistor R_T should be 1k Ω or above.
4. **Reference voltage pin (pin ④)**
The highest voltage for the oscillation level is supplied at this pin. The voltage should be set at approximately 2/3

- V_{CC} by a variable resistor or resistor division (R_1, R_2). The capacitor C_T is corrected by minutely adjusting this voltage by a variable resistor. The voltage can be set between 0.4V to 0.8V. The oscillator may not function if the voltage is outside this range.
5. **Stabilization zener pin (pin ⑦)**
A zener diode of approximately 6.4V is connected between this pin and pin ⑤ (GND). Set the resistance so that the zener current is 5mA.
6. **Output pin (pin ⑧)**
The output voltage changes from the low state to the high state when the oscillation period is 1024 times of T_D , and the voltage returns from the high state to the low state when the period is 2048 times for one cycle. The voltage can sink or source up to 10mA (typ.). Care must be taken in power supply variation in the integrated circuit when the output load current is large.

APPLICATION EXAMPLE

Monostable operation



The delay time in the above application example is given by :

$$T_D \approx 900C_T \cdot R_T (\text{sec})$$

The timing resistor R_T should be 1k Ω or above

Precautions for use

1. CR input pin (pin ③) must not be directly connected to V_{CC} to avoid destruction of the integrated circuit.
2. The voltage at the reference voltage pin (pin ④) should be set between 0.4V to 0.8V. The oscillator may not function if the voltage is outside this range.
3. This integrated operates at ultra-low supply voltage. Therefore it is very sensitive to external noise and is subject to misoperation. A capacitor must be connected near the V_{CC} pin (pin ⑥) to avoid noise.
4. The specifications are subject to change.

M58479P/M58482P

CMOS COUNTER/TIMERS

DESCRIPTION

The M58479P and M58482P are electronic timer ICs developed by aluminum-gate CMOS technology. Use of these ICs makes possible timer devices without mechanical elements, which have reduced power dissipation, superior reliability, and higher noise immunity. The M58479P is specifically designed for high noise immunity, while the M58482P particularly features low power dissipation.

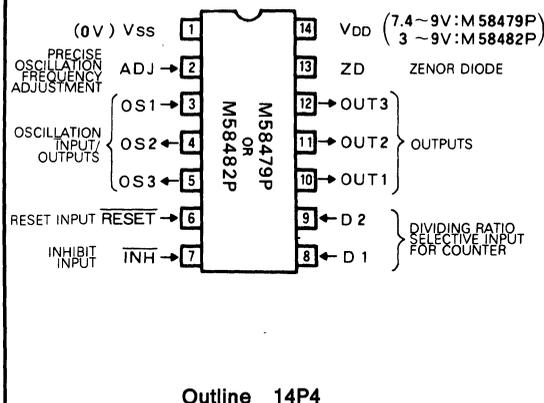
FEATURES

- Low power dissipation
M58479P: 2mW (typ.), 7.5mW (max.)
M58482P: 200 μ W (typ.), 750 μ W (max.)
- Superior noise immunity
- Single power supply with a zenor diode
- Internal RC oscillator
- Precise oscillation frequency regulating capability
- Extremely broad time-delay range (50ms~4800h)
- Time-delay settable to 10, 60, or 600 times fundamental time (1024 times oscillation period)
- M58479P has automatic-reset function during power engagement
- Built-in reset and inhibit functions
- Residual time display possible by adding Mitsubishi's M53290P and M53242P IC

APPLICATION

- Electronic timer or counter with broad time-delay range (50ms~4800h)

PIN CONFIGURATION (TOP VIEW)

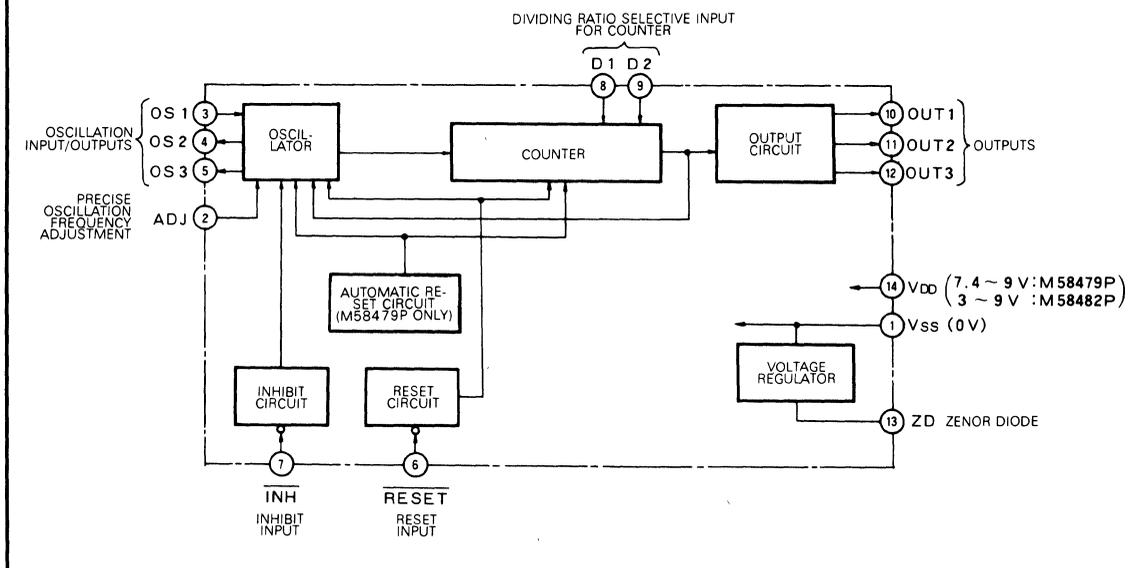


FUNCTIONAL DESCRIPTION

These devices make possible extremely long clock performance, by counting pulse signals from the RC oscillator. It has precise oscillation frequency adjustment, automatic-reset, reset, and inhibit functions.

There are three outputs. When the time duration is up, OUT1 turns from low to high and OUT2 from high to low. OUT3 can be connected to M53290P and M53242P TTLs for residual time display.

BLOCK DIAGRAM



CMOS COUNTER/TIMERS

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Conditions	Ratings	Unit
V _{DD}	Supply voltage	With respect to V _{SS}	-0.3 ~ 9.5	V
V _I	Input voltage		V _{SS} ≤ V _I ≤ V _{DD}	V
P _d	Maximum power dissipation	T _a = 25°C	250	mW
T _{opr}	Operating free-air temperature range		-30 ~ 75	°C
T _{stg}	Storage temperature range		-40 ~ 125	°C

RECOMMENDED OPERATING CONDITIONS (T_a = -30 ~ 75°C, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V _{DD}	Supply voltage	M58479P	7.4	9	V
		M58482P	3	9	V
I _{ZD}	Zenor current			10	mA
R _{FC}	Feedback resistance	0.005		10	MΩ
C _{FC}	Oscillation capacitance	0.001		1	μF
R _{FC}	Resistance for fine-adjustment of oscillation frequency	0		100	kΩ
V _{IH}	High-level input voltage, RESET, INH, D ₁ , D ₂	0.7 × V _{DD}	V _{DD}	V _{DD}	V
V _{IL}	Low-level input voltage, RESET, INH, D ₁ , D ₂	0	0	0.3 × V _{DD}	V

ELECTRICAL CHARACTERISTICS (T_a = 25°C, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{ZD}	Zenor voltage	I _{ZD} = 2 mA	7.4	8.2	9	V
		I _{ZD} = 10 mA	7.5	8.2	9	V
I _{DD}	Supply current	M58479P V _{DD} = 7.5V, C _{FC} = 0.01 μF, R _{FC} = 1MΩ R _{ADJ} = 0Ω, Input/output open		0.25	1	mA
		M58482P V _{DD} = 7.5V, C _{FC} = 0.01 μF, R _{FC} = 1MΩ R _{ADJ} = 0Ω, Input/output open		25	100	μA
V _{RE}	Supply voltage at the time of automatic-reset release	M58479P	3	1	5.4	V
V _{TR}	Transition voltage of first inverter in the oscillator	V _{DD} = 7.5V, R _{ADJ} = 0Ω	2	9	4.8	V
R _I	Pull-up resistance RESET, INH, D ₁ , D ₂ inputs	M58479P	10	20	30	kΩ
		M58482P	25	50	75	
I _{OH}	High-level output current OUT1 and OUT2 outputs	V _{DD} = 7.5V, V _O = 0V	-5	-10		mA
I _{OL}	Low-level output current OUT1, OUT2 and OUT3 outputs	V _{DD} = 7.5V, V _O = 7.5V	10	20		mA
I _{OZH}	Off-state output current OUT3 output	V _{DD} = 7.5V, V _O = 7.5V			1	μA
I _{OL}	Low-level output current OUT1, OUT2 and OUT3 outputs	V _{DD} = 7.5V, V _O = 0.4V	1.6			mA
I _{OL}	Low-level output current OUT1, OUT2 and OUT3 outputs	M58482P V _{DD} = 4.5V, V _O = 0.4V	1.6			mA
V _{OL}	Low level output voltage OUT1, OUT2 and OUT3 outputs	V _{DD} = 7.5V			0.1	V

FUNCTIONAL DESCRIPTION

Voltage Regulator

A zenor diode is on-chip, making it easy to obtain a constant voltage regulator circuit. Since the zenor diode terminal (ZD) is independent of the power terminal (V_{DD}), it can be used as a constant voltage power supply for the total system.

Oscillator

Oscillation is obtained by connecting an external resistor (feedback resistor R_{FC}) between terminals OS1 and OS3 and an external capacitor (oscillation capacitor C_{FC}) between terminals OS1 and OS2. The values of the external resistor and capacitor can then be changed to vary the oscillation period and thus change the time delay. Oscillation period T_0 is obtained by the following equation:

$$T_0 = -R_{FC} \cdot C_{FC} \left\{ \ln \frac{V_{TR}}{V_{DD} + V_{BE}} + \ln \frac{V_{DD} - V_{TR}}{V_{DD} + V_{BE}} \right\} \dots (1)$$

Where,

R_{FC} : Resistance of external resistor

C_{FC} : Capacitance of external capacitor

V_{TR} : Transition voltage of the first inverter in the oscillation circuit

V_{DD} : Supply voltage

V_{BE} : Forward rising voltage of the diode in terminal OS1 (0.3~0.7V)

Automatic-Reset Function

The M58479P has a power-supply voltage-detection circuit on-chip, so that the counter is automatically reset by the rising edge of the supply voltage when power is turned on. The reset is then released, making the oscillator ready to function and the counter ready to start counting.

The M58482P can also be provided with the same automatic-reset function by connecting capacitor between terminals $\overline{\text{RESET}}$ and V_{SS} .

Reset Function

When the $\overline{\text{RESET}}$ input turns low (V_{SS}), oscillation of the oscillator can be stopped and the counter reset.

Inhibit Function

When terminal $\overline{\text{INH}}$ turns low (V_{SS}) while the timer is in action, the oscillation halts. When input $\overline{\text{INH}}$ is turned high or returned to OPEN afterwards, it starts to count residual time.

Counter

This counter consists of an 11-stage 1/2 frequency divider, a 2-stage 1/10 frequency divider and a 1-stage 1/6 frequency divider. As shown in the table below, timer duration can be changed by varying the number of pulses counted according to the combination of the input levels on terminals D1 and D2.

D1	D2	Number of pulses counted	Time delay	Typical time delay applied
H	H	1024	T_1	1 min
L	H	1024×10	$T_1 \times 10$	10 min
H	L	$1024 \times 10 \times 6$	$T_1 \times 10 \times 6$	1 h
L	L	$1024 \times 10 \times 6 \times 10$	$T_1 \times 10 \times 6 \times 10$	10 h

Where, $T_1 = T_0 \times 1024$

T_0 is the value obtained from equation (1)

Output Circuits

The chips have three outputs: OUT1 changes from low to high and OUT2 from high to low as soon as the time duration is up. Either can be used to drive a transistor by connecting it to the transistor base. OUT1 can drive a thyristor when connected to the thyristor gate.

OUT3 is an open-drain output with period 1/8 of the time delay, and can be used to drive a TTL in a separate (5V) power supply line. Thus, if a M53290P counter and a M53242P binary-to-decimal decoder are connected to OUT3, with their output connected to a light-emitting diode, residual time will be displayed on the LED. When not in use, OUT3 should be connected to V_{SS} .

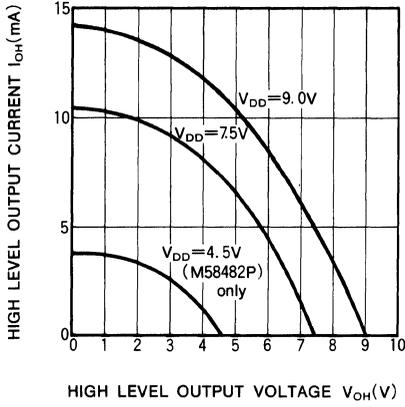
Fine Adjustment of Oscillation Period

A variable resistor can be connected between terminals ADJ and V_{SS} , enabling precise adjustment of the period of the oscillator. However, when not used for fine adjustment, ADJ should be connected to V_{SS} .

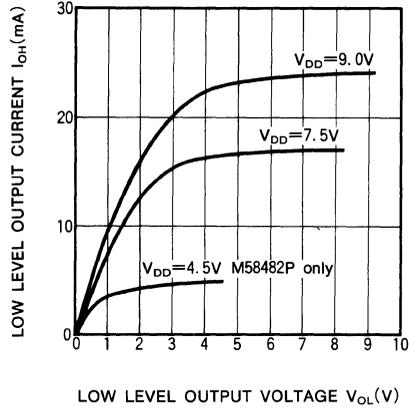
CMOS COUNTER/TIMERS

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

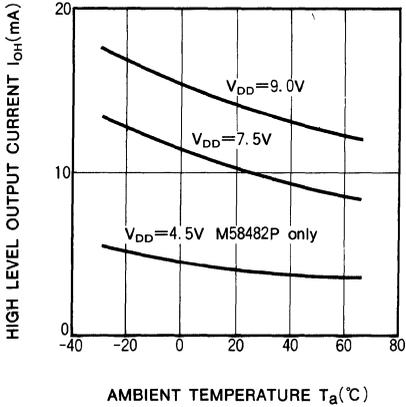
See "9. ELECTRICAL CHARACTERISTICS" for absolute values
 $I_{OH}-V_{OH}$ (OUT 1, OUT 2)



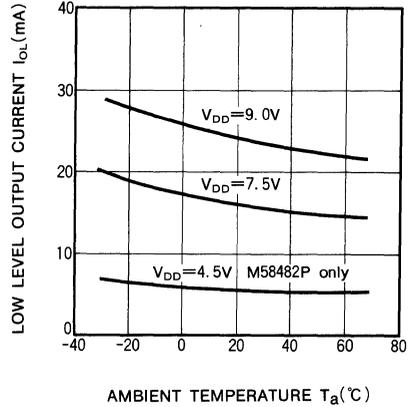
$I_{OH}-V_{OL}$ (OUT 1, OUT 2, OUT 3)



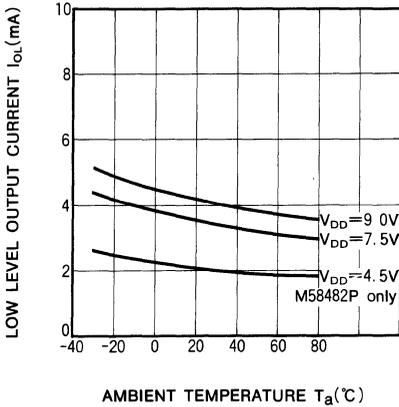
$I_{OH}-T_a$ (OUT 1, OUT 2) $V_{OH} = 0\text{V}$



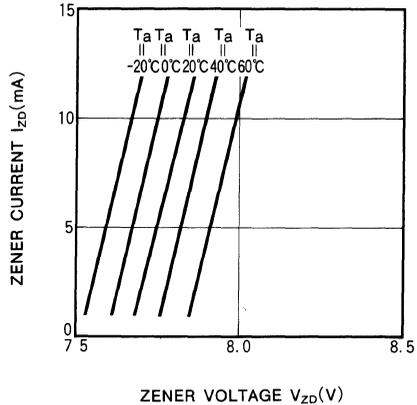
$I_{OL}-T_a$ (OUT 1, OUT 2, OUT 3)
 $V_{OL} = V_{DD}$



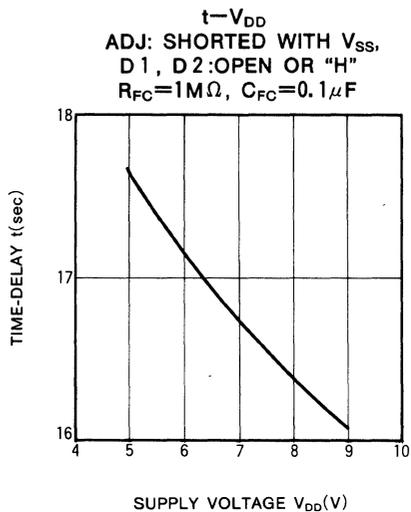
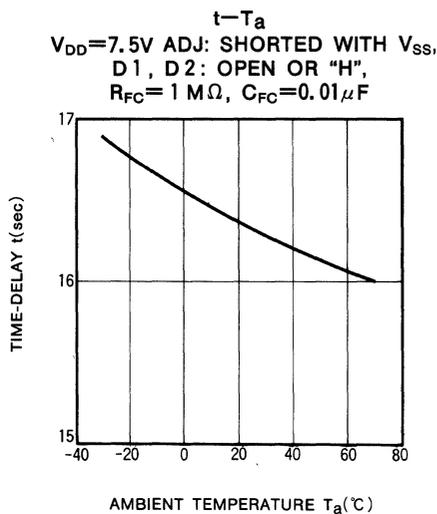
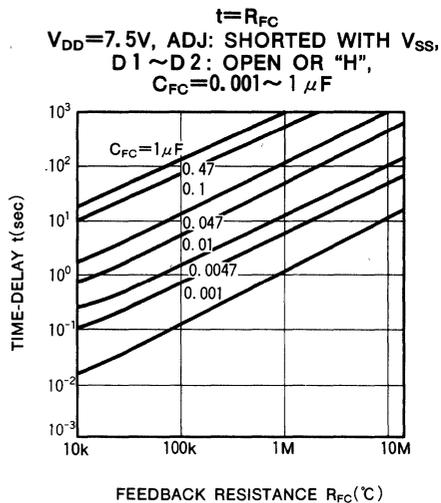
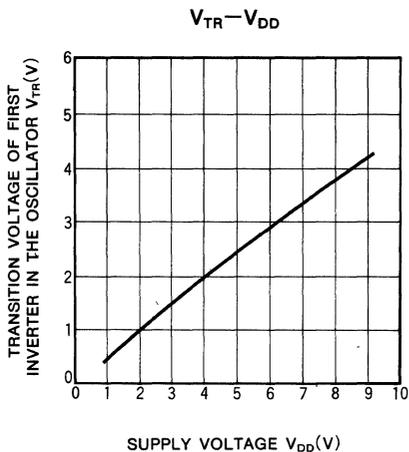
$I_{OL}-T_a$ (OUT 1, OUT 2, OUT 3)
 $V_{OL} = 0.4\text{V}$



$I_{ZD}-V_{ZD}$



CMOS COUNTER/TIMERS



OSCILLATION FREQUENCY

The oscillation period of M58479P and M58482P are formula-
 rized as follows.

$$T_O = -R_{FC}C_{FC} \left\{ \ell n \frac{V_{TR}}{V_{DD}+V_{BE}} + \ell n \frac{V_{DD}-V_{TR}}{V_{DD}+V_{BE}} \right\} \quad (1)$$

The value in { } of (1) takes the maximum value at $V_{TR} = V_{DD}/2$. For example, under the condition of $V_{DD} = 7.5V$, the relation of the V_{TR} and the value in { } is shown in Figure 1.

Regarding the Figure 1, the value in { } of (1) at $V_{DD} = 7.5V$ is, in a range of $V_{BE} = 0.3 \sim 0.7V$ and $V_{TR} = 2.9 \sim 4.8V$, $-1.647 \sim -1.464$.

The oscillation period can be figured out theoretically by the (1) formula; however, as the oscillation is executed by the charge and discharge of R_{FC} , C_{FC} , the correction parameter R_{FC} by the output impedance of OS 2 and OS 3 is added in the (1) formula as:

$$T_O = -(R_{FC} + \Delta R_{FC})C_{FC} \left\{ \ell n \frac{V_{TR}}{V_{DD}+V_{BE}} + \ell n \frac{V_{DD}-V_{TR}}{V_{DD}+V_{BE}} \right\} \quad (2)$$

At this time, the value of the correction parameter ΔR_{FC} will be around $5.5 \pm 2.5k\Omega$.

For the circuit designing, set the oscillation constant regarding to the above matters.

TIMER ADJUSTMENT

Following is the method of adjusting time-delay keeping the external resistance R_{FC} and capacitor C_{FC} fixed.

(1) The method to verify R_{ADJ} value with inserting the parallelly connected R_{ADJ} and C_{ADJ} into ADJ-VSS

As described already, the oscillation period T_O is calculated with (1) formula, as the relation of V_{TR} and the minimum value when $V_{TR} = V_{DD}/2$. This means the T_O can be varied by changing the V_{TR} value. This method is performed by adjusting the time-delay by the V_{TR} .

The ADJ is connected to a N-Channel-FET source of the first inverter of oscillator as Figure 3 illustrates. When the parallelly connected resistance R_{ADJ} and capacitance C_{ADJ} are inserted between ADJ and V_{SS} and the R_{ADJ} changes its value, the voltage of the ADJ varies by the current in the R_{ADJ} , and this results the change of V_{TR} .

As the R_{ADJ} value gets larger, the value of the V_{TR} is increased from that at $R_{ADJ} = 0\Omega$. The value of V_{TR} at $R_{ADJ} = 0\Omega$ is in the range of $2.9 \sim 4.8V$ ($V_{DD} = 7.5V$). Therefore, as Figure 2 indicates, the variation way and the variation rate of the oscillation period T_O when the resistance R_{ADJ} gets larger are found according to the V_{TR} value at $R_{ADJ} = 0\Omega$ and are not constant.

The capacitance C_{ADJ} to be parallelly inserted into the resistance R_{ADJ} has a function of making a variation rate of the T_O toward R_{ADJ} large.

On the resistance R_{ADJ} and the capacitance C_{ADJ} , please follow the ranges below.

$$R_{ADJ} = 0 \sim 100k\Omega$$

$$C_{ADJ} = 100 \sim 1000pF$$

When the ADJ is not used for the oscillation period adjustment, short to the V_{SS} .

Fig. 1. V_{TR} VS $\ell n \frac{V_{TR}}{V_{DD}+V_{BE}} + \ell n \frac{V_{DD}-V_{TR}}{V_{DD}+V_{BE}}$

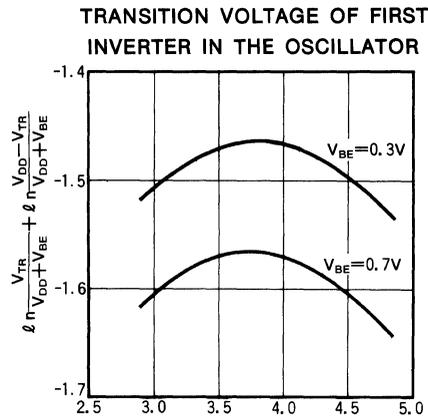


Fig. 2. Oscillation period (T_O) VS Transition Voltage (V_{TR})

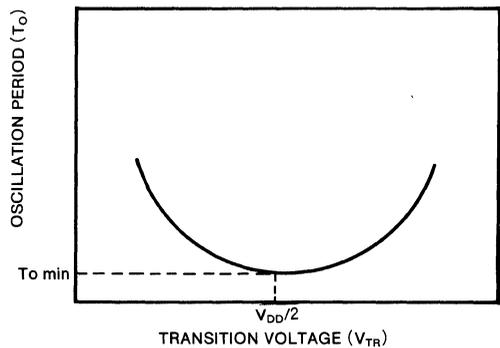
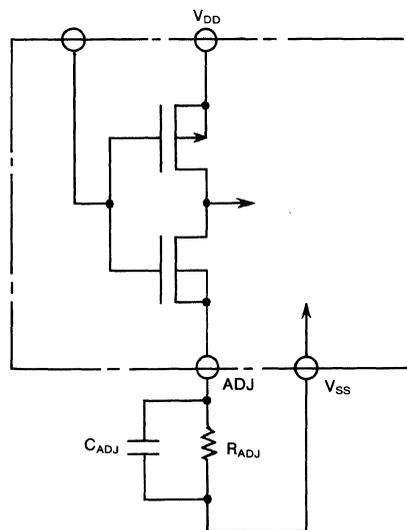


Fig.3 External connection diagram of oscillation frequency adjustment method (1)



(2) The method to verify resistance R_B value with inserting the resistance R_B and capacitance C_B connected in series between OS 1 and OS 2.

The oscillation period T_o is found by the same method as the (1) formula in principle, but a little more complicated

In principle, the variation way, and the variation rate of the oscillation period T_o with the resistance R_B are constant, and not be different by process parameter (V_{TR} etc).

Figure 4 illustrates the external connection diagram of resistances R_{FC} , R_B and capacitance C_{FC} , C_B . In addition, the Figure 5 shows the relation of the time-delay T ($=T_o \times 1024$) with R_B at $C_{FC}=C_B$, $R_{FC}=1\text{ M}\Omega$, and the time-delay variation rate ΔT at $R_B=250\text{ k}\Omega$. As shown in Figure 6, the T_o takes the maximum value near $R_B=250\text{ k}\Omega$ according to $C_{FC}=C_B=10^3, 10^4, 10^5\text{ pF}$.

The change of the time-delay T with the resistance R_{FC} keeping the R_B constant will take poor linearity as the value of R_B increases. Therefore, try to keep the resistance R_B in a range of $0 \sim 150\text{ k}\Omega$.

For that, take the $R_B=50\text{ k}\Omega$ first and change its value in the range of $0 \sim 150\text{ k}\Omega$ to adjust the time-delay at the maximum value of R_{FC} , so the adjustment of $\pm 7\%$ becomes possible.

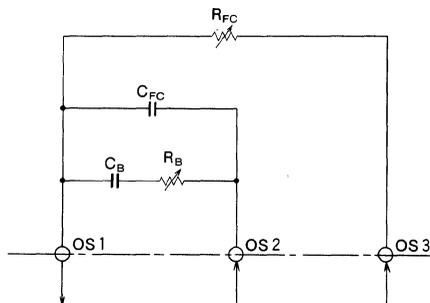


Fig. 4. External connection diagram of oscillation frequency adjustment method (2)

Fig. 5. (a) R_B-T (Method 2)

$C_{FC}=C_B=10^3\text{ pF}$
 $R_{FC}=1\text{ M}\Omega$
 ADJ is shorted with V_{SS}

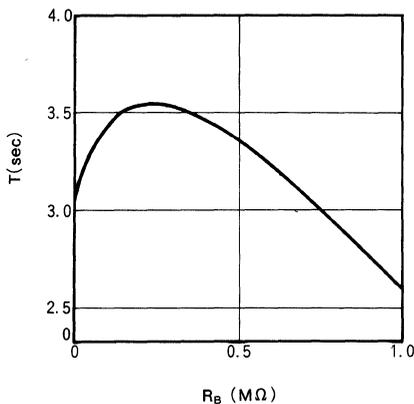


Fig. 5. (b) R_B-T (Method 2)

$C_{FC}=C_B=10^4\text{ pF}$
 $R_{FC}=1\text{ M}\Omega$
 ADJ is shorted with V_{SS}

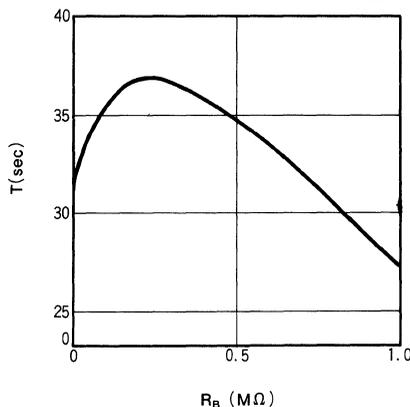


Fig. 5. (c) R_B-T (Method 2)

$O_{FC}=C_H=10^5 pF$
 $R_{FC}=1M\Omega$
 ADJ is shorted with V_{SS}

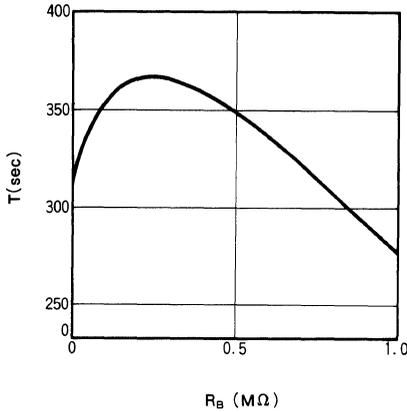
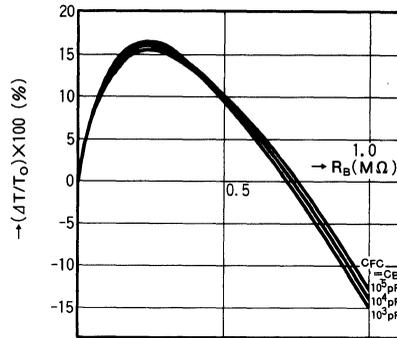


Fig. 6. $R_B-(\Delta T/T_O) \times 100$ (Method 2)

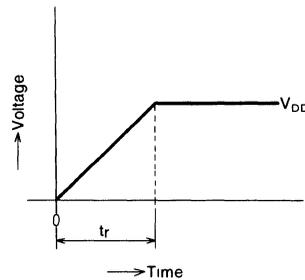
$R_{FC}=1M\Omega$
 $\Delta T=T-T_O$
 $T_O=T(R_b-0\Omega)$



POWER-ON RESET FUNCTION

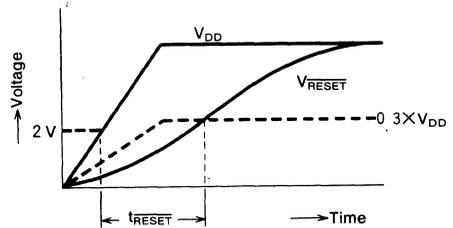
(1) M58479P

The power-on reset function will start when the power is on since the M58479P builds the supply voltage detection circuit in it; however, it is necessary to keep the rising time of power (t_r) more than 1 ms as shown below.

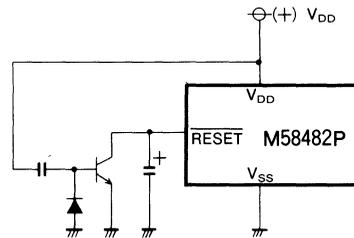


(2) M58482P

The power-on reset function will start by inserting the capacitance between the RESET and V_{SS} when the power is on as same as the M58479P. In order to have an accurate performance on the power-on reset function, keep t_{RESET} over 1msec on the condition of $V_{RESET} \leq 0.3 \times V_{DD}$ when V_{DD} is over 2 V, as illustrated below.

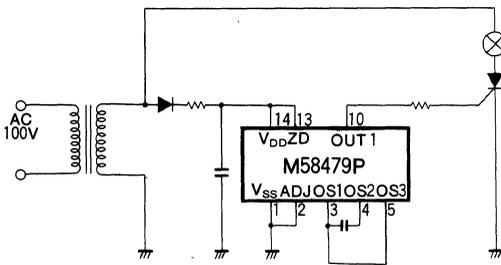


In case the power is on again after it is off and the voltage of RESET V_{RESET} is not perfectly down, the t_{RESET} must be also kept in over 1msec, which was mentioned in the above diagram. When the prescribed condition is not satisfied, add the circuit illustrated below to the RESET and make the power-on reset function accurately. In this case, make sure to select an external capacitance to satisfy the $V_{RESET} \geq 1$ msec.

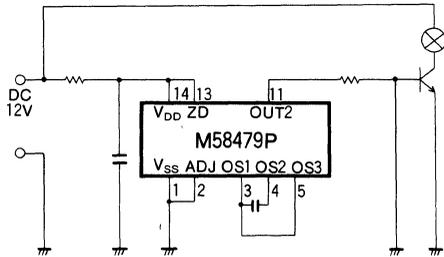


APPLICATION CIRCUITS

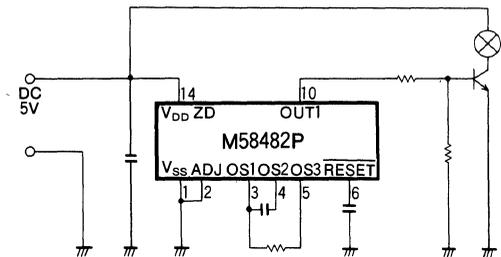
(1) Use of AC supply



(2) Use of DC supply



(3) Use of DC supply (low supply voltage)



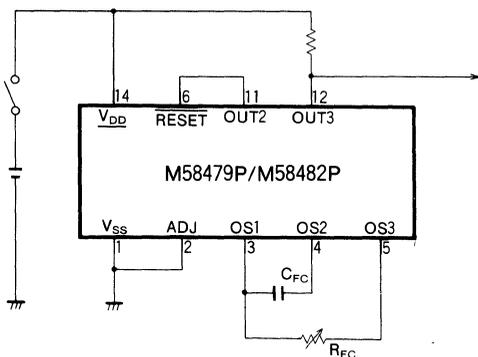
Both M58479P and M58483P build zener diodes in them so that they can adopt AC supply (100V), DC supply (12V) according to external circuits.

If the supply voltage is relatively high, when a power-on reset is required without an external circuit, employ the M58479P.

On the other hand, if a low supply voltage or low power dissipation is required, or if a power supply with a heavy fluctuation on lower voltage is used, employ the M58482P (M58479P may have a reset when V_{DD} is below 5.4V).

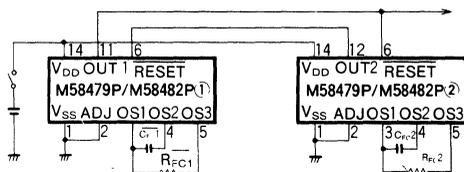
(4) While power is being on,

A pulse of 50% duty of which period is defined by R_{FC} and C_{FC} , is output from OUT 3.



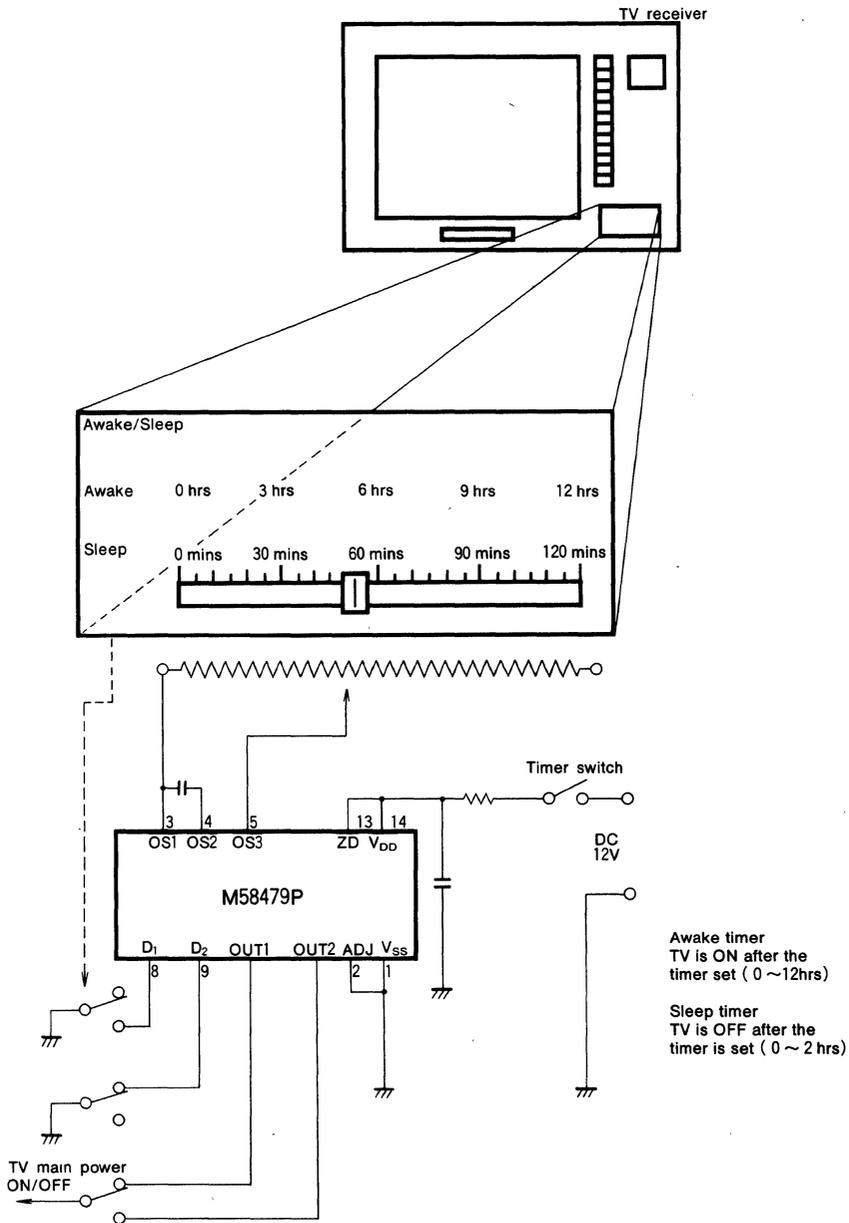
(5) While power is being on,

a (duty changeable) pulse of which a "L" period is defined by R_{FC1} , C_{FC1} , and a "H" period is by R_{FC2} , C_{FC2} , is output from OUT 1 of M58479P/M58482P.



CMOS COUNTER/TIMERS

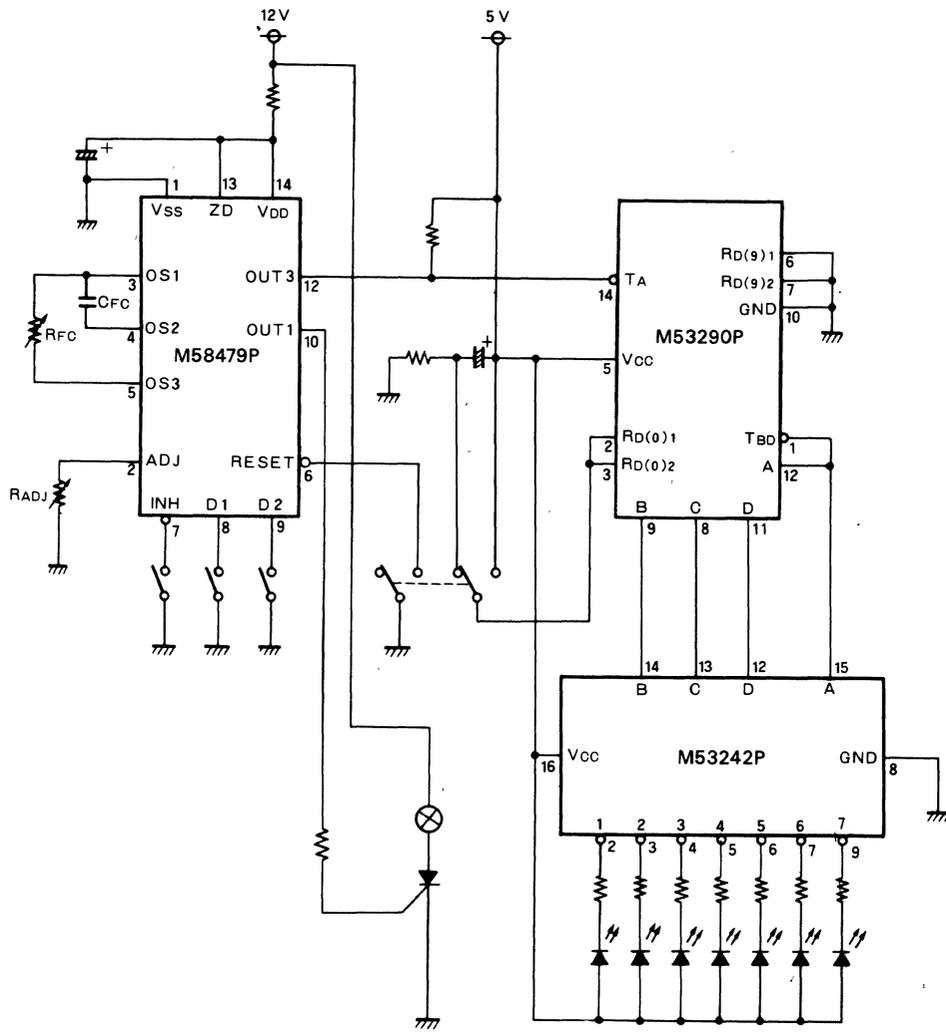
(6) An example of awake/sleep timer



Only one M58479P/M58482P is needed to have one switchover for awake/sleep timer

The application above is just a one example and the M58479P/M58482P can be widely used for home entertainment and industry.

(8) Circuit to display a timer in process



M51845L

COUNTER TIMER

DESCRIPTION

The M51845L is a semiconductor integrated circuit designed for controller of long time delay, consisting of 11 stage divider by 1/L. The time base period of oscillator is extended by 1024 times, determined by the the 11 stage divider by 1/L, and the maximum output period is 50 hours.

FEATURES

- Timing from 100ms through 50 hr
- Timing can be set by one resistor and capacitor
- Built-in 1/L divider with low power dissipation
- Built-in stabilization zener
- Built-in power on reset
- Direct drive of TTL possible

APPLICATION

- Precision timer for consumer and home-use equipments, time delay generation for measuring instruments, Ultra-low-frequency oscillator

RECOMMENDED OPERATING CONDITIONS

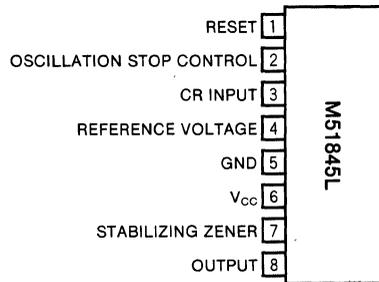
Supply voltage range

..... 4.5V~ V_z (V_z =pin⑦ Zener voltage)

Rated supply voltage ... 5V±10% (No zener diode used)

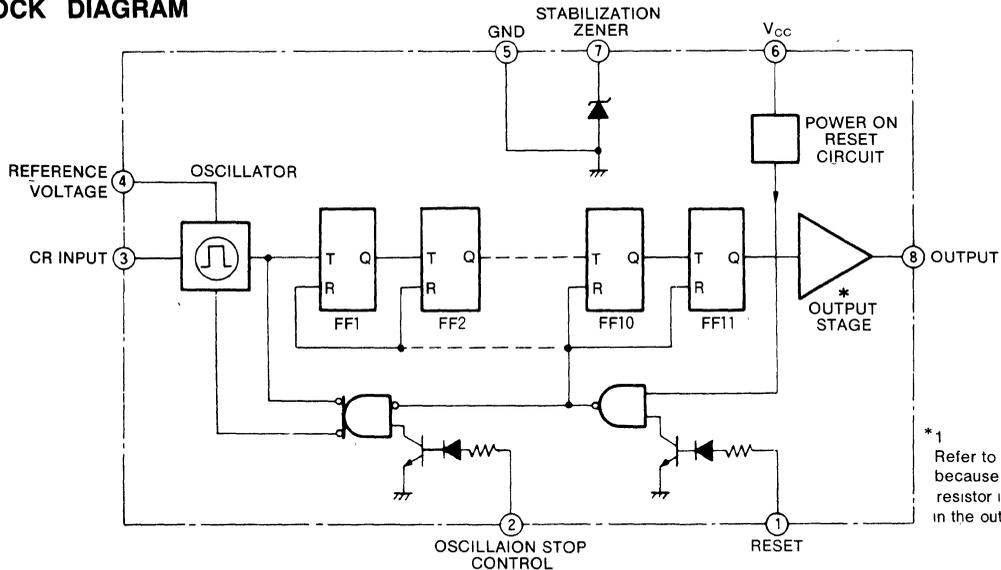
9V, 12V (Stabilization zener diode used)

PIN CONFIGURATION



8-pin molded plastic SIP

BLOCK DIAGRAM



*1 Refer to V_{OH} because a control resistor is built-in in the output stage

COUNTER TIMER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
$V_{CC⑥}$	Supply voltage		6.5	V
I_Z	Zener current		20	mA
I_O	Output sink current		15	mA
P_D	Power dissipation		360	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	3.6	mW/°C
T_{opr}	Operating temperature		0~+60	°C
T_{str}	Storage temperature		-40~+125	°C

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=5\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{CC⑥}$	Circuit current at pin⑥	No load t_a pin⑧		6.5	9.0	mA
V_Z	Zener voltage	$I_Z=1\text{mA}$	5.3	5.8	6.3	V
$I_{IH①}$	Input current at pin①	Pin① voltage 1=3V		0.8	2.3	mA
$I_{IL①}$		Pin① voltage 1=0.4V			1.0	μA
$I_{IH②}$	Input current at pin②	Pin② voltage 2=3V		0.8	2.3	mA
$I_{IL②}$		Pin② voltage 2=0.4V			1.0	μA
V_{OH}	Output voltage	$I_{SOURCE}=2\text{mA}$	2.0	2.65		V
V_{OL}		$I_{SINK}=5\text{mA}$		0.1	0.4	V
f_{max}	Maximum oscillation frequency	CR oscillation part	10	100		kHz

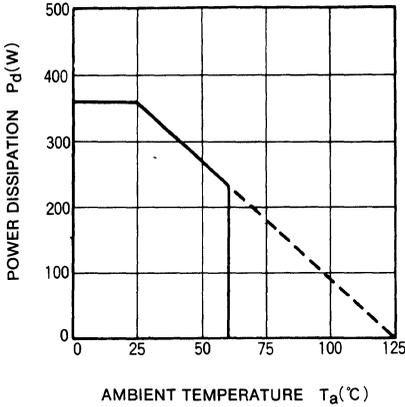
FUNCTION TABLE

Input conditions		Operating and output conditions		
RESET	OSCILLATION STOP CONTROL	OSCILLATOR	11 STAGE DIVIDER	OUTPUT
H	H	STOP	CLEAR (LOW LEVEL)	L
	L			
L	H	STOP	MAINTAIN PREVIOUS CONDITION	MAINTAIN PREVIOUS CONDITION
	L	OSCILLATION	COUNT	COUNT

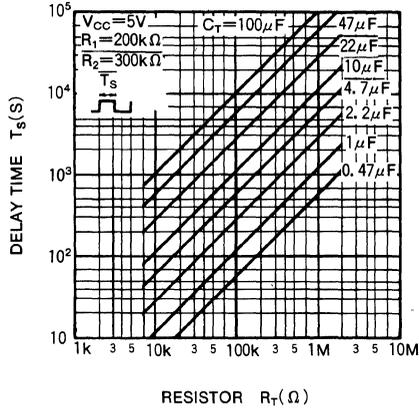
$H \geq 1.4\text{V}$, $L \leq 0.4\text{V}$

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

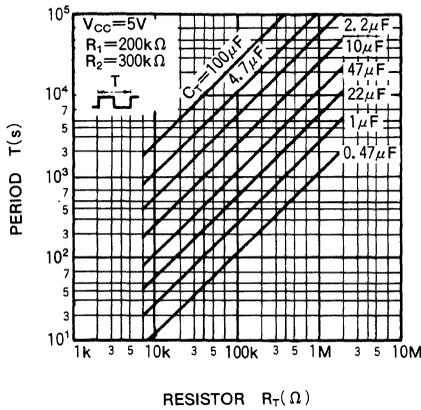
**THERMAL DERATING
(MAXIMUM RATING)**



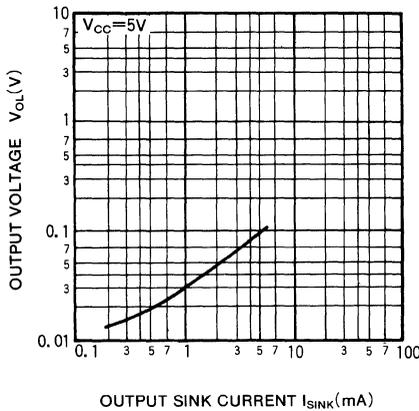
RESISTOR VS. DELAY TIME



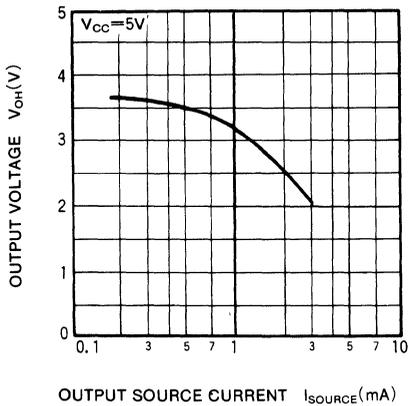
RESISTOR VS. PERIOD



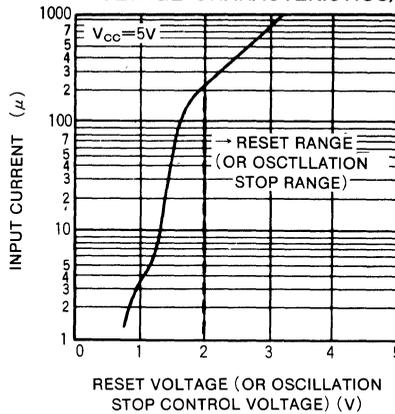
**OUTPUT VOLTAGE VS.
OUTPUT SINK CURRENT**



**OUTPUT VOLTAGE VS.
OUTPUT SOURCE CURRENT**



**INPUT CURRENT VS. RESET VOLTAGE
(OR OSCILLATION STOP CONTROL
VOLTAGE CHARACTERISTICS)**



PIN DESCRIPTION

1. Reset pin (pin ①)

This pin is used to stop counter operation. If the pin is in the high state, the counter is cleared and oscillation is stopped. Connect this pin to GND, if not used.

2. Oscillation stop control pin (Pin ②)

If the pin is in the low state, oscillation is enabled and if the pin is in the high state, oscillation stops. If the feedback is applied to this pin from the output, the output stays in the high state, when it changes from the low to high state. (The function of monostable multivibrator is obtained.) Connect this pin to GND, if not used.

3. CR input pin (pin ③)

The capacitor C_T and resistor R_T are connected to this pin.

The oscillator period T_0 is given in the following equation.

$$T_0 \approx -C_T R_T \ln \frac{R_1 V_{CC} - 0.6(R_1 + R_2)}{(V_{CC} - 1.2)(R_1 + R_2)}$$

(The V_{CC} is pin 6 voltage.)

The period T_0 given in the equation changes depending on the elements. Therefore, a resistor R_T must be connected in serial for fine adjustment if the required setting accuracy is $\pm 25\%$.

4. Reference voltage pin (pin ④)

The highest voltage for the oscillation level is supplied at this pin. The voltage should be set at approximately $2/3 V_{CC}$ by a variable resistor or resistor division (R_1, R_2). The capacitor C_T is corrected by minutely adjusting this voltage by a variable resistor.

5. Supply voltage pin (pin ⑥)

Connect a capacitor (0.1μ to $10\mu F$ depending on the noise) between this pin and GND to avoid external noise from the power supply. The internal power on reset circuit resets the device whenever power is turned on.

6. Stabilization zener pin (pin ⑦)

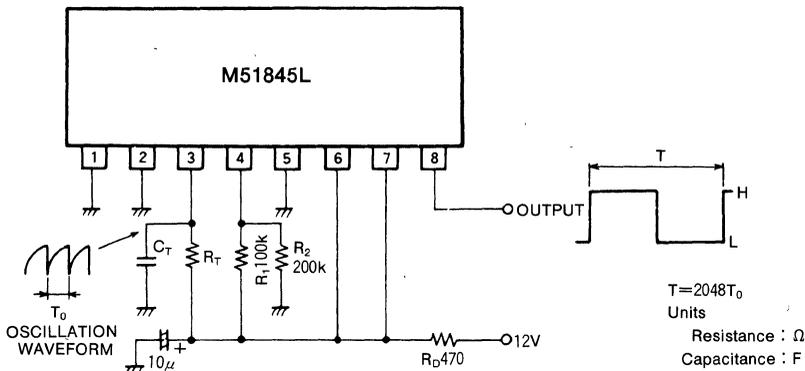
A zener diode of approximately 5.8V is connected between this pin and pin ⑤ (GND). Connection of this pin to pin ⑥ stabilizes power supply of the device and forms 5V supply voltage regulator with externally connected transistors.

7. Output pin (pin ⑧)

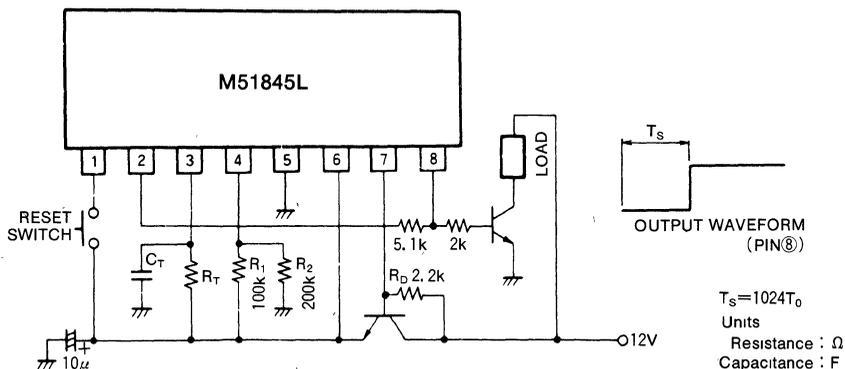
The output voltage changes from the low state to high state when oscillation period is 1024 times T_0 , and the voltage returns from the high state to the low state when the period is 2048 times for one cycle. Use oscillation stop control pin ② to form a monostable multivibrator.

APPLICATION EXAMPLES

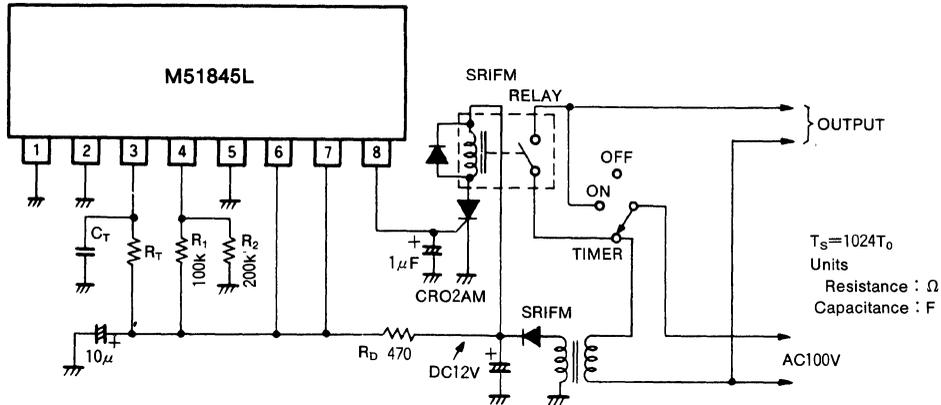
(1) Ultra-low-frequency oscillator



(2) Self-preserving timer



(3) Sleep timer



Precautions for use

1. CR input pin (pin③) must not be directly connected to V_{CC} to avoid destruction of the integrated circuit.
2. The voltage at the reference voltage pin (pin④) should be set to less than 3.5V. If it reaches 3.5V or above, the oscillator may not function.
3. This integrated circuit consisting of a divider by 1/11 and gate circuit operates at low supply voltages. Therefore it is extremely sensitive to external noise and is subject to misoperation. Care should be taken to the following items.
 - 1) Place the device away from any electromagnetic noise generator or protect it with a shield wire.
 - 2) A capacitor must be connected near the V_{CC} pin (pin⑥) to avoid noise.
4. V_{OH} changes according to load because output current control resistor varies depending on the load. Therefore, if the high output current is required, an external output transistor or the M51849L must be used.
5. The minimum resistance for timing resistor R_T is $7k\Omega$. Use the M51849L to have a wide variable range of the timing resistor.

As an improved version of the M51845L, the M51849L has superior characteristics in following.

(compatible in pin configuration)

 1. Large output current (I_{OH} is approximately $10mA_{max}$)
 2. A timing resistor can be set by a wide range of resistance. ($R_{Tmin} = 1k\Omega$)
 3. Excellent temperature stability for timer setting

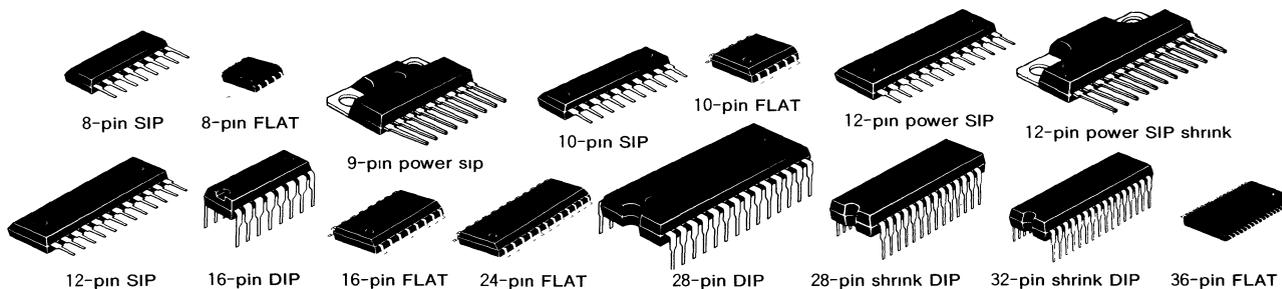
MOTOR DRIVER

QUICK REFERENCE TABLE OF MOTOR DRIVERS

Applica- tion	Type No.	Applicable motor	Speed control method	Function			Features	Supply voltage range	Maximum ratings		Package outline
				FG input circuit	Hall amplifier gain	Motor drive transistor			Output current	Power dissipation	
Speed control	M51722P, FP	DC motor with FG	Sample and hold	PNP input operational amplifier	—	External	<ul style="list-style-type: none"> Built-in multiplexer Built-in FG amplifier 	10V	12mA	770mW 430mW	16-pin DIP 16-pin FLAT
	M51723P, FP	DC motor with FG	Sample and hold	PNP input voltage comparator	—	External	<ul style="list-style-type: none"> Voltage/current outputs possible Start/stop 	20V	6mA	770mW 430mW	16-pin DIP 16-pin FLAT
	M51728L, FP	DC motor with FG	PLL	NPN input voltage comparator	—	External	<ul style="list-style-type: none"> Built-in zener diode (supply voltage clamp $\approx 6.9V$) Regulation against load Null 	—	2mA	700mW 350mW	8-pin SIP 8-pin FLAT
	M51970L	DC motor with FG	Integration	PNP input voltage comparator	—	External	<ul style="list-style-type: none"> Supply voltage regulation $\pm 0.1\%$ typ. Load regulation $\pm 0.1\%$ typ. Temperature coefficient of revolution $\pm 10\text{ppm}/^\circ\text{C}$ typ. 	2.5~18V	40mA	550mW	8-pin SIP
	M51971L, FP	DC motor with FG	Integration	PNP input operational amplifier	—	External	<ul style="list-style-type: none"> Supply voltage regulation $\pm 0.005\%/V$ typ. Load regulation $\pm 0.01\%$ typ. (full load) Temperature coefficient of revolution $7\text{ppm}/^\circ\text{C}$ typ. 	4~17.5V	20mA	880mW 450mW	10-pin SIP 10-pin FLAT
Hall motor driver	M51712P, SP, FP	3-phase hall motor	—	—	40dB	Built-in	<ul style="list-style-type: none"> Low noise, low torque ripple due to linear drive 	26V	1.2A	2.1W 2.0W 1.2W	28-pin DIP 28-pin shrink DIP 36-pin FLAT
	M51719P	1-phase hall motor	—	PNP input operational amplifier	—	Built-in	<ul style="list-style-type: none"> Low cost system possible with a single sensor Built-in FG amplifier 	7V 26V	1.5A	2.5W	16-pin DIP
	M51724P, FP	3-phase hall motor	—	—	—	External	<ul style="list-style-type: none"> Balanced current distribution 	20V	10mA	900mW 540mW	16-pin DIP 16-pin FLAT
	M51718FP	3-phase hall motor	—	—	—	External	<ul style="list-style-type: none"> Designed for low power dissipation and low voltage Applicable to both spindle and access motor F/R, start/stop 	7V 16V	25mA	650mW	24-pin FLAT
Speed control Hall motor driver	M51781SP, FP	2-phase hall motor with FG	Sample and hold	PNP input operational amplifier	—	Built-in	<ul style="list-style-type: none"> Current drive (low W/F) F/R 	18V	1.2A	2.0W 1.2W	32-pin shrink DIP 36-pin FLAT
	M51785P, FP	3-phase hall motor with FG	Digital servo	PNP input operational amplifier	—	Built-in	<ul style="list-style-type: none"> Voltage drive (high speed response) Adjustment free 	15V	1.2A	3.0W 1.2W	32-pin shrink DIP 36-pin FLAT

QUICK REFERENCE TABLE OF MOTOR DRIVERS

Applica- tion	Type No.	Package outline	Speed	Additional function		Features	Maximum ratings			Supply voltage range
				Brake	Thermal shut down		Output current		Power dissipation	
							Continuous	Peak		
Bi-directional motor driver	M51714P	16-pin DIP	4	○		2 bi-directional motor drive	400mA	1.2A	2.5W	5~25V
	M54540AL	8-pin SIP	1			Small mounting space	100mA	0.6A	0.8W	6~11V
	M54541L	9-pin power SIP	1			Built-in clamp diode	200mA	0.8A	0.9W	6~15V
	M54542L	9-pin power SIP	1			High current drive	300mA	1.2A	1.0W	6~15V
	M54543L	9-pin power SIP	1	○		Low output saturation voltage	300mA	1.2A	1.15W	4~15V
	M54543AL	9-pin power SIP	1	○	○	Low power dissipation	300mA	1.5A	1.15W	4~15V
	M54544L	9-pin power SIP	1	○		Low output saturation voltage	300mA	1.2A	1.15W	4~15V
	M54544AL	9-pin power SIP	1	○	○	Low power dissipation	300mA	1.5A	1.15W	4~15V
	M54545L	9-pin power SIP	1	○		Low power dissipation	300mA	1.2A	1.15W	3~15V
	M54546L	10-pin SIP	1	○		Small mounting space	100mA	0.7A	0.6W	4~15V
	M54547P	16-pin DIP	1	○		Built-in operational amplifier, transistor array	100mA	0.6A	1.06W	4~15V
	M54548L	12-pin power SIP	1	○		Built-in operational amplifier for output voltage control	300mA	1.2A	1.2W	4~16V
	M54548AL	12-pin power SIP shrink	1	○		The M54548L equivalent in shrink package	300mA	1.2A	1.1W	4~16V
	M54549L	12-pin power SIP	2	○	○	2 motor drive	300mA	2A	1.2W	4~16V
	M54549AL	12-pin SIP	1	○	○	The M54549L equivalent in small package	100mA	0.8A	0.83W	4~16V
	M54640P	16-pin DIP	1	○	○	Bipolar stepper motor drive with chopper control	—	1A	1.2W	4.75~5.25V
	M54641L	8-pin SIP	1	○	○	Small mounting space	150mA	0.8A	0.58W	4~10V
	M54642L	10-pin SIP	1	○	○	Small mounting space	150mA	0.8A	0.58W	4~10V
	M54643L	10-pin SIP	1	○	○	Low output saturation voltage by externally PNP transistors	200mA	0.8A	0.7W	4~16V
	M54644BL	9-pin power SIP	1	○	○	Small control current	600mA	2A	1.15W	4~16V
M54648L	12-pin power SIP	1	○	○	The M54548L equivalent of 3A output	300mA	3A	1.1W	4~18V	
M54648AL	12-pin power SIP shrink	1	○	○	The M54648L equivalent in shrink package	300mA	3A	1.1W	4~18V	
M54649L	10-pin SIP	2	○	○	Small package for 2 motor drive	600mA	1.6A	0.7W	4~18V	



M51722P,FP

F-V CONVERTER WITH MULTIPLEXER

DESCRIPTION

The M51722P,FP are semiconductor integrated circuits developed for frequency-voltage (F-V) converter.

In combination with predriver ICs, the devices constitute high-accuracy 2-phase, 3-phase, F-servo motor control system.

FEATURES

- Built-in frequency multiplexer controls motors with high accuracy.
- Operates at a low supply voltage (supply voltage range 4.5~10V)
- Built-in FG amplifier circuit enables operation by weak signals
- Large final stage output current (current source and current sink)

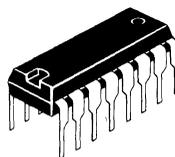
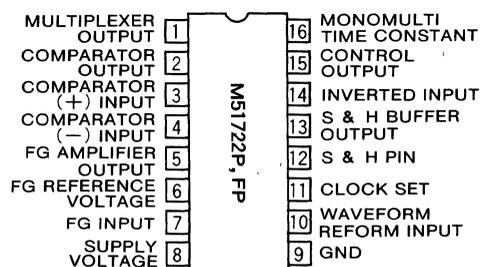
APPLICATION

VTRs, floppy-disk drive, etc.

RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 4.5~10V
- Rated supply voltage 9V

PIN CONFIGURATION (TOP VIEW)

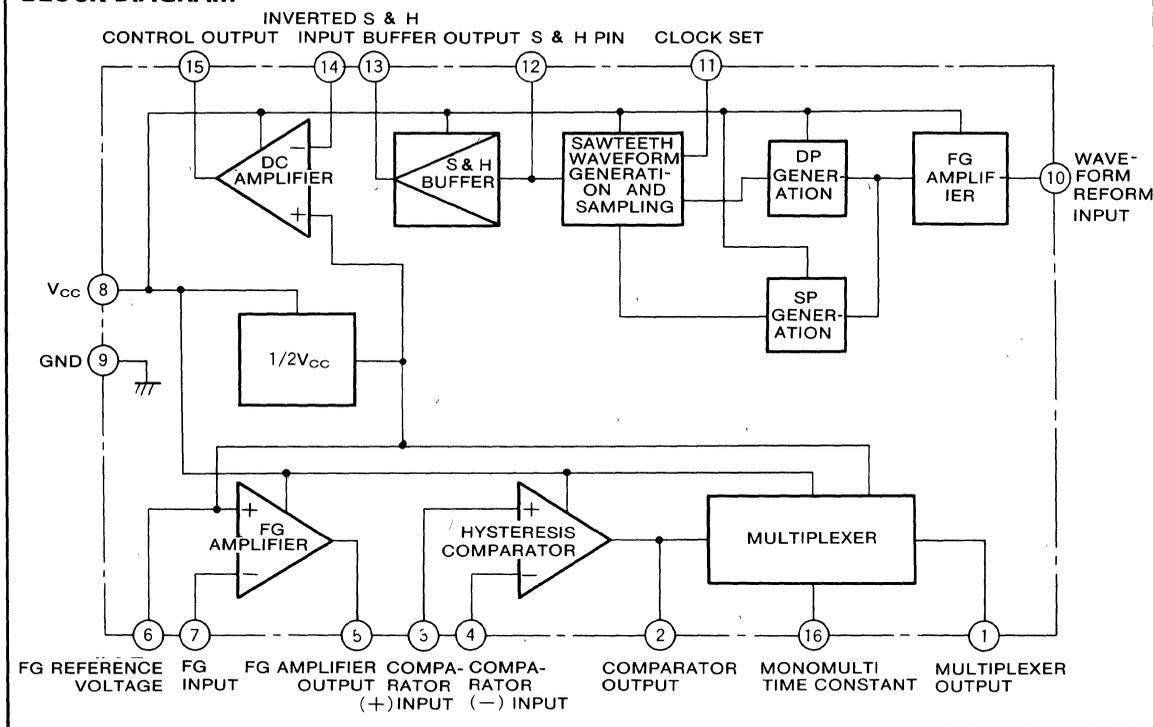


16-pin molded plastic DIP



16-pin molded plastic FLAT (C type)

BLOCK DIAGRAM



F-V CONVERTER WITH MULTIPLEXER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
$I_{(2)}$	Pin ② output current		8	mA
$I_{(5)}$	Pin ⑤ output current		2	mA
$I_{(1)}$	Pin ① source current		8	mA
$I_{(15)}$	Pin ⑮ sink current		12	mA
$I_{(13)}$	Pin ⑬ output current		1	mA
$V_{(7)}$	Pin ⑦ input voltage		$1.5 \sim V_{CC}$	V
$V_{(10)}$	Pin ⑩ input voltage		$0 \sim V_{CC}$	V
P_{dF}	Power dissipation		770(430)	mW
K_θ	Thermal derating		7.7(4.3)	mW/°C
T_{opr}	Operating temperature		$-20 \sim +75$	°C
T_{stg}	Storage temperature		$-40 \sim +125$	°C

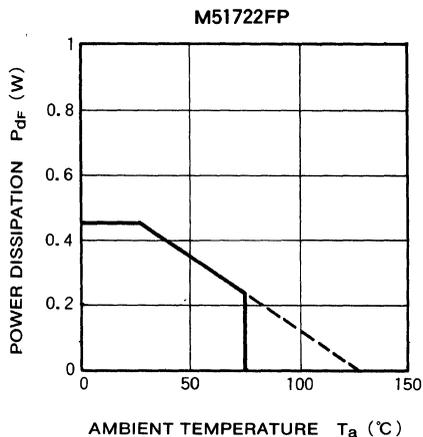
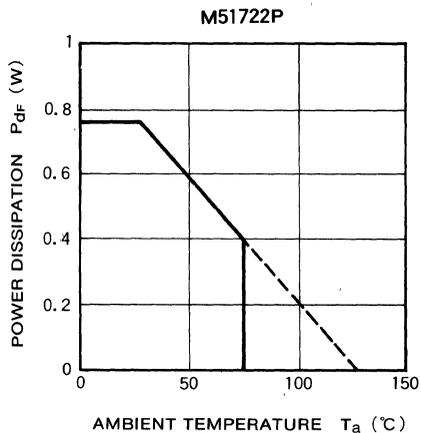
Note () = M51722FP

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=9\text{V}$, unless otherwise noted)

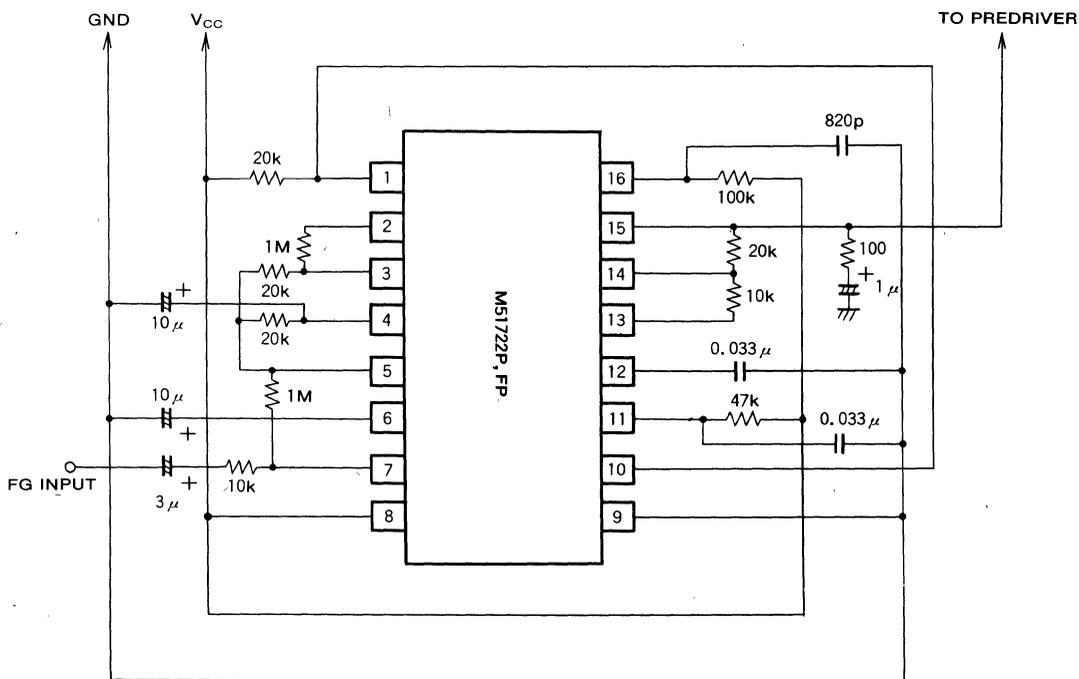
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{CC(1)}$	Circuit current (1)	$V_{CC}=4.5\text{V}$		4.5	9	mA
$I_{CC(2)}$	Circuit current (2)	$V_{CC}=10\text{V}$		4.5	9	mA
$V_{ref(1)}$	Reference voltage of output voltage (1)	Pin ⑥ Open	4.4	4.5	4.6	V
$V_{ref(2)}$	Reference voltage of output voltage (2)	$10\text{k}\Omega$ between pin ⑥ and GND	4.4	4.5	4.6	V
$V_{OH(5)}$	Pin ⑤ high-level output	$3.9\text{k}\Omega$ between pin ⑤ and GND	6.0	6.8		V
$V_{OL(5)}$	Pin ⑤ low-level output	$3.9\text{k}\Omega$ between V_{CC} and pin ⑤		1.0	1.5	V
$V_{OH(2)}$	Pin ② high-level output	$2\text{k}\Omega$ between pin ② and GND	7.5	8.0		V
$V_{OL(2)}$	Pin ② low-level output	$2\text{k}\Omega$ between V_{CC} and pin ②		1.0	1.5	V
$V_{OH(15)}$	Pin ⑮ high-level output	$1.5\text{k}\Omega$ between pin ⑮ and GND	6.6	7.5		V
$V_{OL(15)}$	Pin ⑮ low-level output	$1.5\text{k}\Omega$ between V_{CC} and pin ⑮		1.0	1.5	V
$V_{offset(6-7)}$	Pin ⑥ - pin ⑦ offset voltage				± 6	mV
$V_{offset(3-4)}$	Pin ③ - pin ④ offset voltage				± 6	mV
$V_{offset(12-13)}$	Pin ⑫ - pin ⑬ offset voltage				± 6	mV
$T_{(1)}$	Pin ① pulse width	Pin ⑮ time constant $100\text{k}\Omega + 820\text{pF}$	45	55	65	μs

F-V CONVERTER WITH MULTIPLEXER

THERMAL DERATING (MAXIMUM RATING) ($T_a=25^\circ\text{C}$, unless otherwise noted)



APPLICATION EXAMPLE

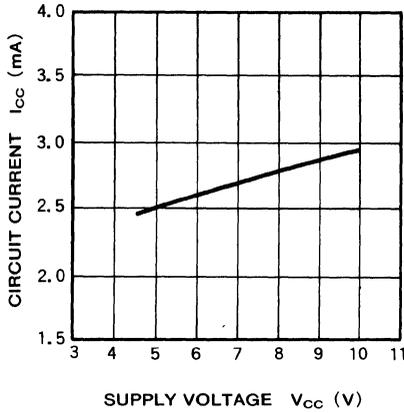


Unit Resistance : Ω
 Capacitance : F

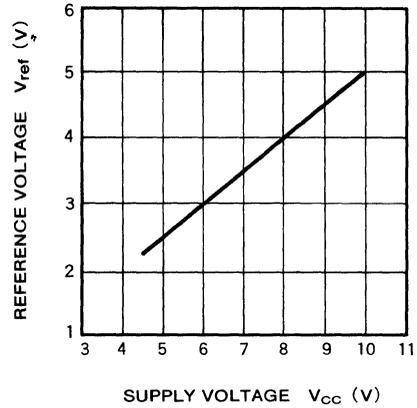
F-V CONVERTER WITH MULTIPLEXER

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

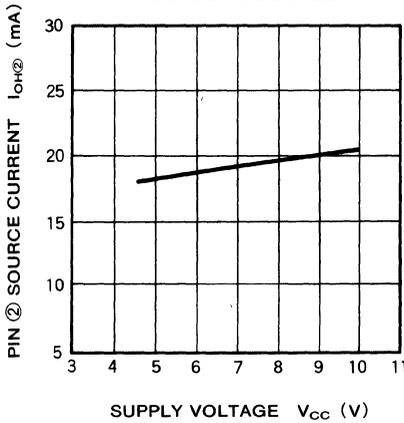
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



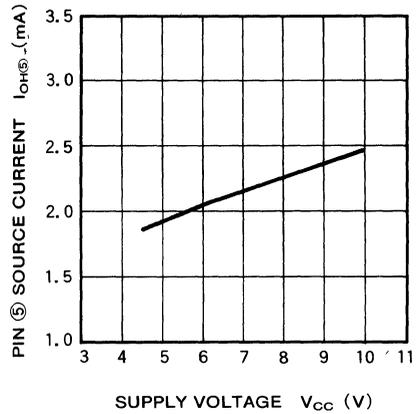
REFERENCE VOLTAGE VS. SUPPLY VOLTAGE



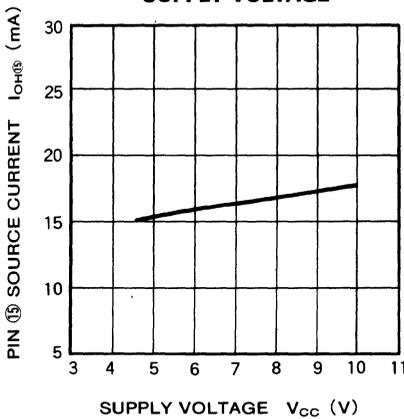
PIN ② SOURCE CURRENT VS. SUPPLY VOLTAGE



PIN ⑤ SOURCE CURRENT VS. SUPPLY VOLTAGE

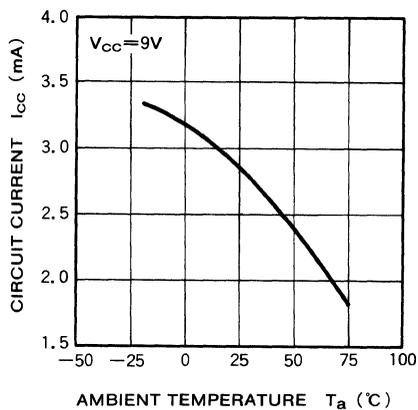


PIN ⑬ SOURCE CURRENT VS. SUPPLY VOLTAGE

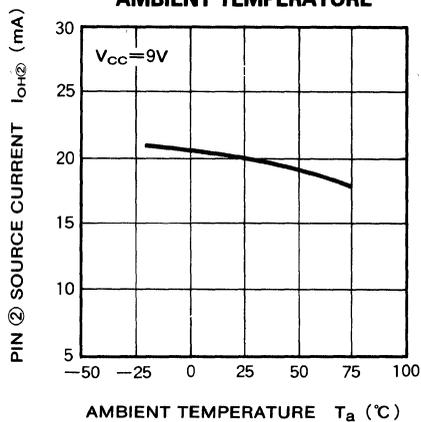


F-V CONVERTER WITH MULTIPLEXER

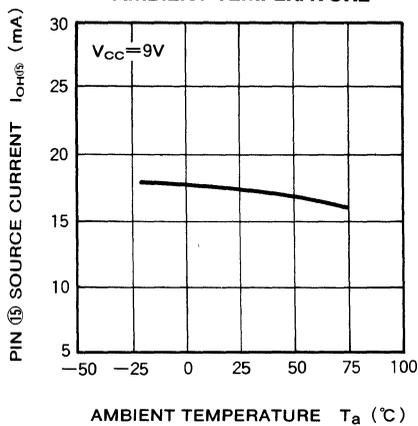
CIRCUIT CURRENT VS. AMBIENT TEMPERATURE



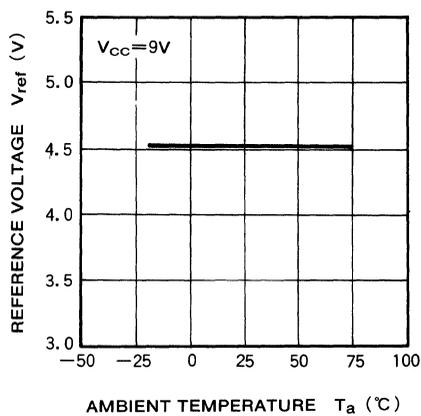
PIN ② SOURCE CURRENT VS. AMBIENT TEMPERATURE



PIN ⑮ SOURCE CURRENT VS. AMBIENT TEMPERATURE



REFERENCE VOLTAGE VS. AMBIENT TEMPERATURE



MITSUBISHI LINEAR ICs M51723P,FP

FREQUENCY-VOLTAGE (F-V) CONVERTER

DESCRIPTION

The M51723P,FP are semiconductor integrated circuits designed for use in frequency-voltage (F-V) converting.

The devices consist of an FG amplifier, sample and hold circuit, error amplifier and sawteeth-wave generating circuits.

The M51723P,FP constitute frequency-servo motor control system in combination with the brushless motor pre-driver, M51724P,FP or other pre-driver ICs.

FEATURES

- Low power dissipation
- Suitable for both current output (current source or current sink) and voltage output
- Start/stop changeover terminal

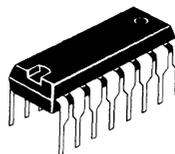
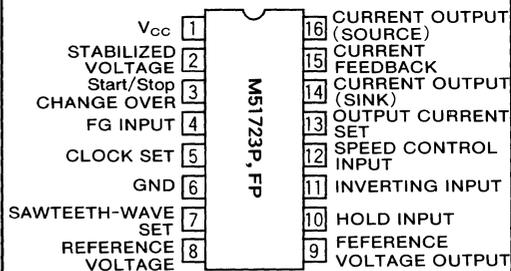
APPLICATION

VTR, floppy disk drive, etc.

RECOMMENDED OPERATING CONDITIONS

Supply voltage range 7.2~20V
Rated supply voltage 12V

PIN CONFIGURATION (TOP VIEW)

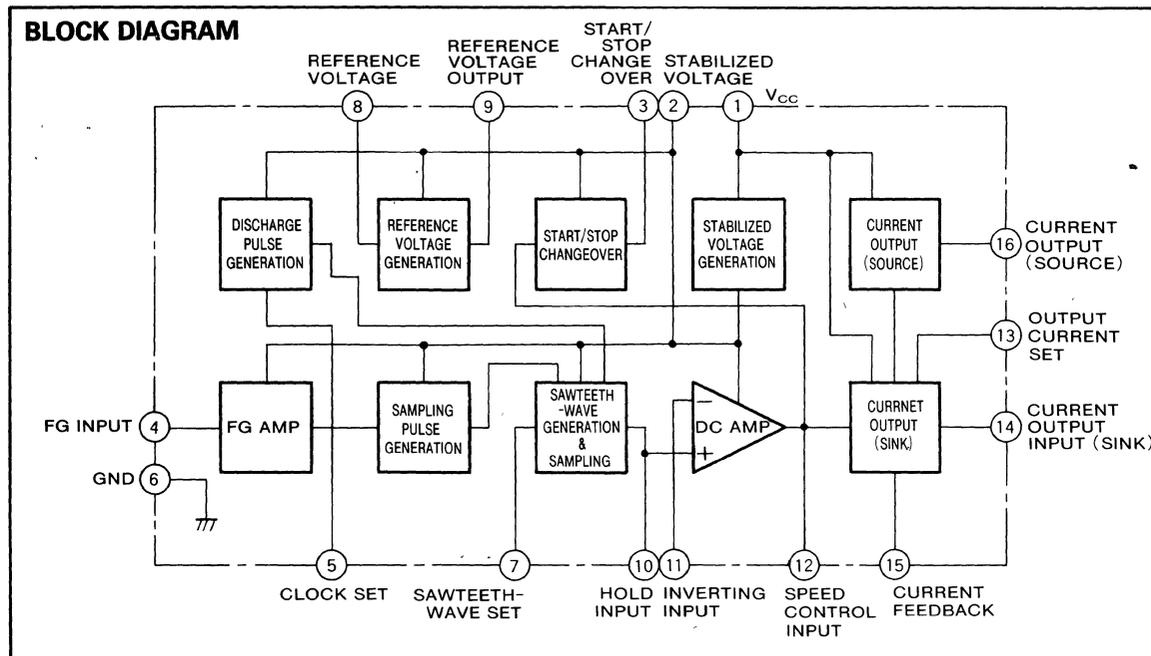


16-pin molded plastic DIP



16-pin molded plastic FLAT (C type)

BLOCK DIAGRAM



FREQUENCY-VOLTAGE (F-V) CONVERTER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		20	V
I_{stb}	Pin ② output current		20	mA
$V_{(4)}$	Pin ④ input voltage		$-0.2 \sim V_{stb}$	V
$I_{OL(14)}$	Pin ⑭ source current		6	mA
$I_{OL(16)}$	Pin ⑰ sink current		6	mA
P_{dF}	Power dissipation		770(430)	mW
T_{opr}	Operating temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^\circ\text{C}$

Note () = M51723FP

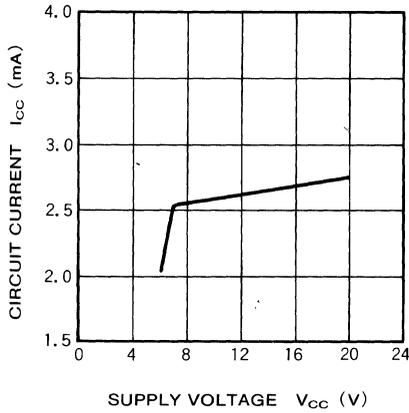
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current		2	3	5	mA
V_{stb}	Stabilized output voltage		5.3	5.8	6.3	V
V_{ref}	Reference voltage	$V_{CC}=V_{stb}=5.6\text{V}$	2.70	2.81	2.94	V
$V_{TH(1)}$	Clock threshold voltage (1)	$V_{CC}=V_{stb}=5.6\text{V}$	1.78	1.90	2.00	V
$V_{TH(2)}$	Clock threshold voltage (2)	$V_{CC}=V_{stb}=5.6\text{V}$	2.67	2.81	2.95	V
$V_{TH(3)}$	Clock threshold voltage (3)	$V_{CC}=V_{stb}=5.6\text{V}$	3.01	3.17	3.33	V
$V_{TH(4)}$	Clock threshold voltage (4)	$V_{CC}=V_{stb}=5.6\text{V}$	3.95	4.17	4.37	V
V_{STOP}	Stop circuit operating voltage	$V_{CC}=V_{stb}=5.6\text{V}$		2.6	3.0	V
$V_{offset(4)}$	Pin ④ input offset voltage	$V_{CC}=V_{stb}=5.6\text{V}$		0	± 6	mV
$V_{offset(7-10)}$	Pin ⑦ - pin ⑩ offset voltage	$V_{CC}=V_{stb}=5.6\text{V}$		0	± 10	mV
$V_{offset(8-9)}$	Pin ⑧ - pin ⑨ offset voltage	$V_{CC}=V_{stb}=5.6\text{V}$		0	± 10	mV
$V_{offset(10-11)}$	Pin ⑩ - pin ⑪ offset voltage	$V_{CC}=V_{stb}=5.6\text{V}$		0	± 10	mV
$I_{S(14)}$	Pin ⑭ sink current	20k Ω between V_{CC} and pin ⑭	440	550	660	μA
$I_{SO(16)}$	Pin ⑰ source current	20k Ω between V_{CC} and pin ⑰	570	720	860	μA

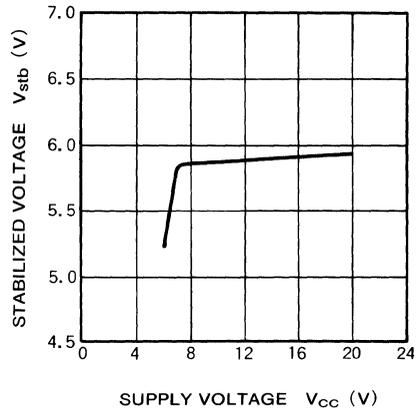
FREQUENCY-VOLTAGE (F-V) CONVERTER

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

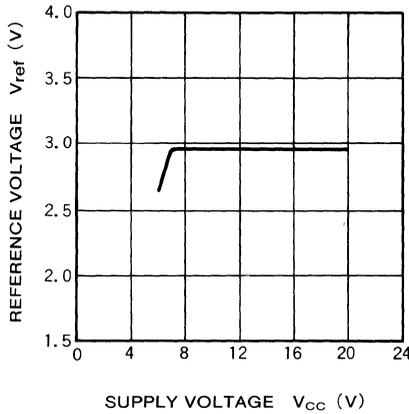
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



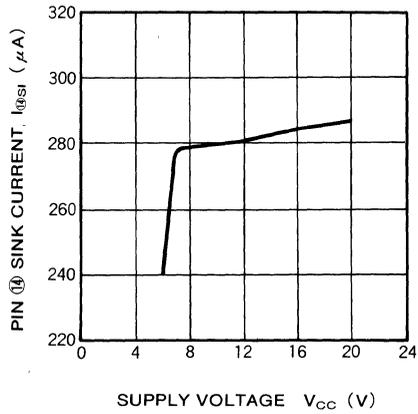
STABILIZED VOLTAGE VS. SUPPLY VOLTAGE



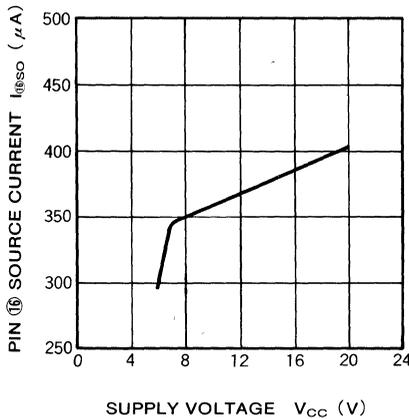
REFERENCE VOLTAGE VS. SUPPLY VOLTAGE



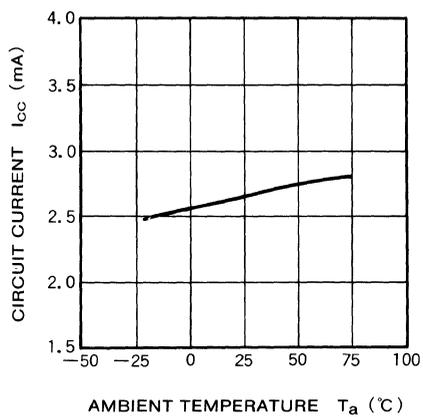
PIN ⑭ SINK CURRENT VS. SUPPLY VOLTAGE



PIN ⑮ SOURCE CURRENT VS. SUPPLY VOLTAGE

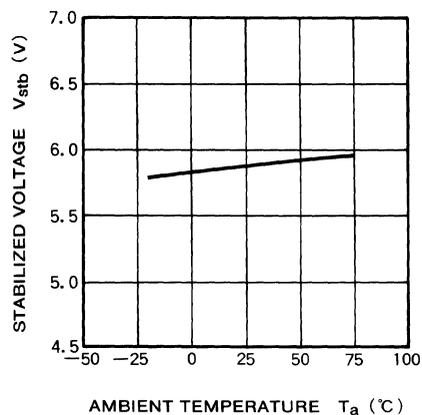


CIRCUIT CURRENT VS. AMBIENT TEMPERATURE

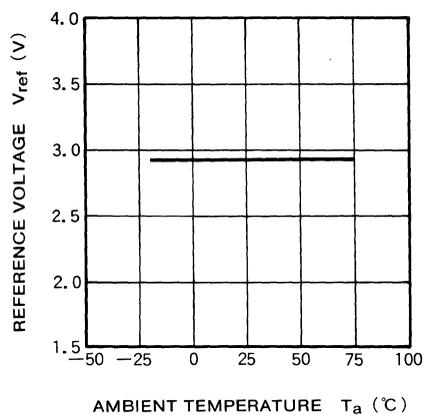


FREQUENCY-VOLTAGE (F-V) CONVERTER

STABILIZED VOLTAGE
VS. AMBIENT TEMPERATURE



REFERENCE VOLTAGE
VS. AMBIENT TEMPERATURE



M51728L,FP

PLL MOTOR SPEED CONTROL

DESCRIPTION

The M51728L,FP are semiconductor integrated circuits designed for motor speed control circuit.

The precision speed control can be obtained because of the PLL circuit.

It controls DC motors with a TG (tacho generator).

FEATURES

- Built-in zener diode6.9V(typ.)
- Supply voltage regulation (oscillating frequency) $\pm 0.01\%$ (typ.)(9~18V, $R_S=910\Omega$)
- Load regulation of motor speed Nil
- Temperature coefficient (oscillating frequency) $\pm 50\text{ppm}/^\circ\text{C}$ (typ.)(-20~+75°C)
- Motor speed can be easily set

APPLICATION

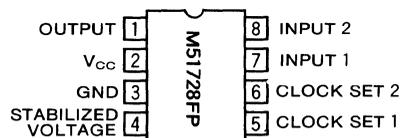
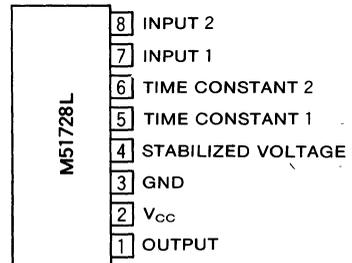
Speed control of motors in audio equipment or terminals.

RECOMMENDED OPERATING CONDITIONS

(V_S when dropper resistor $R_S=910\Omega$)

- Supply voltage range9~18V
- Rated supply voltage13V

PIN CONFIGURATION (TOP VIEW)

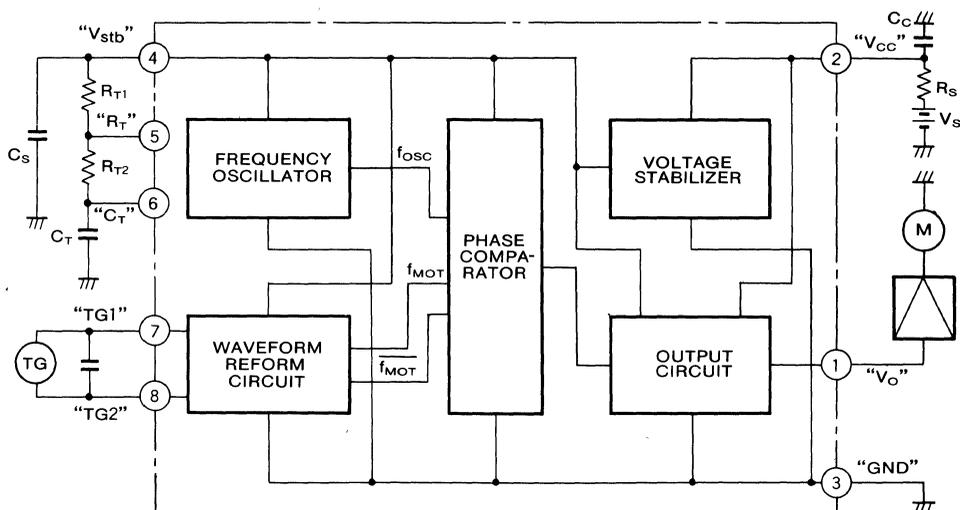


8-pin molded plastic SIP



8-pin molded plastic FLAT (MINI FLAT)

BLOCK DIAGRAM



PLL MOTOR SPEED CONTROL

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

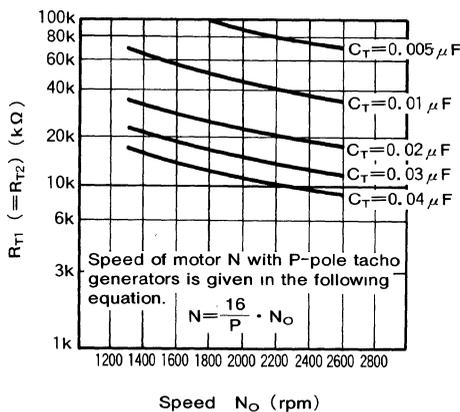
Symbol	Parameter	Conditions	Ratings	Unit
I _{CC}	Circuit current	Source current at Pin ②	20	mA
V _{IN(⑦-⑧)}	Pin ⑦-⑧ applied voltage		5.7	V _{P-P}
I _⑤	Pin ⑤ source current		2	mA
I _O	Output current		Note ±2	mA
V _⑤	Pin ⑤ applied voltage		7	V
V _⑥	Pin ⑥ applied voltage		2.7	V
V _①	Pin ① applied voltage		7	V
P _d	Power dissipation		500(400)	mW
K _θ	Thermal derating		200(250)	°C/W
T _{opr}	Operating temperature		-20~+75	°C
T _{stg}	Storage temperature		-40~+125	°C

Note 1. "+" indicated sink current to Pin ① and "-" indicates source current from Pin ①.

2. ()=M51728FP

ELECTRICAL CHARACTERISTICS (T_a=25°C, V_S=13V, R_S=910Ω, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Circuit voltage	Pin ② voltage	6.2	6.9	7.6	V
I _{CC}	Circuit current	V _{CC} =6V	2.0	3.7	6.0	mA
V _{stb}	Stabilized output voltage	V _{CC} =6V	2.4	2.7	3.0	V
V _{OH}	Pin ① high level output	I _O =-1mA	4.2	5.5		V
V _{OL}	Pin ① low level output	I _O =+1mA		0.04	0.3	V
V _{⑦(V⑧)}	Pin ⑦ (Pin ⑧) voltage	V _{stb} =V _{CC} =3V	1.0	1.3	1.5	V
I _{⑦-⑧}	Current between Pin ⑦ and Pin ⑧	5V applied between Pin ⑦ and Pin ⑧	1.6	2.5	3.6	mA
ΔT _{in}	Difference of TG amplifier output period	Measured under conditions of 5kHz, 150mV _{P-P} SIN input between Pin ⑦ and Pin ⑧, and 10kHz rectangular wave input at Pin ⑥ and output at Pin ①	-20	0	+20	μsec
f _{OSC}	Oscillating output frequency	R _{T1} =R _{T2} =47kΩ, C _T =0.01μF	975	1025	1075	Hz
Δf _(V_S)	Oscillating frequency supply voltage regulation	V _S =9~18V		±0.01		%
Δf _(T)	Oscillating frequency supply voltage regulation	T _a =-20~+75°C, IC only		±50		ppm/°C
Δf _(T_a)	Oscillating frequency time regulation	t=0.5sec~30min		±0.02		%
V _{in(min)}	Minimum operating voltage between Pin ⑦ and Pin ⑧	Voltage satisfying ΔT _{in} when f _{in} (between Pin ⑦ and Pin ⑧)=0.1~5kHz		150		mV _{P-P}
V _{in(N)}	Input sensitivity between Pin ⑦ and Pin ⑧	Measure existence of Pin ① output for the input between Pin ⑦ and Pin ⑧		3		mV _{P-P}



The fig. shows relationship between oscillator time constants (T_{T1}, R_{T2}, C_T) and speed N_O of a motor with sixteen-pole tachometer generator. The limits in the above are obtained when R_{T1}=R_{T2}=47kΩ, C_T=0.01μF.

A trimmer resistor can be added in serial with R_{T1} or R_{T2} for fine adjustment.

The calculating equation is as follows;

$$f_{OSC} = \frac{1}{0.693 \cdot C_T (R_{T1} + 2R_{T2})}$$

$$N = \frac{60 \cdot f_{TG}}{P} = \frac{60 \cdot f_{OSC} / 2}{P} = \frac{30}{P} f_{OSC}$$

where N: Speed

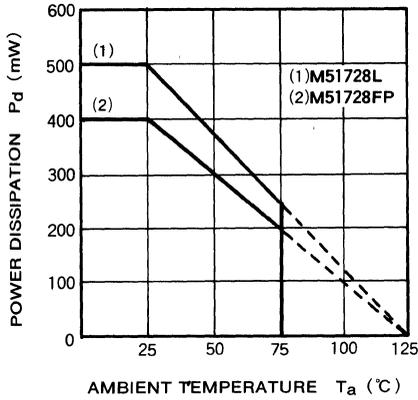
P: number of poles of tacho generator

f_{TG}: output frequency of tacho generator

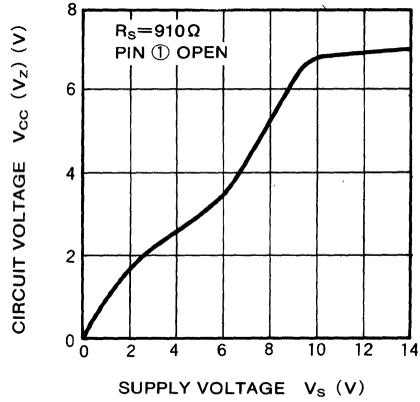
PLL MOTOR SPEED CONTROL

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_s=13\text{V}$, $R_s=910\Omega$, unless otherwise noted)

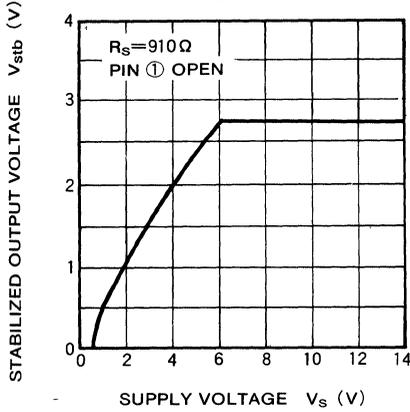
THERMAL DERATING (MAXIMUM RATING)



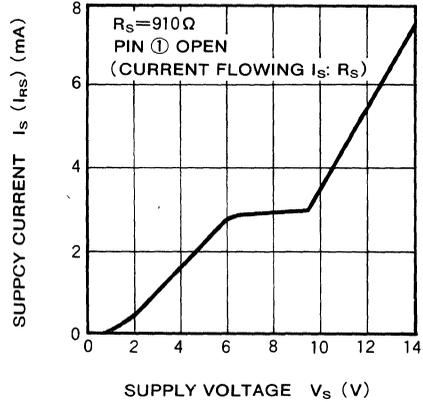
$V_{cc}-V_s$



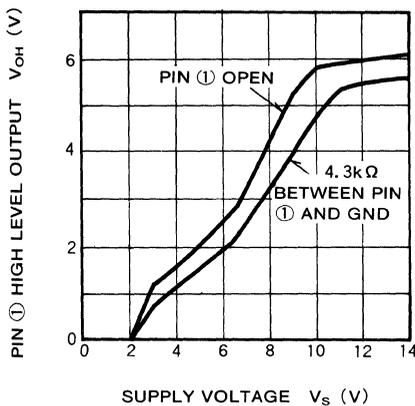
$V_{stb}-V_s$



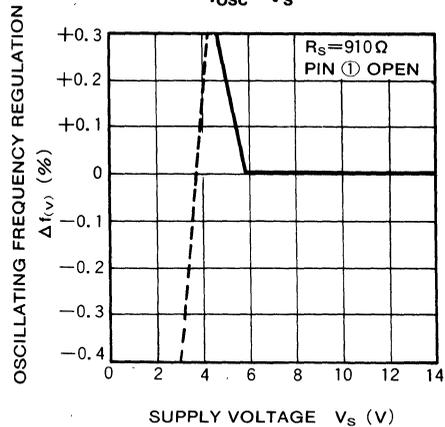
I_s-V_s



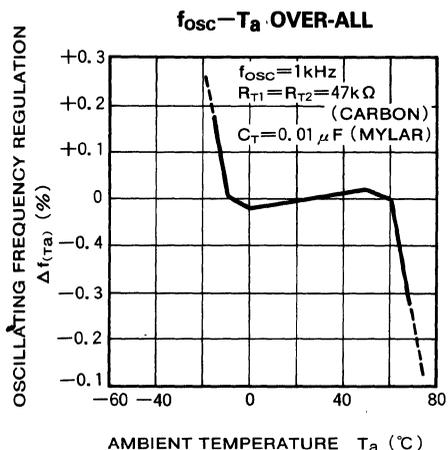
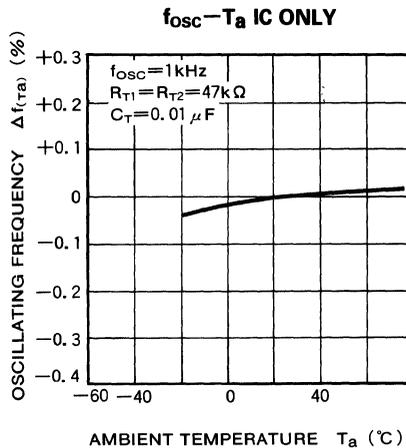
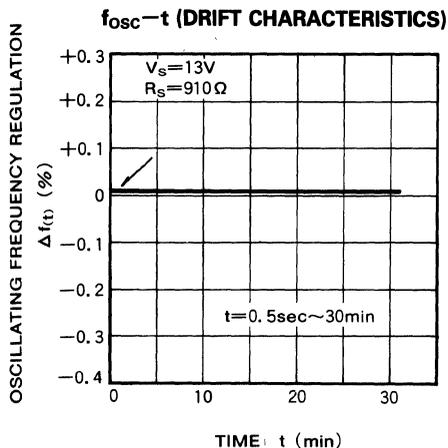
$V_{OH}-V_s$ (PIN ① OUTPUT)



$f_{osc}-V_s$

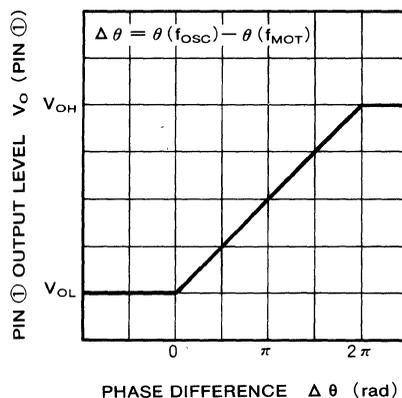


PLL MOTOR SPEED CONTROL

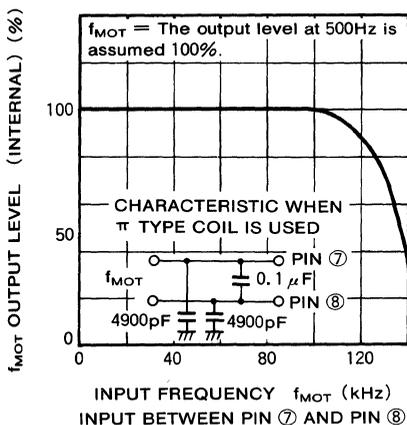


OUTPUT CHARACTERISTICS OF PHASE COMPARATOR (PIN ①)

(Note: The Y-axis indicates the integrated output.)



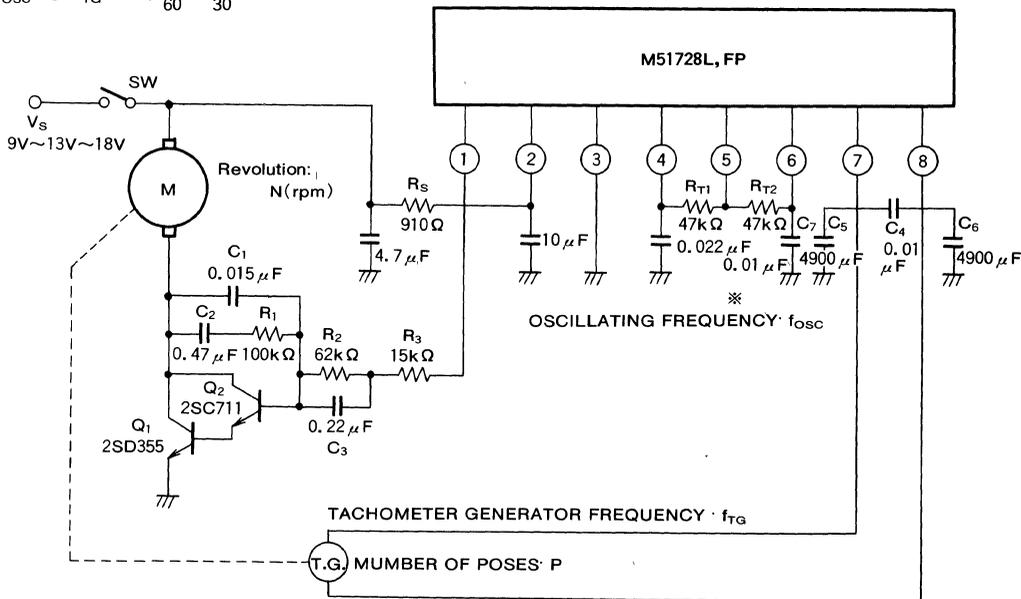
OUTPUT CHARACTERISTICS OF WAVEFORM REFORM CIRCUIT (INTERNAL)



PLL MOTOR SPEED CONTROL

APPLICATION EXAMPLE

Note. $f_{osc} = 2 \cdot f_{TG} = 2 \cdot P \cdot \frac{N}{60} = \frac{PN}{30}$

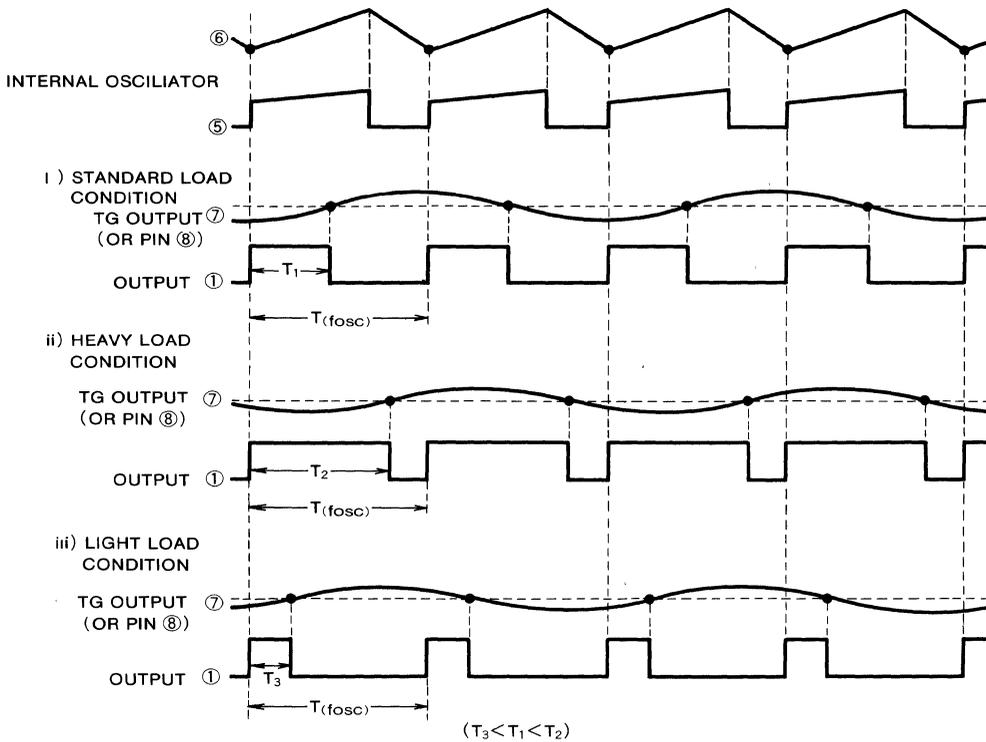


※ External signals of crystal oscillator can be input at Pin ⑥ as oscillating frequency. The input threshold value is as follows.

$$V_{ref(H)} = \frac{2}{3} V_{stb}$$

$$V_{ref(L)} = \frac{1}{3} V_{stb}$$

TIMING CHART OF THE M51728L,FP

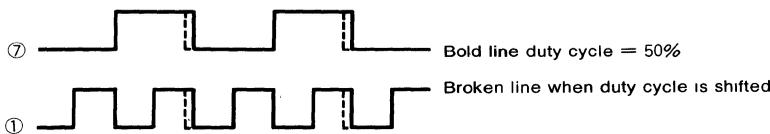


1. Differential type of input is used for tachogenerator frequency input Pin ⑦ and Pin ⑧. Trigger signal is generated for phase comparator when the AC frequency width of the input signal crosses at near zero.
2. Pulse feedback signal generated by encoder must be input to Pin ⑦. The high input level should be less than $V_{stb} - 0.3V$. ($V_{stb} = 2.7V$ typ.)
3. The duty of phase comparator output at Pin ① differs for every cycle if the duty of input signal at Pin ⑦ and Pin ⑧ is less than 50%.

4. An external oscillator, such as crystal oscillator instead of internal oscillator of oscillating frequency, should be connected to Pin ⑥. The input level must be less than the level shown below;

$$\text{for high level } V_{in(H)} \quad \frac{2}{3}V_{stb} < V_{in(H)} < (V_{stb} - 0.3V)$$

$$\text{and for low level } V_{in(L)} \quad 0 < V_{in(L)} < \frac{1}{3}V_{stb}$$



M51970L

SPEED CONTROL FOR DC MICRO MOTOR

DESCRIPTION

The M51970L is a monolithic IC designed for use of speed control for DC micro motor. It controls constantly speed of DC micro motor, connecting the signal of frequency-generation detector to the IC. It consists of an input signal amplifier, a monostable multivibrator, an integrator, an output current amplifier with current limiter, an overshoot protector and an internal voltage regulator.

FEATURES

- Wide supply voltage range 2.5~18V (-20~+75°C)
- High stability vs supply voltage $\pm 0.01\%/V$ (typ.)
- High stability vs temperature $\pm 10\text{ppm}/^\circ\text{C}$ (typ.)
- High stability vs load $\pm 0.1\%$ (typ.)
- Provides DC output drive for minimum RFI.
- Includes overshoot protection circuit for quick start response of motor with less overshoot.

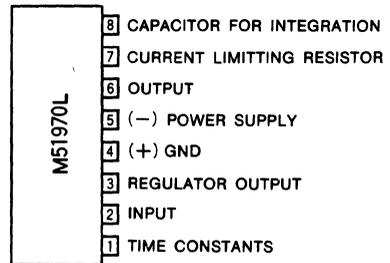
APPLICATIONS

8m/m movie camera, Floppy disk driver, Record player, Tape recorder, Car stereo, Motor driven equipment

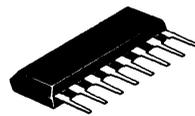
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 2.5~17V
- Rated supply voltage 9V

PIN CONFIGURATION (TOP VIEW)

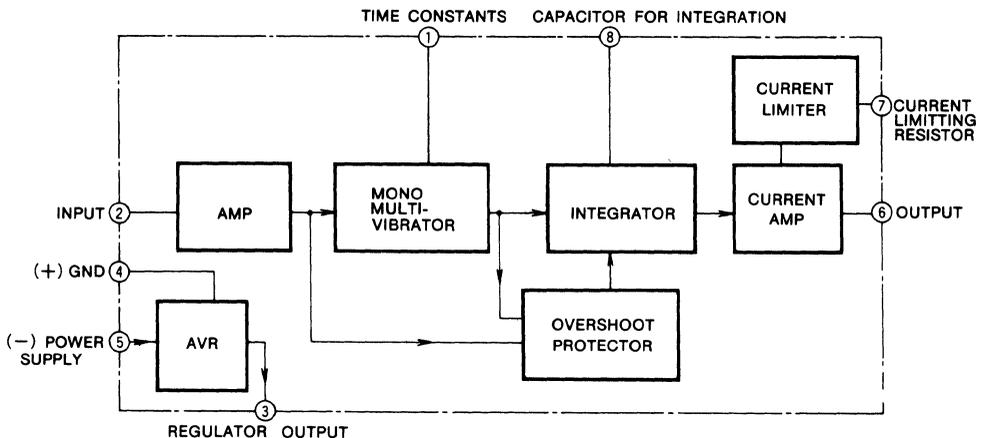


Outline 8P5



8-pin molded plastic SIL

BLOCK DIAGRAM



SPEED CONTROL FOR DC MICRO MOTOR

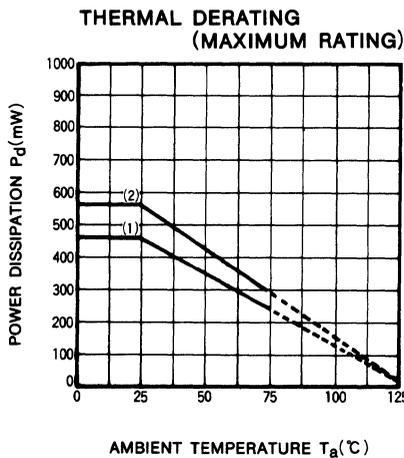
ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		18	V
I _⑥	Sink current into ⑥		40	mA
I _③	Issued current from ③		-3	mA
P _d	Power dissipation	Mounted on the P-C board (Cu foil area 4.5X5.5cm, t=35μ, thickness of the P-C board 2mm)	550	mW
K _θ	Derating		5.5	mW/°C
T _{opr}	Operating temperature		-20~+75	°C
T _{stg}	Storage temperature		-40~+125	°C

ELECTRICAL CHARACTERISTICS (T_a=25°C, V_{CC}=9V, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage range	T _a =-20~+75°C	2.5		18	V
I _{CC}	Circuit current	Except output current	3	4.5	8	mA
V _S	Regulated output voltage	Between ④ and ③	1.8	2.0	2.2	V
V _{TH②}	Input threshold voltage		-50	0	50	mV
R _{IN②}	Input impedance		4.2	7.9	12	kΩ
I _{SC⑤}	Limited output current	R _{SC} =27Ω	20	27	35	mA
T _τ	Pulse width of mono. -multi.	R _τ =75k, C _τ =4700pF	375	395	415	μs
R _{eg-V_{CC}}	Motor speed stability for V _{CC}	V _{CC} =4~15V		±0.1		%
R _{eg-L}	Motor speed stability for load			±0.1		%
TC _N	Motor speed stability for temperature	T _a =-20~+75°C		±10		ppm/°C

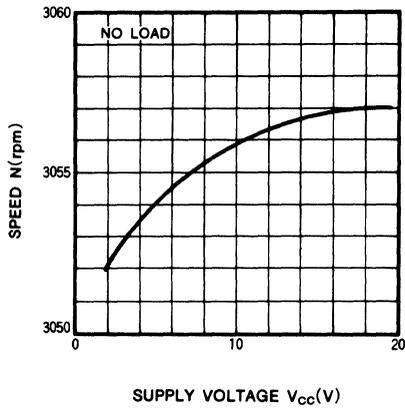
TYPICAL CHARACTERISTICS



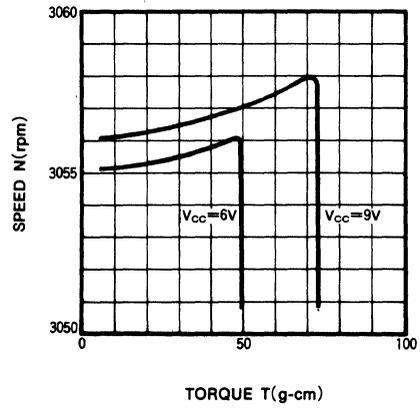
(1) IC only
 (2) With printed circuit board of 2mm thick cover with Cu area of 4.5cmW×5.5cmL×35μt
 Following datas come from the "APPLICATION CIRCUIT" applying the following components.
 R₁=100kΩ, R₂=30kΩ, C_{F1}=1μF, C_{F2}=4.7μF, R_F=4.7kΩ,
 R_τ=75kΩ, C_τ=22,000pF, R_{SC}=56Ω, No. of tachogeneratorpoles=10. Components R, C, were located outside of the temperature test chamber in case of measuring "Speed VS ambient temperature".

SPEED CONTROL FOR DC MICRO MOTOR

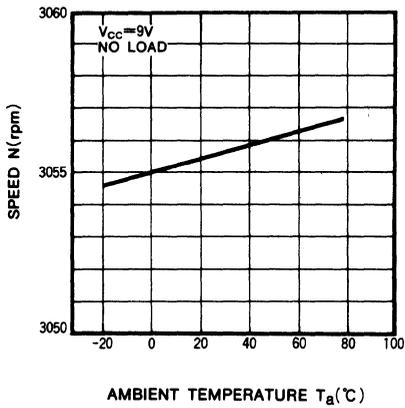
SPEED VS SUPPLY VOLTAGE



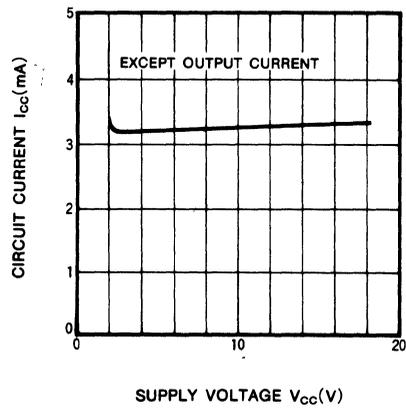
SPEED VS MOTOR TORQUE



SPEED VS AMBIENT TEMPERATURE

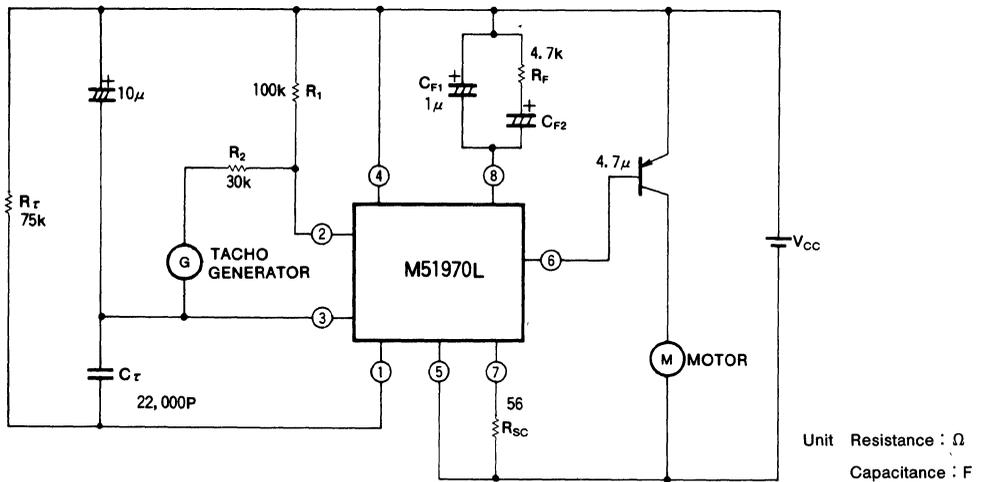


CIRCUIT CURRENT VS POWER SUPPLY

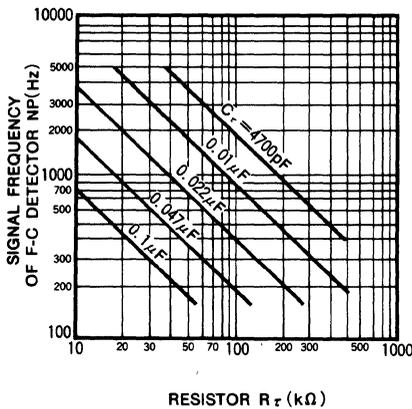


SPEED CONTROL FOR DC MICRO MOTOR

APPLICATION EXAMPLE



● HOW TO DETERMINE R_r , C_r



R_r , C_r determine the speed of motor

$$NP \approx \frac{1}{1.17 R_r C_r}$$

where N; the speed of motor

P; number of pole pair of F-G detector.

The desirable range of R_r is usually 10k to 500kΩ.

APPLICATION HINTS

(1) HOW TO DECIDE THE CONSTANTS OF FILTER AT (8) PIN

The dynamic characteristics of a motor is determined by the relationships between the constants which a motor originally has, (such as, mechanical constant or inertia and tachometer frequency) and circuit constants (C_{F1} , C_{F2} , R_F in the typical application circuit)

The following relationships will be recommendable to choose circuit constants when a cycle of tachometer generator is $T_G (=1/f_G)$, and the mechanical time constant of a motor is τ_M .

1. C_{F1} should be a smaller value in order to improve the circuit response. (But if the value is too small, peak to peak value of ramp shaped driving wave increases and becomes pulsive driven).

$$F_G \times C_{F1} = 50 \sim 150 \text{ Hz} \cdot \mu\text{F}$$

2. The relationship between the time period of a tachometer generator, and the motor constant should be following, because the control is more unstable if speed information is less,

$$T_G / \tau_M \ll 1 (T_G < \frac{1}{8} \tau_M \text{ at least})$$

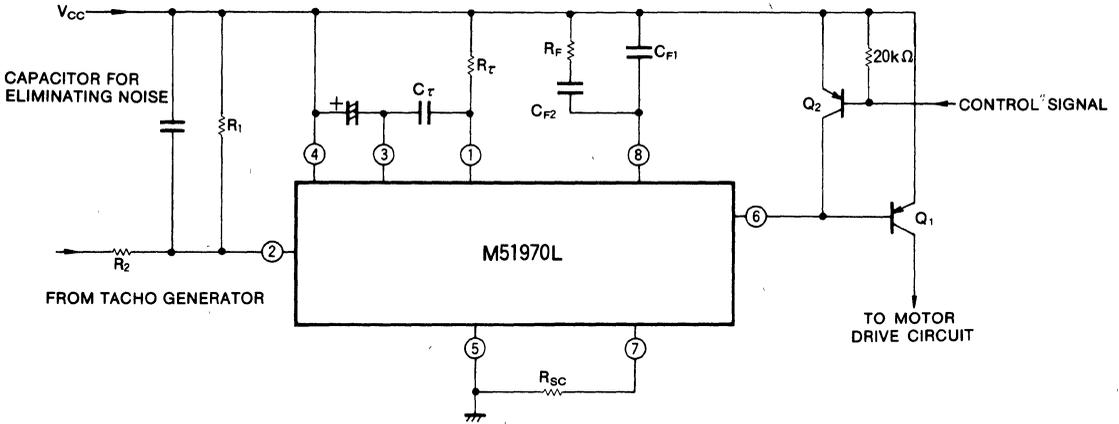
3. Relationship between R_F , C_{F2} and τ_M

$$\tau_M \sim R_F \times C_{F2}$$

SPEED CONTROL FOR DC MICRO MOTOR

(2) HOW TO GET A MOTOR ON OR OFF BY THE CONTROL SIGNAL.

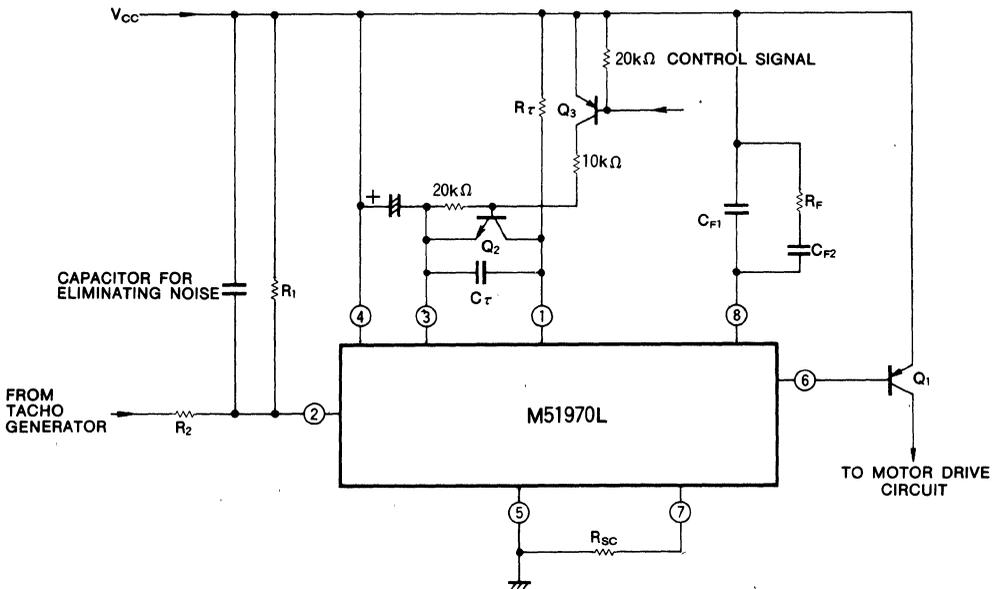
Example 1



The motor is off when Q_2 is on.

Choose the constants as the current driving ability of $Q_2 \geq V_{BE}/R_{SC} \sim 0.7V/R_{SC}$

Example 2

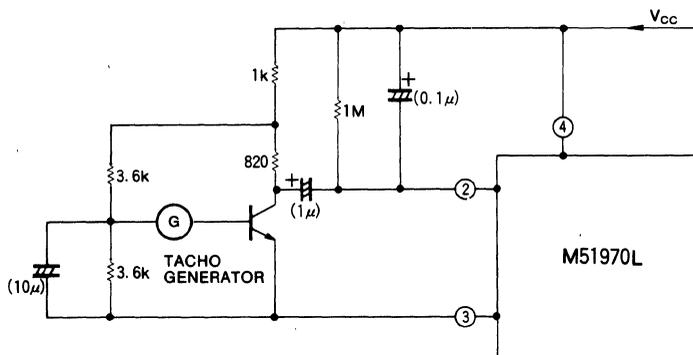


The motor is off when Q_2 and Q_3 are on.

Use example 2 in case overshoot is large in example 1.

SPEED CONTROL FOR DC MICRO MOTOR

(3) HOW TO AMPLIFY THE SIGNAL FROM TACHO GENERATOR WHEN IT IS SMALL.

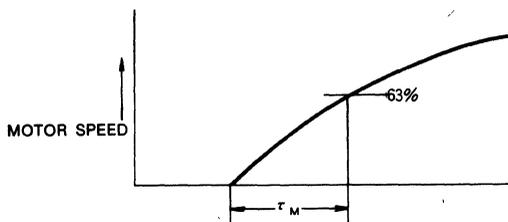
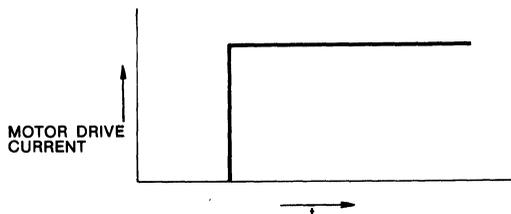


Unit Resistance : Ω
Capacitance : F

The above constants of capacitors are example in case $F_G=500\text{Hz}$.

(4) HOW TO GET THE APPROXIMATE VALUE OF THE MECHANICAL CONSTANT OF A MOTOR, τ_M .

First, drive the stationary motor by the step current. Then measure the time until the motor speed reaches 63% of the final value as following.



MITSUBISHI LINEAR ICs

M51971L, FP

SPEED CONTROL FOR MOTOR

DESCRIPTION

The M51971L,FP are semiconductor integrated circuits designed for use as speed controls for motors.

They include a high-gain F-G amplifier and can be used for a speed detector (F-G detector) for a wide range of speeds.

They control the speed of DC motors with high-precision, using few externally connected components.

FEATURES

- Wide supply voltage range 4~17.5V
- Supply voltage regulation 0.005%/V(typ.)
- Load regulation 0.01%(typ. full load range)
- Temperature coefficient of speed 7ppm/°C(typ.)
- Including a high-performance F-G amplifier

APPLICATION

Speed control of motors for floppy-disk drives, players, tape recorders, car stereo.

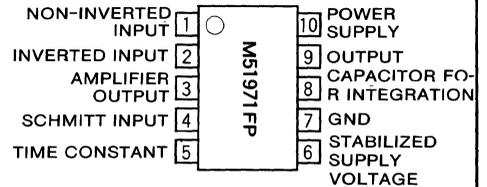
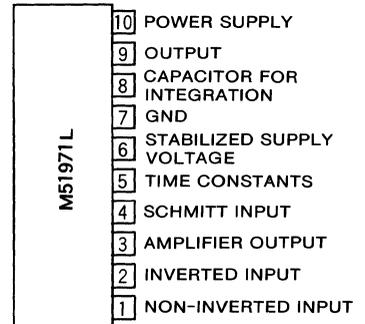
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 4~17.5V
- Rated supply voltage 9V
- Pin ① input voltage range 0.4V~ V_{CC} (Note 1)
- Pin ④ input voltage range 0.4V~ V_{CC}
- Maximum tachometer frequency 2.5kHz
- Minimum trigger pulse width (Pin ④ input pulse) 40 μ s(Note 2)

Note 1. Linear operation range $-0.4\sim+0.4$ V

Note 2. This condition is applicable to both the periods from rise-time to fall-time and from fall-time to rise-time.

PIN CONFIGURATION (TOP VIEW)

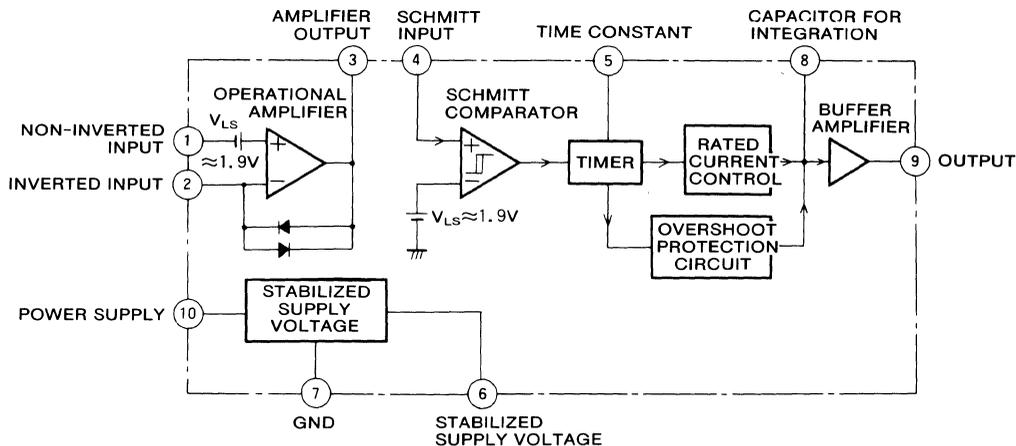


10-pin molded plastic FLAT



10-pin molded plastic SIP

BLOCK DIAGRAM



SPEED CONTROL FOR MOTOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
$V_{①}$	Pin ① applied voltage		$-3 \sim V_{CC}$	V
$I_{③}$	Pin ③ sink current		-5	mA
$I_{⑥}$	Pin ⑥ sink current		-5	mA
$V_{④}$	Pin ④ applied current		$0 \sim V_{CC}$	V
$I_{⑨}$	Pin ⑨ sink current		-20	mA
P_{dF}	Power dissipation		880(M51971L)	mW
			450(M51971FP)	
$K_{\theta F}$	Thermal derating	$T_a \geq 25^\circ\text{C}$	8.8(M51971L)	mW/°C
			4.5(M51971FP)	
T_{opr}	Operating temperature		$-20 \sim +75$	°C
T_{stg}	Storage temperature		$-40 \sim +125$	°C

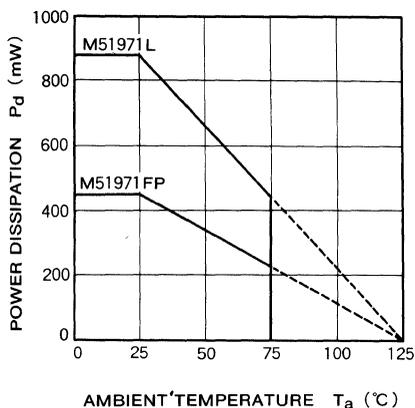
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=9\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage range		4.0		17.5	V
I_{CC}	Circuit current			3.2	6.0	mA
V_S	Stabilized supply voltage	Pin ⑥ voltage	2.44	2.71	2.98	V
$I_{①}$	Pin ① input current	$V_{①}=0\text{V}$	-3.0	-0.5		μA
$I_{②}$	Pin ② input current	$V_{①}=0\text{V}$	-180	-30		nA
$V_{①LS}$	Pin ① level shift voltage	$V_{①}=0\text{V}$	1.51	1.89	2.27	V
A_V	FG amplifier voltage gain	$V_{①}=0.2\text{mVrms}$, $f=500\text{Hz}$, externally set gain=60dB	54	59	64	dB
$I_{④}$	Pin ④ input current	$V_{④}=2.5\text{V}$		0.4	2.0	μA
$V_{④TH}$	Pin ④ threshold voltage	Reference is the Pin ① level shift voltage	0	16	40	mV
$V_{④HY}$	Pin ④ hysteresis width		20	37	55	mV
$V_{⑤S}$	Pin ⑤ saturation voltage	$R_\tau = 75\text{k}\Omega$		3	20	mV
T_τ	One-shot pulse width	$R_\tau = 75\text{k}\Omega$, $C_\tau = 4700\text{pF}$	375	395	415	μsec
$I_{⑧C}$	Pin ⑧ charge current	$V_{⑧}=1\text{V}$	-260	-190	-140	μA
$I_{⑧D}$	Pin ⑧ discharge current	$V_{⑧}=1\text{V}$	-14.5	-11.6	-9.0	
$R_{⑨}$	Pin ⑨ output protection resistor	$I_{⑨} = -20\text{mA}$	65	100	150	Ω
$V_{⑨max}$	Pin ⑨ maximum voltage		2.9	3.2		V
$V_{⑨min}$	Pin ⑨ minimum voltage			50	200	mV
V_{BO}	Buffer amplifier offset voltage	$V_{⑧}=1\text{V}$, $V_{⑨} = -V_{⑨}$	0	100	200	mV
$Reg-V_{CC}$	Supply voltage regulation	$V_{CC}=4 \sim 17.5\text{V}$		0.07		%
$Reg-L$	Supply voltage load	full load range		0.01		%
TC_N	Temperature coefficient of speed	$T_a = -20 \sim +75^\circ\text{C}$		7		ppm/°C

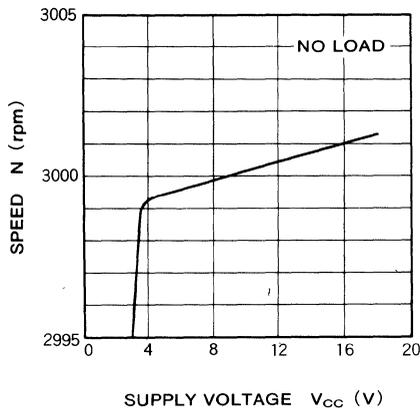
SPEED CONTROL FOR MOTOR

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

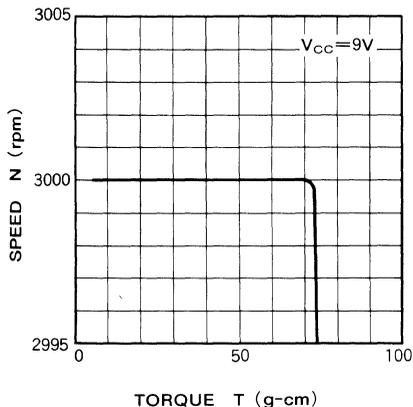
THERMAL DERATING (MAXIMUM RATING)



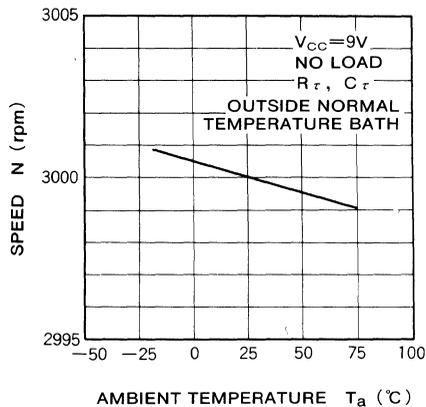
SPEED VS. SUPPLY VOLTAGE



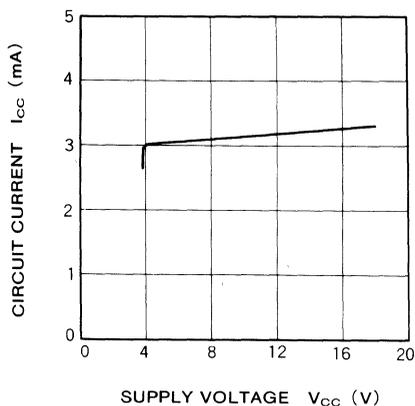
SPEED VS. TORQUE



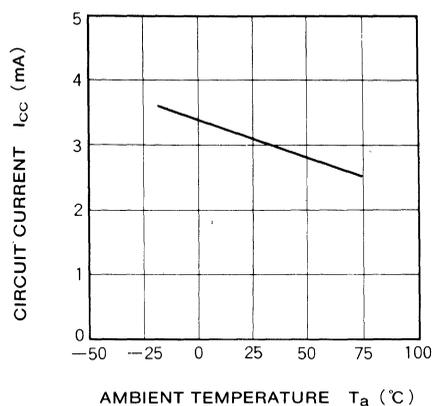
SPEED VS. AMBIENT TEMPERATURE



CIRCUIT CURRENT VS. SUPPLY VOLTAGE

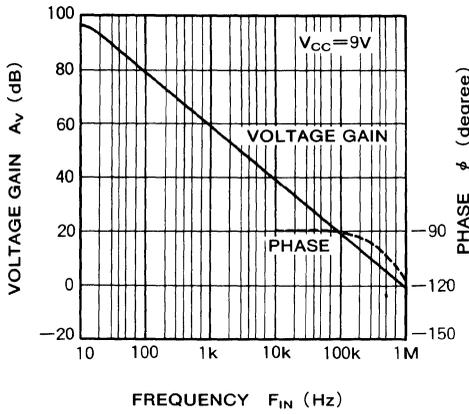


CIRCUIT CURRENT VS. AMBIENT TEMPERATURE

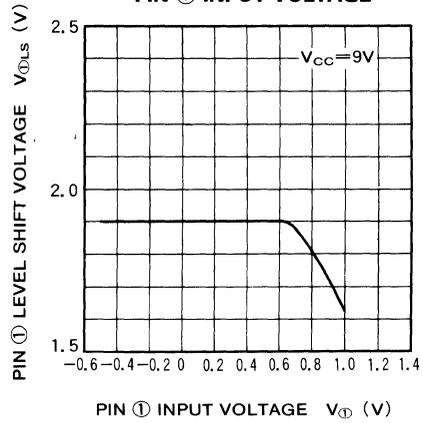


SPEED CONTROL FOR MOTOR

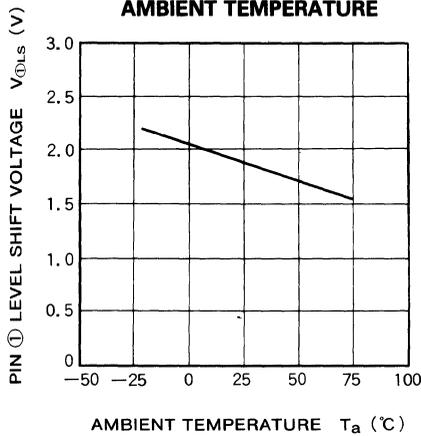
**FG AMPLIFIER OPEN LOOP
 VOLTAGE GAIN VS. PHASE SHIFT**



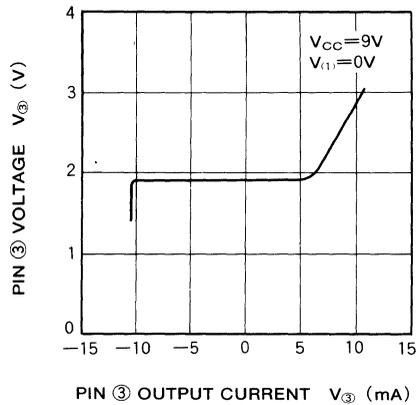
**PIN ① LEVEL SHIFT VOLTAGE VS.
 PIN ① INPUT VOLTAGE**



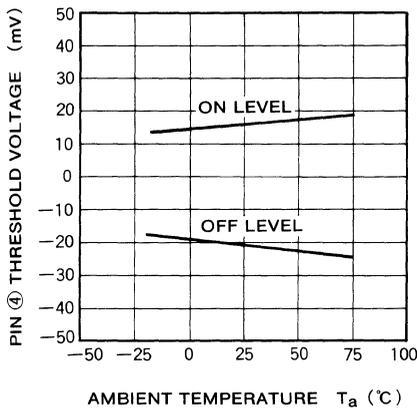
**PIN ① LEVEL SHIFT VOLTAGE VS.
 AMBIENT TEMPERATURE**



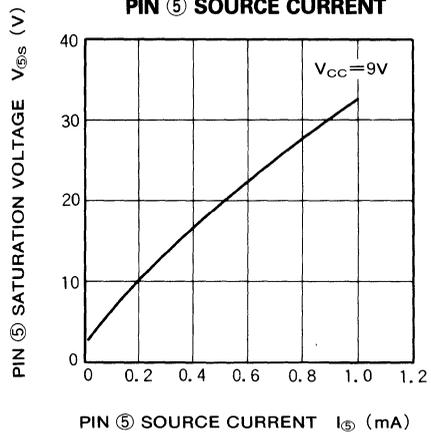
**PIN ③ VOLTAGE VS.
 PIN ③ OUTPUT CURRENT**



**PIN ④ THRESHOLD VOLTAGE VS.
 AMBIENT TEMPERATURE**

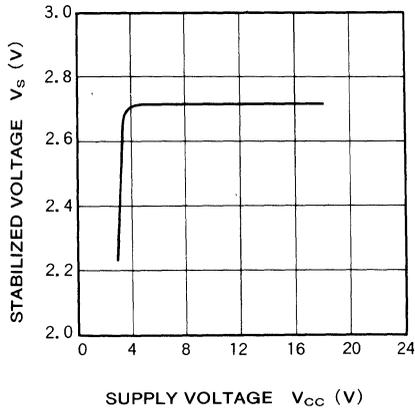


**PIN ⑤ SATURATION VOLTAGE VS.
 PIN ⑤ SOURCE CURRENT**

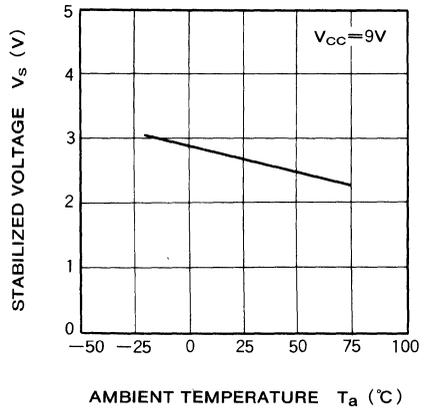


SPEED CONTROL FOR MOTOR

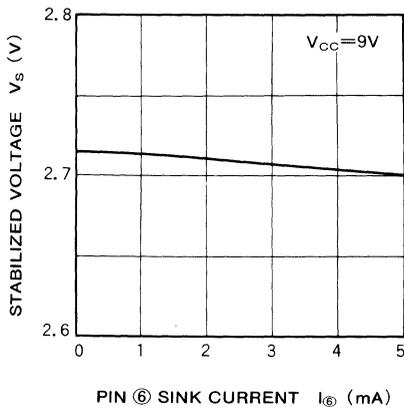
STABILIZED VOLTAGE VS. SUPPLY VOLTAGE



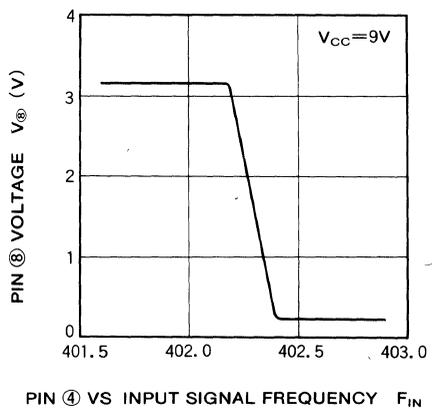
STABILIZED VOLTAGE VS. AMBIENT TEMPERATURE



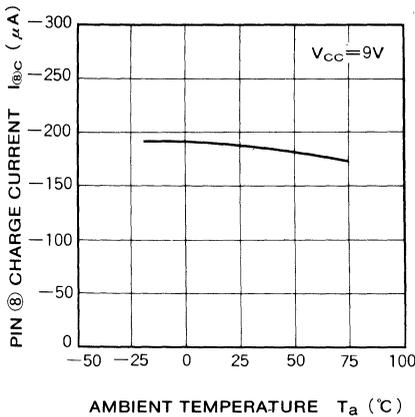
STABILIZED VOLTAGE VS. PIN ⑥ SINK CURRENT



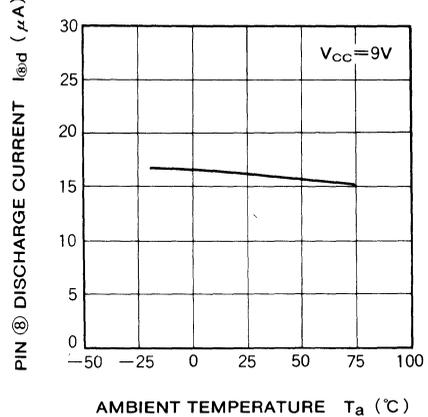
PIN ⑧ VOLTAGE VS. INPUT SIGNAL FREQUENCY



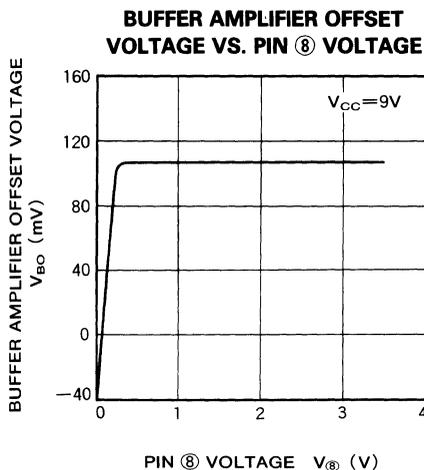
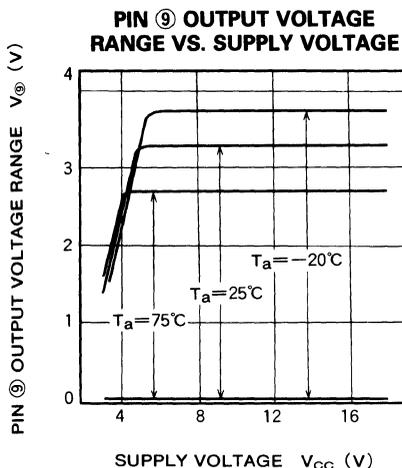
PIN ⑧ CHARGE CURRENT VS. AMBIENT TEMPERATURE



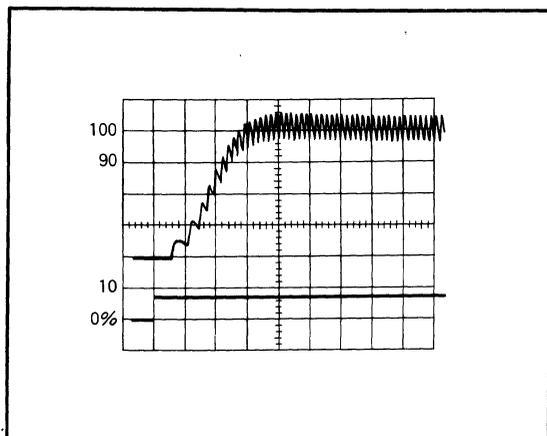
PIN ⑧ DISCHARGE CURRENT VS. AMBIENT TEMPERATURE



SPEED CONTROL FOR MOTOR



RESPONSE CHARACTERISTICS

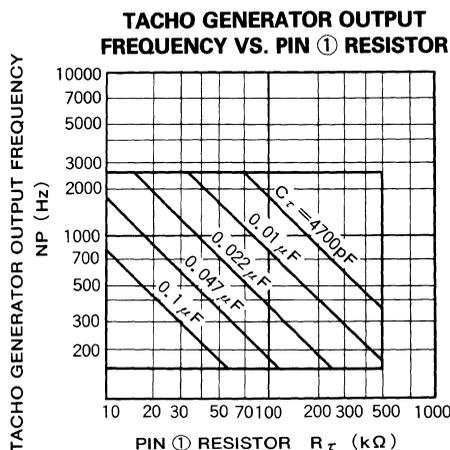


Upper : Motor speed (IF/V converting waveform of tachogenerator frequency)
 Lower : Supply voltage
 X-axis : 20ms/div
 Time constant of motor $\approx 100\text{ms}$

R_τ, C_τ

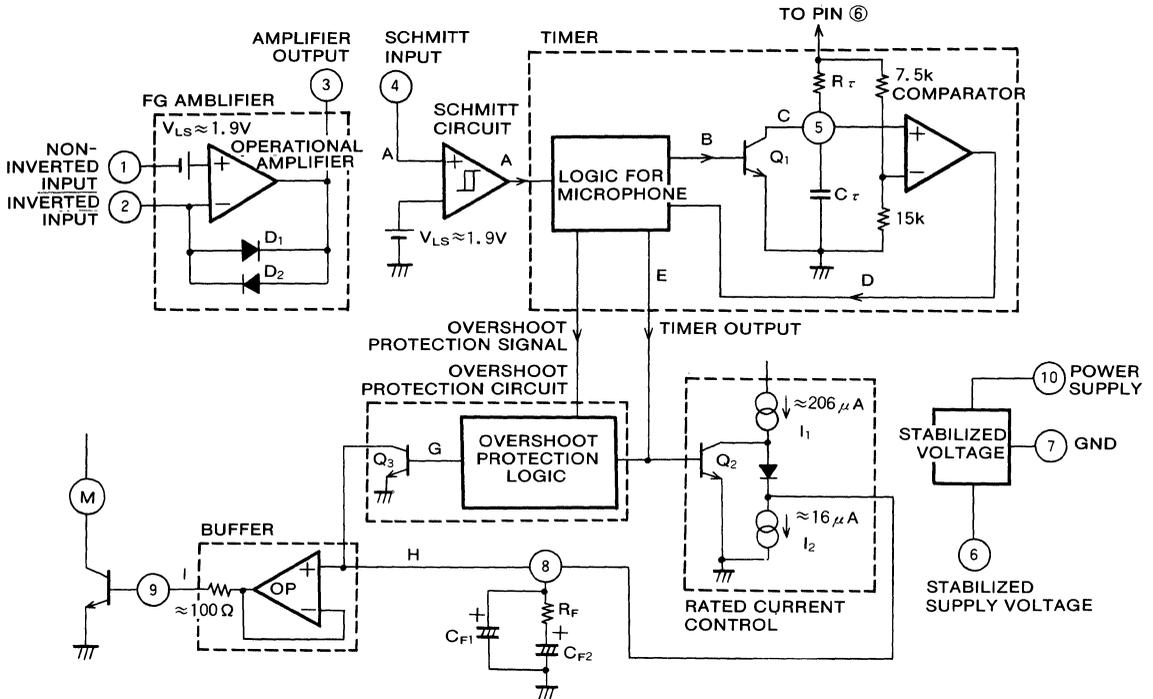
R_τ, C_τ determine the speed of motor. The following equation is obtained where N is the speed of motor and P is the number of pole pairs of tachogenerator. The desirable range of R_τ is $10\text{k}\Omega$ to $500\text{k}\Omega$. Care should be taken for leak current on the surface of printed circuit board when a large resistor is used.

$$NP \approx \frac{1}{1.20 \cdot R_\tau \cdot C_\tau}$$



SPEED CONTROL FOR MOTOR

OPERATIONAL DESCRIPTION OF THE M51971
DESCRIPTION OF EACH BLOCK



F-G amplifier

The F-G amplifier consists of an operational amplifier, level shift circuit and diode for waveform clipping.

The output DC voltage at Pin ③ is increased from the DC voltage at Pin ① by the V_{LS} ($\approx 1.9V \approx 3V_{BE}$) when a DC protection capacitor is connected to Pin ②.

AC signals near the GND can be easily amplified. The clipper diode can limit the output signal width within $\pm 0.7V(V_{BE})$, and rapidly charges a capacitor for DC protection at power on.

Schmitt Circuit

The schmitt circuit is a comparator with hysteresis, and has the ON level of $V_{LS} + 20mV$, and the OFF level of $V_{OS} - 20mV$.

Timer

The timer circuit generates reference time required for speed control of motor. The timer consists of a one-shot circuit, triggered by the input signal, and generates pulse with pulse width of $1.1C\tau R\tau$.

Rated Current Limiting Circuit

The rated current limiting circuit is controlled by the output from the timer circuit and generates source current of $I_1 - I_2$ ($\approx 190\mu A$) when one-shot pulse is not gener-

ated and generates source current of I_2 ($\approx 190\mu A$) when one-shot pulse is generated. The ratio of I_1 against I_2 is unique for each IC and the tachogenerator frequency to be set is determined by the one-shot pulse width and this ratio of currents ($I_1/I_2 \approx 12.6$),

$$T_G = T\tau \times \frac{I_1}{I_1 - I_2} \approx 1.09 \times T\tau$$

where T_G : Tachogenerator frequency period (set value)

$T\tau$: One-shot pulse width

Overshoot Protection Circuit

The overshoot protection circuit operates when the overshoot is exceptionally large, for example, when motor is released from locked condition.

The Q_3 is turned ON for approximately the period of one-shot pulse width ($T\tau$), when the signal period of tachogenerator becomes shorter than one-shot pulse width. C_{F1} is discharged because usually $R_F \cdot C_{F2} \gg T\tau$.

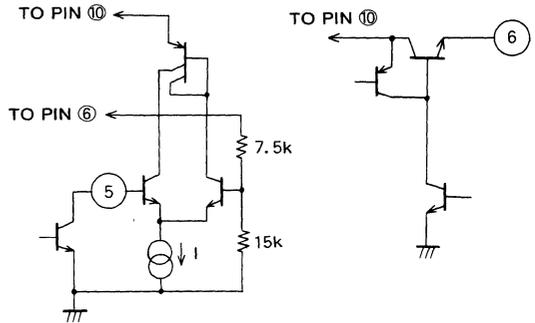
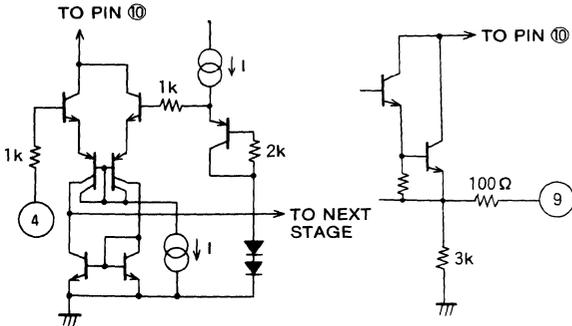
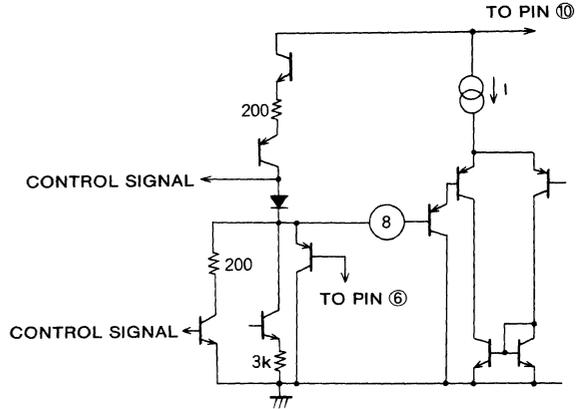
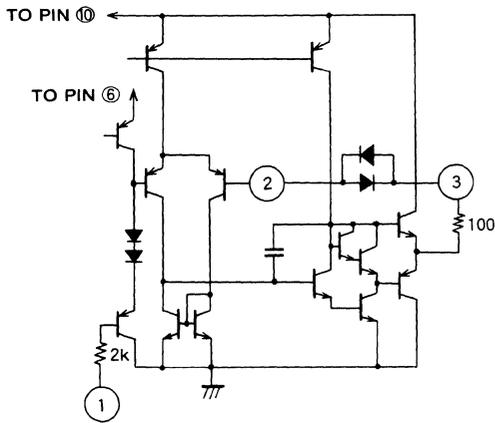
Buffer Amplifier

The buffer amplifier circuit consists of a voltage-follower circuit using operational amplifier. The input current is very low (within $10nA$), and the driven output current is $20mA$.

SPEED CONTROL FOR MOTOR

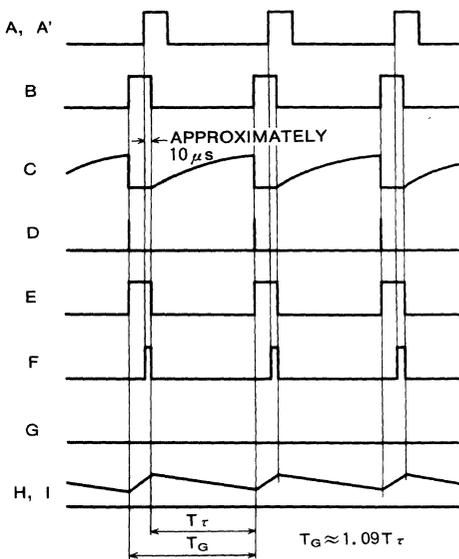
INPUT/OUTPUT CIRCUITS

Unit . Ω

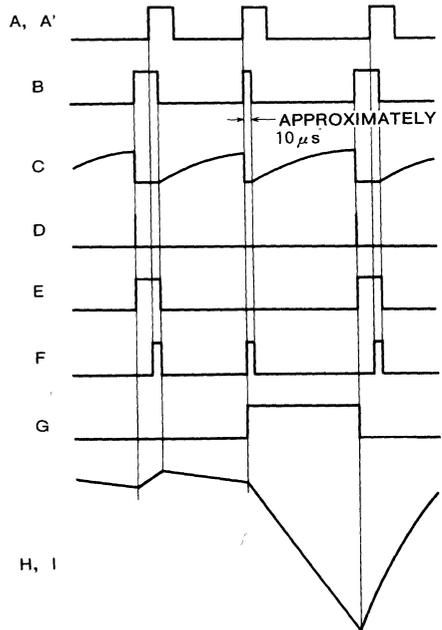


TIMING CHARTS

I. NORMAL OPERATION



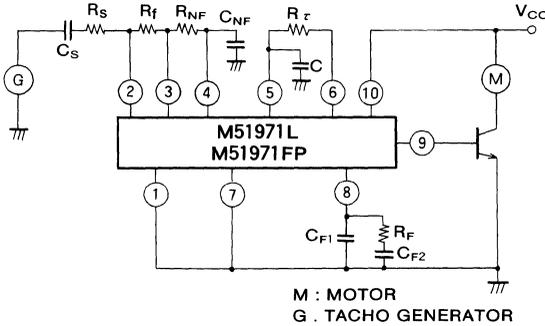
II. NORMAL OPERATION TO QUICK DISCHARGE OPERATION



SPEED CONTROL FOR MOTOR

APPLICATION EXAMPLES

I. When output impedance of tacho generator is low



C_S : Coupling capacitor for AC amplifier
 R_S, R_f : Resistor for setting FG amplifier gain
 R_{NF}, C_{NF} : Filter for eliminating noise
 R_r, C_r : Time constants for setting speed of motor
 C_{F1}, C_{F2}, R_F : Capacitor and resistor for phase compensation to stabilize integration and speed control systems

Note

1. The tacho generator signal width of set speed of motor should be less than $1mV_{P.P.}$.

$$2. \text{FG amplifier gain} \approx \sqrt{\frac{1 + \omega_G^2 C_S^2 (R_S + R_f)^2}{1 + \omega_G^2 C_S^2 R_S^2}}$$

ω_G : Tacho generator width angle frequency

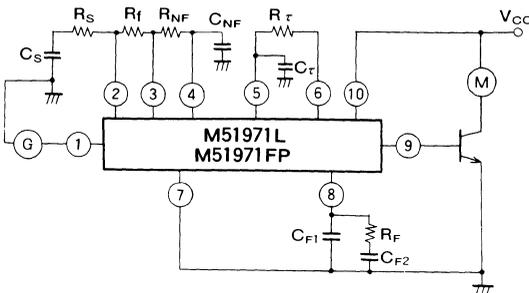
3. The desirable values for C_S, R_S, R_{NF}, C_{NF} are as follows.

$$C_S \leq 4.7 \mu F$$

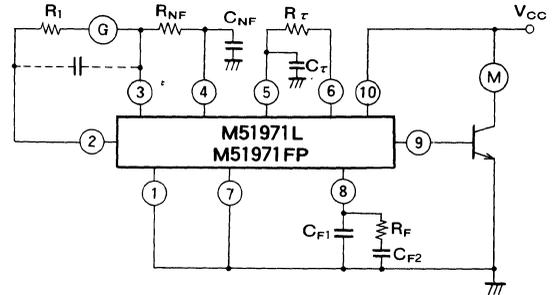
$$\frac{2}{\omega_G} \geq C_S R_S \geq \frac{1}{\omega_G}$$

$$R_{NF} \cdot C_{NF} \leq \frac{1}{\omega_G}$$

II. When the output impedance of tacho generator is high and the signal width is small

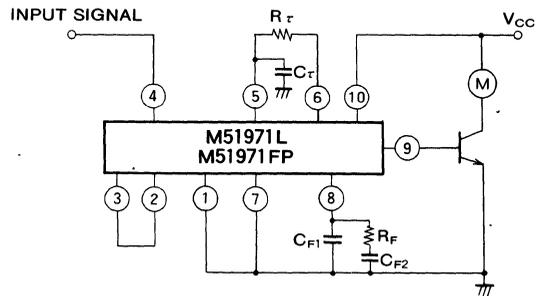


III. When the signal width of tacho generator is large



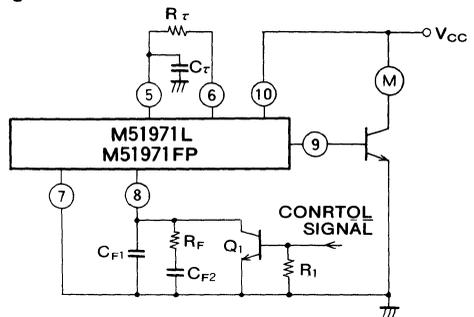
The output waveform exceeding the value of $V_f(0.7V)$ in the above three examples are clipped by an internal waveform clipping diode.

IV. When the input waveform is pulsive



NOTE. The Pin ④ threshold voltage to GND is approximately 1.9V.

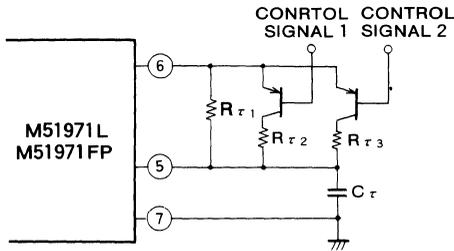
V. When a motor is turned ON/OFF by the control signal



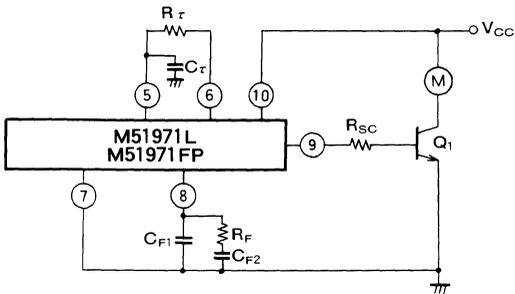
The motor is off when Q_1 is ON.
 The speed control is on when Q_1 is OFF.

SPEED CONTROL FOR MOTOR

VI. When the set speed needs to be changed gradually by the control signal



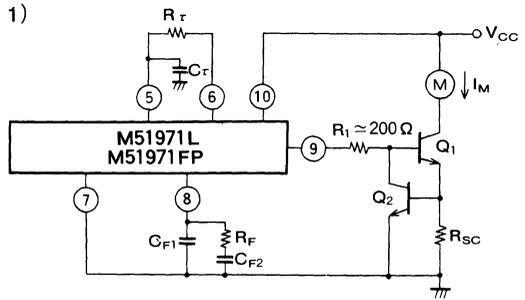
VII. To limit the Pin 9 output current to suppress the heat of the IC



$$I_{9max} = \frac{V_{9max}}{R_{9} + R_{SC}} \quad V_{9max} \approx 3.2V, R_{9} \approx 100\Omega$$

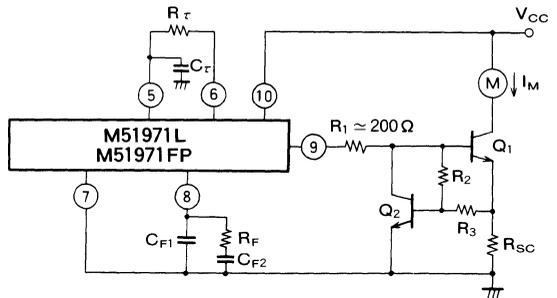
(Refer to ELECTRICAL CHARACTERISTICS and TYPICAL CHARACTERISTICS)

VIII. To limit the drive current of motor



$$I_{Mmax} = \frac{V_{BE2}}{R_{SC}} \approx \frac{0.7V}{R_{SC}}$$

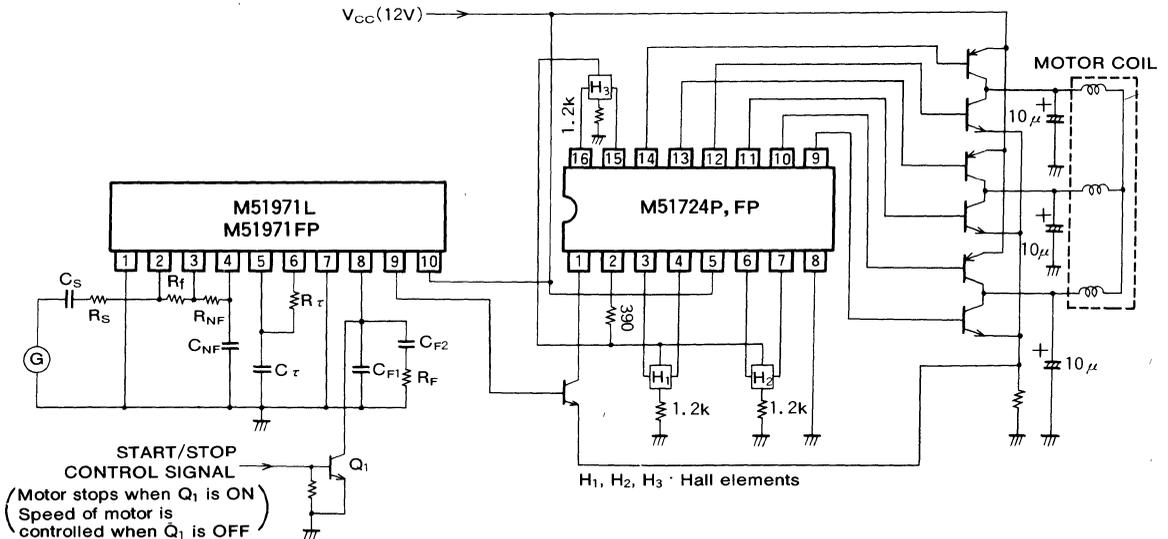
2) To minimize the power loss of current limiting resistor



$$I_{Mmax} = (V_{BE2} - V_{BE1} \times \frac{R_3}{R_2 + R_3}) / R_{SC}$$

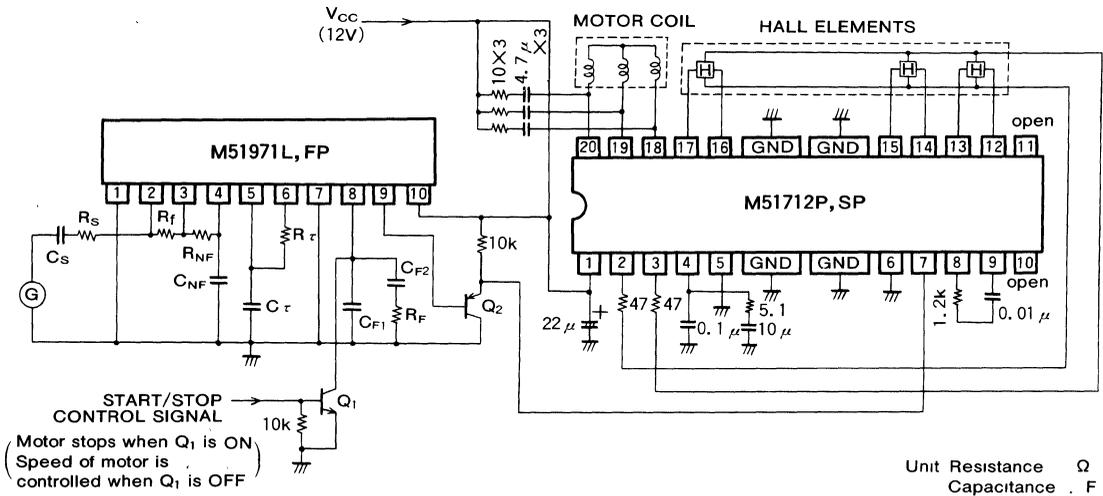
$$\approx \frac{0.7V \times R_2}{(R_2 + R_3) \cdot R_{SC}}$$

IX. Speed control drive circuit for 3-phase brushless motor, using the M51971 and M51924

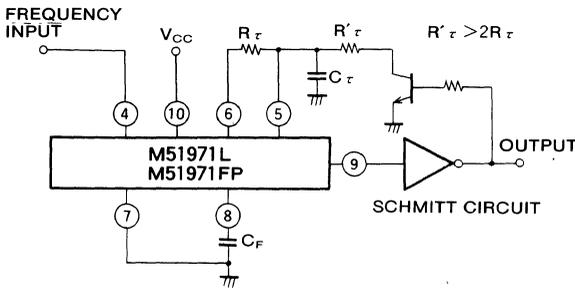


SPEED CONTROL FOR MOTOR

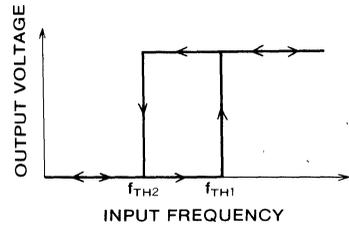
X. Speed control drive circuit for 3-phase brushless motor, using the M51971 and M51712



XI. Frequency comparator



Input/output transfer characteristics



Note. The hysteresis of Schmitt circuit should be higher than ripple voltage at Pin ⑧ (to prevent chattering)

$$f_{TH1} \approx \frac{1}{1.20 \times R_{\tau} \cdot C_{\tau}}$$

$$f_{TH2} \approx \frac{1}{1.09 \times R_{\tau} // R'_{\tau} \times C_{\tau} \times \ln \left\{ \frac{3(R_{\tau} + R'_{\tau})}{R'_{\tau} - 2R_{\tau}} \right\}}$$

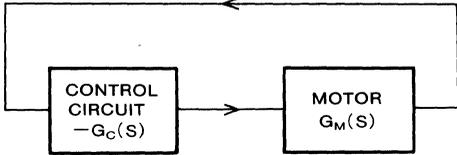
SPEED CONTROL FOR MOTOR

SOME HINTS TO DESIGN STABLE SPEED CONTROL SYSTEM

(How to determine filter constants at Pin ⑧ (C_{F1}, C_{F2}, R_F))

The filter constants at Pin ⑧ must be determined to satisfy the stability of the system.

1. Transfer function of speed control system for motor



SPEED CONTROL SYSTEM FOR MOTOR

The speed control system for the motor is a negative feedback system, including control circuit and motor.

To achieve stable negative feedback, the phase stays 180° or less in the frequency range in which the gain of cyclic transfer function (G_C(S) · G_M(S)) is greater than 1.

2. Transfer function of motor

With the armature current of motor as I_a and angle speed as ω_v, the following equation is obtained.

$$\Delta T_g = K_T \cdot \Delta I_a = (S J + D) \cdot \Delta \omega_v \dots (1)$$

where T_g: torque generated by motor

K_T: proportional constant between torque generated by motor and armature current

J: inertia moment of motor and load

D: viscous friction coefficient

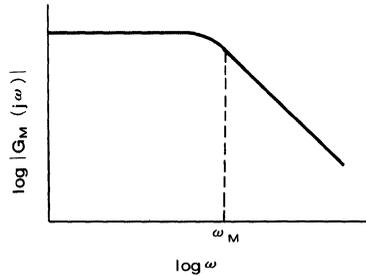
As the relationship between tacho generator angle frequency ω and motor angle speed ω_v, with number of pole pairs of tacho generator ad P, is expressed as ω = P · ω_v, the transfer function of motor (transfer function including motor and tacho generator) G_M(S) is a single-pole transfer function, and is given by

$$G_M(S) = \frac{\Delta \omega}{\Delta I_a} = \frac{P \cdot K_T}{D \cdot (1 + S \cdot \frac{J}{D})} \dots (2)$$

$$= \frac{K_M}{1 + \frac{S}{\omega_M}} \dots (3)$$

where $K_M = \frac{P \cdot K_T}{D} \dots (4)$

$$\omega_M = \frac{D}{J} \dots (5)$$



Simulated transfer function of the motor

3. Transfer function of control circuit using the M51971

Under the assumption that input information is continuously given, (tacho generator frequency is assumed to be infinitely high), the transfer function from the input at Pin ④ to the output at Pin ⑨ is given as follows;

$$G_{C(M51971)}(S) = \frac{\Delta(\text{PIN ⑨ OUTPUT VOLTAGE})}{\Delta(\text{PIN ④ INPUT FREQUENCY})} = \frac{T_\tau (|I_{\text{⑨c}}| + |I_{\text{⑨d}}|)}{C_{F1} + C_{F2}} \times \frac{1 + S/\omega_{F1}}{S(1 + S/\omega_{F2})} \dots (6)$$

where T_τ: pulse width of timer ≈ 1.10 × R_τ × C_τ

I_{⑨c}: charge current at Pin ⑧

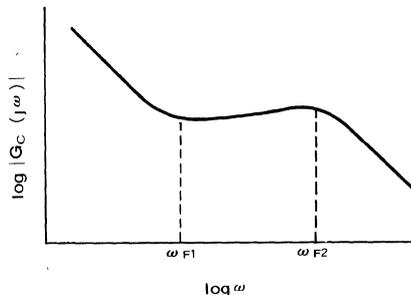
I_{⑨d}: discharge current at Pin ⑧

$$\omega_{F1} = \frac{1}{R_F \cdot C_{F2}}$$

$$\omega_{F2} = \frac{C_{F1} + C_{F2}}{R_F \cdot C_{F1} \cdot C_{F2}}$$

With the gain of the circuit connected after Pin ⑨ of the M51971 as K_{CP}, the transfer function of the whole circuit G_C(S) is given in the following;

$$G_C(S) = K_{CP} \times \frac{T_\tau (|I_{\text{⑨c}}| + |I_{\text{⑨d}}|)}{C_{F1} + C_{F2}} \times \frac{1 + S/\omega_{F1}}{S(1 + S/\omega_{F2})} \dots (7)$$



Simulated transfer function of the control circuit

SPEED CONTROL FOR MOTOR

4. Requirement for stable control

For stable control of motor, the phase characteristics must be less than 180 in the frequency fange where the gain of $G_C(S) \cdot G_M(S)$ is greater than 1.

The relationship between phase and gain is determined by the Baud's theorem when all the poles and zero pints of transfer function are on the left of the complex sphere.

If the $G_C(j\omega) \cdot G_M(j\omega)$ follows the Baud's theorem, the inclination of gain of $G_C(j\omega) \cdot G_M(j\omega)$ should be more than -12dB/oct in the frequency range of $|G_C(j\omega) \cdot G_M(j\omega)| \geq 1$ to ensure stable control.

When circuits constants are chosen so that $\omega_{F1} \approx \omega_M$, the following equation must be established in the frequency range where the inclination of gain of $G_C(j\omega)$ and $G_M(j\omega)$ are both -6dB/oct , in other words, the inclination of gain of $G_C(j\omega) \cdot G_M(j\omega)$ becomes -12dB/oct .

For precision control, the gain of one cycle transfer function must be large in all frequency ranges.

Fluctuation of speed of motor due to external disturbance is reduced if the inclination is -6dB/oct in the frequency range higher than ω_M .

Therefore, capacity of speed control in the frequency range from ω_{F1} to ω_{F2} is determined by the gain of one cycle transfer function when $\omega_{F1} (\approx \omega_{F2})$. The following equation is established as $|G_C(j\omega_{F2}) \cdot G_M(j\omega_{F2})| < 1$ and the inclination of gain of $G_C(j\omega) \cdot G_M(j\omega)$ is approximately -6dB in the frequency range of ω_{F2} or less.

$$|G_C(j\omega_M) \cdot G_M(j\omega_M)| < \frac{\omega_{F2}}{\omega_{F1}} \approx \frac{\omega_{F2}}{\omega_M} \dots\dots\dots (9)$$

Therefore, the following conditions are required to improve accuracy of control in the frequency range from ω_{F1} to ω_{F2} .

$$\omega_{F1} \approx \omega_M \dots\dots\dots (10)$$

$$\frac{\omega_{F2}}{\omega_{F1}} \gg 1 \dots\dots\dots (11)$$

Then, the values of K_{CP} or $C_{F1} + C_{F2}$ must be set to satisfy the equations (4) and (5).

5. Influence of tacho generator frequency on stability

The tacho generator frequency, which is a system controlled by period, is basically a kind of sample hold system controlled by discrete information on the time axis.

The sample hold operation only adds an extra phase delay only to make the system unstable.

$$H^*(j\omega) = \frac{\sin \pi (\omega / \omega_G)}{\pi (\omega / \omega_G)} e^{-\frac{2\pi\omega}{\omega_G} \dots\dots\dots} \sum_{n=-\infty}^{\infty} H(j\omega + jn\omega_G) \dots\dots\dots (12)$$

Here, ω_G : Set value of tacho generator frequency
Additional phase delay of $2\pi \omega / \omega_G$ (rad.) need to be considered.

Therefore, for the angle frequency where $|G_C^*(j\omega) \cdot G_M^*(j\omega)| = 1$ is ω_{odB} , the following equation must be established.

$$\omega_G > 4 \cdot \omega_{odB} \dots\dots\dots (13)$$

With this, if the ω_G is determined, the gain value of one cycle transfer function can be obtained.

$$|G_C(j\omega_M) \cdot G_M(j\omega_M)| < 0.357 \times \frac{\omega_G}{\omega_M} \dots\dots\dots (14)$$

This equation (14) must be satisfied in any control systems using tacho generator frequency irrespective of control method, and, it can be noted that, once the motor and tacho generator are determined, the upper limit value of the control gain in ω_M is determined.

To improve the accuracy of speed control, it is necessary that $|G_C(j\omega_M) \cdot G_M(j\omega_M)| \gg 1$ and, therefore, the following equation must be established.

$$0.357 \cdot \frac{\omega_G}{\omega_M} \gg 1 \dots\dots\dots (15)$$

6. Summary

The following is the method of designing speed control system, making the maximum use of characteristics of motors, based on the above-mentioned theoretical considerations.

$$(1) \omega_{F1} = \frac{1}{R_F \cdot C_{F2}} \approx \omega_M \dots\dots\dots (16)$$

It is desirable that the circuit constant is set closed to the minimum value of ω_M , if the ω_M changes greatly when motor load is changed.

$$(2) \omega_{F2} = \frac{C_{F1} + C_{F2}}{R_F \cdot C_{F1} \cdot C_{F2}} \geq \frac{1}{4} \omega_G \dots\dots\dots (17)$$

The smaller the value of C_{F1} , the less is the influence of ω_{F2} , but the greater the peak-to-peak value of output pin waveform and the drive waveforms becomes more pulsive.

Therefore, both arms of the equation should be equal in most cases of system designing.

(3) Selection of gain

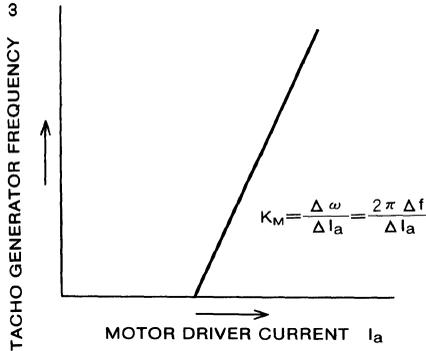
Obtain the value for stable control, changing the values of K_{CP} , C_{F1} or C_{F2} , satisfying at the same time the relationship described in the above (16) and (17).

When setting the stepping motor speed, the speed is usually unstable at slow speed, ad therefore, experiments should be made at the slowest speed.

SPEED CONTROL FOR MOTOR

How to obtain approximate value of transfer function of motor

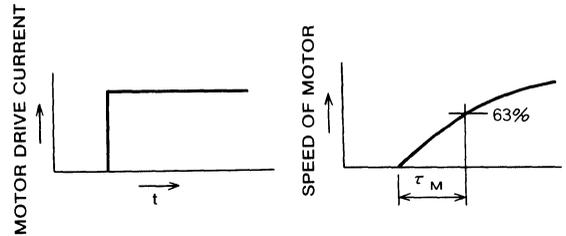
(1) How to obtain K_M



Plot the relationship between motor drive current and tacho generator frequency and obtain the inclination.

(2) How to obtain ω_M

The ω_M can be obtained by measuring frequency response of motor but usually this method takes long time. In stead, the approximate value can be simply obtained by measuring the step response.



The method is to apply stepping current to halted motor and measure the time τ_M when the speed of motor reaches 63% of the final speed and the τ_M is obtained from the following equation.

$$\tau_M = \frac{1}{\tau_M} \dots\dots\dots (18)$$

MITSUBISHI LINEAR ICs M51724P,FP

3-PHASE BRUSHLESS MOTOR PRE-DRIVER

DESCRIPTION

The M51724P,FP are semiconductor integrated circuits designed for use in 3-phase DC brushless motor.

FEATURES

- Suitable for various kind of motor system by selecting the external power transistors
- Internal current distribution circuit
- Good balance in output current between each phase
- Few externally connected parts

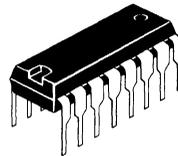
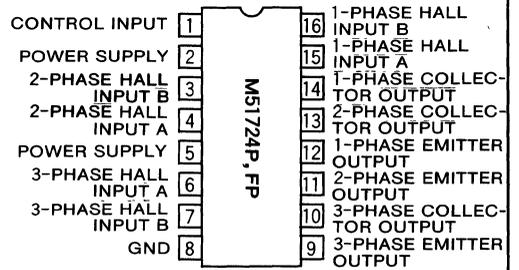
APPLICATION

Brushless motor driver for VTR, cassette tape deck, floppy-disk drive.

RECOMMENDED OPERATING CONDITIONS

Supply voltage range 10V~20V
 Rated supply voltage 15V

PIN CONFIGURATION (TOP VIEW)

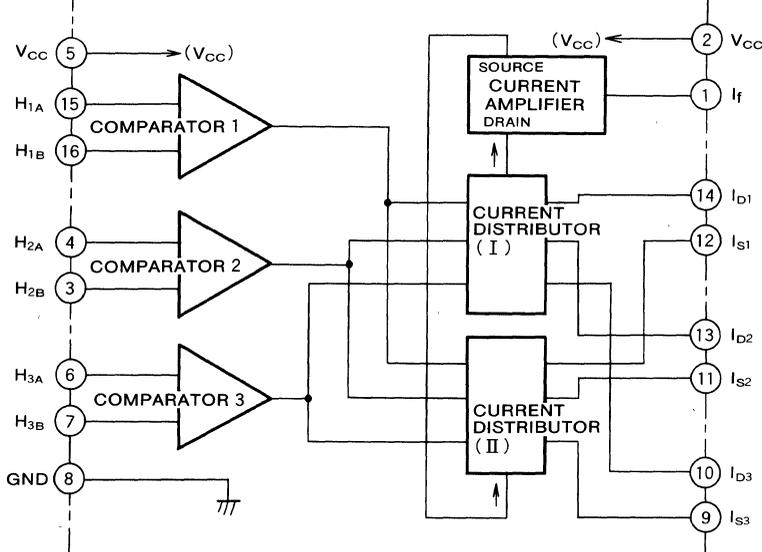


16-pin molded plastic DIP



16-pin molded plastic FLAT

BLOCK DIAGRAM



3-PHASE BRUSHLESS MOTOR PRE-DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$, unless otherwise noted)

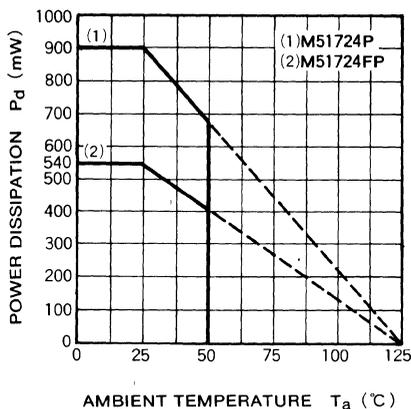
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		20	V
I_f	Control input current		1	mA
V_D	Applied voltage at collector output pin		24	V
V_S	Applied voltage at emitter output pin		6.5	V
V_H	Applied voltage at hall output pins		6.5	V
f_{in}	Hall input frequency		DC~1	kHz
P_d	Power dissipation	()=M51724FP	900(540)	mW
K_{θ}	Thermal derating ($T \geq 25^{\circ}\text{C}$)	()=M51724FP	110(185)	$^{\circ}\text{C}/\text{W}$
T_{opr}	Operating temperature		-10~+50	$^{\circ}\text{C}$
T_{stg}	Storage temperature		-40~+125	$^{\circ}\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, $V_{CC}=15\text{V}$, unless otherwise noted)

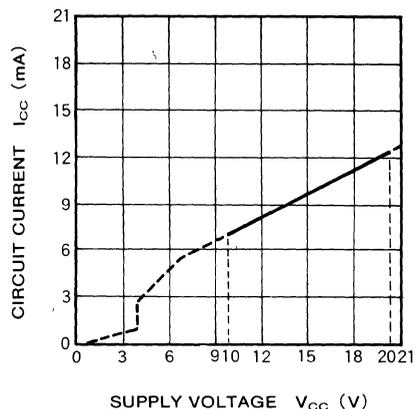
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current		5	12	27	mA
V_{offset}	Comparator input offset voltage			5	20	mV
K_D	Collector output current gain		10	15	20	A/A
K_S	Emitter output current gain		11	16	21	A/A
$I_{D(max1)}$	Maximum output current of collector output (1)		3	4		mA
$I_{D(max2)}$	Maximum output current of collector output (2)		5.5	7		mA
$I_{S(max1)}$	Maximum output current of emitter output (1)		3	4		mA
$I_{S(max2)}$	Maximum output current of emitter output (2)		5	7		mA
M_D	Current gain ratio between collector output phases		0.75	1	1.33	A/A
M_S	Current gain ratio between emitter output phases		0.75	1	1.33	A/A
I_{in}	Comparator input current		0.1	1.5	6	μA
I_{LD}	Collector output leak current				200	nA
I_{LS}	Emitter output leak current				200	nA

TYPICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, unless otherwise noted)

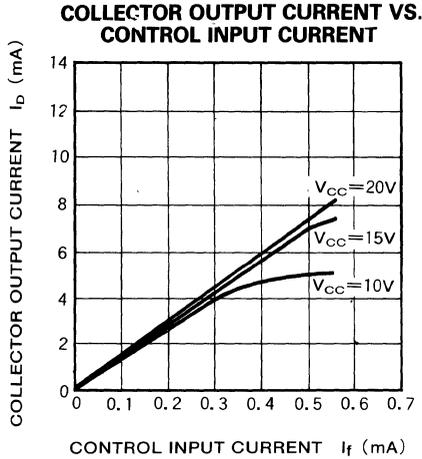
THERMAL DERATING (MAXIMUM RATING)



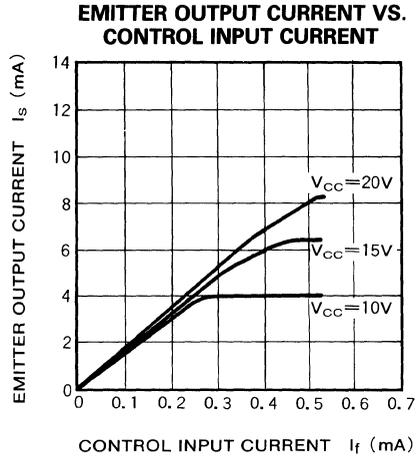
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



3-PHASE BRUSHLESS MOTOR PRE-DRIVER

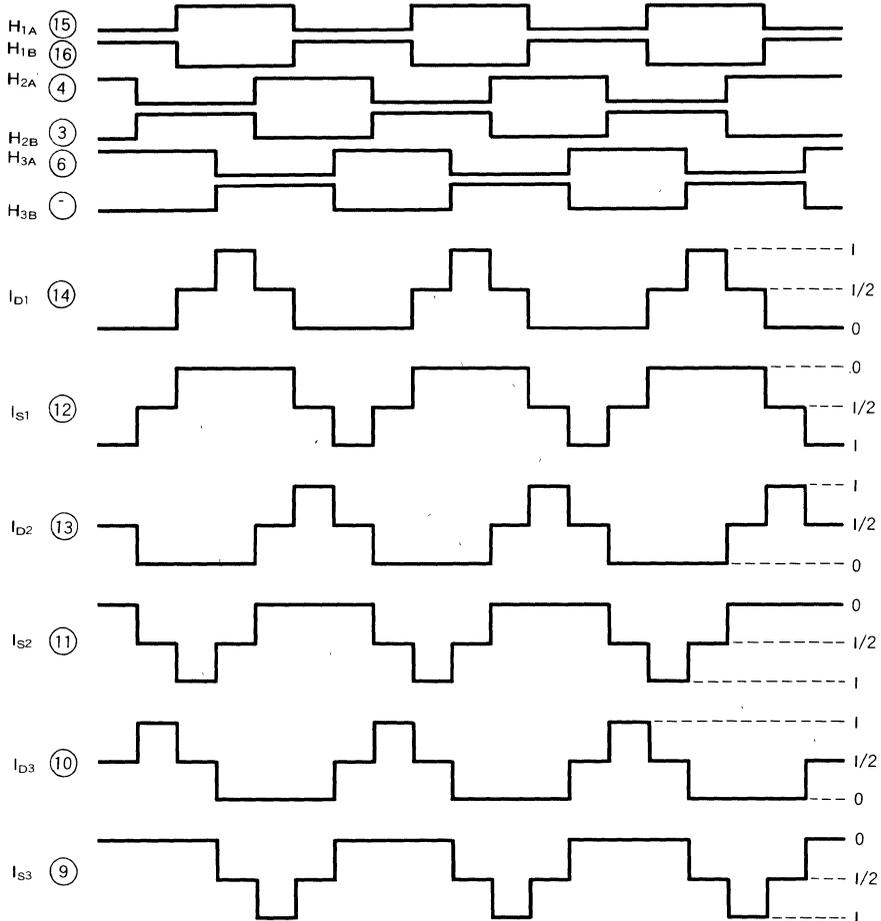


Note. when collector output 1 system and emitter output 2 system are ON



Note when collector output 2 system and emitter output 1 system are ON

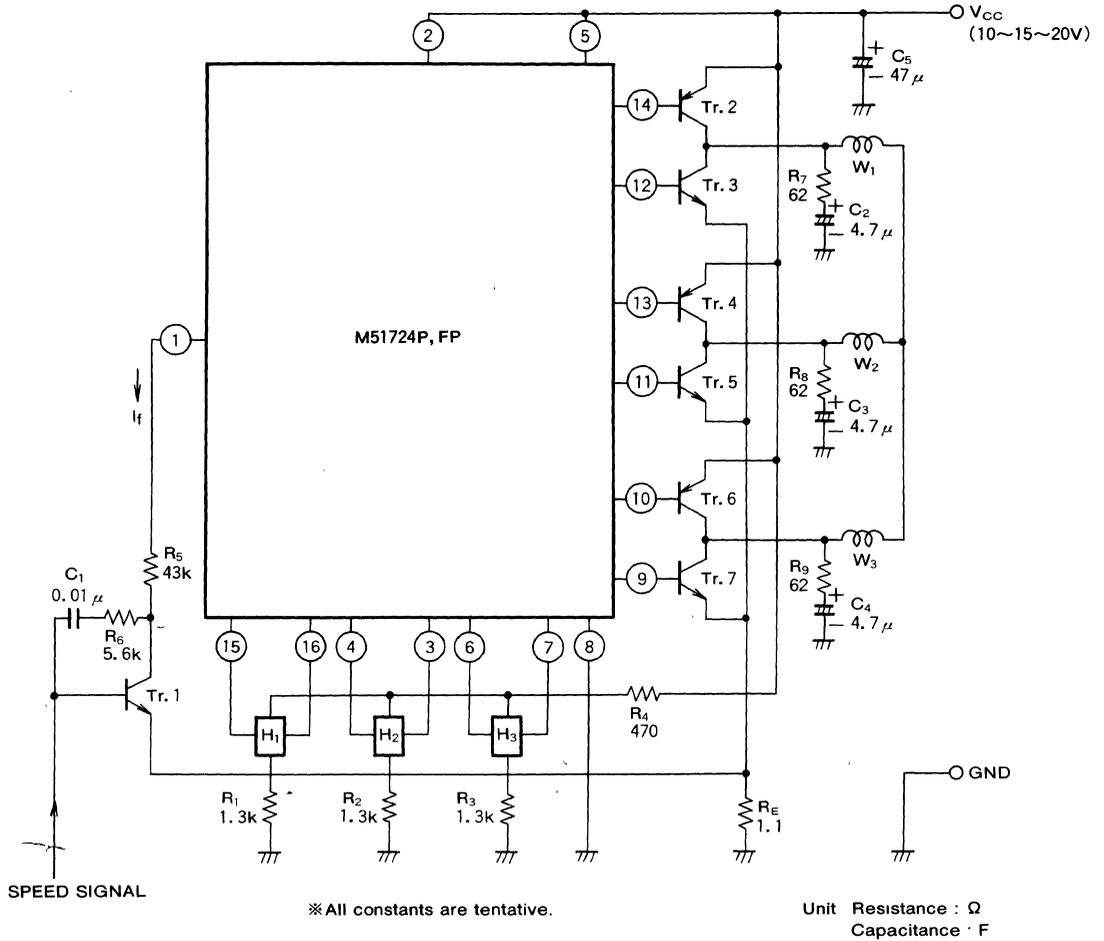
INPUT/OUTPUT TIMING CHART OF THE M51724P,FP



- Note :
1. In the output current value (I_b , I_s), + indicates source current and - indicates sink current
 2. All the input pins are biased.
 3. Care must be taken to connect a load (low impedance) to all the output pins according to the current at the control input Pin ①

3-PHASE BRUSHLESS MOTOR PRE-DRIVER

APPLICATION EXAMPLE



NOTE :

1. $R_1 \sim R_4$: For hall element bias
2. R_5 : For output (input) current limiting
3. R_6, C_1 : For oscillation prevention
4. $R_7 \sim R_9, C_2, C_4$: For reduction of driver noise
5. C_5 : For power supply stabilizing
6. R_E : For current feedback
7. Tr. 1 : Control transistor (S. S.)
8. Tr. 2-Rr. 7 : Power transistors
9. $H_1 \sim H_3$: Hall elements for position detection
10. The same power supply (V_{CC}) must be connected to Pin ② and Pin ⑤

SPEED CONTROL AND DRIVER FOR 2-PHASE BRUSHLESS MOTOR

DESCRIPTION

The M51781SP is a semiconductor integrated circuit developed for use in controlling 2-phase DC brushless motor. Both control and driver systems are contained in a single chip. It employs hall elements for position detection and consists of an FG amplifier, F-V converter, sample and hold circuit, constituting a complete circuit of F-serve motor driver system in a single chip.

FEATURES

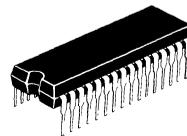
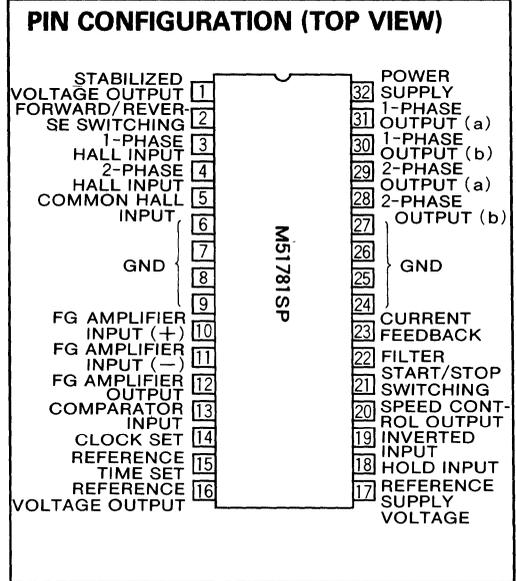
- Includes both control system and driver system in a single chip
- Drives all waves of 2-phase motors
- High output current
- Including FG amplifier, hysteresis comparator, F-V converter, sample and hold circuit
- Gain adjustment of control amplifier of the final stage of servo system possible
- Internal reference voltage generator
- Built-in output current limiter circuit
- Including forward/reverse direction control pins
- Start/stop control pins provided

APPLICATION

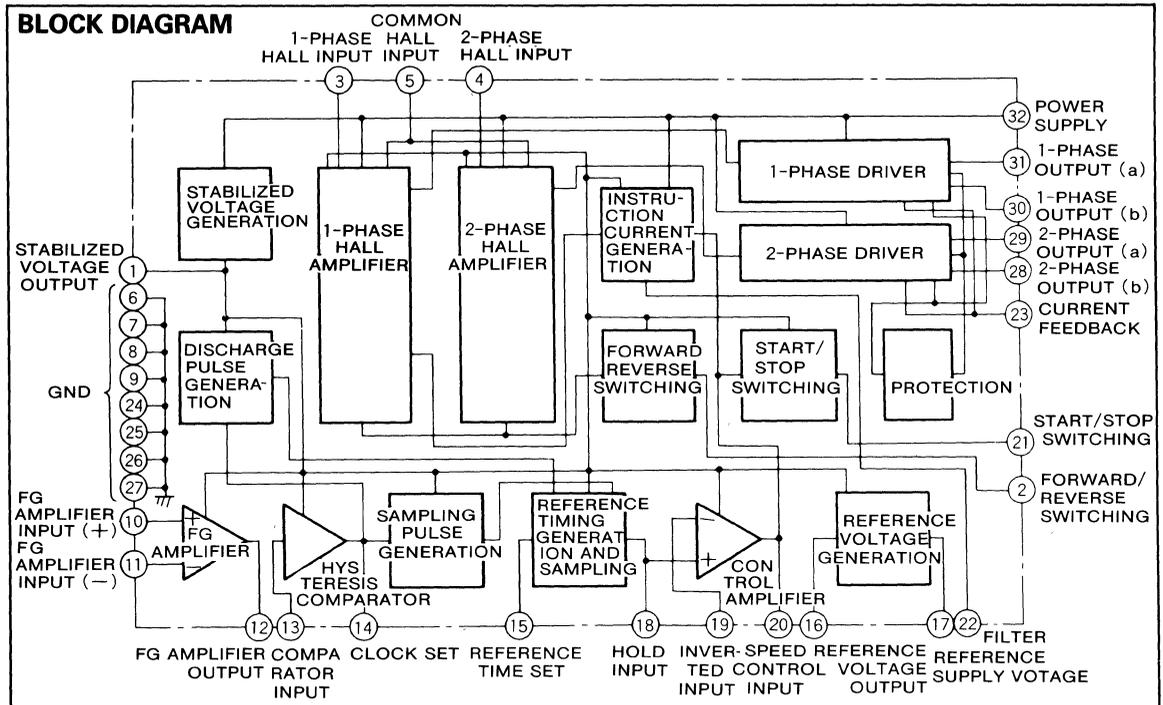
FDD, etc.

RECOMMENDED OPERATING CONDITIONS

Supply voltage range 7.2V~18V
 Rated supply voltage12V



32-pin molded plastic DIP shrink



SPEED CONTROL AND DRIVER FOR 2-PHASE BRUSHLESS MOTOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
I_{Lmax}	Charge current		1.2	A/PHASE
I_L	Output current		0.4	A/PHASE
I_{stb}	Pin ① output current		20	mA
$I_{OL⑭}$	Pin ⑭ source current	When discharge of time constant capacitor is finished	10	mA
$I_{OL⑮}$	Pin ⑮ source current	When discharge of time constant capacitor is finished	10	mA
V_H	Applied voltages between pins ③-⑤ and pins ④-⑤		5	V_{P-P}
$V_{⑩-⑪}$	Applied current between pin ⑩ and pin ⑪		± 0.7	V
$V_{H(C)}$	Hall input common phase voltage	Between pins ③-⑤ and pins ④-⑤	$1 \sim V_{stb} - 1$	V
P_{dF}	Power dissipation	Heat sink of infinite size is used	4	W
T_{opr}	Operating temperature		$-20 \sim +70$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^\circ\text{C}$

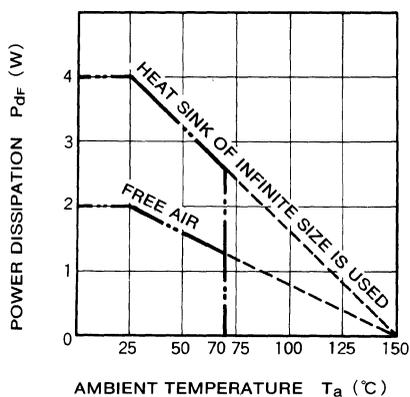
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{CC(1)}$	Circuit current (1)		2.2	3.5	5.5	mA
$I_{CC(2)}$	Circuit current (2)	$V_{CC}=18\text{V}$	2.4	3.9	6.6	mA
$V_{stb(1)}$	Stabilized output voltage (1)		5.4	5.8	6.2	V
$V_{stb(2)}$	Stabilized output voltage (2)	270Ω between V_{stb} and GND	5.2	5.8	6.2	V
V_{ref}	Output voltage of reference voltage		2.72	2.90	3.02	V
$I_{⑩}$	Pin ⑩ input current	$V_{CC}=V_{stb}=5.8\text{V}$, $V_{⑩}=0\text{V}$		0.5	3.0	μA
$I_{⑪}$	Pin ⑪ input current	$V_{CC}=V_{stb}=5.8\text{V}$, $V_{⑪}=0\text{V}$		30	180	nA
$V_{OL⑭}$	Pin ⑭ low level voltage	$V_{CC}=V_{stb}=5.8\text{V}$, $10\text{k}\Omega$ between V_{stb} and pin ⑭		80	120	mV
$V_{OL⑮}$	Pin ⑮ low level voltage	$V_{CC}=V_{stb}=5.8\text{V}$, $10\text{k}\Omega$ between V_{stb} and pin ⑮		50	90	mV
$V_{TH(1)}$	Clock threshold voltage (1)	$V_{CC}=V_{stb}=5.8\text{V}$	1.83	1.95	2.05	V
$V_{TH(2)}$	Clock threshold voltage (2)	$V_{CC}=V_{stb}=5.8\text{V}$	2.75	2.90	3.05	V
$V_{TH(3)}$	Clock threshold voltage (3)	$V_{CC}=V_{stb}=5.8\text{V}$	3.12	3.28	3.44	V
$V_{TH(4)}$	Clock threshold voltage (4)	$V_{CC}=V_{stb}=5.8\text{V}$	4.10	4.32	4.54	V
$V_{OL⑳}$	Pin ⑳ low level voltage	$V_{CC}=V_{stb}=5.8\text{V}$, $5.1\text{k}\Omega$ between V_{stb} and pin ⑳		0.75	1.00	V
$V_{offset⑮-⑩}$	Offset voltage between pin ⑮ and pin ⑩	$V_{CC}=V_{stb}=5.8\text{V}$			± 10	mV
$V_{offset⑩-⑮}$	Offset voltage between pin ⑩ and pin ⑮	$V_{CC}=V_{stb}=5.8\text{V}$			± 10	mV
$I_{ba(max)}$	Maximum bias current of a output	Measured at pin ㉓	9	13		mA
$I_{bb(max)}$	Maximum bias current of b output	Measured at pin ㉓	9	13		mA
$K_{Ib(max)}$	Maximum current bias ratio	Calculated by $I_{ba(max)}/I_{bb(max)}$	0.9	1.0	1.1	A/A
$V_{sat(u)}$	Output saturation voltage (upper)	$I_O=0.8\text{A}$, applicable to pins ㉔~㉑		1.5	2.2	V
$V_{sat(D)}$	Output saturation voltage (lower)	$I_O=0.8\text{A}$, applicable to pins ㉔~㉑		1.0	1.5	V

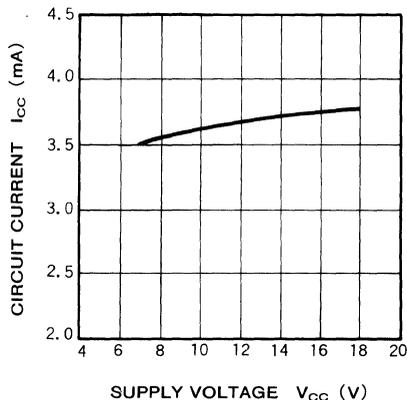
SPEED CONTROL AND DRIVER FOR 2-PHASE BRUSHLESS MOTOR

TYPICAL CHARACTERISTICS

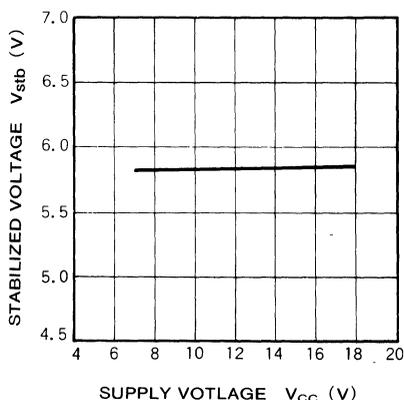
THERMAL DERATING (MAXIMUM RATING)



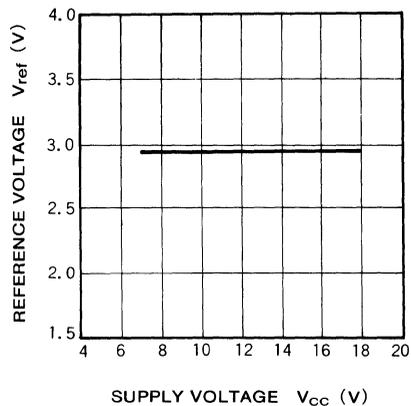
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



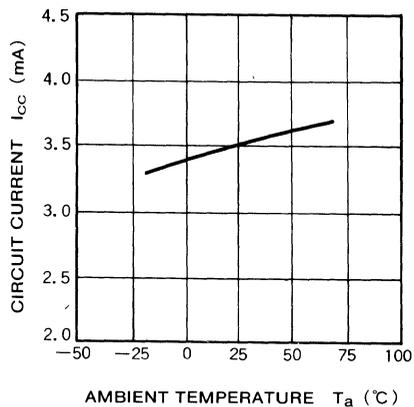
STABILIZED VOLTAGE VS. SUPPLY VOLTAGE



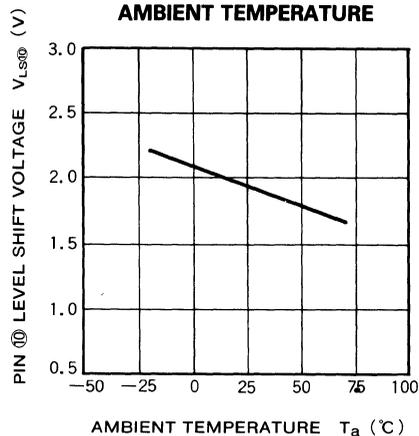
REFERENCE VOLTAGE VS. SUPPLY VOLTAGE



CIRCUIT CURRENT VS. AMBIENT TEMPERATURE

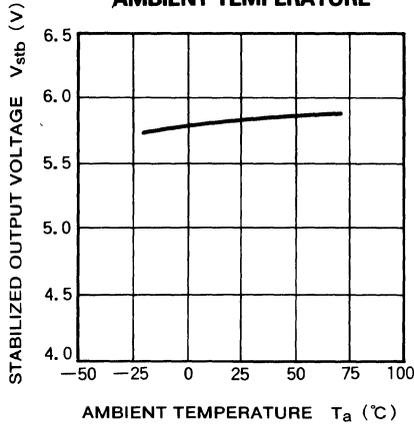


PIN ⑩ LEVEL SHIFT VOLTAGE VS. AMBIENT TEMPERATURE

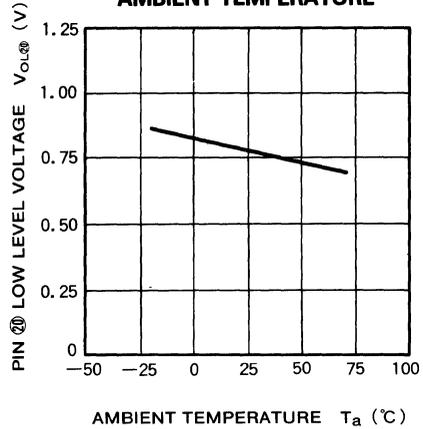


SPEED CONTROL AND DRIVER FOR 2-PHASE BRUSHLESS MOTOR

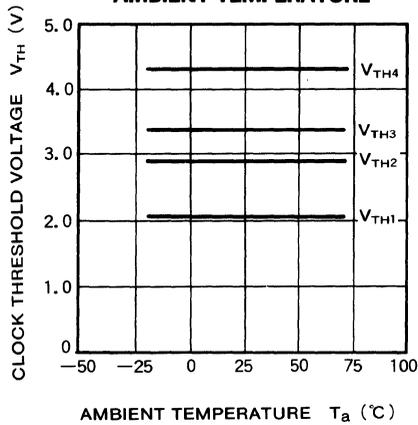
STABILIZED OUTPUT VOLTAGE VS. AMBIENT TEMPERATURE



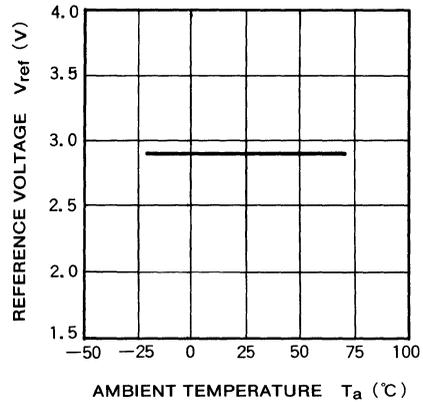
PIN ⑩ LOW LEVEL VOLTAGE VS. AMBIENT TEMPERATURE



CLOCK THRESHOLD VOLTAGE VS. AMBIENT TEMPERATURE

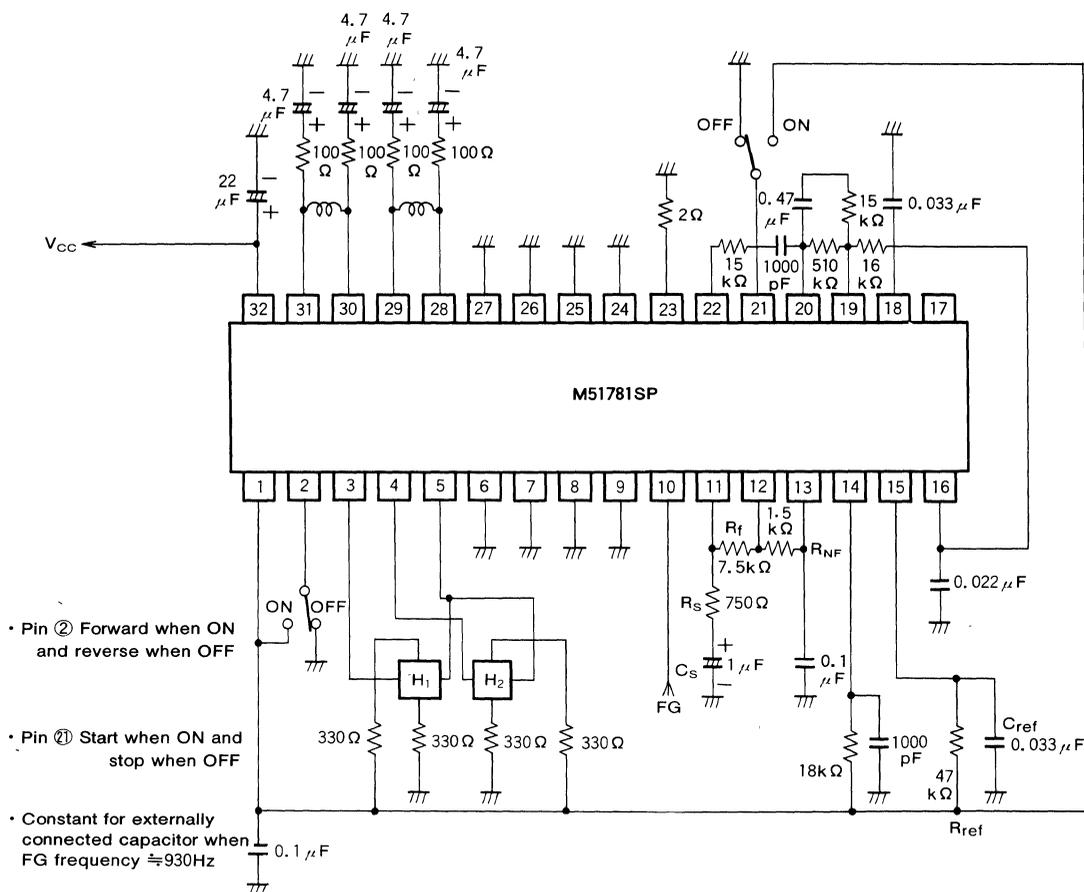


REFERENCE VOLTAGE VS. AMBIENT TEMPERATURE



SPEED CONTROL AND DRIVER FOR 2-PHASE BRUSHLESS MOTOR

APPLICATION EXAMPLE



Note 1. The signal amplitude for the input tachogenerator should be $1\text{mV}_{\text{P-P}}$ or more.

First stage amplifier gain \sim

$$\frac{1 + \omega_G^2 C_S^2 (R_S + R_f)^2}{1 + \omega_G^2 C_S^2 R_S^2}$$

where ω_G : input tachogenerator angle frequency

C_S must be chosen $4.7\mu\text{F}$ or less. The desirable range of R_S is obtained in the following.

$$\frac{2}{\omega_G} \geq C_S \cdot R_S \geq \frac{1}{\omega_G}$$

Note 2. R_{NF} and C_{NF} are resistor and capacitor for noise filtering. If the time constant is too large, the input signal is weakened. R_{NF} and C_{NF} should be chosen as follows;

$$R_{\text{NF}} \cdot C_{\text{NF}} \leq \frac{1}{\omega_G}$$

Note 3. R_{ref} and C_{ref} are resistor and capacitor for setting speed of motor. R_{ref} and C_{ref} should be chosen as follows;

$$R_{\text{ref}} \cdot C_{\text{ref}} = \frac{1}{0.693 \cdot f_G}$$

where f_G : input tachogenerator frequency

MITSUBISHI LINEAR ICs M51781FP

SPEED CONTROL AND DRIVER FOR 2-PHASE BRUSHLESS MOTOR

DESCRIPTION

The M51781FP is a semiconductor integrated circuit developed for use in controlling 2-phase DC brushless motor. Both control and driver systems are contained in a single chip. It employs hall elements for position detection and consists of an FG amplifier, F-V converter, sample and hold circuit, constituting a complete circuit of F-serve motor driver system in a single chip.

FEATURES

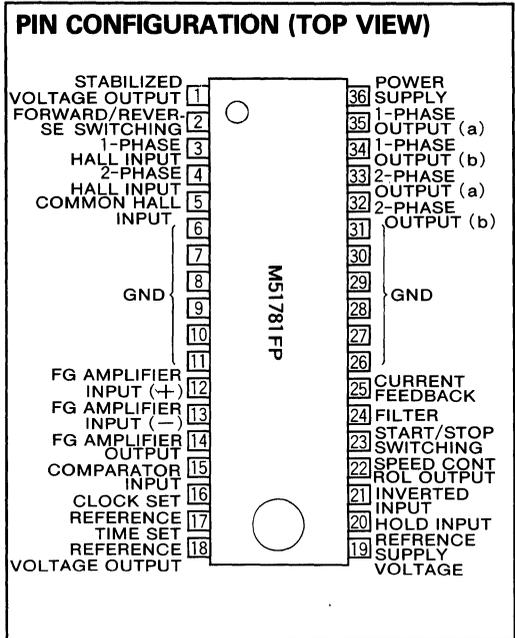
- Includes both control system and driver system in a single chip
- Drives all waves of 2-phase motors
- High output current
- Including FG amplifier, hysteresis comparator, F-V converter, sample and hold circuit
- Gain adjustment of control amplifier of the final stage of servo system possible
- Internal reference voltage generator
- Built-in output current limiter circuit
- Including forward/reverse direction control pins
- Start/stop control pins provided

APPLICATION

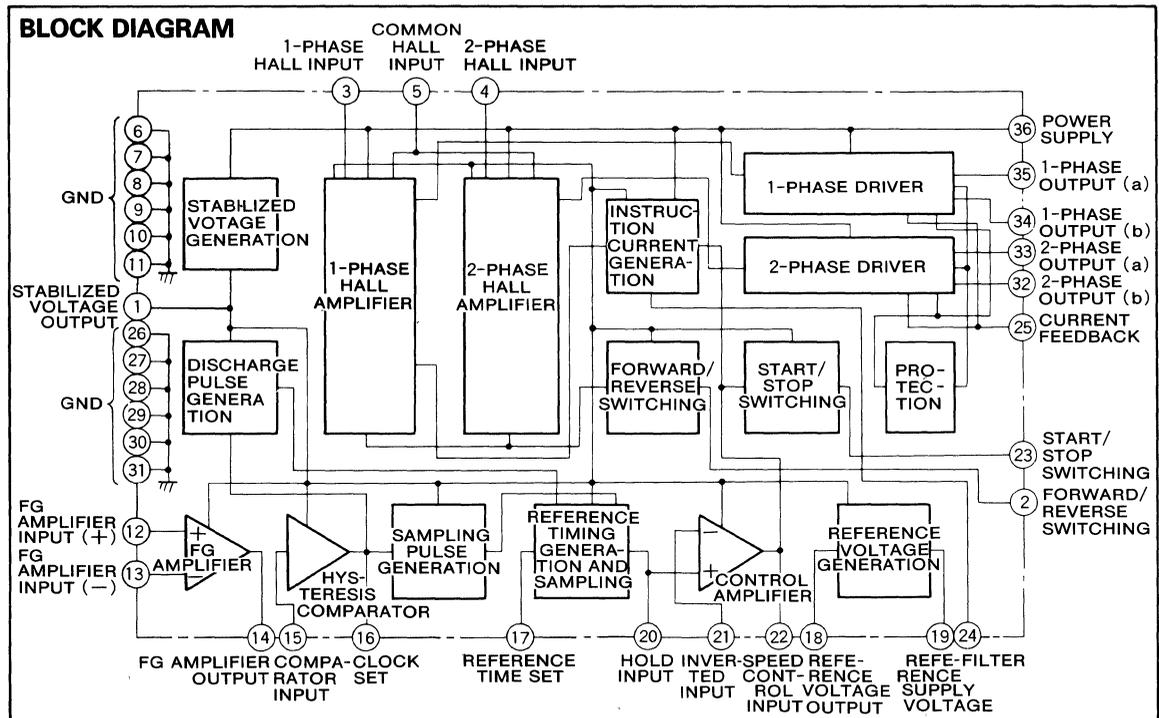
FDD, etc.

RECOMMENDED OPERATING CONDITIONS

Supply voltage range 7.2V~18V
Rated supply voltage12V



36-pin molded plastic FLAT (A type)



SPEED CONTROL AND DRIVER FOR 2-PHASE BRUSHLESS MOTOR

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		18	V
I _{Lmax}	Charge current		1.2	A/PHASE
I _L	Output current		0.4	A/PHASE
I _{stb}	Pin ① output current		20	mA
I _{OL⑩}	Pin ⑩ source current	When discharge of time constant capacitor is finished	10	mA
I _{OL⑪}	Pin ⑪ source current	When discharge of time constant capacitor is finished	10	mA
V _H	Applied voltages between pins ③-⑤ and pins ④-⑤		5	V _{P-P}
V _{⑩-⑪}	Applied current between pin ⑩ and pin ⑪		±0.7	V
V _{H(C)}	Hall input common phase voltage	Between pins ③-⑤ and pins ④-⑤	1~V _{stb} -1	V
P _{dF}	Power dissipation	Heat sink of infinite size is used	3.5	W
T _{opr}	Operating temperature		-20~+70	°C
T _{stg}	Storage temperature		-40~+125	°C

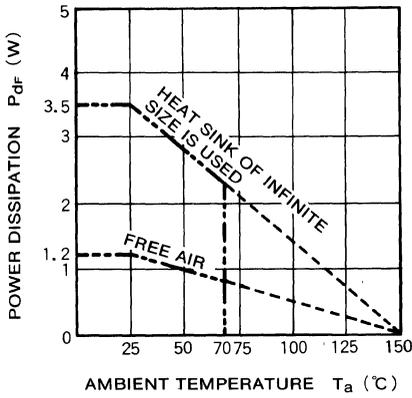
ELECTRICAL CHARACTERISTICS (T_a=25°C, V_{CC}=12V, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I _{CC(1)}	Circuit current (1)		2.2	3.5	5.5	mA
I _{CC(2)}	Circuit current (2)	V _{CC} =18V	2.4	3.9	6.6	mA
V _{stb(1)}	Stabilized output voltage (1)		5.4	5.8	6.2	V
V _{stb(2)}	Stabilized output voltage (2)	270Ω between V _{stb} and GND	5.2	5.8	6.2	V
V _{ref}	Output voltage of reference voltage		2.72	2.90	3.02	V
I _⑩	Pin ⑩ input current	V _{CC} =V _{stb} =5.8V, V _{⑩2} =0V		0.5	3.0	μA
I _⑪	Pin ⑪ input current	V _{CC} =V _{stb} =5.8V, V _{⑪3} =0V		30	180	nA
V _{OL⑩}	Pin ⑩ low level voltage	V _{CC} =V _{stb} =5.8V, 10kΩ between V _{stb} and pin ⑩		80	120	mV
V _{OL⑪}	Pin ⑪ low level voltage	V _{CC} =V _{stb} =5.8V, 10kΩ between V _{stb} and pin ⑪		50	90	mV
V _{TH(1)}	Clock threshold voltage (1)	V _{CC} =V _{stb} =5.8V	1.83	1.95	2.05	V
V _{TH(2)}	Clock threshold voltage (2)	V _{CC} =V _{stb} =5.8V	2.75	2.90	3.05	V
V _{TH(3)}	Clock threshold voltage (3)	V _{CC} =V _{stb} =5.8V	3.12	3.28	3.44	V
V _{TH(4)}	Clock threshold voltage (4)	V _{CC} =V _{stb} =5.8V	4.10	4.32	4.54	V
V _{OL⑫}	Pin ⑫ low level voltage	V _{CC} =V _{stb} =5.8V, 5.1kΩ between V _{stb} and pin ⑫		0.75	1.00	V
V _{offset⑩-⑪}	Offset voltage between pin ⑩ and pin ⑪	V _{CC} =V _{stb} =5.8V			±10	mV
V _{offset⑫-⑬}	Offset voltage between pin ⑫ and pin ⑬	V _{CC} =V _{stb} =5.8V			±10	mV
I _{ba(max)}	Maximum bias current of a output	Measured at pin ⑮	9	13		mA
I _{bb(max)}	Maximum bias current of b output	Measured at pin ⑮	9	13		mA
K _{Ib(max)}	Maximum current bias ratio	Calculated by I _{ba(max)} /I _{bb(max)}	0.9	1.0	1.1	A/A
V _{sat(U)}	Output saturation voltage (upper)	I _O =0.8A, applicable to pins ⑳~㉕		1.5	2.2	V
V _{sat(D)}	Output saturation voltage (lower)	I _O =0.8A, applicable to pins ⑳~㉕		1.0	1.5	V

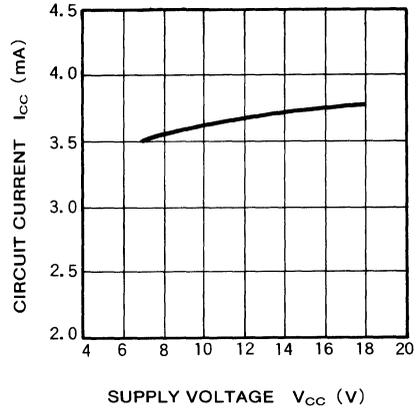
SPEED CONTROL AND DRIVER FOR 2-PHASE BRUSHLESS MOTOR

TYPICAL CHARACTERISTICS

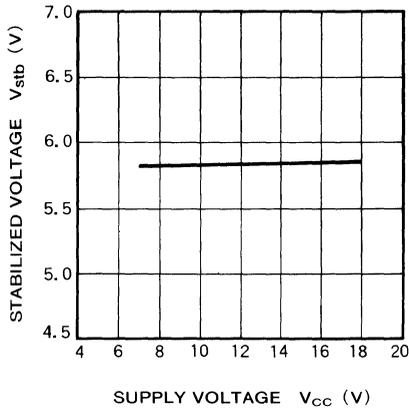
THERMAL DERATING (MAXIMUM RATING)



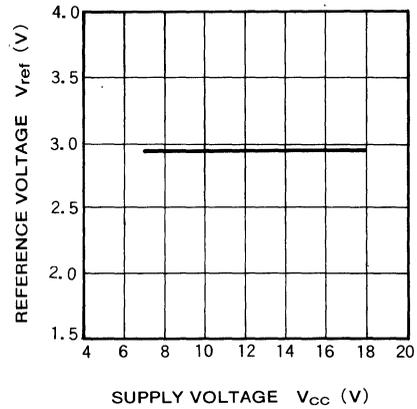
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



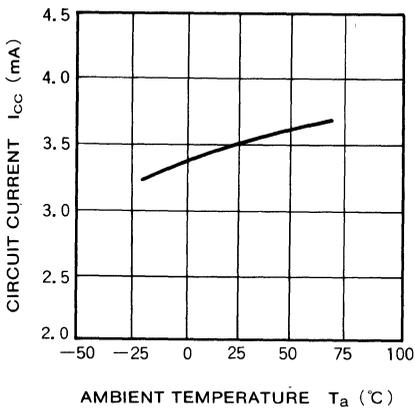
STABILIZED VOLTAGE VS. SUPPLY VOLTAGE



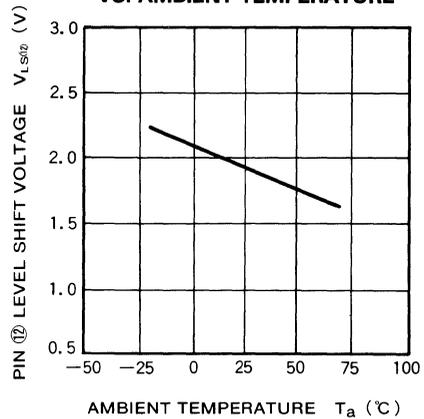
REFERENCE VOLTAGE VS. SUPPLY VOLTAGE



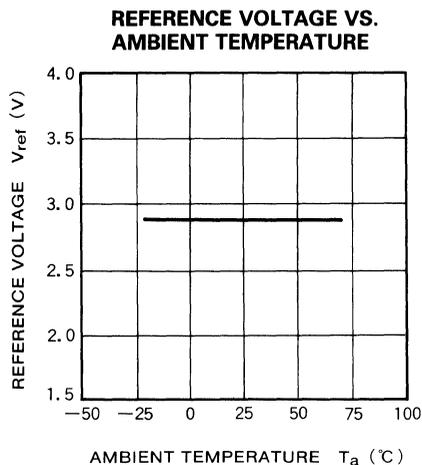
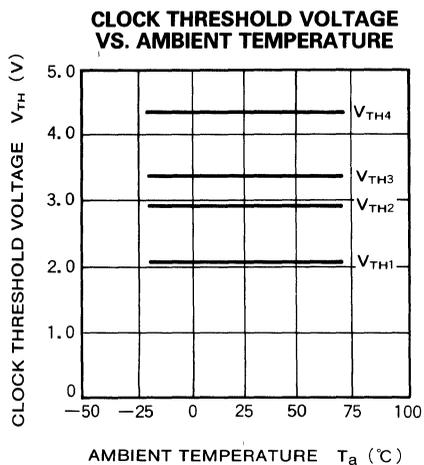
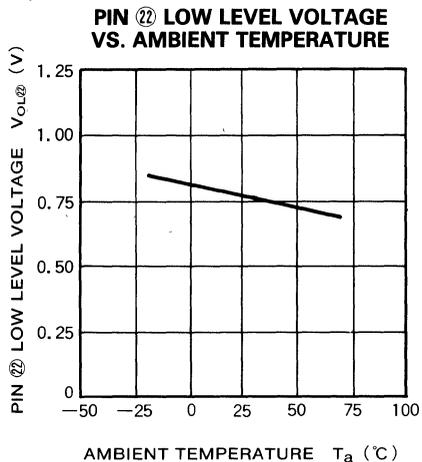
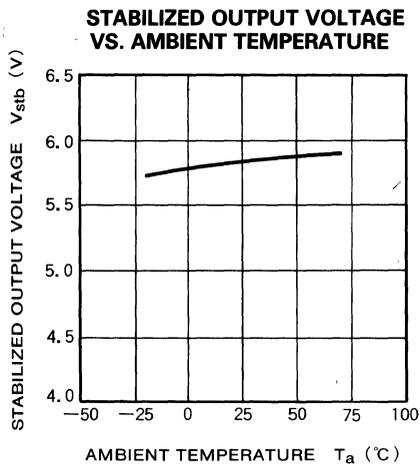
CIRCUIT CURRENT VS. AMBIENT TEMPERATURE



PIN ⑫ LEVEL SHIFT VOLTAGE VS. AMBIENT TEMPERATURE

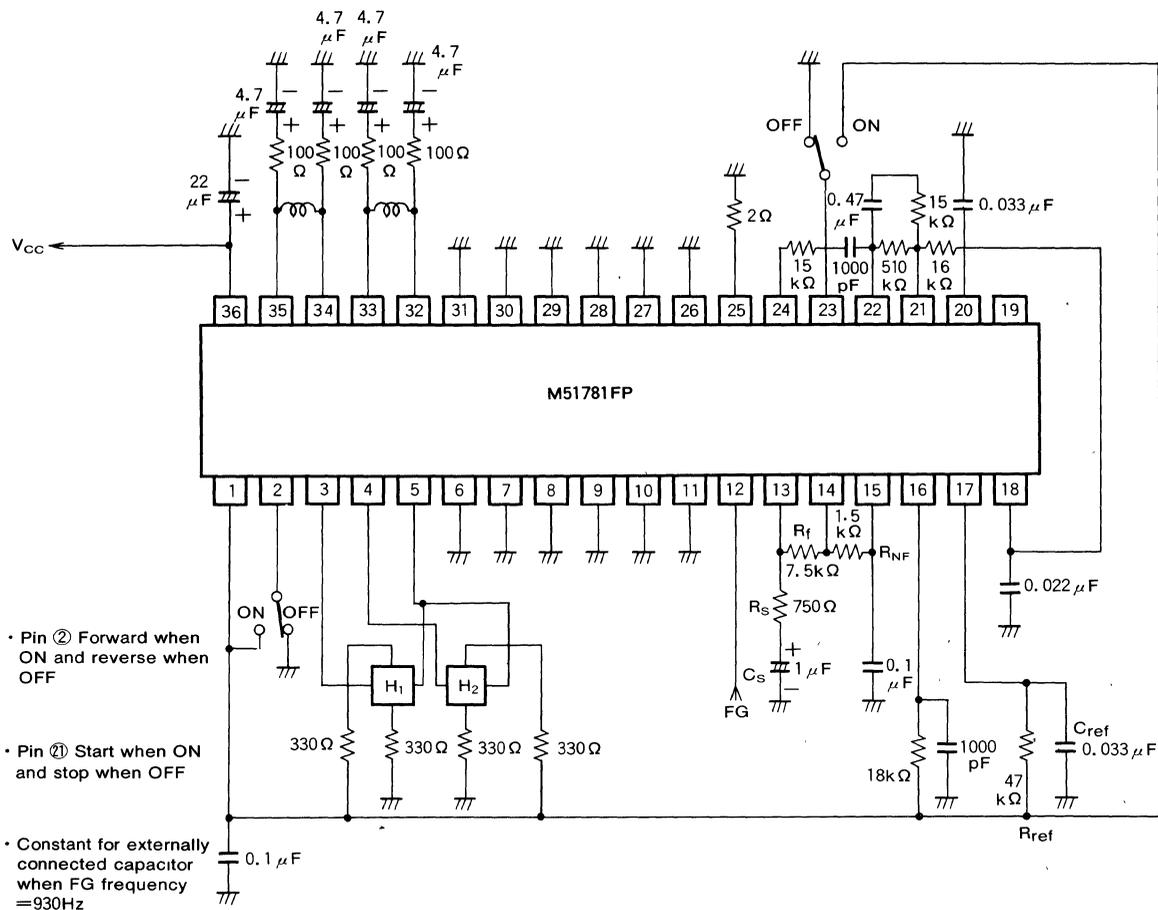


SPEED CONTROL AND DRIVER FOR 2-PHASE BRUSHLESS MOTOR



SPEED CONTROL AND DRIVER FOR 2-PHASE BRUSHLESS MOTOR

APPLICATION EXAMPLE



Note 1. The signal amplitude for the input tachogenerator should be 1mV_{P-P} or more.

First stage amplifier gain ~

$$\frac{1 + \omega_G^2 C_S^2 (R_S + R_f)^2}{1 + \omega_G^2 C_S^2 R_S^2}$$

where ω_G : input tachogenerator angle frequency

C_S must be chosen 4.7 μF or less. The desirable range of R_S is obtained in the following.

$$\frac{2}{\omega_G} \geq C_S \cdot R_S \geq \frac{1}{\omega_G}$$

Note 2. R_{NF} and C_{NF} are resistor and capacitor for noise filtering. If the time constant is too large, the input signal is weakened. R_{NF} and C_{NF} should be chosen as follows;

$$R_{NF} \cdot C_{NF} \leq \frac{1}{\omega_G}$$

Note 3. R_{ref} and C_{ref} are resistor and capacitor for setting speed of motor. R_{ref} and C_{ref} should be chosen as follows;

$$R_{ref} \cdot C_{ref} = \frac{1}{0.693 \cdot f_G}$$

where f_G : input tachogenerator frequency

MITSUBISHI LINEAR ICs
M51785P,SP

3-PHASE BRUSHLESS MOTOR CONTROL

DESCRIPTION

The M51785P/SP is a semiconductor integrated circuit designed for a single-chip controller for FDD spindle motor, consisting of power amplifier, Hall amplifier, FG amplifier, oscillator and speed discriminator and various protection circuits.

The device shows superiority in speed switching function of 1 : 1.2 which enables miniaturization of motor sets and cost reduction.

FEATURES

- High-accuracy, high-stability, and adjustment-free controller is possible by digital servo
- Speed switch of 1 : 1.2 possible..... MOD
- $I_o(\text{peak})=1.2A$
- 2 ENABLE systems EN, \overline{EN}

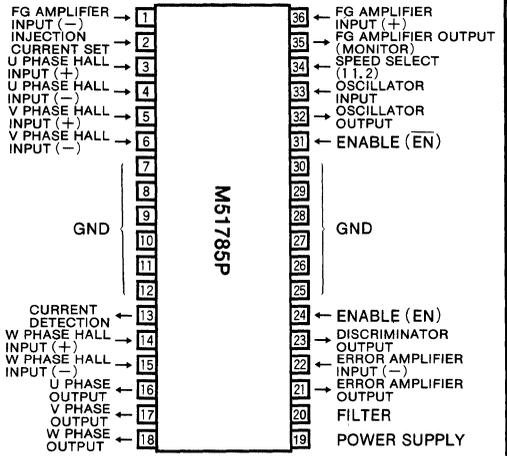
APPLICATION

FDD spindle motor (5", 3.5")

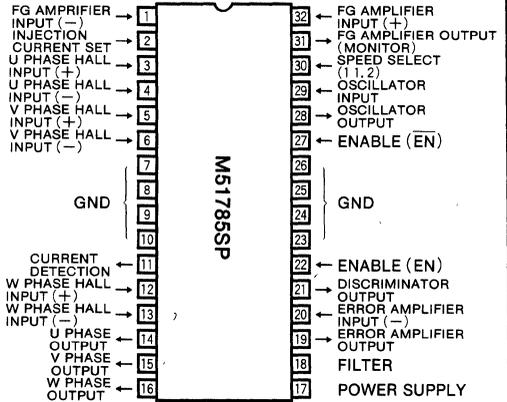
RECOMMENDED OPERATING CONDITIONS

- Supply voltage 10.8-12-13.2V
- Oscillating frequency 400-650kHz
- Ingector current 2.5-3-7mA
- Maximum output current 800mA
- FG OUT Load resistance 100k Ω
- FG amplifier input signal level 5 or above mV_{P-P}
- Hall amplifier input signal level 50-100-150 mV_{P-P}

PIN CONFIGURATION (TOP VIEW)



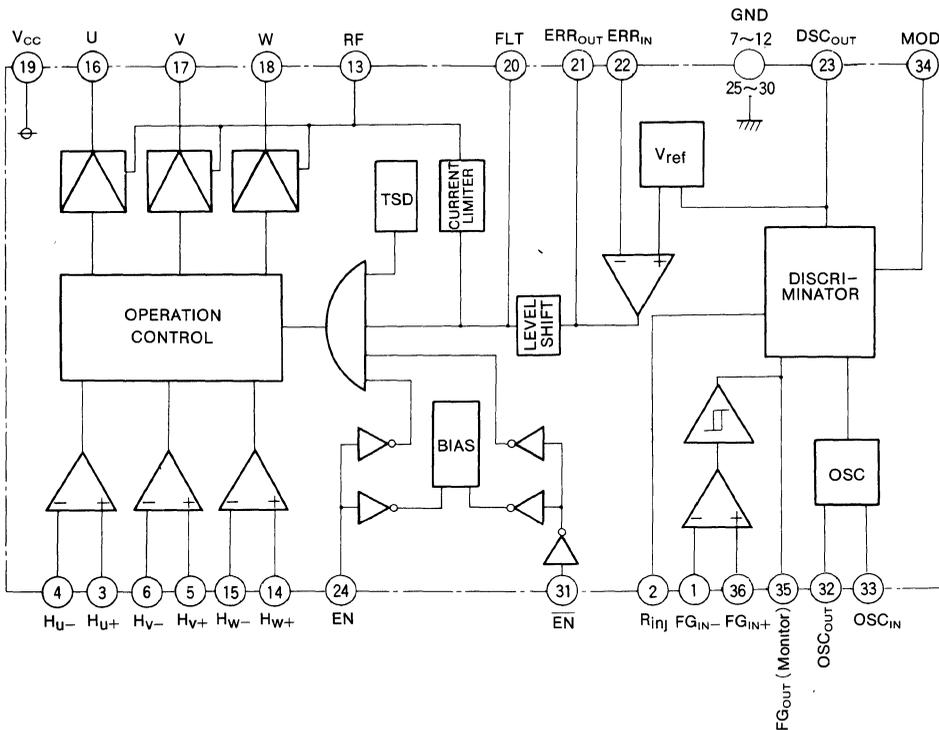
36-pin molded plastic FLAT (shrink)



32-pin molded plastic DIP (shrink) with fin

3-PHASE BRUSHLESS MOTOR CONTROL

BLOCK DIAGRAM



Note Pin No = M51785P.

ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CCA}	Operating supply voltage		15	V
I_O	Output current		1, 2	A
V_{HD}	Hall amplifier differential input voltage	3-4, 5-6, 14-15(Pin no.)	5	V
I_{SS}	Source/sink current	20, 21, 23, 32, 33, 36(Pin no.)	± 3	mA
V_{IN}	Pin applied voltage	1, 3, 4, 5, 6, 14, 15, 22, 24, 31, 34(Pin no)	$0 \sim V_{CC}$	V
I_{inj}	Injection current		20	mA
V_{RF}	⑬ pin applied voltage		1	V
P_t	Power dissipation	Heatsink of infinite size used	4.5(8)	W
$K \theta$	Thermal derating	Heatsink of infinite size used	27.8(15.6)	$^{\circ}\text{C}/\text{W}$
T_j	Junction temperature		150	$^{\circ}\text{C}$
T_{opr}	Operating temperature		$-20 \sim +75$	$^{\circ}\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^{\circ}\text{C}$
V_{CCB}	Quiescent supply voltage	EN-Lo, EN-Hi.	16	V

() Shows the value of M51785SP

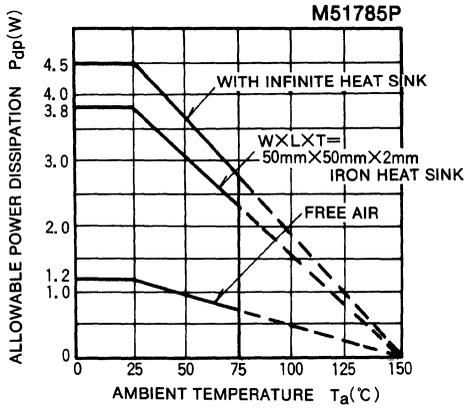
3-PHASE BRUSHLESS MOTOR CONTROL

ELECTRICAL CHARACTERISTICS ($V_{CC}=12V$, $T_a=25^{\circ}C$, unless otherwise noted)

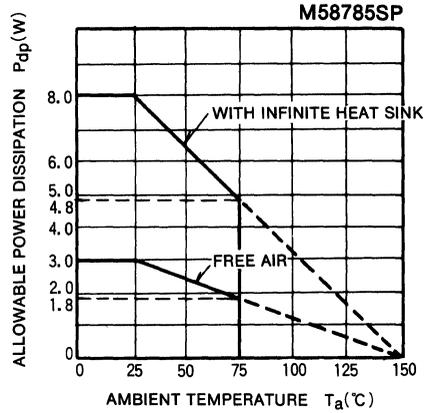
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{CC}(H)$	Circuit current (EN ON)	EN=2.5V, MOD=EN=0.8V, excluding injection current and FG monitor pin current. No load	9	18	28	mA
$I_{CC}(L)$	Circuit current (EN OFF)	Connect injection setting pin and FG monitor pin directly to V_{CC} . No load, EN=0.8V, EN=2.5V	—	90	300	μA
$V_{CC}(OP)$	Operating supply voltage		9	12	15	V
$I_{INH A}$	Hall amplifier input current		—	0.4	4	μA
V_N	Phase output middle point voltage		5.3	6.5	7.3	V
ΔV_N	Difference of middle point voltage between phases		—	—	0.2	V
V_{sat}	Output saturation voltage	Current flow U→V, V→W, W→U. Total of V_{sat} of T_r on both sides, $l_o=0.7A$	—	2.3	3.3	V
V_{TH}	Control input reference voltage	FLT pin voltage producing output	1.0	1.1	1.2	V
G_V	Voltage gain between control input and output	Source	16.65	18.06	26.81	dB
		Sink	20.82	23.80	26.81	
ΔG_V	Difference of voltage gain between phases		—	—	2	dB
V_{ref}	Error amplifier reference voltage	Measure middle level of discriminator output	2.0	2.2	2.4	V
I_{INEA}	Error amplifier input current		-2.0	-0.02	—	μA
V_{OEA}	Error amplifier output level	Hi	2.2	2.5	3.1	V
		Lo	0.6	0.8	1.05	
V_{CL}	Current limiter reference voltage	R_F pin voltage when FLT pin voltage is reduced to less than 1.5V	0.36	0.40	0.44	V
V_{IN}	Function input threshold value	Hi 24, 31, 34	2.5	—	—	V
		Lo	—	—	0.8	
I_{IN}	Input current at function input pin	$V_{IN}=12V$ 24, 34	500	700	1000	μA
		$V_{IN}=0V$ 31	-150	-100	-70	
V_{inj}	Injection pin voltage	$I_{inj}=6mA$	0.6	0.9	1.5	V
V_{ODSC}	Discriminator output level	Hi	4.1	4.8	5.3	V
		Lo	0.5	0.8	1.2	
ΔT	Discriminator count error	+ for deceleration, - for acceleration $f_{osc}=610.2kHz$	-6	1	6	μsec
f_{OSC}	Oscillating frequency	$f_{osc}=610.2kHz$	-0.2	—	+8.2	%
I_{injMAX}	Maximum injection operating current	$f_{osc}=610.2kHz$	17	—	—	mA
I_{injMIN}	Minimum injection operating current	$f_{osc}=610.2kHz$	—	—	4	mA
$V_{OL}(FG)$	FG amplifier output low level (monitor)	$I_L=200\mu A$	—	0.1	0.2	V
$I_I(FG)$	Leak current at FG amplifier output (monitor) pin	12V is applied	—	—	1.0	μA
$V_{CC}(SD)$	Over-voltage protection operating voltage			16.3		V
$T_{(SD)}$	Thermal shutdown protection operating temperature			150		$^{\circ}C$
$\Delta T_{(SD)}$	Thermal shutdown protection hysteresis			25		$^{\circ}C$
$V_{IN}(FG)MIN$	FG amplifier operating minimum input voltage	Measure at FG monitor pin		2.5		mV _{p-p}
$V_{FG}(NM)$	FG amplifier input noise margin			1.0		mV _{p-p}
N_{CLK}	Discriminator count no	MOD=Lo	Count error is specified in section 19 of ELECTRICAL CHARACTERISTICS		1695	—
		MOD=Hi			2034	
f_{FGL1}	Synchronous frequency 1	MOD=Hi, $f_{osc}=610.2kHz$		300.0		Hz
f_{FGL2}	Synchronous frequency 2	MOD=Lo, $f_{osc}=610.2kHz$		360.0		Hz

3-PHASE BRUSHLESS MOTOR CONTROL

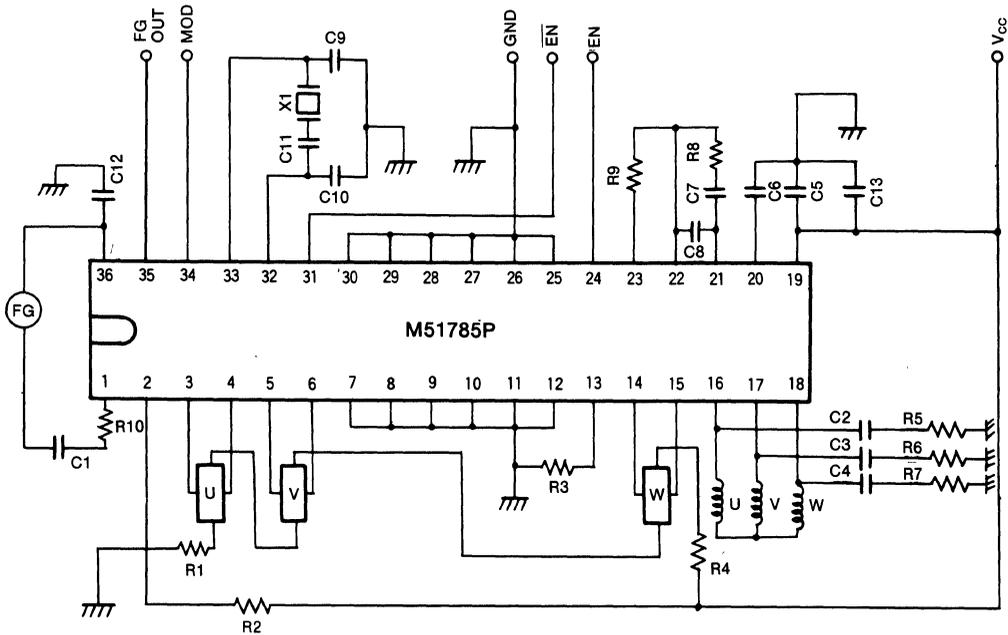
THERMAL DERATING (MAXIMUM RATING)



THERMAL DERATING (MAXIMUM RATING)



APPLICATION EXAMPLE



CONSTANTS

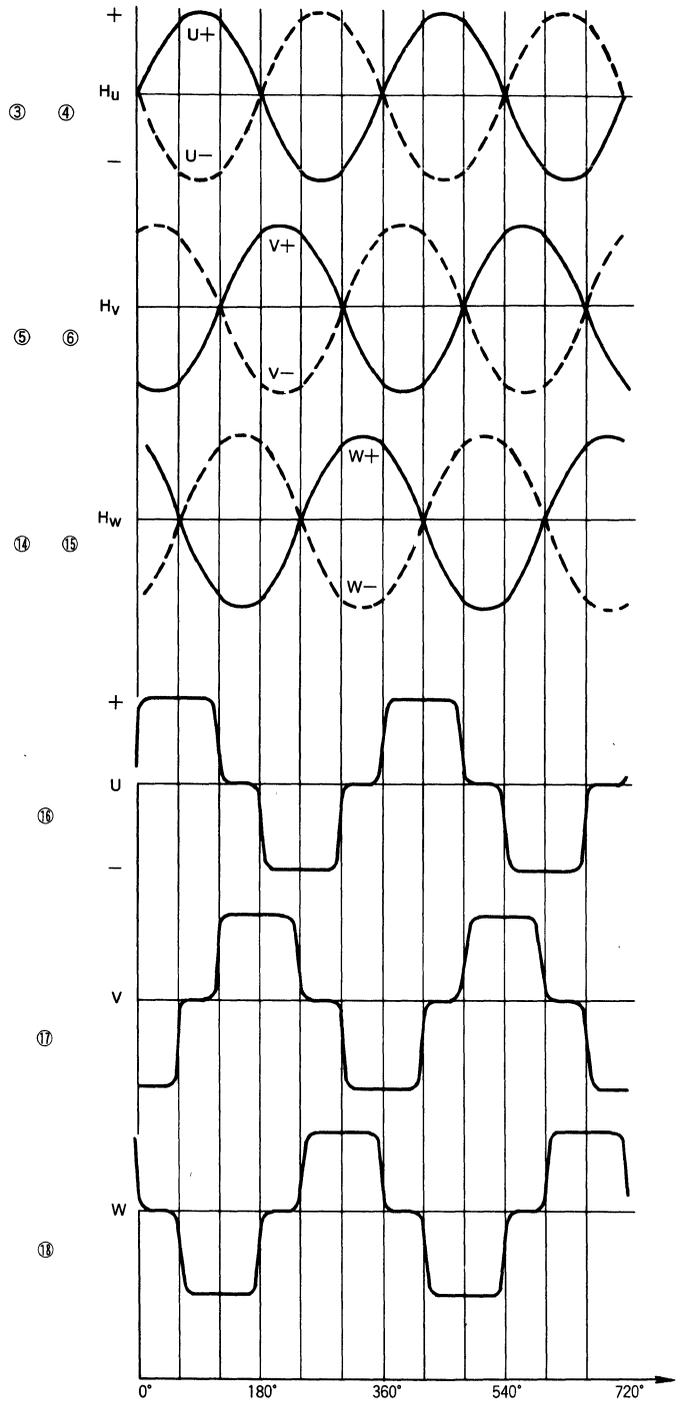
R1	330Ω	C1	4.7μF	X1	610.2kHz
R2	3.6kΩ	C2	0.1μF		
R3	0.5Ω	C3	0.1μF		
R4	330Ω	C4	0.1μF		
R5	4.7Ω	C5	0.1μF		
R6	4.7Ω	C6	0.22μF		
R7	4.7Ω	C7	0.33μF		
R8	75kΩ	C8	0.033pF		
R9	22kΩ	C9	220pF		
R10	330Ω	C10	220pF		
		C11	100pF		
		C12	0.1μF		
		C13	33μF		

Note : Open collector output at FG OUT pin

3-PHASE BRUSHLESS MOTOR CONTROL

TIMING CHART

HALL INPUT

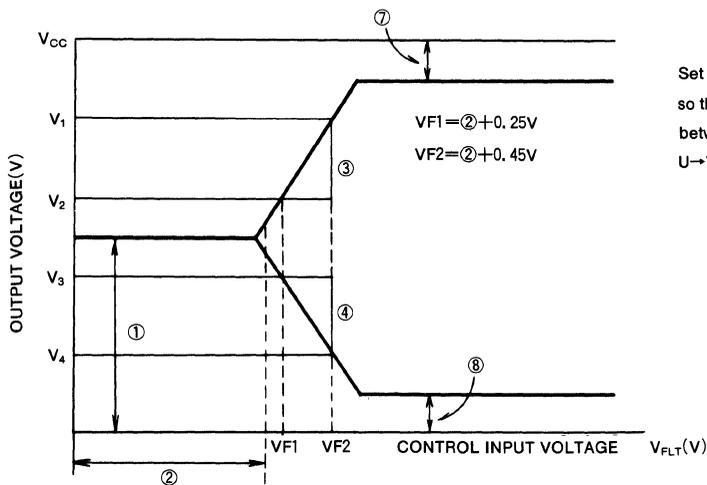


Note 1 . The waveforms shown above are different from those at actual motor operation
 Note 2 . Pin No. =M51785P

3-PHASE BRUSHLESS MOTOR CONTROL

TEST DESCRIPTION

Phase output middle point voltage	→①	
Difference of middle point voltage between phase		; Measure voltage 1 for each phase and ΔV_N is given in, $\Delta V_N = \Delta V_{UV} = V_U - V_V$ $= \Delta V_{VW} = V_V - V_W$ $= \Delta V_{WU} = V_W - V_U$
Output saturation voltage	→⑦+⑧	Load current 0.7A Control input voltage 2.2V
Control input reference voltage	→②	Control input voltage value (V_{FLT}) when the output voltage is ①+100mV.
Voltage gain between control input and output	→③(source) ④(sink)	
		③=20 log $\{(V_1 - V_2)/0.2\}$ ④=20 log $\{(V_3 - V_4)/0.2\}$
Difference of voltage gain between phase		; Measure ③ and ④ for each phase, and ΔG_V is given in, (source and sink) $\Delta G = \Delta G_{UV} = G_{V(U)} - G_{V(V)}$ $= \Delta G_{VW} = G_{V(V)} - G_{V(W)}$ $= \Delta G_{WU} = G_{V(W)} - G_{V(U)}$

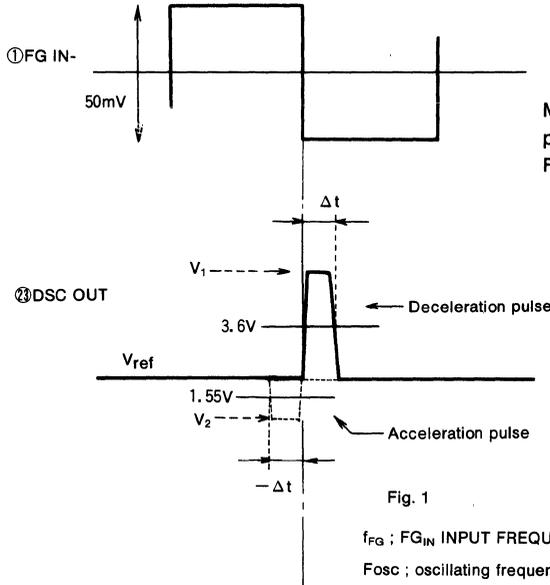


Set the values of ② to ⑧
 so that the output current flows
 between phases
 U→V, V→W, W→U.

3-PHASE BRUSHLESS MOTOR CONTROL

DISCRIMINATOR COUNT ERROR

Measure the pulsewidth at 23 pin DSC OUT. The test value is negative for accelerating pulse.



Measure acceleration or deceleration pulse at DOS_{OUT} output, applying pulse (synchronous with Fosc) divided by Fosc to FG_{IN}-in each mode.

☆Refer to table 1 for the frequency given to FG_{IN}.

Fig. 1

f_{FG} : FG_{IN} INPUT FREQUENCY
 Fosc ; oscillating frequency

Table 1

③MOD	①FG _{IN} input frequency
L 1665 division	F _{osc} /1695
H 2034 division	F _{osc} /2034

DISCRIMINATOR OUTPUT LEVEL

Measure V₁ and V₂ in Fig. 1. V_{O_{DSC}}(Hi)→V₁
 (Lo)→V₂

But, for Low(V₂)level, measure Lo level of f_{FG}=250Hz(acceleration pulse),
 and for High(V₁)level, measure Hi level of f_{FG}=400Hz (deceleration pulse).

ENABLE FUNCTION

Table 2

		EN	
		Lo	Hi
EN	Lo	DISABLE	ENABLE
	Hi	DISABLE	DISABLE

☆EN pin=circuit is operated only when EN pin=Hi and EN pin=Lo

☆EN pin→open=Lo

EN pin→open=Hi

(But anti-noise characteristics may deteriorate if used with EN. • EN=open after mounting on the equipment.)

MITSUBISHI LINEAR ICs

M51714P

DUAL BI-DIRECTIONAL DRIVER

DESCRIPTION

The M51714P is a semiconductor integrated circuit for use in dual bi-directional micro motor.

FEATURES

- Superiority in mounting
- Can be controlled by TTL level interface
- 2-ENABLE facility
- 1.2A maximum output current

APPLICATION

General home-use equipment

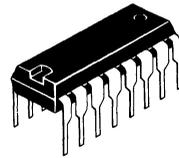
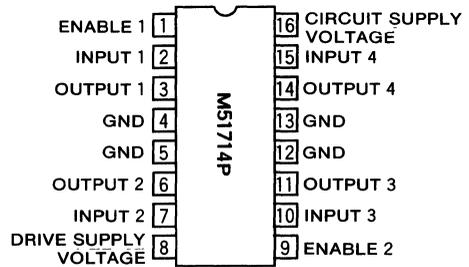
FUNCTIONAL DESCRIPTION

The M51714P has 4 internal amplifiers, and each amplifier operates as current source type when the input is in the high level and as current sink type when the input is in the low level.

Two ENABLE are provided, and \overline{EN}_1 functions in amplifiers 1 and 2 and \overline{EN}_2 in amplifiers 3 and 4. When the ENABLE input is in the high level, the system output is off and when the input is in the low level, and input-related output is produced.

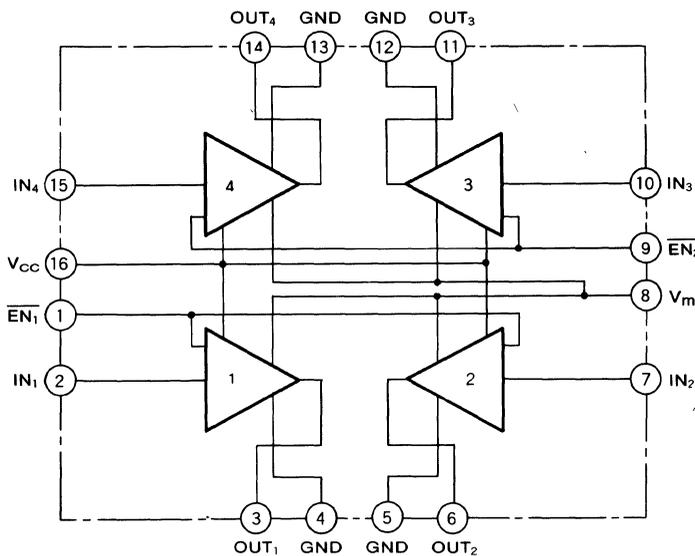
The circuit power supply and drive power supply are independently provided for power supply of the device, facilitating free equipment design.

PIN CONFIGURATION (TOP VIEW)



16-pin molded plastic DIP

BLOCK DIAGRAM



DUAL BI-DIRECTIONAL DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_m	Drive supply voltage		$V_{CC}\sim 26$	V
V_{CC}	Circuit supply voltage		V_m	V
$V_{I(IN)}$	Input voltage		$-0.3\sim 7$	V
$V_{I(EN)}$	Enable input voltage		$-0.3\sim 7$	V
$I_{O(max)}$	Ruch current		1.2	A
I_o	Output current		0.4	A
P_d	Allowable power dissipation		2.5	W
T_{opr}	Operating temperature		$-20\sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40\sim +125$	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_m	Drive supply voltage		5	12	24	V
V_{CC}	Circuit supply voltage	Should be less than V_m		5	12	V
I_o	Output current				300	mA
V_{IH}	High-level input voltage	For both input and enable input	2.8		$V_{CC}-1.5$	V
V_{IL}	Low-level input voltage	For both input and enable input	0		1.5	V

Note. It is desirable that enable input is in the high level and the output is off when the motor is stopped.

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit	
			Min	Typ	Max		
I_{CC}	Circuit current	$V_{CC}=5\text{V}$ bi-directional mode	All enable input="H"		4	8	mA
			All enable input="L"		10	20	
I_m	Drive supply voltage current	$V_m=12\text{V}$, no-load bi-directional mode	All enable input="H"		0		mA
			All enable input="L"		25	50	
$V_{TH(IN)}$	Input threshold value	$V_{CC}=5\text{V}$	1.6	2.2	2.7	V	
$V_{TH(EN)}$	Enable input threshold value	$V_{CC}=5\text{V}$	1.6	2.2	2.7	V	
$I_{IN(IN)}$	Input current	$V_{CC}=5\text{V}$ (flowing to the pin)		3.5	20	μA	
$I_{IN(EN)}$	Enable input current	$V_{CC}=5\text{V}$ (flowing from the pin)		0.5	10	μA	
$V_{sat(U)}$	Output saturation voltage (source)	$I_o=0.2\text{A}$		1.0	1.6	V	
		$I_o=0.4\text{A}$		1.3	1.9		
$V_{sat(D)}$	Output saturation voltage (sink)	$I_o=0.2\text{A}$		0.9	1.5	V	
		$I_o=0.4\text{A}$		1.0	1.6		

DUAL BI-DIRECTIONAL DRIVER

REFERENCE DATA ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Reference value	Unit
t_r	Rise time	Output 10%→90%	150	ns
t_f	Fall time	Output 90%→10%	550	ns
t_{PLH}	Output low-level to high-level propagation delay time	Input 50%→Output 50%	350	ns
t_{PHL}	Output high-level to low-level propagation delay time	Input 50%→Output 50%	400	ns

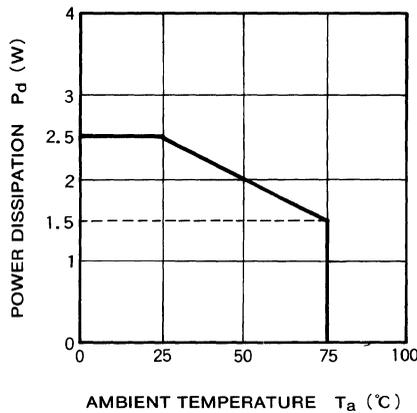
TRUTH TABLE

Input	Enable	Output
"H"	"L"	"H"
"L"	"L"	"L"
"H"	"H"	OFF
"L"	"H"	OFF

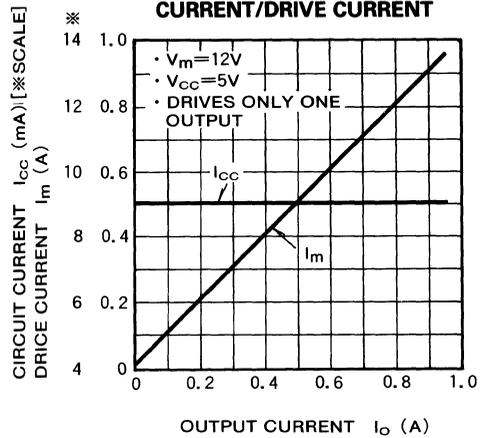
Note 1. ENABLE 1 functions in amplifiers 1 and 2, and ENABLE 2 functions in amplifiers 3 and 4.
 2. OFF means high impedance.

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

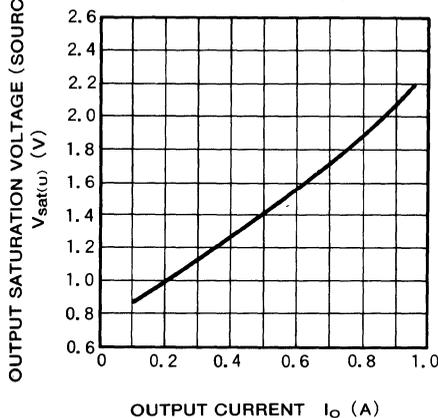
THERMAL DERATING (MAXIMUM RATING)



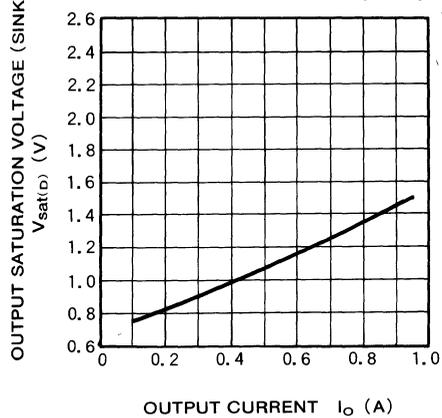
OUTPUT CURRENT VS. CIRCUIT CURRENT/DRIVE CURRENT



OUTPUT CURRENT VS. OUTPUT SATURATION VOLTAGE (SOURCE)

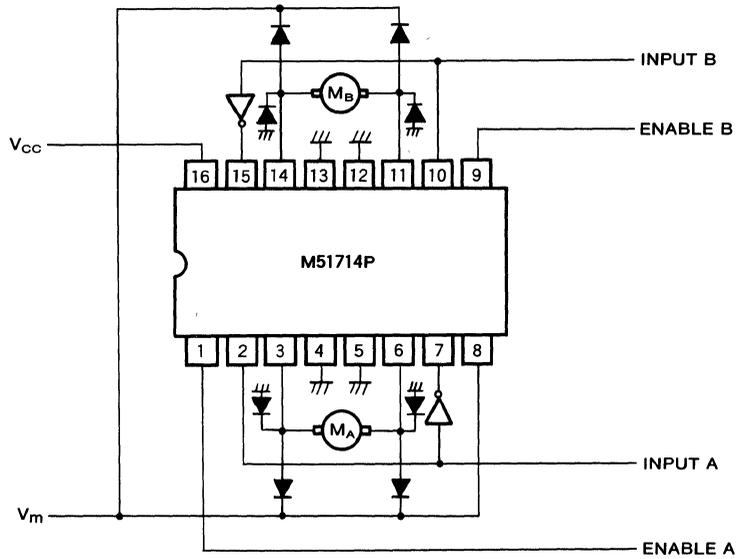


OUTPUT CURRENT VS. OUTPUT SATURATION VOLTAGE (SINK)



DUAL BI-DIRECTIONAL DRIVER

APPLICATION EXAMPLE



Note. A diode of small V_F is used between clamps. (For example, Schottky diode)

M54540AL

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54540AL, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for a low power D-C motor control.

FEATURES

- Small single-in-line package
- Integral diodes for transient suppression
- PMOS compatible input

APPLICATION

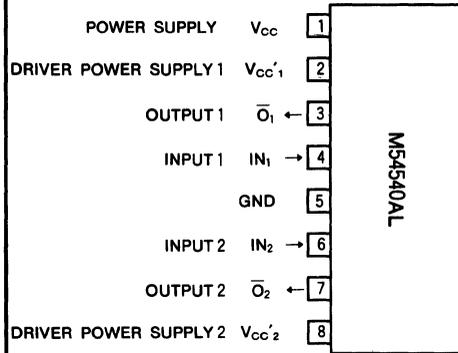
Audio cassette tape recorder

FUNCTION

The M54540AL, full bridge motor driver, has the logic circuitry and non-darlington power drivers for bidirection control of D-C motors operating at currents of up to 600mA.

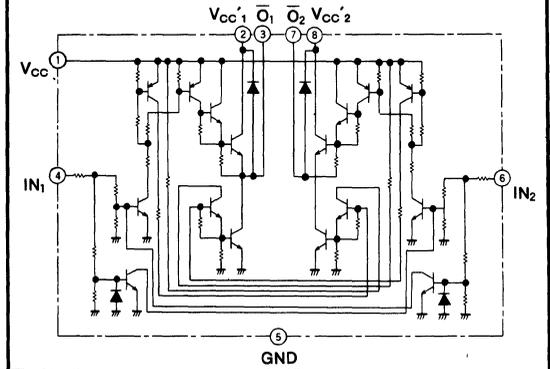
The power supplies for the logic circuitry and the drivers are separated so that the applied voltage to the motor can be controlled by the V_{CC}' of the driver supply voltage.

PIN CONFIGURATION (TOP VIEW)



Outline 8P5

CIRCUIT SCHEMATIC



LOGIC TRUTH TABLE

Input		Output		Note
IN ₁	IN ₂	\bar{O}_1	\bar{O}_2	
L	L	"OFF" state	"OFF" state	Open
H	L	H	L	○
L	H	L	H	○
H	H	"OFF" state	"OFF" state	Open

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		-0.3~+12	V
V _{CC'}	Driver supply voltage		-0.3~V _{CC}	V
V _I	Input voltage		-0.3~V _{CC}	V
V _O	Output voltage		-0.3~V _{CC} +2.5	V
I _{O(max)}	Peak output current	t _{OP} =10ms, Repetitive cycle 0.2Hz max	±600	mA
I _O	Continuous output current		±120	mA
P _d	Power dissipation	T _a =60°C	850	mW
T _{OPR}	Operating ambient temperature range		-10~+60	°C
T _{STG}	Storage temperature range		-55~+125	°C

BI-DIRECTIONAL MOTOR DRIVER

RECOMMENDED OPERATING CONDITIONS (T_a=25°C, unless otherwise noted)

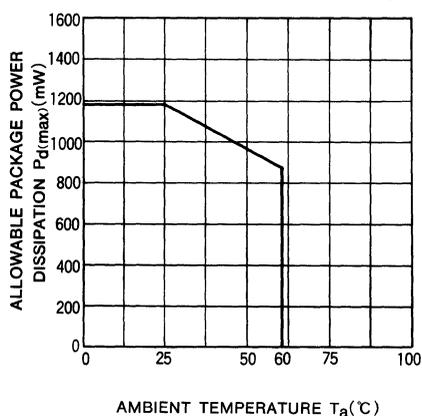
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage		6	10	11	V
I _O	Continuous output current				±100	mA
V _{IH}	"H" Input voltage		3	5	V _{CC}	V
V _{IL}	"L" Input voltage			0	0.4	V
T _{off}	Input switching interval	It is prohibited to switch the inputs at the same time.	10	300		ms

ELECTRICAL CHARACTERISTICS (T_a=25°C, unless otherwise noted)

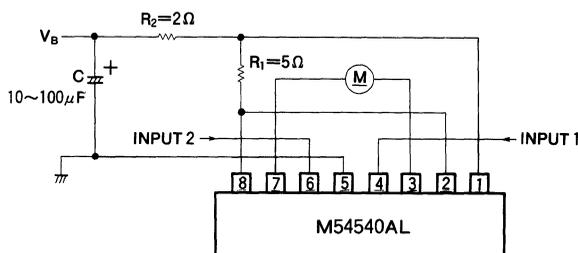
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I _{O(leak)}	Output leakage current	V _{CC} =V _{CC'} =20V V _{I1} =V _{I2} =3V	V _O =12V V _O =0V		100 -100	μA
V _{OH}	"H" Output saturation voltage	V _{CC} =V _{CC'} =10V I _{OH} =-100mA	V _{I1} =3V, V _{I2} =0V V _{I1} =0V, V _{I2} =3V	8		V
V _{OL}	"L" Output saturation voltage	V _{CC} =V _{CC'} =10V I _{OL} =100mA	V _{I1} =3V, V _{I2} =0V V _{I1} =0V, V _{I2} =3V		0.6	V
I _{IH}	"H" Input current	V _{CC} =V _{CC'} =10V	V _{I1} =3V V _{I2} =3V		500	μA
I _{CC}	Supply current	V _{CC} =V _{CC'} =12V	V _{I1} =3V, V _{I2} =0V V _{I1} =0V, V _{I2} =3V V _{I1} =0V, V _{I2} =0V V _{I1} =3V, V _{I2} =3V		28 0 40	mA

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE
POWER DISSIPATION



APPLICATION EXAMPLE



Note

1. It is prohibited to switch the both inputs simultaneously. The inputs should be driven separately to avoid high crossover current.
2. The pins 1, 2 and 8 are separated and shall be connected externally.

M54541L

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54541L, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for D-C motor control.

FEATURES

- 9-pin single-in-line power tab package
- Integral diodes for transient suppression
- 800mA output current
- PMOS compatible input

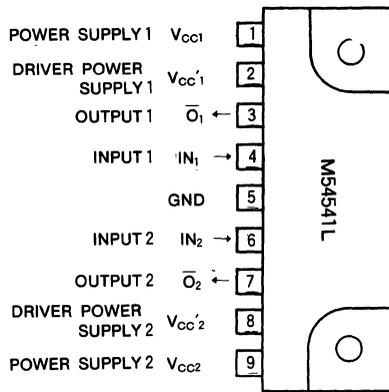
APPLICATIONS

- Audio cassette tape recorder
- Video cassette recorder

FUNCTION

The M54541L, full bridge motor driver, has the logic circuitry and darlington-pair power drivers for bidirectional control of D-C motors operating at currents up to 800mA. The power supplies for the logic circuitry and the drivers are separated so that the applied voltage to the motor can be controlled by the V_{CC} of the driver power supply voltage.

PIN CONFIGURATION (TOP VIEW)

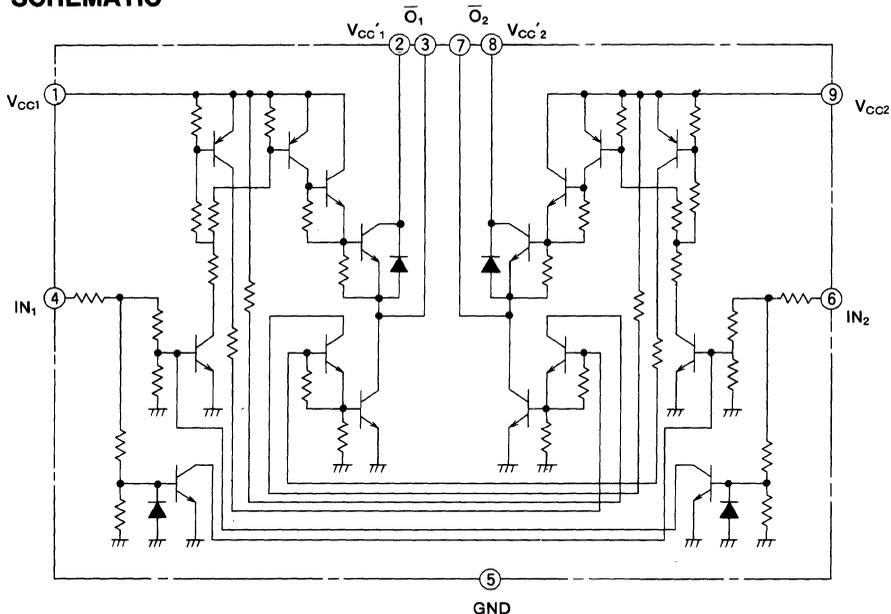


Outline 9P9

LOGIC TRUTH TABLE

Input		Output		Note
IN ₁	IN ₂	O ₁	O ₂	
L	L	"OFF" state	"OFF" state	Open
H	L	H	L	⊙
L	H	L	H	⊙
H	H	"OFF" state	"OFF" state	Open

CIRCUIT SCHEMATIC



BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		$-0.5 \sim +16$	V
$V_{CC'}$	Driver supply voltage		$-0.5 \sim V_{CC}$	V
V_i	Input voltage		$-0.5 \sim V_{CC}$	V
V_o	Output voltage		$-0.5 \sim V_{CC} + 2.5$	V
$I_{O(max)}$	Peak output current	$t_{op}=10\text{ms}$, Repetitive cycle 0.2Hz max	± 800	mA
I_o	Continuous output current		± 220	mA
P_d	Power dissipation	$T_a=60^\circ\text{C}$	900	mW
T_{opr}	Operating ambient temperature range		$-10 \sim +60$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		6	14	15	V
I_o	Continuous output current				± 200	mA
V_{IH}	"H" Input voltage		3	5	V_{CC}	V
V_{iL}	"L" Input voltage			0	0.4	V
T_{off}	Input switching interval	It is prohibited to switch the inputs at the same time	10	300		ms

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

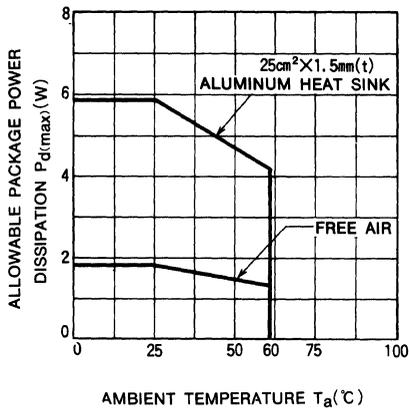
Symbol	Parameter	Test conditions		Limits			Unit
				Min	Typ*	Max	
$I_{O(leak)}$	Output leakage current	$V_{CC}=V_{CC'}=20\text{V}$ $V_{i1}=V_{i2}=3\text{V}$	$V_o=20\text{V}$ $V_o=0\text{V}$			100 -100	μA
V_{OH}	"H" Output saturation voltage	$V_{CC}=V_{CC'}=12\text{V}$ $I_{OH}=-200\text{mA}$	$V_{i1}=3\text{V}, V_{i2}=0\text{V}$ $V_{i1}=0\text{V}, V_{i2}=3\text{V}$	9.9	10.4		V
V_{OL}	"L" Output saturation voltage	$V_{CC}=V_{CC'}=12\text{V}$ $I_{OL}=200\text{mA}$	$V_{i1}=3\text{V}, V_{i2}=0\text{V}$ $V_{i1}=0\text{V}, V_{i2}=3\text{V}$		0.3	0.7	V
I_{IH}	"H" Input current	$V_{CC}=V_{CC'}=12\text{V}$	$V_{i1}=3\text{V}$ $V_{i2}=3\text{V}$			500	μA
I_{CC}	Supply current	$V_{CC}=V_{CC'}=16\text{V}$	$V_{i1}=3\text{V}, V_{i2}=0\text{V}$ $V_{i1}=0\text{V}, V_{i2}=3\text{V}$ $V_{i1}=0\text{V}, V_{i2}=0\text{V}$ $V_{i1}=3\text{V}, V_{i2}=3\text{V}$		35 0	50	mA

* : The all typical values are at $T_a=25^\circ\text{C}$.

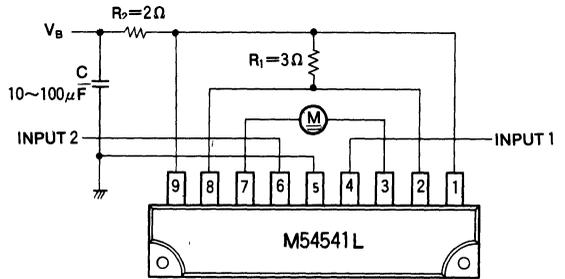
BI-DIRECTIONAL MOTOR DRIVER

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE
POWER DISSIPATION



APPLICATION EXAMPLE



Note

1. It is prohibited to switch the both input simultaneously. The inputs should be driven separately to avoid high crossover current.
2. The pins 1, 9 and 2, 8 are separated and shall be connected externally.

M54542L

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54542L, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for D-C motor control.

FEATURES

- 9-pin single-in-line power tab package
- Integral diodes for transient suppression
- 1.2A output current
- PMOS compatible input

APPLICATIONS

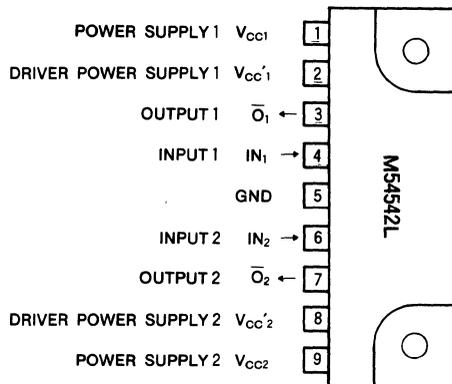
- Audio, video cassette recorders
- Floppy disk driver

FUNCTION

The M54542L, full bridge motor driver, has the logic circuitry and darlington-pair power drivers for bidirectional control of D-C motors operating at currents up to 1.2A.

The power supplies for the logic circuitry and the drivers are separated so that the applied voltage to the motor can be controlled by the V_{CC}' of the driver power supply voltage.

PIN CONFIGURATION (TOP VIEW)

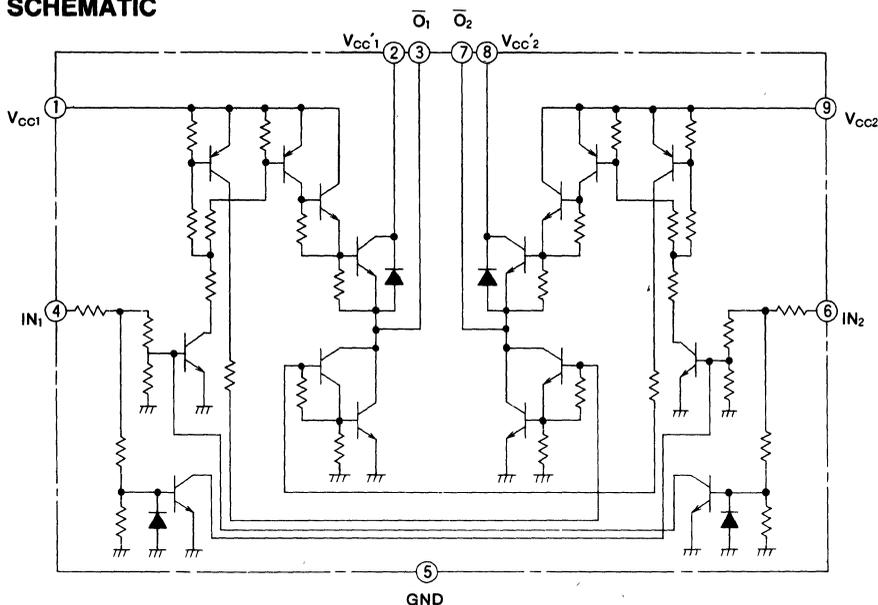


Outline 9P9

LOGIC TRUTH TABLE

INPUT		OUTPUT		NOTE
IN ₁	IN ₂	\bar{O}_1	\bar{O}_2	
L	L	"OFF" state	"OFF" state	Open
H	L	H	L	⊙
L	H	L	H	⊙
H	H	"OFF" state	"OFF" state	Open

CIRCUIT SCHEMATIC



BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		-0.5~+16	V
$V_{CC'}$	Driver voltage		-0.5~ V_{CC}	V
V_I	Input voltage		-0.5~ V_{CC}	V
V_O	Output voltage		-0.5~ $V_{CC}+2.5$	V
$I_{O(max)}$	Peak output current	$t_{op}=10\text{ms}$ Repetitive cycle 0.2Hz max	± 1200	mA
I_O	Continuous output current		± 330	mA
P_d	Power dissipation	$T_a=60^\circ\text{C}$	1000	mW
T_{opr}	Operating ambient temperature range		-10~+60	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		6	14	15	V
I_O	Continuous output current				± 300	mA
V_{IH}	"H" Input voltage		3	5	V_{CC}	V
V_{IL}	"L" Input voltage			0	0.4	V
T_{OFF}	Input switching interval	It is prohibited to switch the inputs at the same time.	10	300		ms

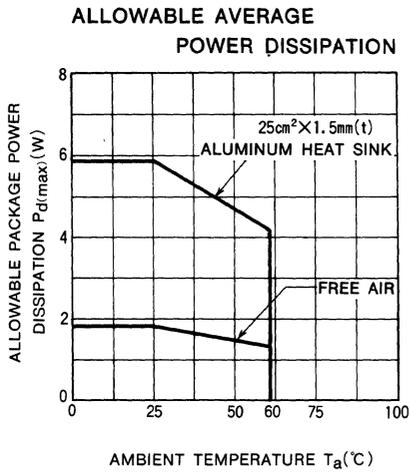
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions		Limits			Unit
				Min	Typ*	Max	
$I_{O(leak)}$	Output leakage current	$V_{CC}=V_{CC'}=20\text{V}$ $V_{I1}=V_{I2}=3\text{V}$	$V_O=20\text{V}$			100	μA
			$V_O=0\text{V}$			-100	
V_{OH}	"H" Output saturation voltage	$V_{CC}=V_{CC'}=12\text{V}$ $I_{OH}=-300\text{mA}$	$V_{I1}=3\text{V}, V_{I2}=0\text{V}$ $V_{I1}=0\text{V}, V_{I2}=3\text{V}$	9.7	10.2		V
V_{OL}	"L" Output saturation voltage	$V_{CC}=V_{CC'}=12\text{V}$ $I_{OL}=300\text{mA}$	$V_{I1}=3\text{V}, V_{I2}=0\text{V}$ $V_{I1}=0\text{V}, V_{I2}=3\text{V}$				0.9
I_{IH}	"H" Input current	$V_{CC}=V_{CC'}=12\text{V}$	$V_{I1}=3\text{V}$ $V_{I2}=3\text{V}$			500	μA
I_{CC}	Supply current	$V_{CC}=V_{CC'}=16\text{V}$	$V_{I1}=3\text{V}, V_{I2}=0\text{V}$		7	10	mA
			$V_{I1}=0\text{V}, V_{I2}=3\text{V}$				
			$V_{I1}=0\text{V}, V_{I2}=0\text{V}$ $V_{I1}=3\text{V}, V_{I2}=3\text{V}$		0		

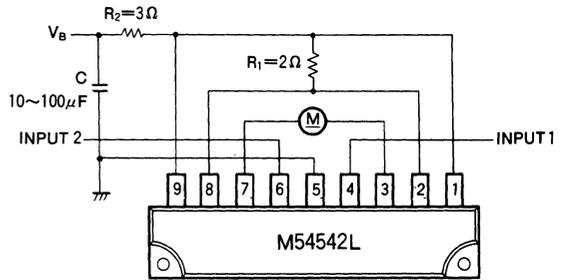
* : The all typical values are at $T_a=25^\circ\text{C}$

BI-DIRECTIONAL MOTOR DRIVER

TYPICAL CHARACTERISTICS



APPLICATION EXAMPLE



Note

1. It is prohibited to switch the both inputs simultaneously. The inputs should be driven separately to avoid high crossover current.
2. The pins 1, 9 and 2, 8 are separated and shall be connected externally.

M54543L

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54543L, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for D-C motor control.

FEATURES

- Wide operating voltage range ($V_{CC} = 4 \sim 16V$)
- TTL, PMOS and CMOS compatible input
- Low output saturation voltage
- Integral diodes for transient suppression
- 1.2A output current
- Braking mode input

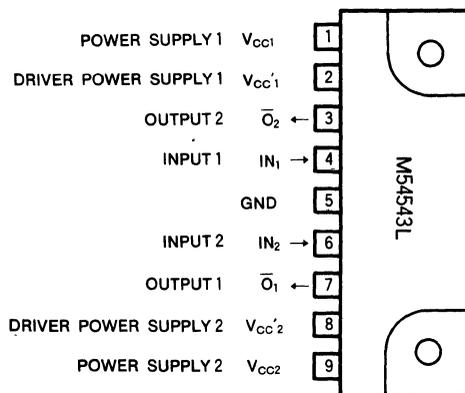
APPLICATION

Audio, video cassette recorder

FUNCTION

The M54543L, full bridge motor driver, has the logic circuitry and non-darlington power drivers for bidirectional control of D-C motors operating at currents up to 1.2A. A braking mode by switching the both inputs high may make easier to control the motor. The both of the separated power supplies for the logic circuitry and the drivers are usable for motor speed control.

PIN CONFIGURATION (TOP VIEW)

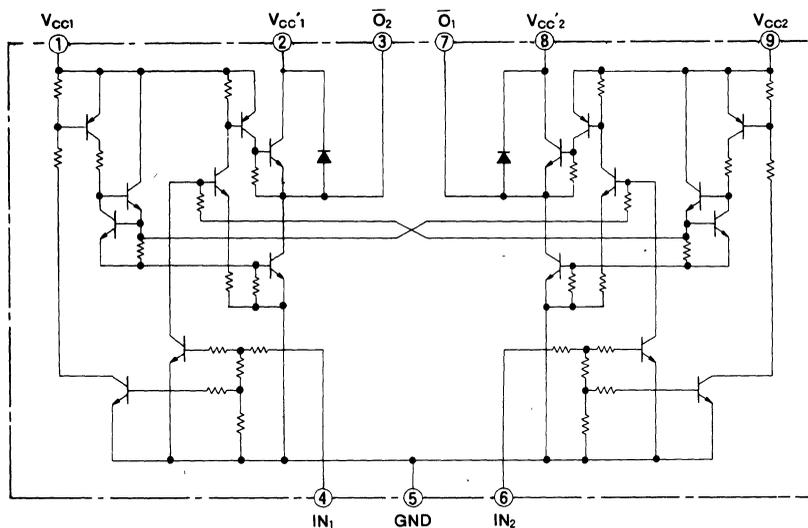


Outline 9P9

LOGIC TRUTH TABLE

INPUT		OUTPUT		NOTE
IN ₁	IN ₂	O ₁	O ₂	
L	L	"OFF" state	"OFF" state	Open
H	L	H	L	○
L	H	L	H	○
H	H	L	L	Braking

CIRCUIT SCHEMATIC



BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
$V_{CC(1)}$	Supply voltage (1)		-0.5~+16	V
$V_{CC(2)}$	Supply voltage (2)	With an external heat sink (3000mm ² X1.5mm)	-0.5~+20	V
$V_{CC'}$	Driver supply voltage		-0.5~+16	V
V_i	Input voltage		0~ V_{CC}	V
V_o	Output voltage		-0.5~ $V_{CC'}+2.5$	V
$I_o(\text{max})$	Peak output current	$t_{op}=10\text{ms}$: Repetitive cycle 0.2Hz max	± 1.2	A
$I_o(1)$	Continuous output current (1)		± 330	mA
$I_o(2)$	Continuous output current (2)	With an external heat sink (3000mm ² X1.5mm)	± 600	mA
P_d	Power dissipation	$T_a=75^\circ\text{C}$	1.15	W
T_{opr}	Operating ambient temperature range		-10~+75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

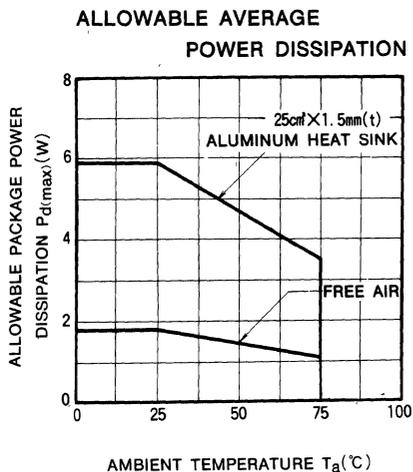
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	12	15	V
I_o	Continuous output current				± 300	mA
V_{IH}	"H" Input voltage		2		V_{CC}	V
V_{iL}	"L" Input voltage		0		0.4	V
t_B	Motor braking interval		100			ms

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions		Limits			Unit
				Min	Typ	Max	
$I_o(\text{leak})$	Output leakage current	$V_{CC}=V_{CC'}=20\text{V}$ $V_{i1}=V_{i2}=0\text{V}$	$V_o=20\text{V}$			100	μA
			$V_o=0\text{V}$			-100	
$V_{OH(1)}$	"H" Output saturation voltage (1)	$V_{CC}=V_{CC'}=12\text{V}$	$V_{i1}=2\text{V}$ $V_{i2}=0\text{V}$	$I_{OH(1)}=-300\text{mA}$	10.8		V
				$I_{OH(1)}=-500\text{mA}$	10.7		
$V_{OH(2)}$	"H" Output saturation voltage (2)	$V_{CC}=V_{CC'}=12\text{V}$	$V_{i1}=0\text{V}$ $V_{i2}=2\text{V}$	$I_{OH(2)}=-300\text{mA}$	10.8		V
				$I_{OH(2)}=-500\text{mA}$	10.7		
$V_{OL(1)}$	"L" Output saturation voltage (1)	$V_{CC}=V_{CC'}=12\text{V}$	$V_{i1}=0\text{V}$ $V_{i2}=2\text{V}$	$I_{OL(1)}=300\text{mA}$		0.5	V
				$I_{OL(1)}=500\text{mA}$		0.65	
						0.65	
$V_{OL(2)}$	"L" Output saturation voltage (2)	$V_{CC}=V_{CC'}=12\text{V}$	$V_{i1}=2\text{V}$ $V_{i2}=0\text{V}$ $V_{i1}=V_{i2}=2\text{V}$	$I_{OL(2)}=300\text{mA}$		0.5	V
				$I_{OL(2)}=500\text{mA}$		0.65	
						0.65	
$I_{IH(1)}$	"H" Input current (1)	$V_{CC}=V_{CC'}=12\text{V}$, $V_{i1}=2\text{V}$, $V_{i2}=0\text{V}$		70		200	μA
$I_{IH(2)}$	"H" Input current (2)	$V_{CC}=V_{CC'}=12\text{V}$, $V_{i1}=0\text{V}$, $V_{i2}=2\text{V}$		70		200	μA
I_{CC}	Supply current	$V_{CC}=V_{CC'}=16\text{V}$	$V_{i1}=2\text{V}$, $V_{i2}=0\text{V}$			40	mA
			$V_{i1}=0\text{V}$, $V_{i2}=2\text{V}$			60	
			$V_{i1}=V_{i2}=2\text{V}$			60	
			$V_{i1}=V_{i2}=0\text{V}$		0		

BI-DIRECTIONAL MOTOR DRIVER

TYPICAL CHARACTERISTICS



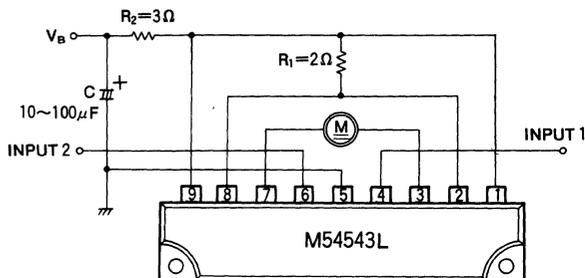
Note

When the $V_{CC'}$ is lower than the V_{CC} , the current will flow from the V_{CC} to the $V_{CC'}$ and may drive the motor.

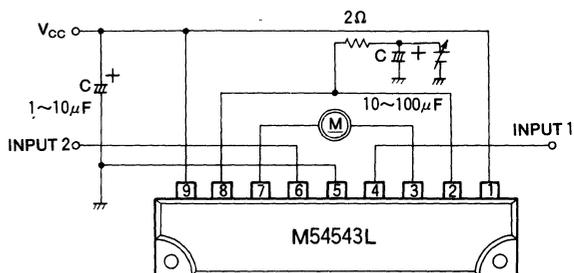
The M54544L may be recommended to have the wider control voltage range of the $V_{CC'}$.

APPLICATION EXAMPLES

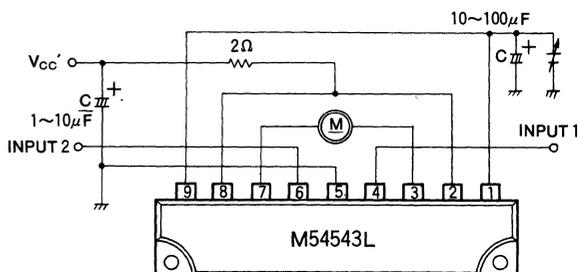
1) Motor speed control by V_{CC} and $V_{CC'}$



2) Motor speed control by the $V_{CC'}$



3) Motor speed control by the V_{CC}



M54543AL

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54543AL, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for D-C motor control.

FEATURES

- Wide operating voltage range ($V_{CC} = 4 \sim 16V$)
- TTL, PMOS and CMOS compatible input
- Low output saturation voltage
- Integral diodes for transient suppression
- 1.2A output current
- Braking mode input
- Internal thermal shutdown protection

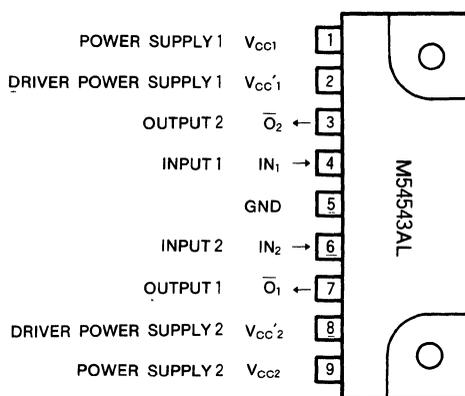
APPLICATION

Audio, video cassette recorder

FUNCTION

The M54543AL, full bridge motor driver, has the logic circuitry and non-darlington power drivers for bidirectional control of D-C motors operating at currents up to 1.2A. A braking mode by switching the both inputs high may make easier to control the motor. The both of the separated power supplies for the logic circuitry and the drivers are usable for motor speed control.

PIN CONFIGURATION (TOP VIEW)

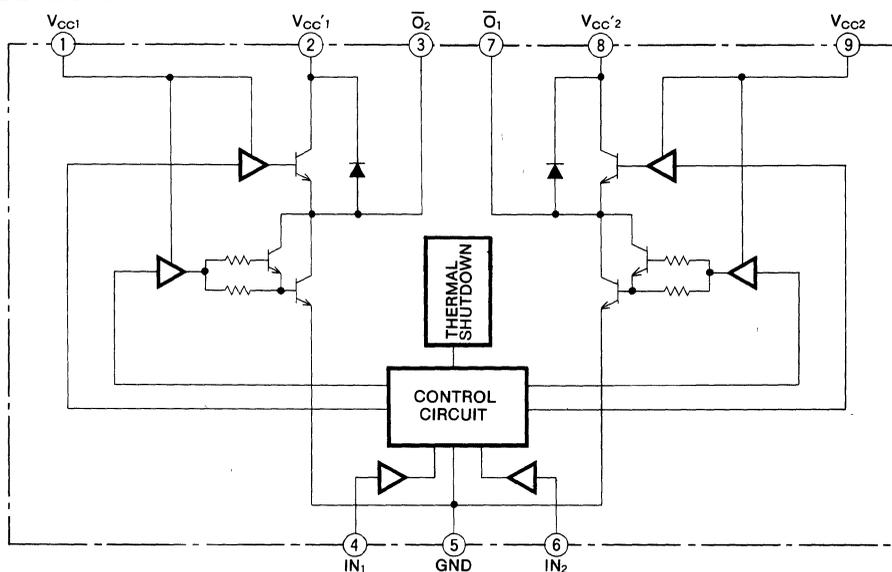


Outline 9P9

LOGIC TRUTH TABLE

INPUT		OUTPUT		NOTE
IN ₁	IN ₂	\bar{O}_1	\bar{O}_2	
L	L	"OFF" state	"OFF" state	Open
H	L	H	L	○
L	H	L	H	○
H	H	L	L	Braking

BLOCK DIAGRAM



BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
$V_{CC(1)}$	Supply voltage (1)		-0.5~+16	V
$V_{CC(2)}$	Supply voltage (2)	With an external heat sink (3000mm ² X1.5mm)	-0.5~+20	V
$V_{CC'}$	Driver supply voltage		-0.5~+16	V
V_i	Input voltage		0~ V_{CC}	V
V_o	Output voltage		-0.5~ $V_{CC'}+2.5$	V
$I_{O(max)}$	Peak output current	$t_{op}=10\text{ms}$; Repetitive cycle 0.2Hz max	± 2	A
$I_{O(1)}$	Continuous output current (1)		± 330	mA
$I_{O(2)}$	Continuous output current (2)	With an external heat sink (3000mm ² X1.5mm)	± 600	mA
P_d	Power dissipation	$T_a=75^\circ\text{C}$	1.15	W
T_{opr}	Operating ambient temperature range		-10~+75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

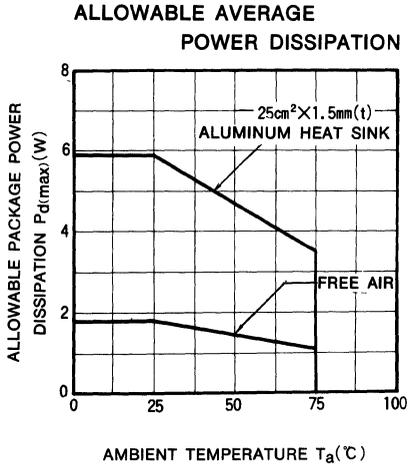
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	12	15	V
I_o	Continuous output current				± 300	mA
V_{IH}	"H" Input voltage		2		V_{CC}	V
V_{IL}	"L" Input voltage		0		0.4	V
t_B	Motor braking interval		100			ms
$T_{j(shut)}$	Thermal shutdown temperature	Junction temperature		150		$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{O(leak)}$	Output leakage current	$V_{CC}=V_{CC'}=20\text{V}$ $V_{i1}=V_{i2}=0\text{V}$	$V_o=20\text{V}$		100	μA
			$V_o=0\text{V}$		-100	
$V_{OH(1)}$	"H" Output saturation voltage (1)	$V_{CC}=V_{CC'}=12\text{V}$	$V_{i1}=2\text{V}$ $V_{i2}=0\text{V}$	$I_{OH(1)}=-200\text{mA}$	10.8	V
				$I_{OH(1)}=-500\text{mA}$	10.7	
$V_{OH(2)}$	"H" Output saturation voltage (2)	$V_{CC}=V_{CC'}=12\text{V}$	$V_{i1}=0\text{V}$ $V_{i2}=2\text{V}$	$I_{OH(2)}=-200\text{mA}$	10.8	V
				$I_{OH(2)}=-500\text{mA}$	10.7	
$V_{OL(1)}$	"L" Output saturation voltage (1)	$V_{CC}=V_{CC'}=12\text{V}$	$V_{i1}=0\text{V}$ $V_{i2}=2\text{V}$	$I_{OL(1)}=200\text{mA}$	0.5	V
				$I_{OL(1)}=500\text{mA}$	1.35	
$V_{OL(2)}$	"L" Output saturation voltage (2)	$V_{CC}=V_{CC'}=12\text{V}$	$V_{i1}=2\text{V}$ $V_{i2}=0\text{V}$	$I_{OL(2)}=200\text{mA}$	0.5	V
				$I_{OL(2)}=500\text{mA}$	1.35	
$I_{IH(1)}$	"H" Input current (1)	$V_{CC}=V_{CC'}=12\text{V}$, $V_{i1}=2\text{V}$, $V_{i2}=0\text{V}$	50	120	μA	
$I_{IH(2)}$	"H" Input current (2)	$V_{CC}=V_{CC'}=12\text{V}$, $V_{i1}=0\text{V}$, $V_{i2}=2\text{V}$	50	120	μA	
I_{CC}	Supply current	$V_{CC}=V_{CC'}=16\text{V}$	$V_{i1}=2\text{V}$, $V_{i2}=0\text{V}$		20	mA
			$V_{i1}=0\text{V}$, $V_{i2}=2\text{V}$		20	
			$V_{i1}=V_{i2}=2\text{V}$		20	
			$V_{i1}=V_{i2}=0\text{V}$		4	

BI-DIRECTIONAL MOTOR DRIVER

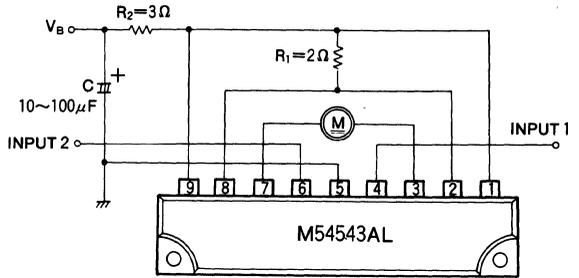
TYPICAL CHARACTERISTICS



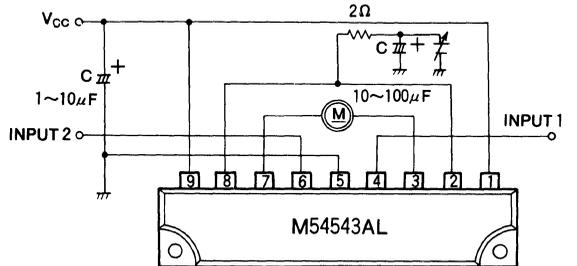
Note
 When the $V_{CC'}$ is lower than the V_{CC} , the current will flow from the V_{CC} to the $V_{CC'}$ and may drive the motor.
 The M54544AL may be recommended to have the wider control voltage range of the V_{CC} .

APPLICATION EXAMPLES

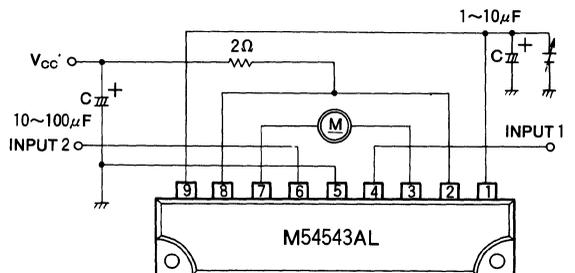
1) Motor speed control by V_{CC} and $V_{CC'}$



2) Motor speed control by the $V_{CC'}$



3) Motor speed control by the V_{CC}



M54544L

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54544L, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for D-C motor control.

FEATURES

- Wide operating voltage range ($V_{CC} = 4 \sim 16V$)
- TTL, PMOS and CMOS compatible input
- Low output saturation voltage
- Integral diodes for transient suppression
- 1.2A output current
- Braking mode input

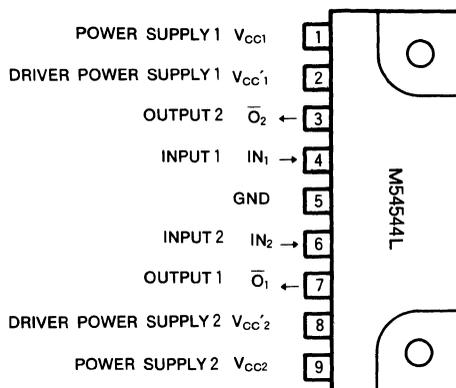
APPLICATION

Audio, video cassette recorders

FUNCTION

The M54544L, full bridge motor driver, has the logic circuitry and non-darlington power drivers for bidirectional control of D-C motors operating at currents up to 1.2A. A braking mode by switching the both inputs high may make easier to control the motor. The both of the separated power supplies for the logic circuitry and the drivers are usable for motor speed control. The power supply for the predriver is connected with the driver power supply to have a wider control range of motor supply voltage.

PIN CONFIGURATION (TOP VIEW)

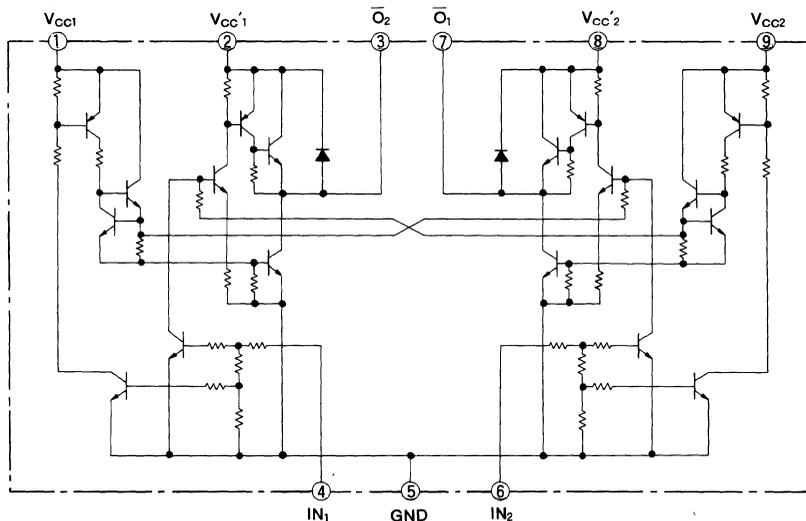


Outline 9P9

LOGIC TRUTH TABLE

Input		Output		Note
IN ₁	IN ₂	O ₁	O ₂	
L	L	"OFF" state	"OFF" state	Open
H	L	H	L	○
L	H	L	H	○
H	H	L	L	Braking

CIRCUIT SCHEMATIC



BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC(1)}	Supply voltage (1)		-0.5~+16	V
V _{CC(2)}	Supply voltage (2)	With an external heat sink (3000mm ² ×1.5mm)	-0.5~+20	V
V _{CC'}	Driver supply voltage		-0.5~+16	V
V _I	Input voltage		0~V _{CC}	V
V _O	Output voltage		-0.5~V _{CC'} +2.5	V
I _{O(max)}	Peak output current	t _{op} =10ms: Repetitive cycle 0.2Hz max	±1.2	A
I _{O(1)}	Continuous output current (1)		±330	mA
I _{O(2)}	Continuous output current (2)	With an external heat sink (3000mm ² ×1.5mm)	±600	mA
P _d	Power Dissipation	T _a =75°C	1.15	W
T _{opr}	Operating ambient temperature range		-10~+75	°C
T _{stg}	Storage temperature range		-55~+125	°C

RECOMMENDED OPERATING CONDITIONS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage		4	12	15	V
I _O	Continuous output current				±300	mA
V _{IH}	"H" Input voltage		2		V _{CC}	V
V _{IL}	"L" Input voltage		0		0.4	V
t _B	Motor braking interval		100			ms

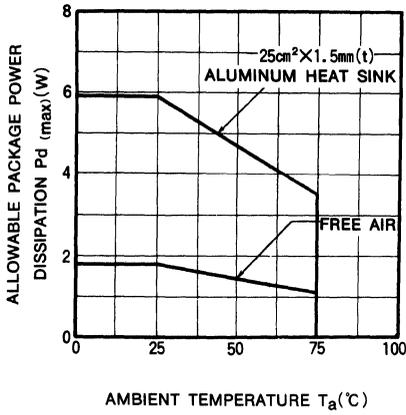
ELECTRICAL CHARACTERISTICS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Test conditions		Limits			Unit
				Min	Typ	Max	
I _{O(leak)}	Output leakage current	V _{CC} =V _{CC'} =20V V _{I1} =V _{I2} =0V	V _O =20V V _O =0V			100 -100	μA
V _{OH(1)}	"H" Output saturation voltage (1)	V _{CC} =V _{CC'} =12V	V _{I1} =2V V _{I2} =0V	I _{OH(1)} =-300mA I _{OH(1)} =-500mA	10.8 10.7		V
V _{OH(2)}	"H" Output saturation voltage (2)	V _{CC} =V _{CC'} =12V	V _{I2} =2V	I _{OH(2)} =-300mA I _{OH(2)} =-500mA	10.8 10.7		V
V _{OL(1)}	"L" Output saturation voltage (1)	V _{CC} =V _{CC'} =12V	V _{I1} =0V V _{I2} =2V V _{I1} =V _{I2} =2V	I _{OL(1)} =300mA I _{OL(1)} =500mA		0.5 0.65	V
V _{OL(2)}	"L" Output saturation voltage (2)	V _{CC} =V _{CC'} =12V	V _{I1} =2V V _{I2} =0V V _{I1} =V _{I2} =2V	I _{OL(2)} =300mA I _{OL(2)} =500mA		0.5 0.65	V
I _{IH(1)}	"H" Input current (1)	V _{CC} =V _{CC'} =12V, V _{I1} =2V, V _{I2} =0V			70	200	μA
I _{IH(2)}	"H" Input current (2)	V _{CC} =V _{CC'} =12V, V _{I1} =0V, V _{I2} =2V			70	200	μA
I _{CC}	Supply current	V _{CC} =V _{CC'} =16V	V _{I1} =2V, V _{I2} =0V V _{I1} =0V, V _{I2} =2V V _{I1} =V _{I2} =2V V _{I1} =V _{I2} =0V			30 60 0	mA

BI-DIRECTIONAL MOTOR DRIVER

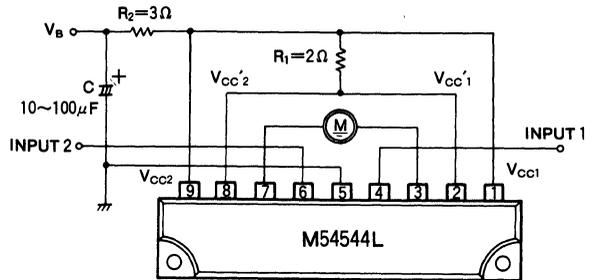
TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE
 POWER DISSIPATION

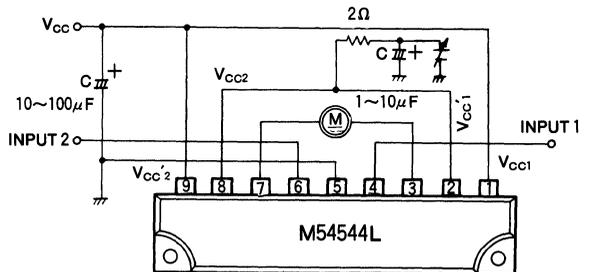


APPLICATION EXAMPLES

1) Motor speed control by V_{CC} and V_{CC}'



2) Motor speed control by the V_{CC}'



M54544AL

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54544AL, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for D-C motor control.

FEATURES

- Wide operating voltage range ($V_{CC} = 4 \sim 16V$)
- TTL, PMOS and CMOS compatible input
- Low output saturation voltage
- Integral diodes for transient suppression
- 1.2A output current
- Braking mode input
- Internal thermal shutdown protection

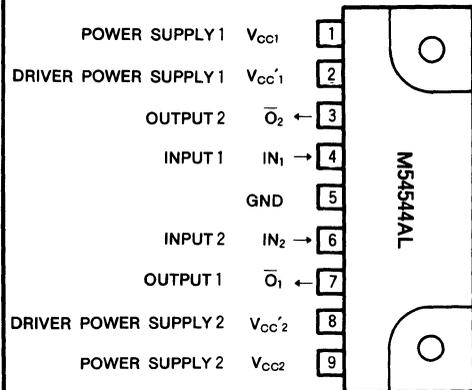
APPLICATION

Audio, video cassette recorders

FUNCTION

The M54544AL, full bridge motor driver, has the logic circuitry and non-darlington power drivers for bidirectional control of D-C motors operating at currents up to 1.2A. A braking mode by switching the both inputs high may make easier to control the motor. The both of the separated power supplies for the logic circuitry and the drivers are usable for motor speed control. The power supply for the predriver is connected with the driver power supply to have a wider control range of motor supply voltage.

PIN CONFIGURATION (TOP VIEW)

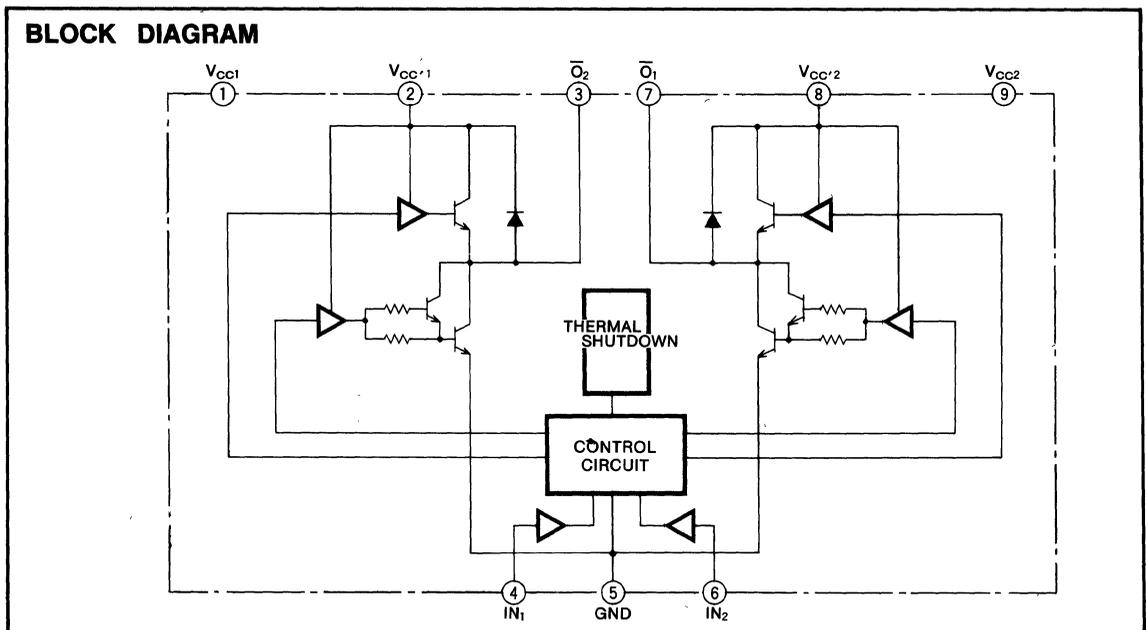


Outline 9P9

LOGIC TRUTH TABLE

Input		Output		Note
IN ₁	IN ₂	O ₁	O ₂	
L	L	"OFF" state	"OFF" state	Open
H	L	H	L	⊙
L	H	L	H	⊙
H	H	L	L	Braking

BLOCK DIAGRAM



BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC(1)}	Supply voltage (1)		-0.5~+16	V
V _{CC(2)}	Supply voltage (2)	With an external heat sink (3000mm ² X1.5mm)	-0.5~+20	V
V _{CC'}	Driver supply voltage		-0.5~+16	V
V _I	Input voltage		0~V _{CC}	V
V _O	Output voltage		-0.5~V _{CC'+2.5}	V
I _{O(max)}	Peak output current	Repetitive cycle 0.2Hz max	±2	A
I _{O(1)}	Continuous output current (1)		±330	mA
I _{O(2)}	Continuous output current (2)	With an external heat sink (3000mm ² X1.5mm)	±600	mA
Pd	Power Dissipation	T _a =75°C	1.15	W
T _{opr}	Operating ambient temperature range		-10~+75	°C
T _{stg}	Storage temperature range		-55~+125	°C

RECOMMENDED OPERATING CONDITIONS (T_a=25°C, unless otherwise noted)

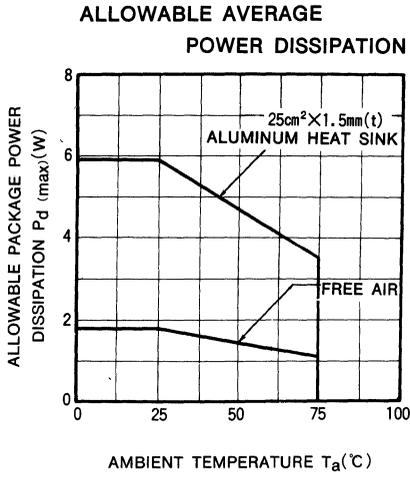
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage		4	12	15	V
I _O	Continuous output current				±300	mA
V _{IH}	"H" Input voltage		2		V _{CC}	V
V _{IL}	"L" Input voltage		0		0.4	V
t _B	Motor braking interval		10	100		ms
T _{J(shut)}	Thermal shutdown temperature	junction temperature		150		°C

ELECTRICAL CHARACTERISTICS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Test conditions		Limits			Unit
				Min	Typ	Max	
I _{O(leak)}	Output leakage current	V _{CC} =V _{CC'} =20V V _{I1} =V _{I2} =0V	V _O =20V			100	μA
			V _O =0V			-100	
V _{OH(1)}	"H" Output saturation voltage (1)	V _{CC} =V _{CC'} =12V	V _{I1} =2V V _{I2} =0V	I _{OH(1)} =-200mA	10.8		V
				I _{OH(1)} =-500mA	10.7		
V _{OH(2)}	"H" Output saturation voltage (2)	V _{CC} =V _{CC'} =12V	V _{I1} =0V V _{I2} =2V	I _{OH(2)} =-200mA	10.8		V
				I _{OH(2)} =-500mA	10.7		
V _{OL(1)}	"L" Output saturation voltage (1)	V _{CC} =V _{CC'} =12V	V _{I1} =0V V _{I2} =2V	I _{OL(1)} =200mA		0.5	V
				I _{OL(1)} =500mA		1.35	
			V _{I1} =V _{I2} =2V			1.35	
V _{OL(2)}	"L" Output saturation voltage (2)	V _{CC} =V _{CC'} =12V	V _{I1} =2V V _{I2} =0V	I _{OL(2)} =200mA		0.5	V
				I _{OL(2)} =500mA		1.35	
			V _{I1} =V _{I2} =2V			1.35	
I _{IH(1)}	"H" Input current (1)	V _{CC} =V _{CC'} =12V, V _{I1} =2V, V _{I2} =0V		50	120	μA	
I _{IH(2)}	"H" Input current (2)	V _{CC} =V _{CC'} =12V, V _{I1} =0V, V _{I2} =2V		50	120	μA	
I _{CC}	Supply current	V _{CC} =V _{CC'} =16V	V _{I1} =2V, V _{I2} =0V			15	mA
			V _{I1} =0V, V _{I2} =2V				
			V _{I1} =V _{I2} =2V			20	
			V _{I1} =V _{I2} =0V			4	

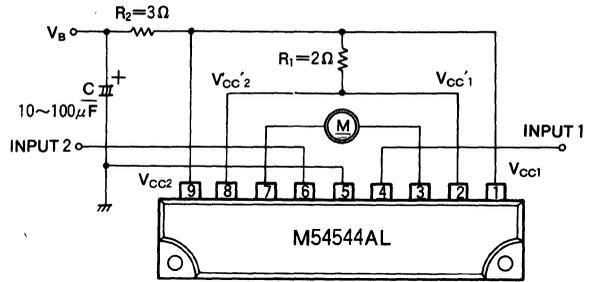
BI-DIRECTIONAL MOTOR DRIVER

TYPICAL CHARACTERISTICS

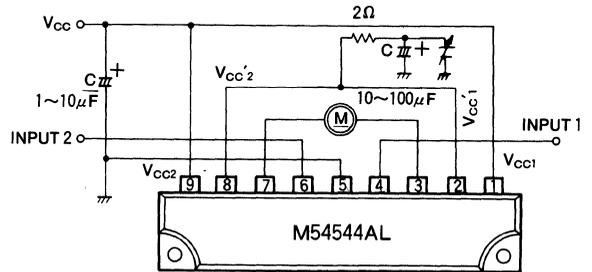


APPLICATION EXAMPLES

1) Motor speed control by V_{cc} and V_{cc}'



2) Motor speed control by the V_{cc}'



M54545L

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54545L, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for D-C motor control.

FEATURES

- Wide operating voltage range ($V_{CC} = 3 \sim 16V$)
- Low output saturation voltage
- Integral diodes for transient suppression
- 1.2A output current
- Braking mode input
- Low standby current

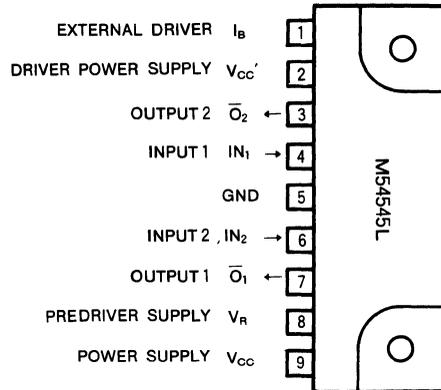
APPLICATION

Audio, video cassette recorder

FUNCTION

The M54545L, full bridge motor driver, has the logic circuitry and the quasi-darlington power driver for bidirectional control of D-C motors operating at currents up to 1.2A. A braking mode by switching the both inputs high may make easier to control the motor. The power supplies for the logic circuitry, the predrivers and the power drivers are separated so that the application circuit with the M54545L can be easily optimized for lower power consumption.

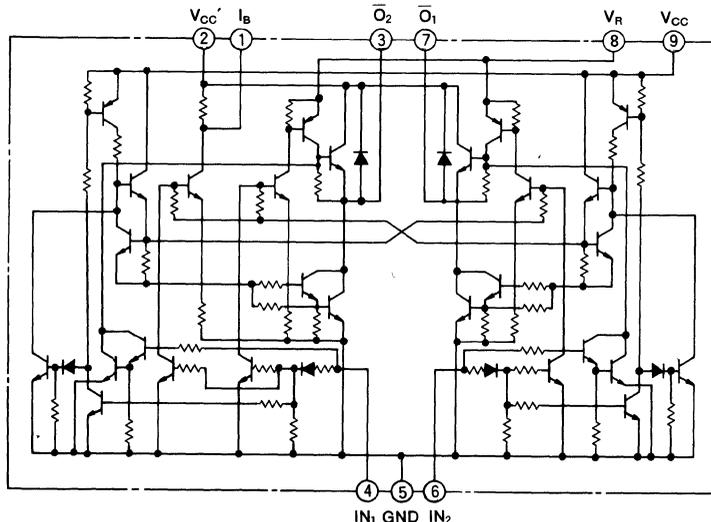
PIN CONFIGURATION (TOP VIEW)



LOGIC TRUTH TABLE

Input		Output			Note
IN ₁	IN ₂	\bar{O}_1	\bar{O}_2	I _B	
L	L	"OFF" state	"OFF" state	H	Off
H	L	H	L	H	○
L	H	L	H	L	○
H	H	L	L	H	Braking

CIRCUIT SCHEMATIC



BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		-0.5~+16	V
V _R	Predriver supply voltage		-0.5~+16	V
V _{CC'}	Driver supply voltage		-0.5~+16	V
V _I	Input voltage		0~V _{CC}	V
V _O	Output voltage		-0.5~V _{CC'+2.5}	V
I _{O(max)}	Peak output current	t _{op} =10ms V _{CC} ≥5V; Repetitive cycle 0.2Hz max	±1.2	A
I _O	Continuous output current		±330	mA
P _D	Power dissipation	T _a =75°C	1.15	W
T _{opr}	Operating ambient temperature range		-10~+75	°C
T _{stg}	Storage temperature range		-55~+125	°C

RECOMMENDED OPERATING CONDITIONS (T_a=25°C, unless otherwise noted)

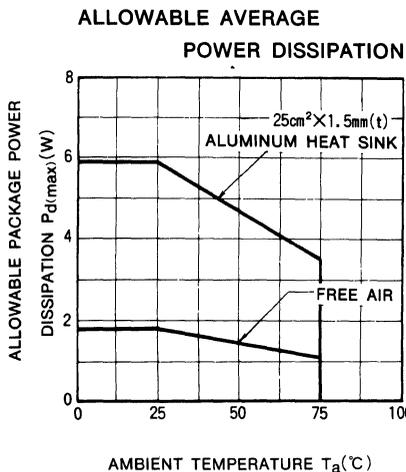
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage		3	12	15	V
I _O	Continuous output current				±200	mA
V _{IH}	"H" Input voltage		3		V _{CC}	V
V _{IL}	"L" Input voltage		0		0.4	V
t _B	Motor braking interval		100			ms

ELECTRICAL CHARACTERISTICS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit	
			Min	Typ	Max		
I _{O(leak)}	Output leakage current	V _{CC} =V _R =V _{CC'} =16V			100	μA	
		V _{I1} =V _{I2} =0V	V _O =16V		-100		
V _{OH(1)}	"H" Output saturation voltage (1)	V _{CC} =V _R =V _{CC'} =12V	V _{I1} =3V, V _{I2} =0V	10.8		V	
V _{OH(2)}	"H" Output saturation voltage (2)	I _{OH} =-200mA	V _{I1} =0V, V _{I2} =3V	10.8		V	
V _{OL(1)}	"L" Output saturation voltage (1)	V _{CC} =V _R =V _{CC'} =12V	V _{I1} =0V, V _{I2} =3V			0.4	V
		I _{OL(1)} =200mA	V _{I1} =3V, V _{I2} =3V				
V _{OL(2)}	"L" Output saturation voltage (2)	V _{CC} =V _R =V _{CC'} =12V	V _{I1} =3V, V _{I2} =0V			0.4	V
		I _{OL(2)} =200mA	V _{I1} =3V, V _{I2} =3V				
I _{IH(1)}	"H" Input current (1)	V _{CC} =V _R =V _{CC'} =12V	V _{I1} =3V, V _{I2} =0V			700	μA
I _{IH(2)}	"H" Input current (2)	V _{CC} =V _R =V _{CC'} =12V	V _{I1} =0V, V _{I2} =3V			700	μA
I _{CC}	Supply current	V _{CC} =V _R =V _{CC'} =12V	V _{I1} =0V, V _{I2} =0V			5	mA
			V _{I1} =3V, V _{I2} =0V			10	
			V _{I1} =0V, V _{I2} =3V				
I _B	I _B Output current	V _{CC} =V _R =V _{CC'} =12V, V _{I8} =12V, V _{I1} =0V, V _{I2} =3V		1.0		15.0	mA

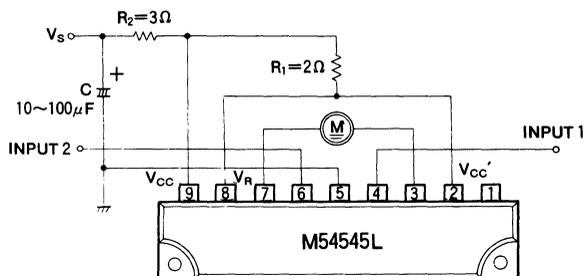
BI-DIRECTIONAL MOTOR DRIVER

TYPICAL CHARACTERISTICS

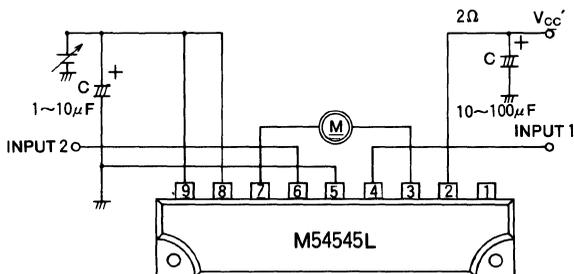


APPLICATION EXAMPLES

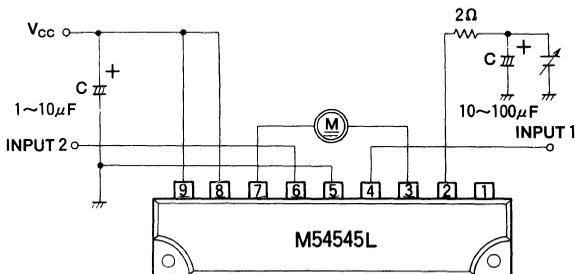
1) Motor speed control by V_{CC} and V_{CC}'



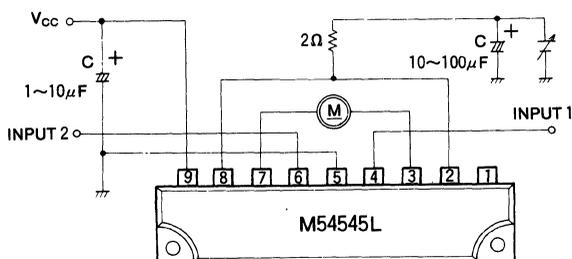
2) Motor speed control by the V_R and V_{CC}



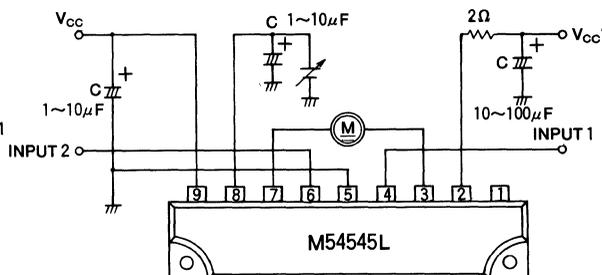
3) Motor speed control by the V_{CC}'



4) Motor speed control by the V_R and V_{CC}'



5) Motor speed control by the V_R



M54546L

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54546L, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for D-C motor control.

FEATURES

- Wide operating voltage range ($V_{CC} = 4 \sim 16V$)
- TTL, PMOS and CMOS compatible input
- Low output saturation voltage
- Integral diodes for transient suppression
- Small single-in-line package
- Braking mode input

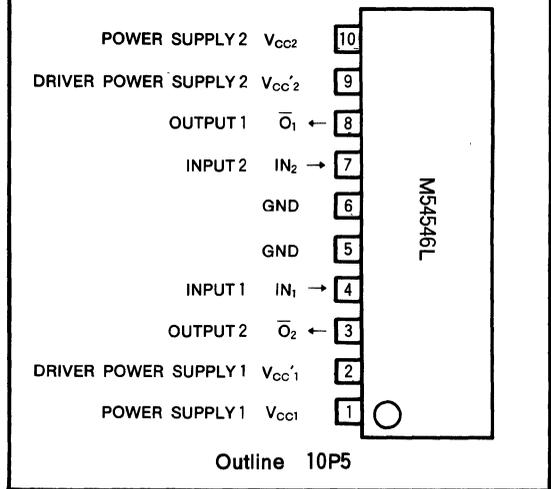
APPLICATION

Audio, video cassette recorder

FUNCTION

The M54546L, full bridge motor driver, has the logic circuitry and non-darlington power drivers for bidirectional control of D-C motors operating at currents up to 700mA. A braking mode by switching the both inputs high may make easier to control the motor. The both of the separated power supplies for the logic circuitry and the drivers are usable for motor speed control. The power supply of the predriver is connected with the driver power supply to have a wider control range of motor supply voltage.

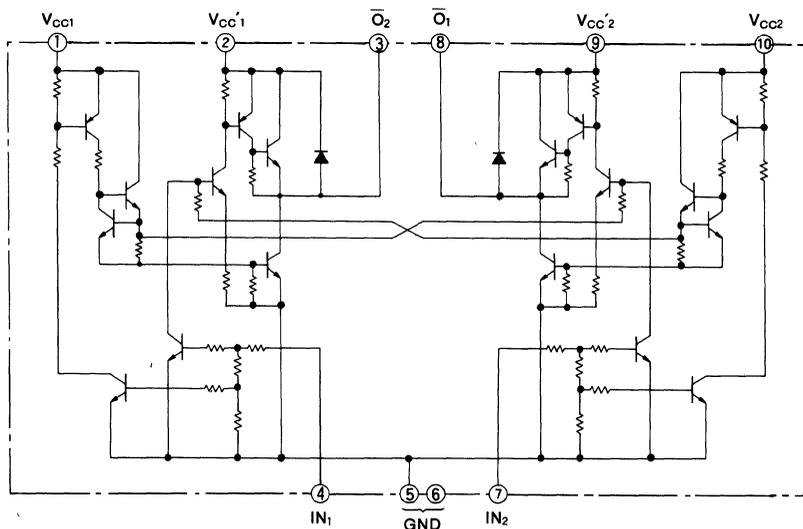
PIN CONFIGURATION (TOP VIEW)



LOGIC TRUTH TABLE

INPUT		OUTPUT		NOTE
IN ₁	IN ₂	O ₁	O ₂	
L	L	"OFF" state	"OFF" state	Open
H	L	H	L	○
L	H	L	H	○
H	H	L	L	Braking

CIRCUIT SCHEMATIC



BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		-0.5~+16	V
V _{CC'}	Driver supply voltage		-0.5~+16	V
V _I	Input voltage		0~V _{CC}	V
V _O	Output voltage		-0.5~V _{CC'} +2.5	V
I _{O(max)}	Peak output current	t _{op} =10ms Repetitive cycle 0.2Hz max	±700	mA
I _O	Continuous output current		±150	mA
P _d	Power dissipation	T _a =75°C	600	mW
T _{opr}	Operating ambient temperature range		-10~+75	°C
T _{stg}	Storage temperature range		-55~+125	°C

RECOMMENDED OPERATING CONDITIONS (T_a=25°C, unless otherwise noted)

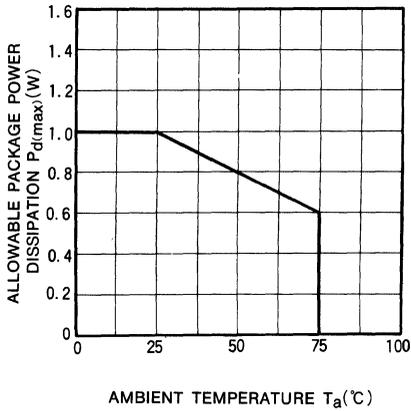
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage		4	12	15	V
I _O	Continuous output current				±300	mA
V _{IH}	"H" Input voltage		2		V _{CC}	V
V _{IL}	"L" Input voltage		0		0.4	V
t _B	Motor braking interval		100			ms

ELECTRICAL CHARACTERISTICS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I _{O(leak)}	Output leakage current	V _{CC} =V _{CC'} =20V V _{I1} =V _{I2} =2V			100 -100	μA
V _{OH(1)}	"H" Output saturation voltage (1)	V _{CC} =V _{CC'} =12V V _{I1} =2V V _{I2} =0V			11.0 10.9	V
V _{OH(2)}	"H" Output saturation voltage (2)	V _{CC} =V _{CC'} =12V V _{I1} =0V V _{I2} =2V			11.0 10.9	V
V _{OL(1)}	"L" Output saturation voltage (1)	V _{CC} =V _{CC'} =12V V _{I1} =0V V _{I2} =2V V _{I1} =V _{I2} =2V			0.3 0.35 0.35	V
V _{OL(2)}	"L" Output saturation voltage (2)	V _{CC} =V _{CC'} =12V V _{I1} =2V V _{I2} =0V V _{I1} =V _{I2} =2V			0.3 0.35 0.35	V
I _{IH(1)}	"H" Input current (1)	V _{CC} =V _{CC'} =12V, V _{I1} =2V, V _{I2} =0V	70		200	μA
I _{IH(2)}	"H" Input current (2)	V _{CC} =V _{CC'} =12V, V _{I1} =0V, V _{I2} =2V	70		200	μA
I _{CC}	Supply current	V _{CC} =V _{CC'} =16V			30 60 0	mA
		V _{I1} =2V, V _{I2} =0V				
		V _{I1} =0V, V _{I2} =2V				
		V _{I1} =V _{I2} =2V				
		V _{I1} =V _{I2} =0V				

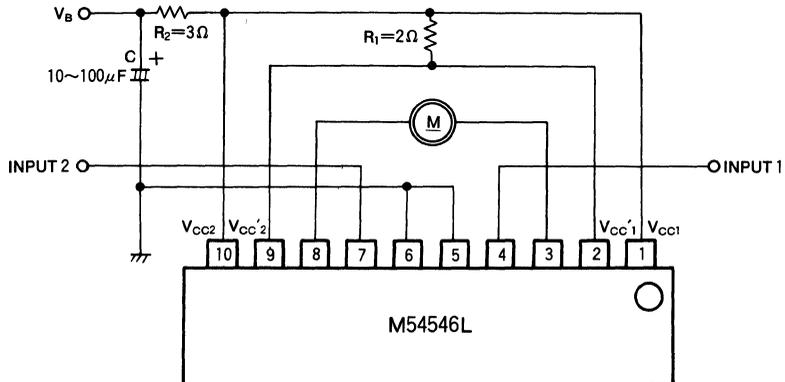
TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE
POWER DISSIPATION

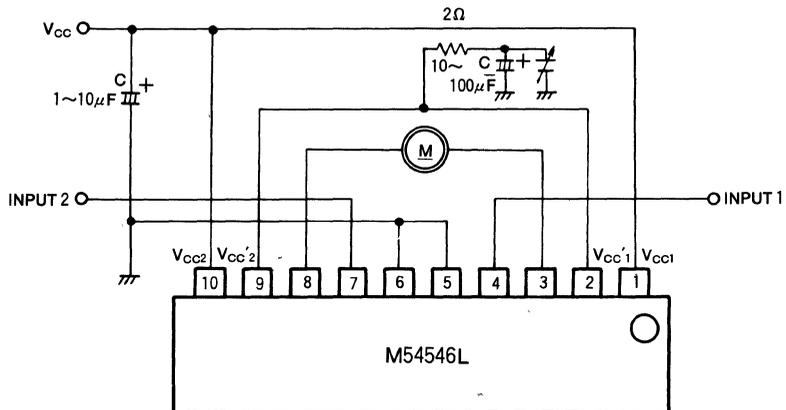


APPLICATION EXAMPLES

1) Motor speed control by V_{CC} and V_{CC}'



2) Motor speed control by the V_{CC}'



M54547P

BI-DIRECTIONAL MOTOR DRIVER WITH OP AMP AND TRANSISTOR ARRAY

DESCRIPTION

The M54547P, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver and dual general purpose NPN darlington pairs.

FEATURES

- 600mA output current
- Braking mode input
- Integral operational amplifier at direction control input
- Output transient suppression

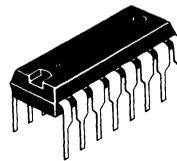
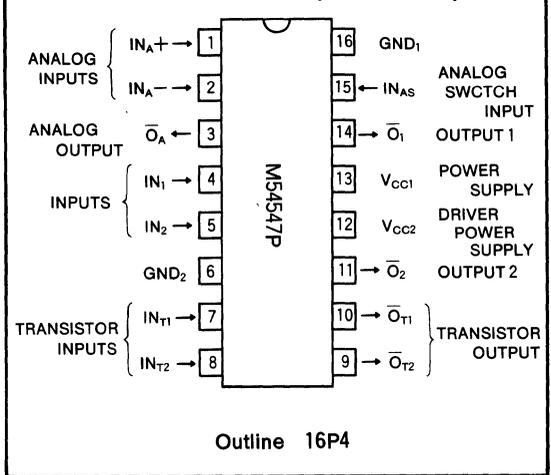
APPLICATION

Audio, video cassette recorder

FUNCTION

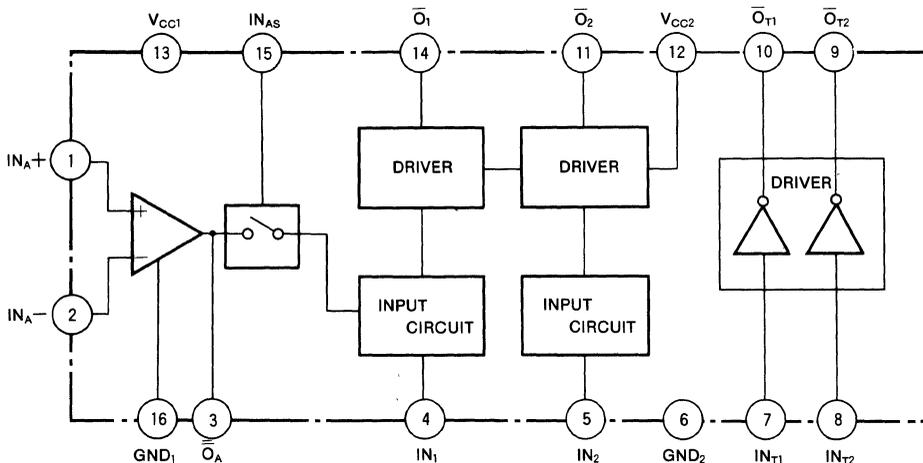
The M54547P, full bridge motor driver, has the logic circuitry and darlington power drivers for bidirectional control of D-C motors operating at currents up to 600mA. The operational amplifier is connected to the direction control input through an analog switch controlled by pin 15 input. By switching the IN_{AS} input high and the IN_1 input low, the output of the amplifier appears at the output O_1 so that the voltage across the bridge output is altered linearly by the amplifier input. The internal NPN darlington pairs are capable of sinking 300mA and will withstand 20V in the OFF state.

PIN CONFIGURATION (TOP VIEW)



16-pin molded plastic DIP

BLOCK DIAGRAM



BI-DIRECTIONAL MOTOR DRIVER WITH OP AMP AND TRANSISTOR ARRAY

LOGIC TRUTH TABLE

Input			Output		Note
IN _{SW}	IN ₁	IN ₂	O ₁	O ₂	
L	L	L	H	H	Braking
L	L	H	H	L	○
L	H	L	L	H	○
L	H	H	L	L	Braking
H	L	L	A*	H	Analog ○
H	L	H	A*	L	Analog ○
H	H	L	L	H	○
H	H	H	L	L	Braking

A* : The output voltage is controlled by the amplifier output.

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC1}	Supply voltage		-0.5~+16	V
V _{CC2}	Driver supply voltage		-0.5~+16	V
V _I , V _{IAS}	Input voltage		0~V _{CC}	V
V _O	Output voltage		-0.5~V _{CC2} +2.5V	V
I _{OP}	Peak output current	t _{OP} =10ms : Repetitive cycle 0.2Hz max	±600	mA
I _O	Continuous output current		±150	mA
V _{CEO}	Collector-emitter applied voltage(transistor array)		20	V
I _C	Collector current(transistor array)		300	mA
V _I	Input voltage(Transistor array)		10	V
P _d	Power dissipation	T _a =25°C	1.47	W
		T _a =60°C	1.06	
T _{OPR}	Operating ambient temperature range		-10~+60	°C
T _{STG}	Storage temperature range		-55~+125	°C

RECOMMENDED OPERATING CONDITIONS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V _{CC1}	Supply voltage		4	12	15	V
I _O	Continuous output current				±100	mA
V _{IH}	Input voltage(motor driver)		3		V _{CC}	V
V _{IL}	(IN ₁ , IN ₂ , IN _{AS})		0		0.6	V
t _B	Motor braking interval		100			ms
V _{IH}	Transistor array input voltage		4		10	V
V _{IL}	(IN _{T1} , IN _{T2})		0		0.6	V

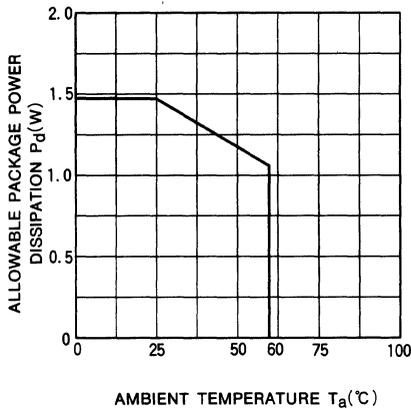
BI-DIRECTIONAL MOTOR DRIVER WITH OP AMP AND TRANSISTOR ARRAY

ELECTRICAL CHARACTERISTICS (T_a=25°C, unless otherwise noted)

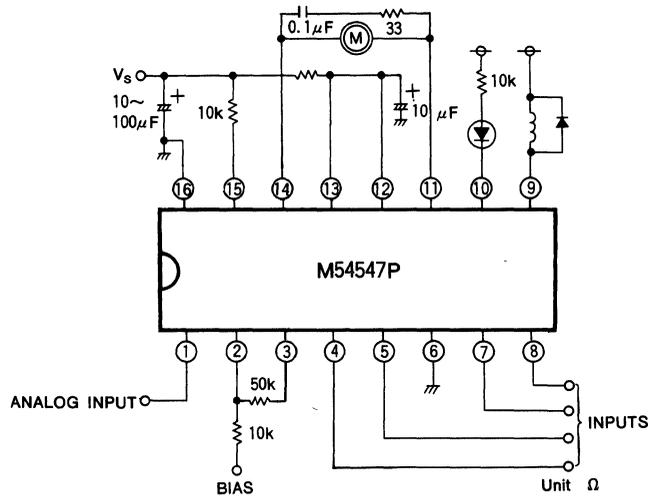
Symbol	Parameter	Test conditions	Limits			Unit	
			Min	Typ	Max		
I _{o(Leak)}	Output leakage current(\bar{O}_1, \bar{O}_2)	V _{CC1} =V _{CC2} =16V V _{IN1} =V _{IN2} =V _{IAS} =0V	V _O =0V V _O =16V			±100	μA
V _{OH}	"H" Output saturation voltage(\bar{O}_1, \bar{O}_2)	V _{CC1} =V _{CC2} =12V I _O =-150mA	V _{IN1} =0V, V _{IN2} =3V V _{IN1} =3V, V _{IN2} =0V	10.3			V
V _{OL}	"L" Output saturation voltage(\bar{O}_1, \bar{O}_2)	V _{CC1} =V _{CC2} =12V I _O =150mA	V _{IN1} =0V, V _{IN2} =3V V _{IN1} =3V, V _{IN2} =0V			1.2	V
I _I	Input current(IN ₁ , IN ₂ , I _{NAS})	V _{CC1} =12V, V _I =3V				0.3	mA
I _{o(Leak)}	Output leakage current(\bar{O}_1, \bar{O}_2)	V _O =30V, V _I =0.6V				100	μA
V _{OC}	"L" Output saturation voltage	V _I =4V	I _C =100mA I _C =200mA			1.3 1.5	V
I _I	Input current	V _I =4V				0.8	mA
A _O	OP Amp open-loop-gain			40			dB
I _{CC1}	Supply current	V _{CC1} =12V, V _{IN1} =V _{IN2} =V _{IAS} =3V				6	mA

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION



APPLICATION EXAMPLE



M54548L

BI-DIRECTIONAL MOTOR DRIVER WITH MOTOR SPEED CONTROL

DESCRIPTION

The M54548L, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for use in a D-C motor control circuit. The internal operational amplifier is capable for controlling the voltage across the bridge outputs.

FEATURES

- Wide operating voltage range
- NMOS and CMOS compatible input
- 1.2A output current
- Integral operational amplifier for output source voltage
- Output transient suppression
- Braking mode input

APPLICATION

Audio, video cassette recorder

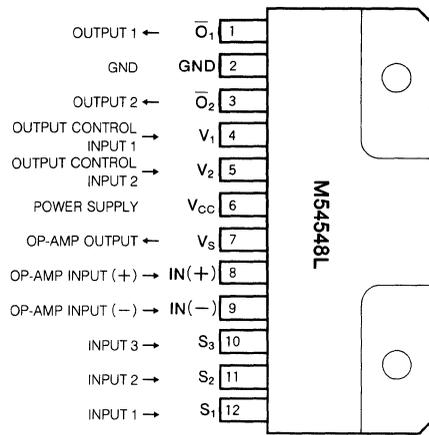
FUNCTION

The M54548L, full bridge motor driver, has the logic circuitry and the quasi-darlington power driver for bidirectional control of D-C motors operating at current up to 1.2A. The inputs, S_1 , S_2 and S_3 , are capable to control the bridge output polarity and also to select the supply voltage of the predriver from the voltages driven by V_1 , V_2 or the output of the operational amplifier.

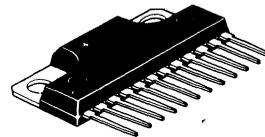
LOGIC TRUTH TABLE

Input			Output		Driver power supply	Note
S_1	S_2	S_3	\overline{O}_1	\overline{O}_2		
L	L	L	"OFF" state	"OFF" state	—	STOP
L	L	H	H	L	OP AMP OUTPUT	PLAY(+)
L	H	L	L	H	OP AMP OUTPUT	PLAY(-)
L	H	H	H	L	V_2	FF(2)
H	L	L	L	H	V_2	REW(2)
H	L	H	H	L	V_1	FF(1)
H	H	L	L	H	V_1	REW(1)
H	H	H	L	L	V_s	BRAKING

PIN CONFIGURATION (TOP VIEW)

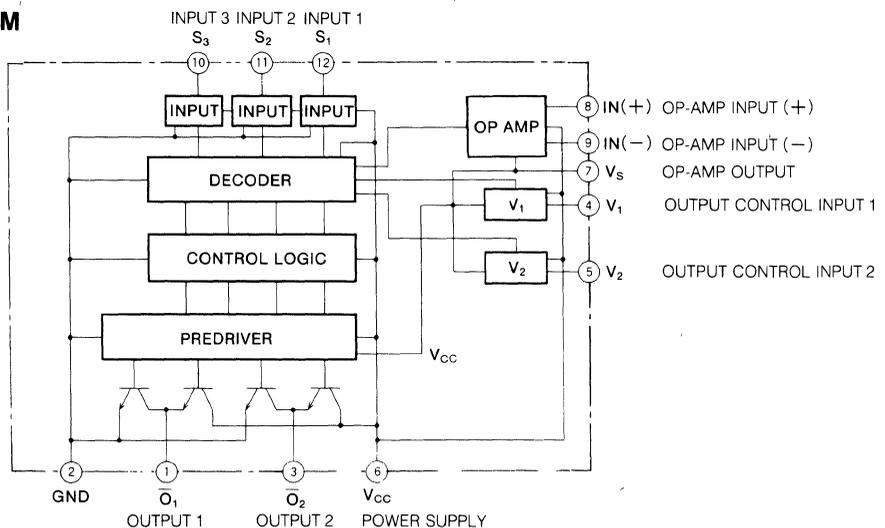


Outline 12P9



12-pin molded plastic SIP

BLOCK DIAGRAM



BI-DIRECTIONAL MOTOR DRIVER WITH MOTOR SPEED CONTROL

ABSOLUTE MAXIMUM RATINGS ($T_a = 25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage	With an external heat sink (3000mm ² ×1.5mm ²)	-0.5~+18	V
V_I	Input voltage	4 Pin, 5 Pin	-0.5~+14 or V_{CC} -0.5~ V_{CC}	V
V_O	Output voltage		-0.5~ $V_{CC}+2.5$	V
$I_{O(max)}$	Peak output current	$t_{op}=10\text{ms}$; Repetitive cycle 0.2Hz max	±1.2	A
$I_{O(1)}$	Continuous output current (1)		±300	mA
$I_{O(2)}$	Continuous output current (2)	With an external heat sink (3000mm ² ×1.5mm ²)	±600	mA
P_d	Power dissipation	$T_a = 75^\circ\text{C}$	1.2	W
T_{opr}	Operating ambient temperature range		-10~+75	°C
T_{stg}	Storage temperature range		-55~+125	°C

RECOMMENDED OPERATING CONDITIONS ($T_a = 25^\circ\text{C}$, unless otherwise noted)

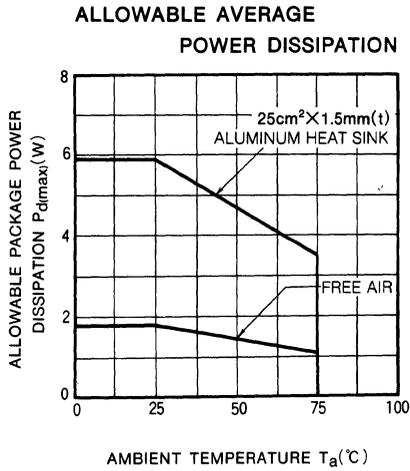
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	12	16	V
I_O	Continuous output current				±200	mA
V_{IH}	"H" Input voltage		3			V
V_{IL}	"L" Input voltage				1	V
t_s	Motor braking interval		100			ms

ELECTRICAL CHARACTERISTICS ($T_a = 25^\circ\text{C}$, unless otherwise noted)

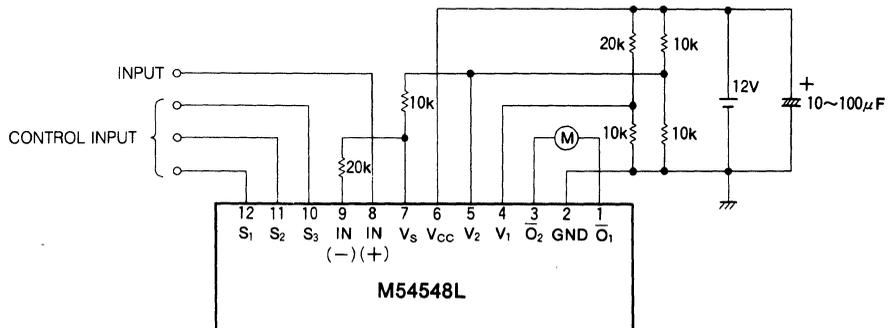
Symbol	Parameter	Test conditions		Limits			Unit
				Min	Typ	Max	
$I_{O(leak)}$	Output leakage current	$V_{S1}=0\text{V}$ $V_{S2}=0\text{V}$ $V_{S3}=0\text{V}$	$V_O=0\text{V}$ $V_{CC}=V_S=20\text{V}$ $V_O=14\text{V}$ $V_{CC}=V_S=14\text{V}$			-100 +100	μA
$V_{OH(1)}$	"H" Output saturation voltage (1)	$V_{CC}=16\text{V}$ $V_{IN(-)}=0\text{V}$ $V_{IN(+)}=3\text{V}$	$V_{S1}=V_{S2}=0\text{V}$ $V_{S3}=3\text{V}$	$I_{OH}=-200\text{mA}$ $I_{OH}=-500\text{mA}$	13 12.8		V
$V_{OH(2)}$	"H" Output saturation voltage (2)	$V_{CC}=16\text{V}$ $V_{IN(-)}=0\text{V}$ $V_{IN(+)}=3\text{V}$	$V_{S1}=V_{S3}=0\text{V}$ $V_{S2}=3\text{V}$	$I_{OH}=-200\text{mA}$ $I_{OH}=-500\text{mA}$	13 12.8		V
$V_{OL(1)}$	"L" Output saturation voltage (1)	$V_{CC}=16\text{V}$ $V_{IN(-)}=0\text{V}$ $V_{IN(+)}=3\text{V}$	$V_{S1}=V_{S3}=0\text{V}$ $V_{S2}=3\text{V}$	$I_{OL}=200\text{mA}$ $I_{OL}=500\text{mA}$		0.5 1.4	V
$V_{OL(2)}$	"L" Output saturation voltage (2)	$V_{CC}=16\text{V}$ $V_{IN(-)}=0\text{V}$ $V_{IN(+)}=3\text{V}$	$V_{S1}=V_{S2}=0\text{V}$ $V_{S3}=3\text{V}$	$I_{OL}=200\text{mA}$ $I_{OL}=500\text{mA}$		0.5 1.4	V
I_{IH}	"H" Input current	$V_{CC}=16\text{V}$, $V_{IS}=3\text{V}$ (S_1, S_2, S_3)				10	μA
I_{IL}	"L" Input current	$V_{CC}=16\text{V}$, $V_{IS}=0\text{V}$ (S_1, S_2, S_3)				-20	μA
I_{CC}	Supply current	$V_{CC}=16\text{V}$, $V_{S1}=V_{S2}=V_{S3}=3\text{V}$				25	mA
A	Op amp open-loop-gain				50		dB

BI-DIRECTIONAL MOTOR DRIVER WITH MOTOR SPEED CONTROL

TYPICAL CHARACTERISTICS



APPLICATION EXAMPLE



Unit Ω

M54549L

DUAL BI-DIRECTIONAL MOTOR DRIVER WITH THERMAL SHUT DOWN FUNCTION

DESCRIPTION

The M54549L, BI-DIRECTIONAL MOTOR DRIVER, consists of the two full bridge power designed for use in are two D-C motors control circuit.

FEATURES

- Two separated full bridge driver (only one circuit can be switched by the SE input)
- Wide operating voltage range ($V_{CC}=4\sim 16V$)
- TTL, PMOS, CMOS outputs, capable of direct drive
- Low output saturation voltage
- Built-in clamp diode
- Large output drive current ($I_{O(max)}=\pm 2A$)
- Braking mode input
- Internal thermal shutdown protection

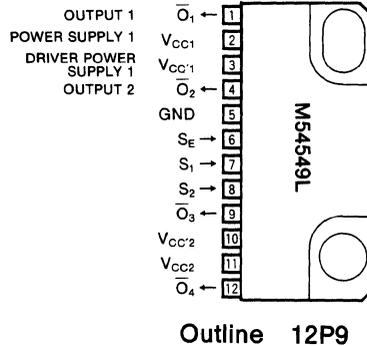
APPLICATION

- Audio tape-deck player, radio/cassette player
- VTR
- Home-use equipment

FUNCTION

The M54549L, two-full bridge motor driver, has the logic circuitry and the darlington power driver for bi-directional control of two D-C motors operating at current up to 2.0A. The input SE selects one of the bridges and S_1, S_2 determine the output polarity.

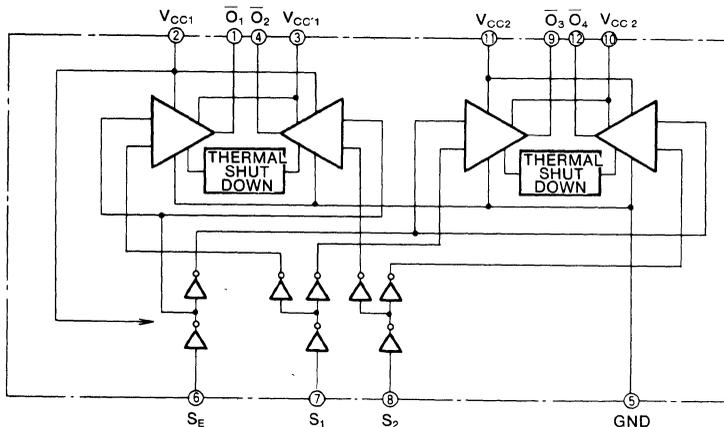
PIN CONFIGURATION (TOP VIEW)



LOGIC TRUTH TABLE

Input			Output				Note	
S_E	S_1	S_2	\bar{O}_1	\bar{O}_2	\bar{O}_3	\bar{O}_4	Output \bar{O}_1, \bar{O}_2	Output \bar{O}_3, \bar{O}_4
0	0	0	OFF	OFF	OFF	OFF	Open	Open
0	1	0	1	0	OFF	OFF	⊙	Open
0	0	1	0	1	OFF	OFF	⊙	Open
0	1	1	0	0	OFF	OFF	Braking	Open
1	0	0	OFF	OFF	OFF	OFF	Open	Open
1	1	0	OFF	OFF	1	0	Open	⊙
1	0	1	OFF	OFF	0	1	Open	⊙
1	1	1	OFF	OFF	0	0	Open	Braking

BLOCK DIAGRAM



DUAL BI-DIRECTIONAL MOTOR DRIVER WITH THERMAL SHUT DOWN FUNCTION

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
$V_{CC(1)}$	Supply voltage (1)		$-0.5 \sim +18$	V
$V_{CC(2)}$	Supply voltage (2)	With an external sink (3000mm ² ×1.5mm)	$-0.5 \sim +18$	V
$V_{CC'}$	Driver supply voltage		$-0.5 \sim +18$	V
V_I	Input voltage		$0 \sim V_{CC}$	V
V_O	Output voltage		$-2 \sim V_{CC} + 2.5$	V
$I_{O(max)}$	Peak output current	$t_{op}=10\text{ms}$, retitive cycle 0.2Hz max	± 2.0	A
$I_{O(1)}$	Continuous output current (1)		± 330	mA
$I_{O(2)}$	Continuous output current (2)	With an external sink (3000mm ² ×1.5mm)	± 600	mA
P_d	Power dissipation	$T_a=75^\circ\text{C}$	1.20	W
T_{opr}	Operating ambient temperature range		$-10 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

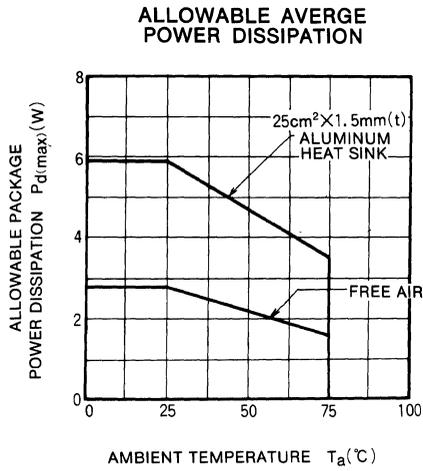
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	12	16	V
I_O	Output current				± 300	mA
V_{IH}	"H" Input voltage	Input S ₁ , S ₂ , S _E	2		V_{CC}	V
V_{IL}	"L" Input voltage	Input S ₁ , S ₂ , S _E	0		0.4	V
t_s	Motor braking interval		10	100		ms
$t_j(\text{shut})$	Thermal shut down temperature	$V_{CC} \geq 7V$		150		$^\circ\text{C}$

ELCTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

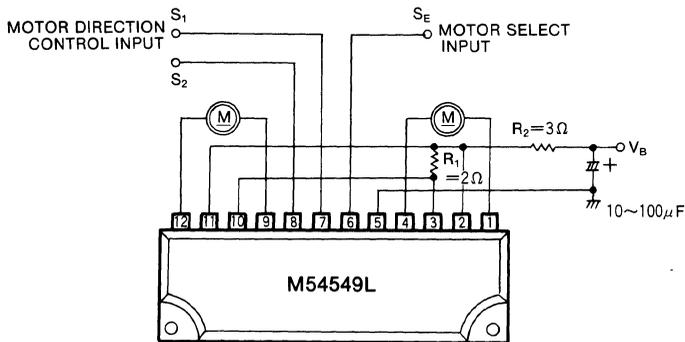
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{O(\text{leak})}$	Output leakage	$V_{CC}=V_{CC'}=18V$ $V_{S1}=V_{S2}=0V$ $V_{SE}=0V$ or $2V$	$V_O=18V$		100	μA
			$V_O=0V$		-10	
V_{OH}	"H" output voltage	$V_{CC}=V_{CC'}=12V$	$I_{OH(1)}=-200\text{mA}$	10.8		V
			$I_{OH(1)}=-500\text{mA}$	10.7		
V_{OL}	"L" output voltage	$V_{CC}=V_{CC'}=12V$	$I_{OL}=200\text{mA}$		0.5	V
			$I_{OL(1)}=500\text{mA}$		1.35	
I_{IH}	"H" input current	$V_{CC}=V_{CC'}=12V$, $V_I=2V$	70		200	μA
I_{IL}	"L" input current	$V_{CC}=V_{CC'}=12V$, $V_I=0V$	70		200	μA
I_{CC}	Supply current	$V_{CC}=V_{CC'}=12V$	$V_{SE}=0V$, $V_{S1}=V_{S2}=0V$		10	mA
			$V_{SE}=0V$, $V_{S1}=V_{S2}=0V$			
			$V_{SE}=0V$, $V_{S1}=0V$, $V_{S2}=2V$		20	

DUAL BI-DIRECTIONAL MOTOR DRIVER WITH THERMAL SHUT DOWN FUNCTION

TYPICAL CHARACTERISTICS



APPLICATION EXAMPLE



M54549AL

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54549AL, BI-DIRECTIONAL MOTOR DRIVER, consists of the two full bridge power designed for use in are two D-C motors control circuit.

FEATURES

- Two separated full-bridge drivers (only one circuit can be switched by the SE input)
- Wide operating voltage range ($V_{CC}=4\sim 16V$)
- TTL, PMOS, CMOS outputs, capable of direct drive
- Low output saturation voltage
- Built-in clamp diode
- Large drive current ($I_{O(max)}=\pm 1.2A$)
- Braking mode input
- Internal thermal shutdown protection

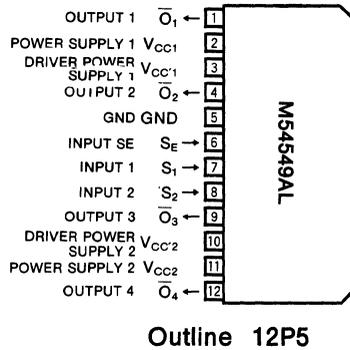
APPLICATION

- Audio tape-deck player, radio cassette player
- VTR
- Home-use equipment

FUNCTION

The M54549AL, two-full bridge motor driver, has the logic circuitry and the darlington power driver for bi-directional control of two D-C motors operating at current up to 2.0A. The input SE selects one of the bridges and S_1, S_2 determine the output polarity.

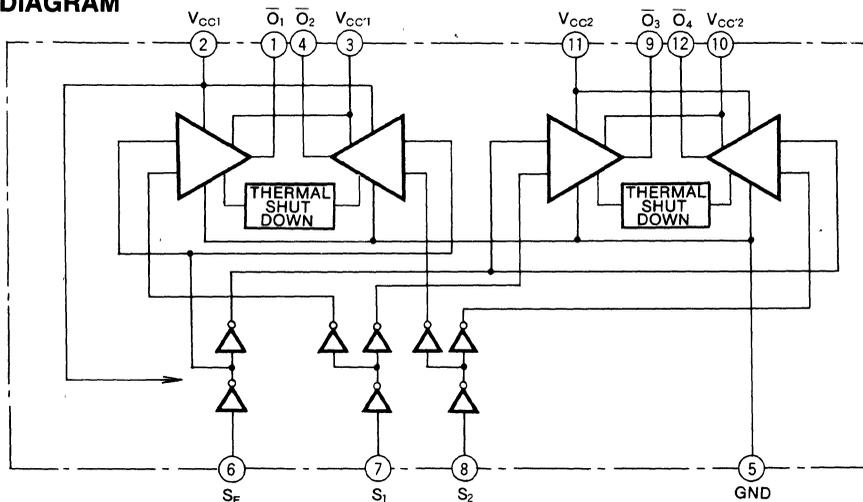
PIN CONFIGURATION (TOP VIEW)



LOGIC TRUTH TABLE

Input			Output				Note	
S_E	S_1	S_2	O_1	O_2	O_3	O_4	Output O_1, O_2	Output O_3, O_4
0	0	0	OFF	OFF	OFF	OFF	Open	Open
0	1	0	1	0	OFF	OFF	⊙	Open
0	0	1	0	1	OFF	OFF	⊙	Open
0	1	1	0	0	OFF	OFF	Braking	Open
1	0	0	OFF	OFF	OFF	OFF	Open	Open
1	1	0	OFF	OFF	1	0	Open	⊙
1	0	1	OFF	OFF	0	1	Open	⊙
1	1	1	OFF	OFF	0	0	Open	Braking

BLOCK DIAGRAM



BI-DIRECTIONAL MOTER DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		-0.5~+18	V
$V_{CC'}$	Driver supply voltage		-0.5~+18	V
V_I	Input voltage		0~ V_{CC}	V
V_O	Output voltage		-2~ $V_{CC}+2.5$	V
$I_{O(max)}$	Peak output current	$t_{op}=10\text{ms}$, retitive cycle 0.2Hz max	± 1.2	A
$I_{O(1)}$	Continuous output current (1)		± 330	mA
P_d	Power dissipation	$T_a=75^\circ\text{C}$	830	mW
T_{opr}	Operating ambient temperature		-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

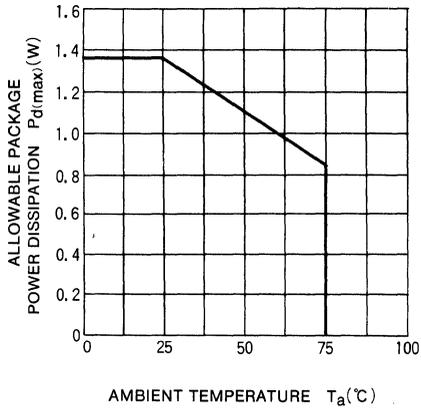
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	12	16	V
I_O	Output current				± 300	mA
V_{IH}	"H" input voltage		2		V_{CC}	V
V_{IL}	"L" input voltage		0		0.4	V
t_B	Motor braking interval		100			ms
$t_{j(shut)}$	Thermal shutdown temperature			150		$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

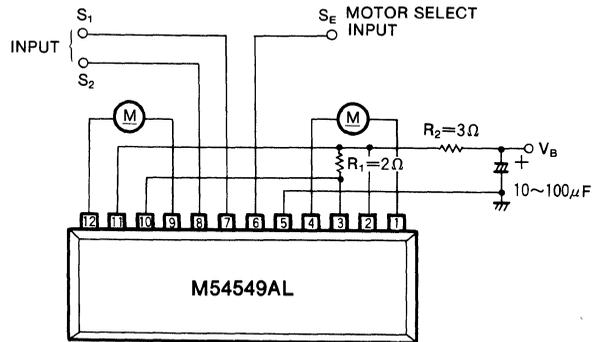
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{O(leak)}$	Output leakage current	$V_{CC}=V_{CC'}=18\text{V}$ $V_{S1}=V_{S2}=0.4\text{V}$, $V_{SE}=0.4\text{V}$ or 2V $V_O=18\text{V}$ $V_O=0\text{V}$			100 -100	μA
V_{OH}	"H" output saturation voltage	$V_{CC}=V_{CC'}=12\text{V}$ $I_{OH(1)}=-200\text{mA}$ $I_{OH(1)}=-500\text{mA}$	10.8 10.7			V
V_{OL}	"L" output saturation voltage	$V_{CC}=V_{CC'}=12\text{V}$ $I_{OL}=200\text{mA}$ $I_{OL(1)}=500\text{mA}$			0.5 1.35	V
I_{IH}	"H" input current	$V_{CC}=V_{CC'}=12\text{V}$, $V_I=2\text{V}$	50		120	μA
I_{CC1}	Supply current (1)	$V_{CC}=V_{CC'}=12\text{V}$ $V_{SE0}=V_{S1}=V_{S2}=0.4\text{V}$ $V_{SE}=V_{S1}=0.4\text{V}$, $V_{S2}=2\text{V}$			10 20	mA
I_{CC2}	Supply current (2)	$V_{CC}=V_{CC'}=12\text{V}$ $V_{SE}=2\text{V}$, $V_{S1}=V_{S2}=0.4\text{V}$ $V_{SE}=V_{S1}=2\text{V}$, $V_{S2}=2\text{V}$			10 20	mA

TYPICAL CHARACTERISTICS

**ALLOWABLE AVERAGE
 POWER DISSIPATION**



APPLICATION EXAMPLE



M54641L

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54641L, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power driver designed for D-C motor control.

FEATURES

- Wide operating voltage range ($V_{CC}=4\sim 10V$, $V_{CC}(\max)=20V$)
- Low output saturation voltage in continuous output motor circuit (High voltage between motors)
- Built-in clamp diode
- Output voltage control pin (V_z)
- Internal thermal shutdown protector ($T_{j(\text{shut})}=120^{\circ}\text{C}_{\text{TYP}}$)

APPLICATION

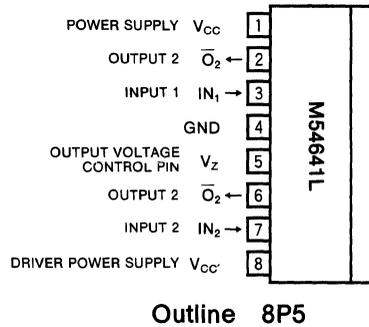
- Audio tape deck, radio cassette player, VTR

FUNCTION

The M54641L, full bridge motor driver, has the logic circuitry and darlington-per power drivers for bidirectional control of D-C motors operating at currents up to 800mA.

The power supplies for the logic circuitry and the drivers are separated so that the applied voltage to the motor can be controlled by the V_z or V_{CC} of the driver power supply voltage.

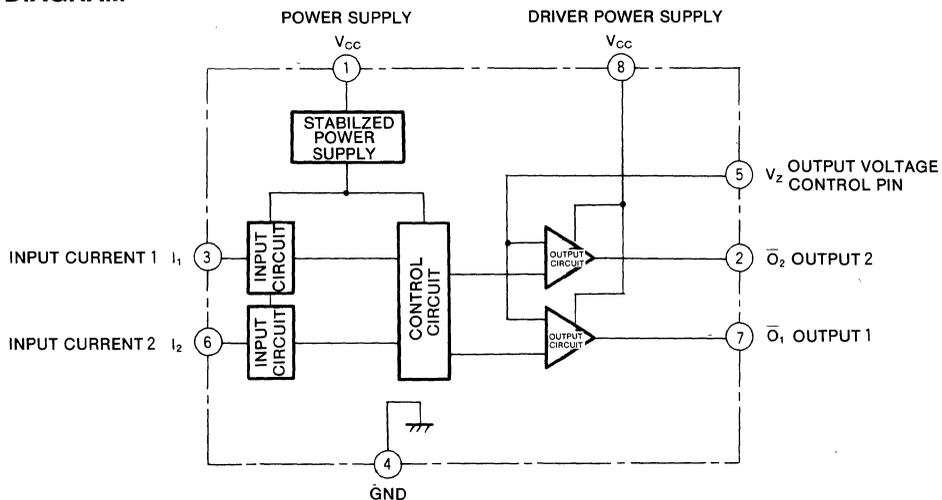
PIN CONFIGURATION (TOP VIEW)



LOGIC TRUTH TABLE

Input		Output		Note
IN_1	IN_2	\bar{O}_1	\bar{O}_2	
L	L	"OFF" state	"OFF" state	Open
H	L	H	L	Clockwise
L	H	L	H	Anti-clockwise
H	H	L	L	Braking

BLOCK DIAGRAM



BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		$-0.5\sim+12$	V
$V_{CC'}$	Driver power supply		$-0.5\sim+20$	V
V_i	Input voltage		$0\sim V_{CC}$	V
V_o	Output voltage		$-0.5\sim V_{CC}+2.5$	V
$I_o(\text{max})$	Peak output current	$t_{op}=10\text{ms}$ $t_{op}=10\text{ms}$ · repetitive cycle 0.2Hz max	± 800	mA
I_o	Continuous output current		± 150	mA
P_d	Power dissipation	$T_a=60^\circ\text{C}$	570	mW
T_j	Junction temperature		100	$^\circ\text{C}$
T_{opr}	Operating ambient temperature		$-10\sim+60$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55\sim+125$	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

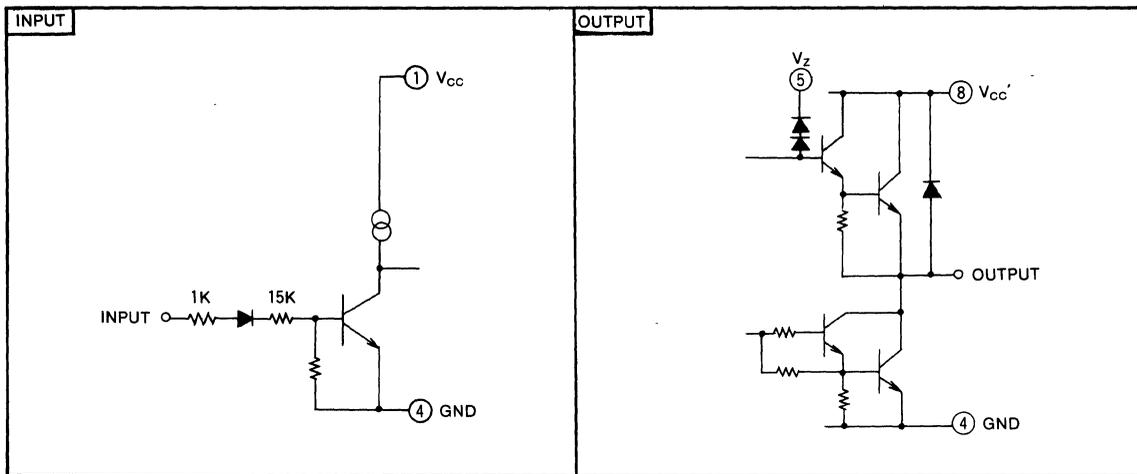
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	5	10	V
I_o	Output current				± 100	mA
V_{IN}	"H" input voltage		3.0		V_{CC}	V
V_{IL}	"L" input voltage		0		0.6	V
t_s	Motor braking interval		10	100		ms
T_s	Thermal shutdown protector operating temperature (Junction temperature)		100	120		$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=5\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions		Limits			Unit
				Min	Typ	Max	
$I_o(\text{leak})$	Output leakage current	$V_{CC}=20\text{V}$ V_z · open	$V_o=20\text{V}$ $V_o=0\text{V}$			100 -100	μA
V_{OH}	"H" output voltage	$V_{CC}=12\text{V}$ V_z · open	$I_{OH}=-50\text{mA}$ $I_{OH}=-100\text{mA}$	10.2 10.0	10.5 10.4		V
V_{OL}	"L" output voltage	$V_{CC}=12\text{V}$ V_z open	$I_{OH}=50\text{mA}$ $I_{OH}=100\text{mA}$		0.1 0.2	0.3 0.4	V
V_{O1-O2}	Voltage between output(1) and output(2) (voltage between motors)	$V_{CC}=12\text{V}$ $V_z=7\text{V}$	$I_o=\pm 100\text{mA}$	6.3	7.0	7.7	V
I_i	Input current	$V_{CC}=12\text{V}$	$V_i=3\text{V}$ $V_i=5\text{V}$			100 240 180 380	μA
I_{CC}	Supply current	$V_{CC}=10\text{V}$ $V_{CC}=12\text{V}$	"OFF" state Clockwise or anti-clockwise Braking		1.2 4.5 7.5	3.0 8.0 12.0	mA

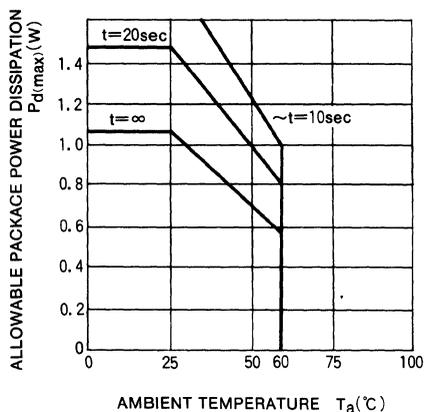
BI-DIRECTIONAL MOTOR DRIVER

INPUT/OUTPUT CIRCUIT



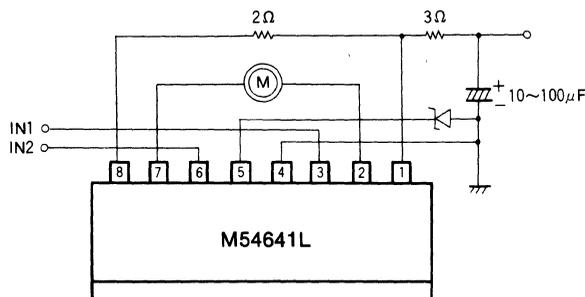
TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION



Note 1 : Mounted on an epoxy PC board with Cu cover on one side (5cm×5cm×0.8mmt)
 Note 2 : t : time for power application

APPLICATION EXAMPLE



M54642L

BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54642L is a semiconductor integrated circuit, capable of directly driving bi-directional micro motor.

FEATURES

- Wide operating voltage range ($V_{CC}=4\sim 10V$, $V_{CC(max)}=20V$)
- Low output saturation voltage in continuous output motor circuit (High voltage between motors)
- Built-in clamp diode
- Output voltage control pin (V_Z)
- Internal thermal shutdown protector ($T_{j(shut)}=120^{\circ}C_{TYP}$)
- Additional interrupt input I_1' , I_2'

APPLICATION

- Audio tape deck, radio cassette player, VTR

FUNCTION

The M54642L is a driver for bi-directional micro motor. The input pin has inputs I_1 and I_2 which are identical to those inputs of the M54641L and interrupt inputs of I_1' and I_2' . The I_1 functions to the I_1' and the I_2 , to the I_2' . The I_1' and the I_2' , if they are "H", operate just like the I_1 and the I_2 input signals in the M54641L. If the "H" signal is input to the input pin I_1 , and the "L" signal is input to, the input pin I_2 , output \bar{O}_1 is "H" and output \bar{O}_2 is "L". If a motor is connected between the output pins of \bar{O}_1 and \bar{O}_2 , the output power supply is current source and the \bar{O}_2 is current sink and the motor is driven. If the reverse signals are input to the I_1 , I_2 , the \bar{O}_1 becomes "L" and the \bar{O}_2 becomes "H", then the motor is driven backward. But, if the I_1 and the I_2 are both "H", and the \bar{O}_1 and the \bar{O}_2 are "L", then the motor is quickly halted. (Braking mode input)

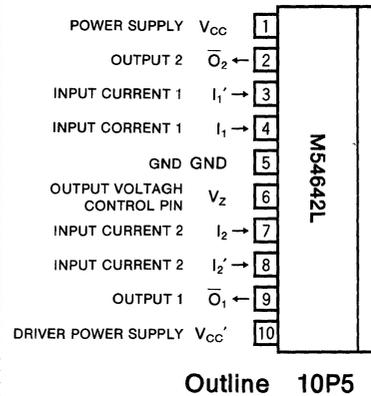
The interrupt input I_1' and I_2' are "L" active inputs and, when the I_1 is "H" and I_2 is "L", the motor is driven, but, if the I_2' is "L", then the outputs \bar{O}_1 and the \bar{O}_2 are both "L" and becomes brake mode.

The speed of motor becomes constant if a zener diode of a certain voltage is added to the V_Z pin, because the output "H" voltage never exceeds this zener voltage. If the V_Z pin is connected to the driver power supply $V_{CC'}$, the speed of motor can be changed by changing the $V_{CC'}$.

The peak output current is $I_{OP(max)}=800mA$, and continuous output current is $I_{O(max)}=150mA$.

Threshold temperature of the internal thermal shutdown protector is min $100^{\circ}C$, and the drive current must be set so that the thermal shutdown protector will not operate.

PIN CONFIGURATION (TOP VIEW)

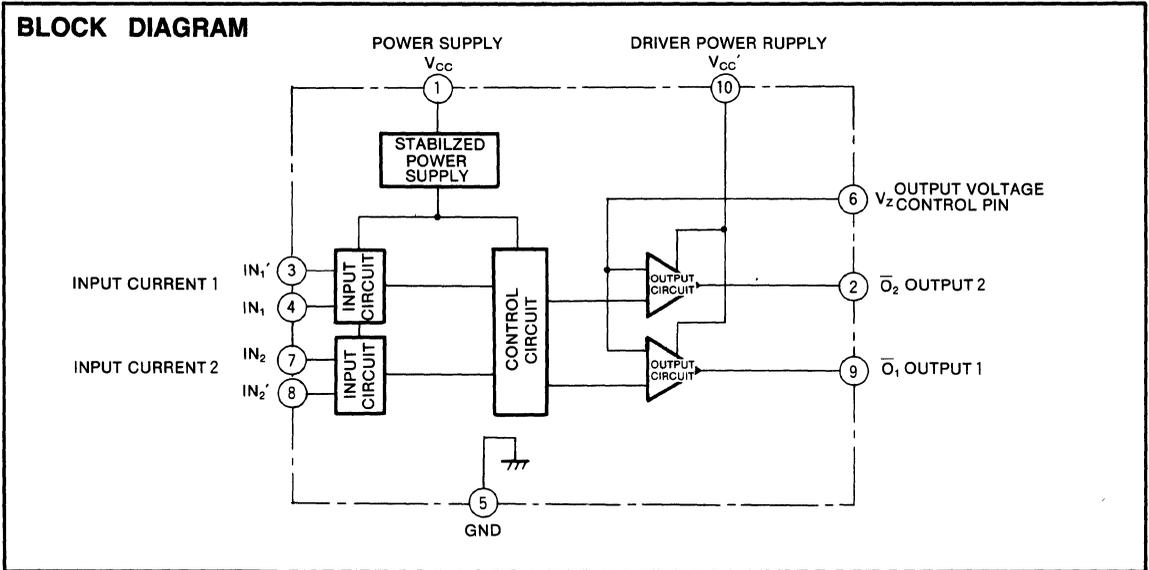


LOGIC TRUTH TABLE

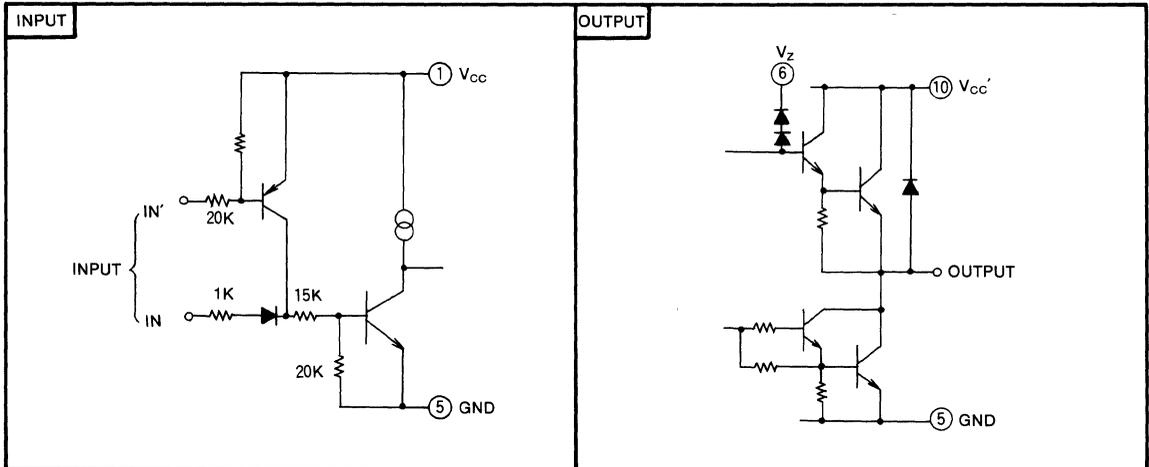
Input				Output		Note
I_1	I_1'	I_2	I_2'	O_1	O_2	
L	H	L	H	"OFF" state	"OFF" state	Open
L	L	L	H	H	L	Clockwise
L	H	L	L	L	H	Anti-clockwise
L	L	L	L	L	L	Braking
H	H	L	H	H	L	Clockwise
H	L	L	H	H	L	Clockwise
H	H	L	L	L	L	Braking
L	H	H	H	L	H	Anti-clockwise
L	H	H	L	L	H	Anti-clockwise
L	L	H	H	L	L	Braking
H	*	H	*	L	L	Braking

* : Irrelevant

BI-DIRECTIONAL MOTOR DRIVER



INPUT/OUTPUT CIRCUIT



BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RAIINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		-0.5~+12	V
$V_{CC'}$	Driver power supply		-0.5~+20	V
V_i	Input voltage		0~ V_{CC}	V
V_o	Output voltage		-0.5~ $V_{CC}+2.5$	V
$I_o(\text{max})$	Peak output current	$t_{op}=10\text{ms}$ $t_{op}=10\text{ms}$: repetitive cycle 0.2Hz max	± 800	mA
I_o	Continuous output current		± 150	mA
P_d	Power dissipation	$T_a=60^\circ\text{C}$	570	mW
T_j	Junction temperature		100	$^\circ\text{C}$
T_{opr}	Operating ambient temperature		-10~+60	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

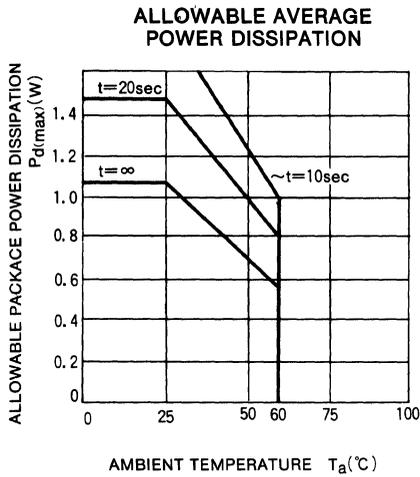
RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	5	10	V
I_o	Output current				± 100	mA
V_{IH}	"H" input voltage	I_1, I_2 input pins	3.0		V_{CC}	V
V_{rH}	"H" input voltage	I_1, I_2 input pins	$V_{CC}-0.5$		V_{CC}	V
V_{IL}	"L" input voltage	I_1, I_2 input pins	0		0.6	V
V_{rL}	"L" input voltage	I_1, I_2 input pins	0		$V_{CC}-3.6$	V
t_s	Motor braking interval		10	100		ms
T_s	Thermal shutdown protector operating temperature (Junction temperature)		100	120		$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=5\text{V}$, unless otherwise noted)

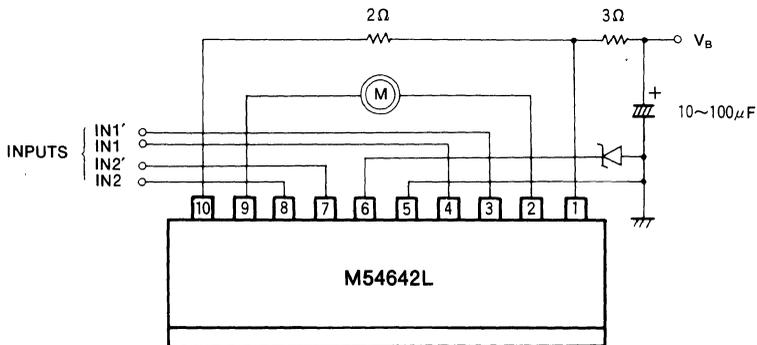
Symbol	Parameter	Test conditions		Limits			Unit
				Min	Typ	Max	
$I_o(\text{leak})$	Output leakage current	$V_{CC}=20\text{V}$ V_z : open	$V_o=20\text{V}$ $V_o=0\text{V}$			100 -100	μA
V_{OH}	"H" output voltage	$V_{CC}=12\text{V}$ V_z : open	$I_{OH}=-50\text{mA}$ $I_{OH}=-100\text{mA}$	10.2 10.0	10.5 10.4		V
V_{OL}	"L" output voltage	$V_{CC}=12\text{V}$ V_z : open	$I_{OH}=50\text{mA}$ $I_{OH}=100\text{mA}$		0.1 0.2	0.3 0.4	V
V_{O1-O2}	Voltage between output (1) and output (2) (voltage between motors)	$V_{CC}=12\text{V}$ $V_z=7\text{V}$	$I_o=\pm 100\text{mA}$	6.3	7.0	7.7	V
I_i	Input current	$V_{CC}=12\text{V}$	$V_1=3\text{V}, V_1'=4.5\text{V}$ $V_1=5\text{V}, V_1'=4.5\text{V}$	Output open		100 240 380	μA
I_i'	Input current	$V_{CC}=12\text{V}$	$V_1=0.4\text{V}, V_1'=4.5\text{V}$ $V_1'=1.4\text{V}, V_1=0.6\text{V}$	Output open		-200 -145 -190	μA
I_{CC}	Supply current	$V_{CC}=10\text{V}$ $V_{CC}=12\text{V}$	"OFF" state Clockwise or anti-clockwise Braking		1.2 4.5 7.5	3.0 8.0 12.0	mA

TYPICAL CHARACTERISTICS



Note 1 : Mounted on an epoxy PC board with Cu cover on one side (5cm×5cm×0.8mmt)
 Note 2 : t : time for power application

APPLICATION EXAMPLE



M54648L

BI-DIRECTIONAL MOTOR DRIVER WITH MOTOR SPEED CONTROL

DESCRIPTION

The M54648L is a semiconductor integrated circuit, capable of directly driving bi-directional micro motor, with a built in speed control circuit.

FEATURES

- Wide operating voltage range ($V_{CC}=4\sim 18V$)
- N MOS, C MOS IC output for direct drive
- Large drive current ($I_{O(max)}=\pm 3.0A$)
- Built-in operational amplifier for "H" output voltage control
- Built-in clamp diode
- Braking mode input

APPLICATION

- Audio tape deck player, radio cassette player
- VTR
- Home use equipment

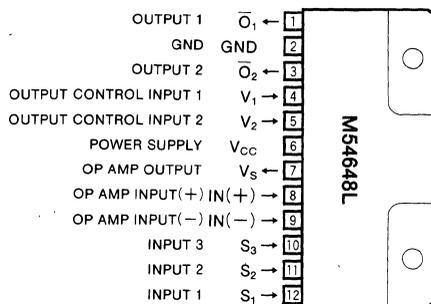
FUNCTION

The M54648L consists of a decoder, operational amplifier and bi-directional motor driver.

The output polarity shown in the logic truth table can be selected by inputs. S_1 , S_2 , and S_3 , and the supply voltage of the predriver can also be selected from the voltages driven by V_1 , V_2 or the output of the operational amplifier for the variable speed of motor. The motor driver has darlington output, capable of driving peak output voltage of $I_{OP(max)}=3.0A$.

The internal thermal shutdown circuit protects the IC from thermal destruction due to blocking of motor, etc.

PIN CONFIGURATION (TOP VIEW)

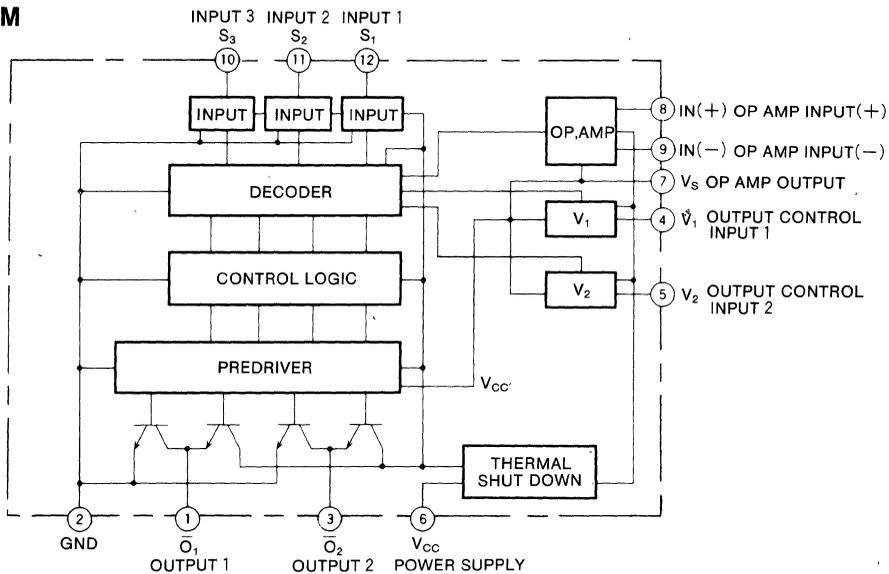


Outline 12P9

LOGIC TRUTH TABLE

Input			Output		Driver power supply (V_{CC})	Note
S_1	S_2	S_3	O_1	O_2		
L	L	L	"OFF" state	"OFF" state	—	STOP
L	L	H	H	L	OP AMP OUTPUT	PLAY(+)
L	H	L	L	H	OP AMP OUTPUT	PLAY(-)
L	H	H	H	L	V_2	FF(2)
H	L	L	L	H	V_2	REW(2)
H	L	H	H	L	V_1	FF(1)
H	H	L	L	H	V_1	REW(1)
H	H	H	L	L	V_S	BRAKE

BLOCK DIAGRAM



BI-DIRECTIONAL MOTOR DRIVER WITH MOTOR SPEED CONTROL

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage	With an external heat sink (3000mm ² X1.5mmt)	-0.5~+20	V
V_I	Input voltage	4Pin, 5Pin	-0.5~+14 or V_{CC}	V
		Inputs pin other than the above	-0.5~ V_{CC}	
V_O	Output voltage		-2.0~ $V_{CC}+2.5$	V
$I_{O(max)}$	Peak output current	top=10ms; repitive cycle 0.2Hz max	± 3.0	A
$I_{O(1)}$	Continuous output current (1)		± 300	mA
$I_{O(2)}$	Continuous output current (2)	With an external hbat sink (3000mm ² X1.5mmt)	± 800	mA
P_d	Power dissipation	$T_a=75^\circ\text{C}$	1.1	W
T_{opr}	Operating ambient temperature		-10~+75	$^\circ\text{C}$
T_{stg}	Storge temperature range		-55~+125	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	12	18	V
I_O	Output current				± 300	mA
V_{IH}	"H" input voltage		3		V_{CC}	V
V_{IL}	"L" input voltage		0		1	V
t_B	Motor braking interval		100			ms
$t_{j(shut)}$	Thermal shutdown temperature		125	150		$^\circ\text{C}$

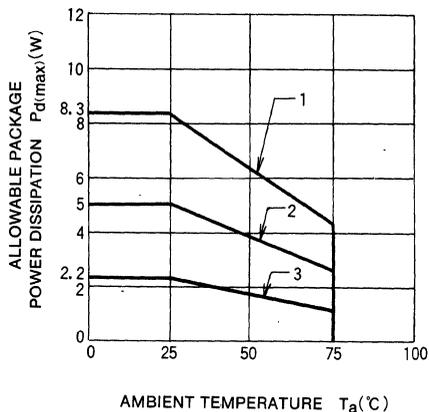
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit	
			Min	Typ	Max		
$I_{O(leak)}$	Output leakage current	$V_{S1}=0V$ $S_{S2}=0V$ $V_{S3}=0V$	$V_O=0V$			-100	μA
			$V_{CC}=V_S=20V$ $V_O=14V$ $V_{CC}=V_S=14V$			+100	
$V_{OH(1)}$	"H" output voltage	$V_{CC}=16V$ $V_{IN(-)}=0V$ $V_{IN(+)}=3V$ $V_{S1}=V_{S2}=0V$ $V_{S3}=3V$	$I_{OH}=-200\text{mA}$	13			V
			$I_{OH}=-500\text{mA}$	12.8			
$V_{OH(2)}$	"H" output voltage (2)	$V_{CC}=16V$ $V_{IN(-)}=0V$ $V_{IN(+)}=3V$ $V_{S1}=V_{S3}=0V$ $V_{S2}=3V$	$I_{OH}=-200\text{mA}$	13			V
			$I_{OH}=-500\text{mA}$	12.8			
$V_{OL(1)}$	"L" output voltage (1)	$V_{CC}=16V$ $V_{IN(-)}=0V$ $V_{IN(+)}=3V$ $V_{S1}=V_{S3}=0V$ $V_{S2}=3V$	$I_{OL}=200\text{mA}$			1.1	V
			$I_{OL}=500\text{mA}$			1.2	
$V_{OL(2)}$	"L" output voltage (2)	$V_{CC}=16V$ $V_{IN(-)}=0V$ $V_{IN(+)}=3V$ $V_{S1}=V_{S2}=0V$ $V_{S3}=3V$	$I_{OL}=200\text{mA}$			1.1	V
			$I_{OL}=500\text{mA}$			1.2	
I_{IH}	"H" input current	$V_{CC}=16V, V_{IS}=3V(S_1, S_2, S_3)$				10	μA
I_{IL}	"L" input current	$V_{CC}=16V, V_{IS}=0V(S_1, S_2, S_3)$				-20	μA
I_{CC}	Supply current	$V_{CC}=16V, V_{S1}=V_{S2}=V_{S3}=3V$				30	mA
A	Op amp open-loop-gain		50				dB

BI-DIRECTIONAL MOTOR DRIVER WITH MOTOR SPEED CONTROL

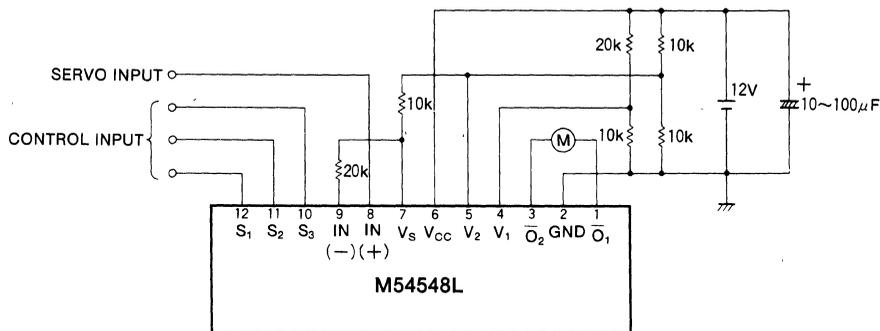
TYPICAL CHARACTERISTICS

**ALLOWABLE AVERAGE
 POWER DISSIPATION**



- 1) WITH HEAT SINK OF INFINITE SIZE
- 2) 25cm²×1.5mm ALUMINUM HEAT SINK
- 3) FREE AIR

APPLICATION EXAMPLE



Unit : Ω

M54648AL

BI-DIRECTIONAL MOTOR DRIVER WITH MOTOR SPEED CONTROL

DESCRIPTION

The M54648AL is a semiconductor integrated circuit, capable of directly driving bi-directional micro motor, with a built in speed control circuit.

FEATURES

- Wide operating voltage range ($V_{CC}=4\sim 18V$)
- N MOS, C MOS IC output for direct drive
- Large output sink current ($I_{O(max)}=\pm 3.0A$)
- Built-in operational amplifier for "H" output voltage control
- Built-in clamp diode
- Braking mode input
- Compact power package requiring small space

APPLICATION

- Audio tape deck player, radio cassette player
- VTR
- Home-use equipment

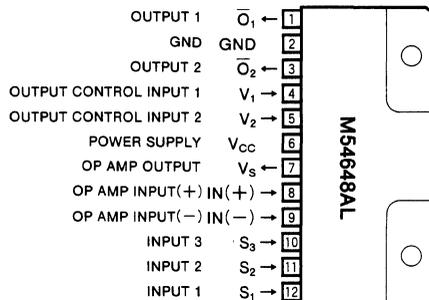
FUNCTION

The M54648AL consists of a decoder, operational amplifier and bi-directional motor driver.

The output polarity shown in the logic truth table can be selected by inputs. S_1 , S_2 , and S_3 , and the supply voltage of the predriver can also be selected from voltages driven by V_1 , V_2 or the output of the operational amplifier for the variable speed of motor. The motor driver has darlington output, capable of driving peak output voltage of $I_{OP(aax)} = 3.0A$.

The internal thermal shutdown protector protects the IC from thermal destruction due to blocking of motor, etc.

PIN CONFIGURATION (TOP VIEW)

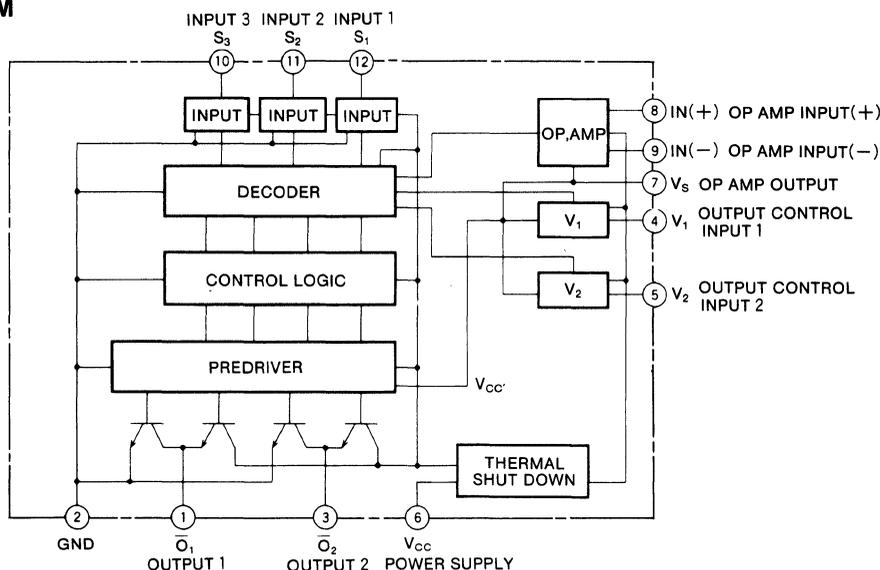


Outline 12P9B

LOGIC TRUTH TABLE

Input			Output		Driver power supply (V_{CC})	Note
S_1	S_2	S_3	\bar{O}_1	\bar{O}_2		
L	L	L	"OFF" state	"OFF" state	—	STOP
L	L	H	H	L	OP AMP OUTPUT	PLAY(+)
L	H	L	L	H	OP AMP OUTPUT	PLAY(-)
L	H	H	H	L	V_2	FF(2)
H	L	L	L	H	V_2	REW(2)
H	L	H	H	L	V_1	FF(1)
H	H	L	L	H	V_1	REW(1)
H	H	H	L	L	V_s	BRAKE

BLOCK DIAGRAM



BI-DIRECTIONAL MOTOR DRIVER WITH MOTOR SPEED CONTROL

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage	With an external heat sink (3000mm ² X1.5mmt)	-0.5~+20	V
V_i	Input voltage	4Pin, 5Pin	-0.5~+14 or V_{CC}	V
		Inputs pin other than the above	-0.5~ V_{CC}	
V_o	Output voltage		-2.0~ $V_{CC}+2.5$	V
$I_{O(max)}$	Peak output current	top=10ms, repitive cycle 0.2Hz max	±3.0	A
$I_{O(1)}$	Continuous output current (1)		±300	mA
$I_{O(2)}$	Continuous output current (2)	With an external hbat sink (3000mm ² X1.5mmt)	±800	mA
P_d	Power dissipation	$T_a=75^\circ\text{C}$	0.8	W
T_{opr}	Operating ambient temperature		-10~+75	°C
T_{stg}	Storage temperature range		-55~+125	°C

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	12	18	V
I_o	Output current				±300	mA
V_{IH}	"H" input voltage		3		V_{CC}	V
V_{IL}	"L" input voltage		0		1	V
t_B	Motor braking interval		100			ms
$t_{j(shut)}$	Thermal shutdown temperature		125	150		°C

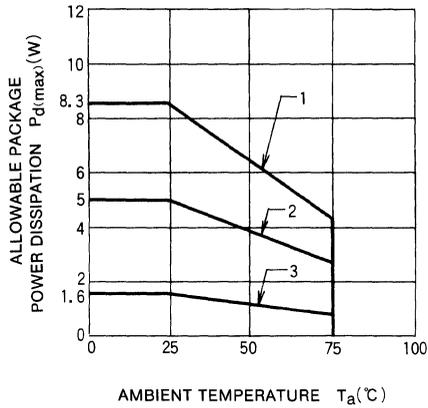
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{O(leak)}$	Output leakage current	$V_{S1}=0V$ $S_{S2}=0V$ $V_{S3}=0V$ $V_o=0V$ $V_{CC}=V_s=20V$ $V_o=14V$ $V_{CC}=V_s=14V$			-100 +100	μA
$V_{OH(1)}$	"H" output voltage (1)	$V_{CC}=16V$ $V_{IN(-)}=0V$ $V_{IN(+)}=3V$ $V_{S1}=V_{S2}=0V$ $V_{S3}=3V$ $I_{OH}=-200mA$ $I_{OH}=-500mA$	13 12.8			V
$V_{OH(2)}$	"H" output voltage (2)	$V_{CC}=16V$ $V_{IN(-)}=0V$ $V_{IN(+)}=3V$ $V_{S1}=V_{S3}=0V$ $V_{S2}=3V$ $I_{OH}=-200mA$ $I_{OH}=-500mA$	13 12.8			V
$V_{OL(1)}$	"L" output voltage (1)	$V_{CC}=16V$ $V_{IN(-)}=0V$ $V_{IN(+)}=3V$ $V_{S1}=V_{S3}=0V$ $V_{S2}=3V$ $I_{OL}=200mA$ $I_{OL}=500mA$			1.1 1.2	V
$V_{OL(2)}$	"L" output voltage (2)	$V_{CC}=16V$ $V_{IN(-)}=0V$ $V_{IN(+)}=3V$ $V_{S1}=V_{S2}=0V$ $V_{S3}=3V$ $I_{OL}=200mA$ $I_{OL}=500mA$			1.1 1.2	V
I_{IH}	"H" input current	$V_{CC}=16V, V_{IS}=3V(S_1, S_2, S_3)$			10	μA
I_{IL}	"L" input current	$V_{CC}=16V, V_{IS}=0V(S_1, S_2, S_3)$			-20	μA
I_{CC}	Supply current	$V_{CC}=16V, V_{S1}=V_{S2}=V_{S3}=3V$			30	mA
A	Op amp open-loop-gain		50			dB

BI-DIRECTIONAL MOTOR DRIVER WITH MOTOR SPEED CONTROL

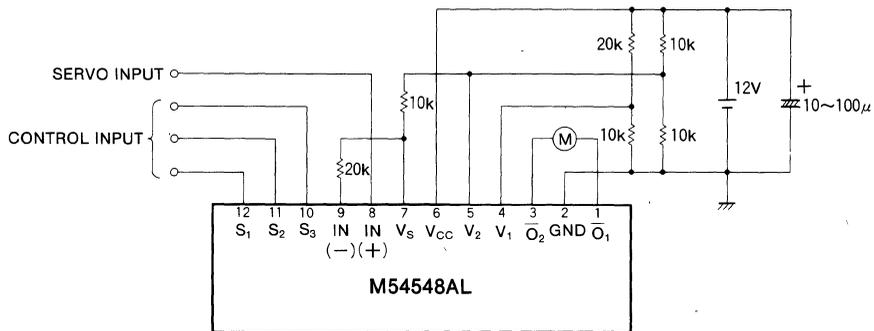
TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION



- 1) WITH HEAT SINK OF INFINITE SIZE
- 2) 25cm²×1.5mm ALUMINUM HEAT SINK
- 3) FREE AIR

APPLICATION EXAMPLE



Unit : Ω

M54649L

DUAL BI-DIRECTIONAL MOTOR DRIVER

DESCRIPTION

The M54649L, BI-DIRECTIONAL MOTOR DRIVER, consists of a full bridge power designed for use in are two DC-motor control circuit. The internal operational amplifier is capable for controlling the voltage across the bridge outputs.

FEATURES

- Capable of driving two bi-directional motors
- "H" Output voltage control pin
- Internal thermal shutdown protector
- Large output sink current ($I_{O(max)}=1.6A$)
- Wide operating supply voltage ($V_{CC}=4\sim 18V$)
- CMOS IC output for direct drive

APPLICATION

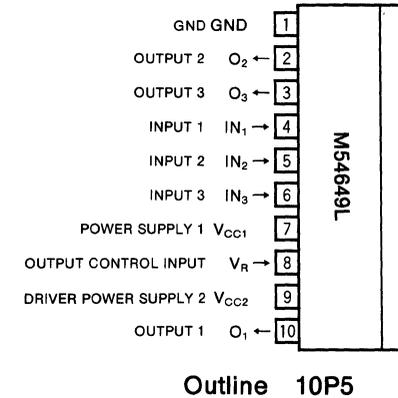
- Audio tape deck, radio cassette player
- VTR
- Home-use equipment

FUNCTION

The M54649L, full bridge motor driver, has the logic circuitry and the darlington power driver for bi-directional control of two DC motors operating at current up to 1.6A.

The input IN1, IN2 and IN3 are capable to control the bridge output polarity.

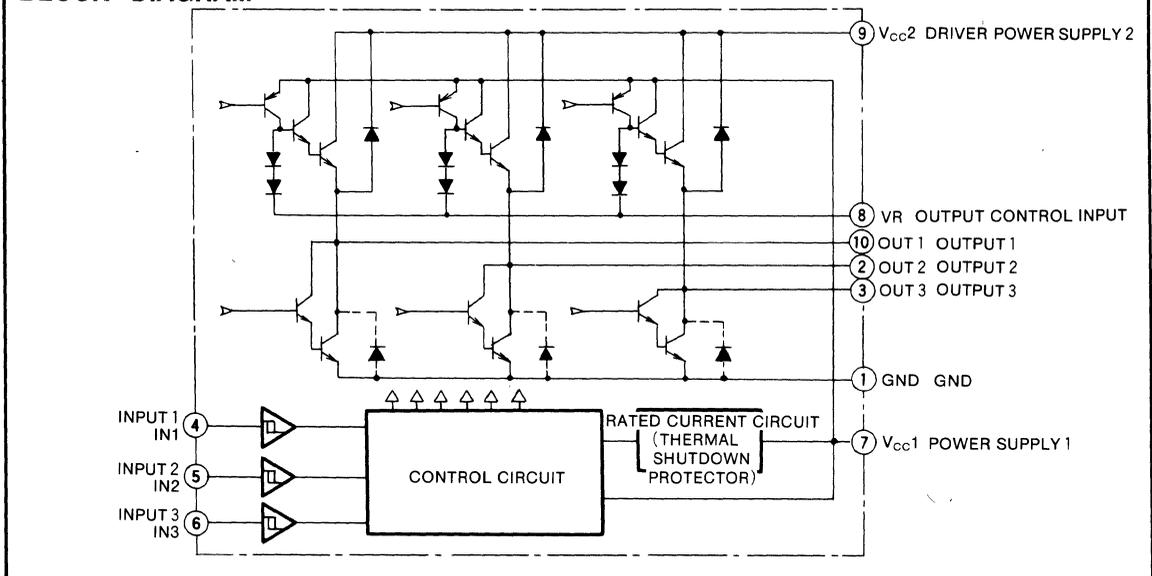
PIN CONFIGURATION (TOP VIEW)



LOGIC TRUTH TABLE

Input			Output			Note
4 PIN (IN 1)	5 PIN (IN 2)	6 PIN (IN 3)	10 PIN (OUT 1)	2 PIN (OUT 2)	3 PIN (OUT 3)	
L	L	L	L	L	L	Braking
H	L	H	H	L	OPEN	○
H	L	H	L	H	OPEN	○
L	H	L	H	OPEN	L	○
L	H	H	L	OPEN	H	○
H	H	L	L	L	L	Braking
		H				

BLOCK DIAGRAM



DUAL BI-DIRECTIONAL MOTOR DRIVER

ABSOLUTE MAXIMUM RATING ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
$V_{CC(1)}$	Supply voltage (1)		-0.5~+20.0	V
$V_{CC(2)}$	Supply voltage (2)		-0.5~+22.0	V
V_I	Input voltage		-0.5~+7.0	V
V_O	Output voltage		-2.0~ $V_{CC}+2.5$	V
I_{OP}	Peak output current	$t_{OP} \geq 50\text{ms}$, duty cycle 1/50	± 1.60	A
I_O	Continuous output current	Note 1	± 600	mA
P_d	Power dissipation	Time of power application 10s or less	2.78	W
T_{opr}	Operating ambient temperature		-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
$V_{CC(1)}$	Supply voltage (1)		4.0	12.0	18.0	V
$V_{CC(2)}$	Supply voltage (2)		0.0		22.0	V
I_O	Output current				± 600	mA
V_{IH}	"H" input voltage		3.5		V_{CC}	V
V_{IL}	"L" input voltage		0.0		1.0	V
V_R	Control voltage		0.0		18.0	V
T_{ON}	Thermal shutdown temperature		125	150		$^\circ\text{C}$
ΔT_{ON-OFF}	Hysteresis temperature width			50		$^\circ\text{C}$

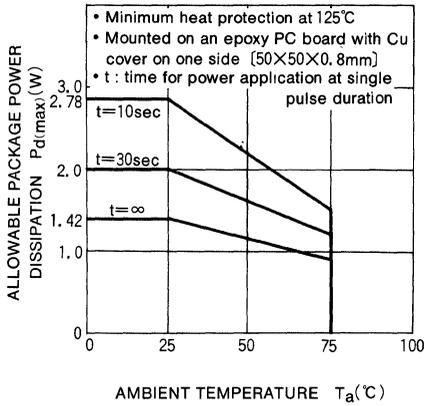
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit	
			Min	Typ	Max		
$I_{O(leak)}$	Output leakage current	Output open $V_O=0$ or 20V			± 100	μA	
V_{OL}	"L" output saturation voltage	$I_{OL}=500\text{mA}$			1.5	V	
V_{OH}	"H" output saturation voltage	$I_{OH}=-500\text{mA}$			10.0	V	
ΔV_O	"H" input current	$I_O=\pm 500\text{mA}$ $V_R=6.0\text{V}$			-0.5	0.5	V
I_R	"L" input current	$I_O \pm 500\text{mA}$ $V_R=6.0\text{V}$			0.2	1.5	mA
I_{CC1}	Supply current	$V_{IN1,2,3}=1.0\text{V}$ $I_O=0\text{mA}$			8.0	24.0	mA

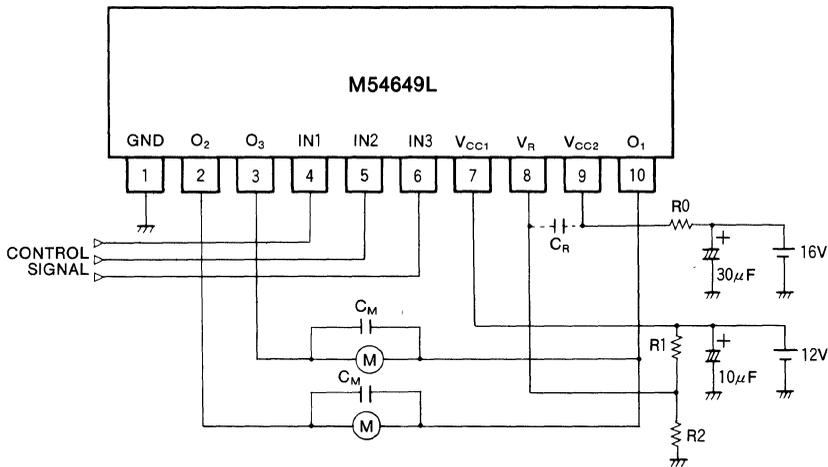
DUAL BI-DIRECTIONAL MOTOR DRIVER

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION



APPLICATION EXAMPLE



CM : Noise absorbing capacitor when the motor is driven should be less than $0.1\mu\text{F}$.

RO : Current limiting resistor when output is shorted.

(R1, R2) : The "H" output voltage $V_{O(H)}$ is given in,

$$V_{O(H)} = V_{CC1} \times \frac{R_2}{R_1 + R_2}$$

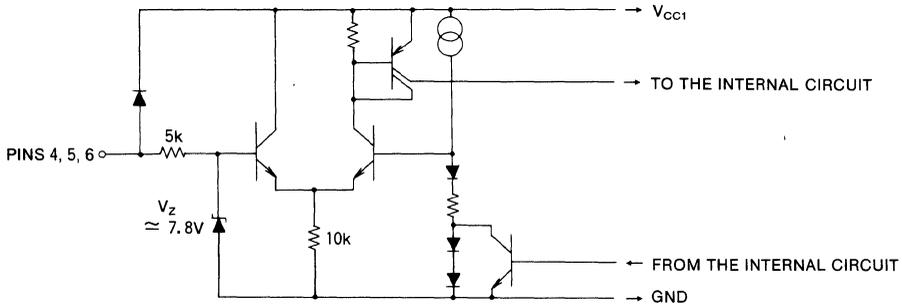
If the resistance of the R_1 and R_2 , the output current $V_{O(H)}$ is higher than the V_R (pin 8).

CR : If separate power supply is used for the V_R (pin 8), the output may oscillate. In this case, a capacitor C_R ($0.01\mu\text{F}$) must be connected between the V_R (pin 8) and V_{CC2} (pin 9).

DUAL BI-DIRECTIONAL MOTOR DRIVER

PRECAUTIONS FOR USE

1. Input circuit schematic diagram



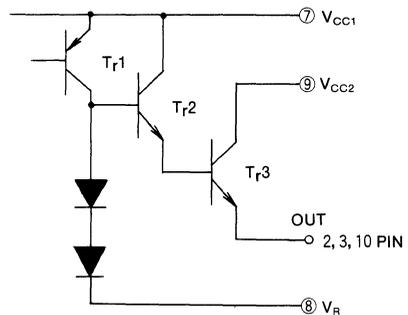
Apply $3.5V - V_{CC}$ to the ON voltage (V_{IH}) and $0 - 1V$ to the OFF voltage (V_{IL}) of the input.
 If the input voltage reaches approximately 7.8V or above, the impedance changes to approximately $5k\Omega$ and, therefore the voltage should be kept below 7V.

2. Output voltage control method

The output control circuit using the pin 8 is described on the right, and the voltage which is almost equal to the 8 pin voltage is output from the circuit. ("H" side)
 If pin 8 is open, the maximum output voltage is available when $V_{CC2} > V_{CC1}$ and the voltage is given in,

$$V_O = V_{CC1} - V_{sat}(Tr1) - V_{BE}(Tr2) - V_{BE}(Tr3)$$

The output voltage can be controlled by varying the V_{CC1} in this condition.



3. Precaution for braking

Care must be taken to braking mode input because the motor may affect other motors at the moment when it is switched from driving condition to braking condition.

4. Allowable power dissipation

The allowable power dissipation of the IC (P_d), when $V_{CC2} > V_{CC1}$, is given by.

$$P_d = V_{CC1} \times I_{CC1} + \downarrow Q \{V_{CC2} - V_{OH} + V_{OL}\}$$

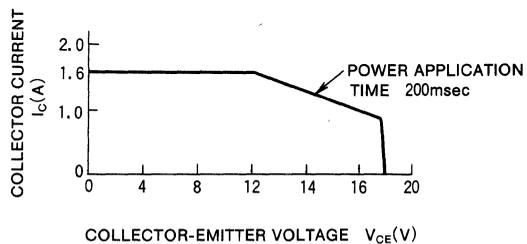
The equipment must be so designed as not exceed this limit.

5. Thermal shutdown

The internal thermal shutdown circuit is provided to protect the IC from overheating when excessive power is applied. The protection circuit functions when the temperature of the IC reaches $150^\circ C$ (Min. $125^\circ C$), and all outputs are in the OPEN mode, and is canceled when the temperature is decreased to $100^\circ C$ (Max. $125^\circ C$).

DUAL BI-DIRECTIONAL MOTOR DRIVER

6. Aso Characteristics



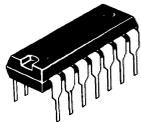
7. Others

Capacitors which are connected between power supply and GND should be placed as close to the IC as possible. Care should be taken as the capacitors may cause oscillation otherwise.

ANALOG SWITCHES

QUICK REFERENCE TABLE OF ANALOG SWITCHES

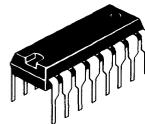
Structure	Type	No of switches		Features	Maximum ratings			Electrical characteristics (typ.)					Package Outline
		No of inputs	No of circuits		Supply voltage	Analog signal input voltage	Allowable power dissipation	Crosstalk	Frequency bandwidth	Offset voltage	Distortion @1kHz	Output noise	
CMOS	M4016BP	1	4	On resistance: 250 Ω typ. ($V_{DD}=15V$)	20V	$V_{SS}-0.5V$ $V_{DD}+0.5V$	—	—	12MHz	—	0.3%	—	14-pin DIP
	M4051	8	1	On resistance: 50 Ω typ. ($V_{DD}=15V$)	20V	$V_{EE}-0.5V$ $V_{DD}+0.5V$	—	—	25MHz	—	0.07%	—	16-pin DIP 16-pin FLAT
	M4052	4	2	On resistance: 50 Ω typ. ($V_{DD}=15V$)	20V	$V_{EE}-0.5V$ $V_{DD}+0.5V$	—	—	25MHz	—	0.07%	—	16-pin DIP 16-pin FLAT
	M4053	2	3	On resistance: 50 Ω typ. ($V_{DD}=15V$)	20V	$V_{EE}-0.5V$ $V_{DD}+0.5V$	—	—	25MHz	—	0.07%	—	16-pin DIP 16-pin FLAT
	M4066	1	4	On resistance: 50 Ω typ. ($V_{DD}=15V$)	20V	$V_{SS}-0.5V$ $V_{DD}+0.5V$	—	—	25MHz	—	0.07%	—	14-pin DIP 14-pin FLAT
Bipolar	M51320P	2	3	3 circuits independent switching	14V	6V	1.25W	55dB at 5MHz	10MHz (Video switch)	100mV	0.02%	50 μ Vrms (Audio switch)	16-pin DIP
	M51321P	3	3	3 circuits simultaneous switching	14V	6V	1.25W	55dB at 5MHz	10MHz (Video switch)	100mV	0.02%	50 μ Vrms (Audio switch)	18-pin DIP
	M51326P	2	3	3 circuits independent switching	14V	6V	1.25W	55dB at 5MHz	10MHz (Video switch)	100mV	0.02%	50 μ Vrms (Audio switch)	16-pin DIP
	M51327P	3	3	3 circuits simultaneous switching	14V	6V	1.25W	55dB at 5MHz	10MHz (Video switch)	100mV	0.02%	50 μ Vrms (Audio switch)	18-pin DIP
	M51329P	3	3	Audio and visual independent switching	14V	6V	1.25W	55dB at 5MHz	10MHz (Video switch)	100mV	0.02%	50 μ Vrms (Audio switch)	18-pin DIP
	M51330P	3	3	Audio and visual independent switching	14V	6V	1.25W	55dB at 5MHz	10MHz (Video switch)	100mV	0.02%	50 μ Vrms (Audio switch)	18-pin DIP
	M51551	2	2	Low distortion: 0.006% Built-in channel indication LED driver	24V	—	620mW	52dB at 1kHz	100kHz	—	0.006%	5.5 μ Vrms	14-pin DIP 14-pin FLAT



14-pin DIP



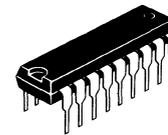
14-pin FLAT



16-pin DIP



16-pin FLAT



18-pin DIP

M4016BP

QUADRUPLE BILATERAL SWITCH

DESCRIPTION

The M4016BP is a semiconductor integrated circuit consisting of four independent bilateral analog switches.

FEATURES

- ON resistance: 250Ω typ. ($V_{DD}=15V$)
- High OFF resistance: $10^9\Omega$ or greater (typ)
- Small differences in ON resistance between each switch in the package: 10Ω typ. ($V_{DD}=15V$)
- Wide operating voltage range: $V_{DD}=3\sim 18V$
- Wide operating temperature range: $T_a=-40\sim +85^\circ C$

APPLICATION

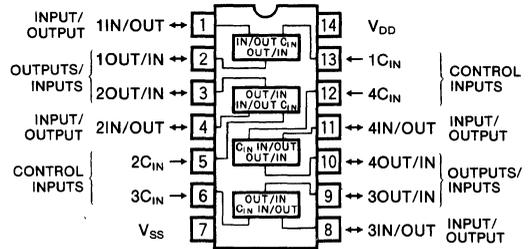
General purpose, for use in industrial and consumer digital equipment.

FUNCTIONAL DESCRIPTION

The control input (C_{IN}) can be used to change the input-to-output impedance (IN/OUT-OUT/IN) of the switches. When (C_{IN}) is made high, the input-to-output switch impedance is low and when set to low, this impedance is high.

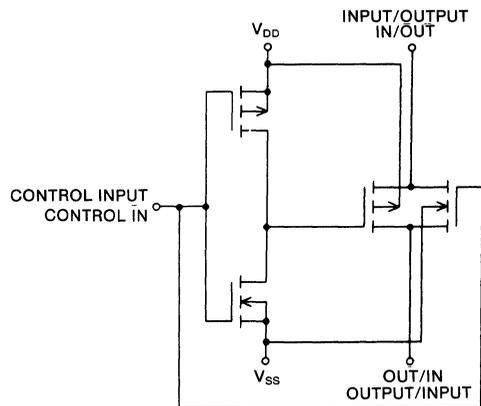
Input	Input/output and output/input resistance ($V_{DD}=10V, 15V$)
C_{IN}	
H	$2\sim 20 \times 10^2 \Omega$
L	$> 10^9 \Omega$ (typ)

PIN CONFIGURATION (TOP VIEW)

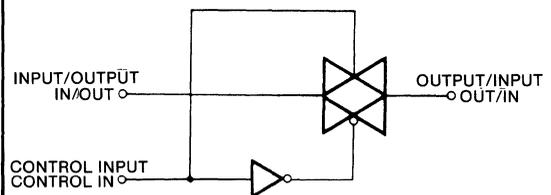


Outline 14P4

CIRCUIT SCHEMATIC (EACH SWITCH)



LOGIC DIAGRAM (EACH SWITCH)



ABSOLUTE MAXIMUM RATINGS ($T_a=-40\sim +85^\circ C$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{DD}	Supply voltage		$V_{SS}-0.5\sim V_{SS}+20$	V
V_i	Input voltage		$V_{SS}-0.5\sim V_{DD}+0.5$	V
I_i	Input current	Control inputs	± 10	mA
I_o	Output current		± 10	mA
T_{opr}	Operating temperature range		$-40\sim +85$	$^\circ C$
T_{stg}	Storage temperature range		$-65\sim +150$	$^\circ C$

QUADRUPLE BILATERAL SWITCH

RECOMMENDED OPERATING CONDITIONS ($T_a = -40 \sim +85^\circ\text{C}$, $V_{SS} = 0\text{V}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{DD}	Supply voltage	3		18	V
V_I	Input voltage	0		V_{DD}	V

ELECTRICAL CHARACTERISTICS ($V_{SS} = 0\text{V}$)

Symbol	Parameter	Test conditions	Limits								Unit	
					-40°C		25°C			85°C		
			$V_{SS}(\text{V})$	$V_{DD}(\text{V})$	Min	Max	Min	Typ	Max	Min		Max
V_{IH}	"H" input voltage (C_{IN})	Input-to-output current=10 μA	0	5	3.5		3.5			3.5		V
			0	10	7.0		7.0			7.0		
			0	15	11.0		11.0			11.0		
V_{IL}	"L" input current (C_{IN})	Input-to-output current=10 μA	0	5		1.0					1.0	V
			0	10		1.0					1.0	
			0	15		1.0					1.0	
R_{ON}	ON resistance	$V_I = 5\text{V}$	0	5				600				Ω
		$V_I = 2.5\text{V}$	0	5				6000				
		$V_I = 0.25\text{V}$	0	5				600				
		$V_I = 10\text{V}$	0	10		600			700		900	
		$V_I = 5\text{V}$	0	10		1300			1500		2000	
		$V_I = 0.25\text{V}$	0	10		600			700		900	
		$V_I = 15\text{V}$	0	15		430			500		650	
		$V_I = 7.5\text{V}$	0	15		800			950		1200	
		$V_I = 0.25\text{V}$	0	15		430			500		650	
		$V_I = 5\text{V}$	-5	5		600			700		900	
		$V_I = \pm 0.25\text{V}$	-5	5		1300			1500		2000	
		$V_I = -5\text{V}$	-5	5		600			700		900	
		$V_I = 7.5\text{V}$	-7.5	7.5		430			500		650	
		$V_I = \pm 0.25\text{V}$	-7.5	7.5		800			950		1200	
Test circuit 1	$V_I = -7.5\text{V}$	-7.5	7.5		430			500		650		
ΔR_{ON}	ON resistance variations between switches of the same package		-5	5				15			Ω	
			-7.5	7.5				10				
I_{OFF}	Input/output off-state leakage current	$V_{I/O} = 10\text{V}$, $V_{O/I} = 0\text{V}$	0	10					125		nA	
		$V_{I/O} = 0\text{V}$, $V_{O/I} = 10\text{V}$	0	10					-125			
		$V_{I/O} = 18\text{V}$, $V_{O/I} = 0\text{V}$	0	18		250			250			1000
		$V_{I/O} = 0\text{V}$, $V_{O/I} = 18\text{V}$	0	18		-250			-250			-1000
I_{DD}	Quiescent supply current	$V_I(C_{IN}) = V_{DD}$, V_{SS}	0	5		1			1		7.5	
			0	10		2			2		15	
			0	15		4			4		30	
I_{IH}	"H" input current (C_{IN})	$V_I = 18\text{V}$	0	18		0.3			0.3		1.0	
I_{IL}	"L" input current (C_{IN})	$V_{IL} = 0\text{V}$	0	18		-0.3			-0.3		-1.0	

QUADRUPLE BILATERAL SWITCH

SWITCHING CHARACTERISTICS (T_a=25°C)

Symbol	Parameter	Test conditions		Limits			Unit	
				V _{SS} (V)	V _{DD} (V)	Min		Typ
f _{max(I/O)}	Maximum transfer frequency	R _L =2kΩ	C _L =15pF Test circuit 2	-5	5		18	MHz
		R _L =10kΩ		-5	5		15	
		R _L =100kΩ		-5	5		12	
f _{max(C_{IN})}	Maximum control frequency	R _L =300Ω	C _L =15pF Test circuit 3	0	5		4	MHz
				0	10		10	
				0	15		12	
t _{PLH}	"L-H" and "H-L" output propagation time (IN/OUT-OUT/IN)	R _L =10kΩ C _L =50pF Test circuit 4	0	5			100	ns
			0	10			40	
			0	15			30	
t _{PHL}	"L-H" and "H-L" output propagation time (CONTROL IN-OUT/IN)	R _L =10kΩ C _L =50pF Test circuit 5	0	5			140	ns
			0	10			60	
			0	15			50	
—	Sine-wave distortion	R _L =10kΩ f _i =1kHz Test circuit 2	-5	5		0.3		%
—	Feedthrough (switch off)	R _L =1kΩ Test circuit 6	-5	5		600		kHz
—	Crosstalk (CONTROL IN-OUT/IN)	R _L =1kΩ R _L =10kΩ C _L =15pF Test circuit 7	0	5		80	mV	
			0	10		150		
			0	15		210		
C _I	Input capacitance	Control input				5	pF	
		Switch input/output				4		

TEST CIRCUITS

1 ON resistance (R_{ON})

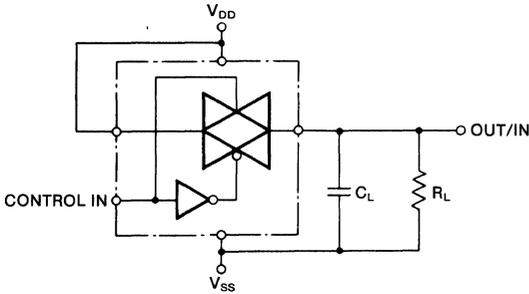
$$R_{ON} = 10 \times \frac{(V_i - V_o)}{V_o} \text{ (k}\Omega\text{)}$$

2 Maximum transfer frequency (f_{max(I/O)})
Sine-wave distortion

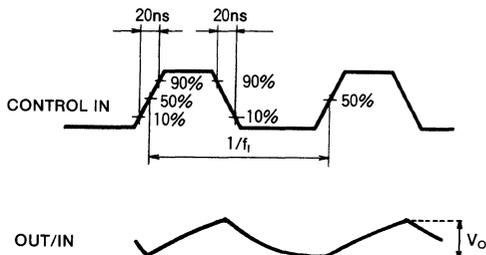
f_{max(I/O)} is taken as that frequency f_i at which, using a sine-wave input of 2.5V_{P-P}, 20 log₁₀(V_o/V_i) = -3dB

QUADRUPLE BILATERAL SWITCH

3 Maximum control frequency ($f_{max}(C_{IN})$)

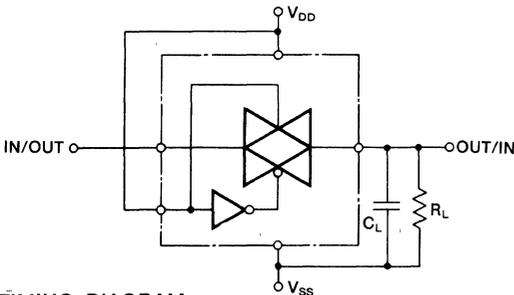


TIMING DIAGRAM

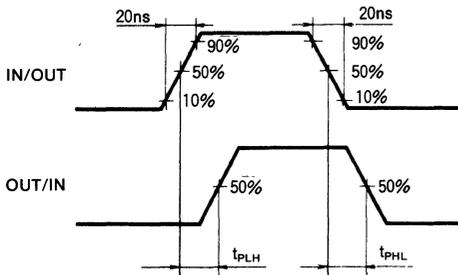


$f_{max}(C_{IN})$ is taken as that frequency f_i at which the output amplitude (V_o) is 1/2 that at 1 kHz.

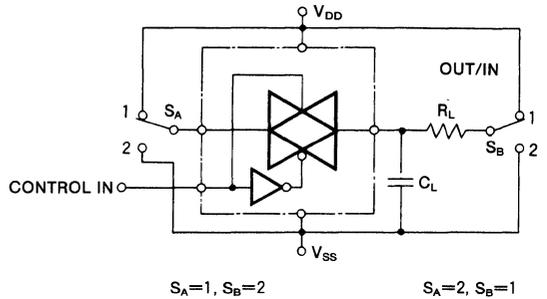
4 "L-H" and "H-L" output propagation time (IN/OUT-OUT/IN)



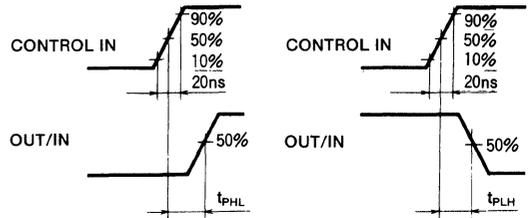
TIMING DIAGRAM



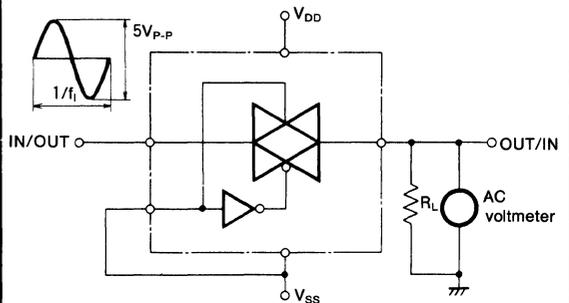
5 "L-H" and "H-L" output propagation time (CONTROL IN-OUT/IN)



TIMING DIAGRAM

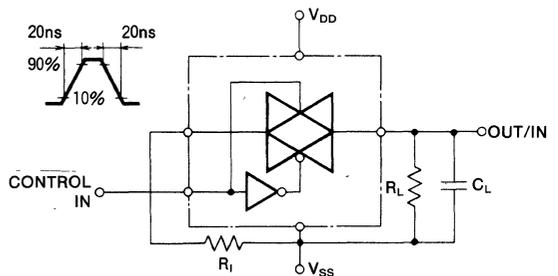


6 Feedthrough



The feedthrough is taken as that frequency f_i at which, using a sine-wave input of $2.5V_{P-P}$, $20 \log_{10}(V_o/V_i) = -50dB$.

7 Crosstalk



M4051BP M4051BFP

8-CHANNEL ANALOG MULTIPLEXER/DEMULTIPLEXER

DESCRIPTION

The M4051BP is a semiconductor integrated circuit consisting of a multiplexer/demultiplexer which uses a 3-bit digital input to perform selection of eight analog switches.

FEATURES

- Low ON resistance: 50Ω typ. ($V_{DD}=15V$)
- High OFF resistance: $10^9\Omega$ or greater (typ)
- Small differences in ON resistance between each switch in the package: 10Ω typ. ($V_{DD}=7.5V$, $V_{SS}=-7.5V$)
- Linearized transfer characteristics: 0.07% distortion (typ)
- Signals with amplitude greater than the logic level amplitude of the control inputs may be switched.
- Provided with an inhibit input

APPLICATION

General purpose, for use in industrial and consumer digital equipment.

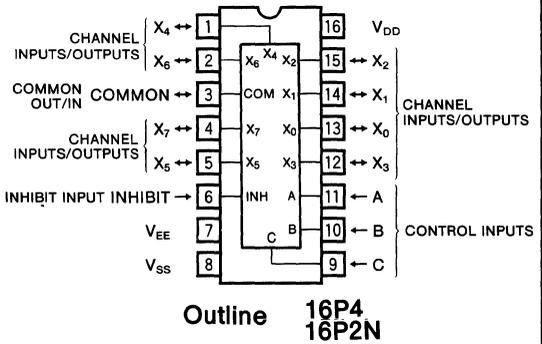
FUNCTIONAL DESCRIPTION

When a 3-bit binary input signal is applied to the control inputs (A, B, and C), the channel number corresponding to the binary value input (X_0 through X_7) is set at low impedance with respect to the (COMMON). All other channels remain at high impedance.

In this operation, if the (INHIBIT) input is held high, all channels (X_0 through X_7) will be put in the high-impedance state, regardless of the state of the other inputs.

It is possible to switch an analog signal of amplitude $V_{DD}-V_{EE}$ if this is greater than the logic level span $V_{DD}-V_{SS}$ for inputs (A, B, and C).

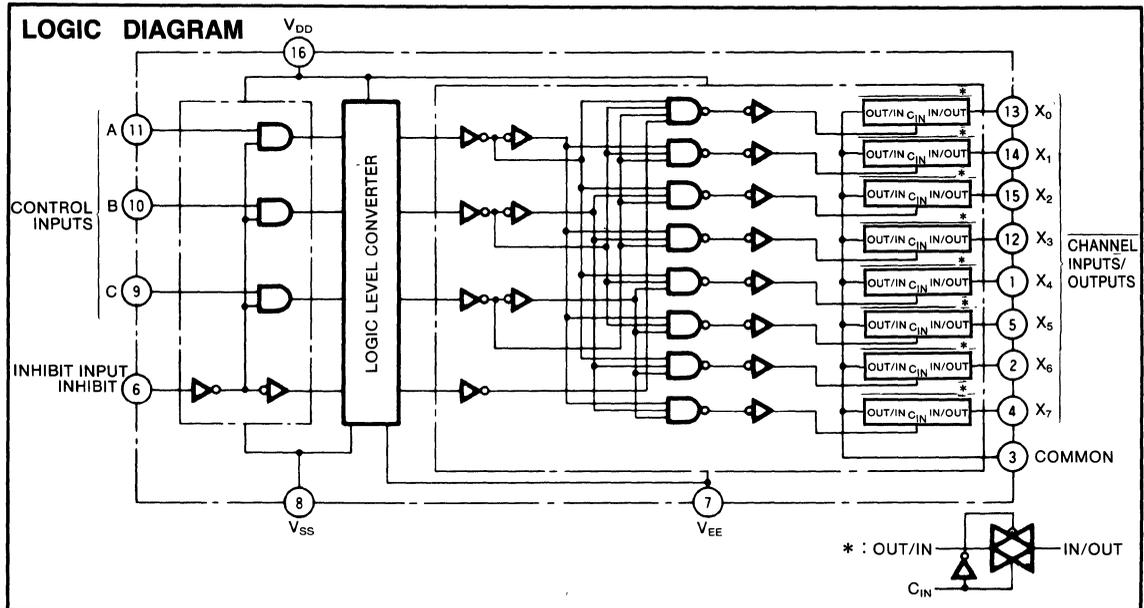
PIN CONFIGURATION (TOP VIEW)



FUNCTION TABLE (Note 1)

Inhibit input	Control inputs			Channel INPUT/OUTPUT to COMMON switch selection								
	INHIBIT	C	B	A	X_0	X_1	X_2	X_3	X_4	X_5	X_6	X_7
L	L	L	L	L	ON	OFF						
L	L	L	L	H	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF
L	L	L	H	L	OFF	OFF	ON	OFF	OFF	OFF	OFF	OFF
L	L	L	H	H	OFF	OFF	OFF	ON	OFF	OFF	OFF	OFF
L	L	H	L	L	OFF	OFF	OFF	OFF	ON	OFF	OFF	OFF
L	L	H	L	H	OFF	OFF	OFF	OFF	OFF	ON	OFF	OFF
L	L	H	H	L	OFF	OFF	OFF	OFF	OFF	OFF	ON	OFF
L	L	H	H	H	OFF	ON						
H	X	X	X	X	OFF							

Note 1 : X : Irrelevant
 ON : Low impedance between X_n and COMMON ($n=0\sim7$)
 OFF : High impedance between X_n and COMMON ($n=0\sim7$)



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8-CHANNEL ANALOG MULTIPLEXER/DEMULTIPLEXER

ABSOLUTE MAXIMUM RATINGS ($T_a = -40 \sim +85^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
$V_{DD}-V_{SS}$	Supply voltage		-0.5~20	V
$V_{DD}-V_{EE}$			-0.5~20	V
V_i	Input voltage	Control and inhibit inputs	$V_{SS}-0.5 \sim V_{DD}+0.5$	V
		Channel and common inputs	$V_{EE}-0.5 \sim V_{DD}+0.5$	
V_{IO}	Input-to-output voltage		± 0.5	V
I_i	Input current	Control and inhibit inputs	± 10	mA
I_o	Output current	Switch-off	± 10	mA
V_o	Output voltage	Channel and common outputs	$V_{EE}-0.5 \sim V_{DD}+0.5$	V
T_{opr}	Operating temperature range		-40~+85	$^\circ\text{C}$
T_{stg}	Storage temperature range		-65~+150	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITING CONDITIONS ($T_a = -40 \sim +85^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
$V_{DD}-V_{SS}$	Supply voltage	3		18	V
$V_{DD}-V_{EE}$		3		18	
V_i	Input voltage	Control and inhibit inputs	V_{SS}	V_{DD}	V
		Channel and common inputs	V_{EE}	V_{DD}	
V_o	Output voltage	V_{EE}		V_{DD}	V

ELECTRICAL CHARACTERISTICS ($V_{SS}=0\text{V}$)

Symbol	Parameter	Test conditions	Limits								Unit	
					-40 $^\circ\text{C}$		25 $^\circ\text{C}$			85 $^\circ\text{C}$		
			$V_{EE}(\text{V})$	$V_{DD}(\text{V})$	Min	Max	Min	Typ	Max	Min		Max
V_{IH}	"H" input voltage (A, B, C, INHIBIT)	Input-to-output current=10 μA	0	5	3.5		3.5			3.5		V
			0	10	7.0		7.0			7.0		
			0	15	11.0		11.0			11.0		
V_{IL}	"L" input current (A, B, C, INHIBIT)	Input-to-output current=10 μA	0	5		1.5			1.5		1.5	V
			0	10		3.0			3.0		3.0	
			0	15		4.0			4.0		4.0	
R_{ON}	ON resistance	$V_i=5\text{V}$	0	5		500			600		800	Ω
		$V_i=2.5\text{V}$	0	5		850			950		1300	
		$V_i=0.25\text{V}$	0	5		500			600		800	
		$V_i=10\text{V}$	0	10		210			250		300	
		$V_i=5\text{V}$	0	10		210			250		300	
		$V_i=0.25\text{V}$	0	10		210			250		300	
		$V_i=15\text{V}$	0	15		140			160		200	
		$V_i=7.5\text{V}$	0	15		140			160		200	
		$V_i=0.25\text{V}$	0	15		140			160		200	
		$V_i=5\text{V}$	-5	5		210			250		300	
		$V_i=\pm 0.25\text{V}$	-5	5		210			250		300	
		$V_i=-5\text{V}$	-5	5		210			250		300	
		$V_i=7.5\text{V}$	-7.5	7.5		140			160		200	
		$V_i=\pm 0.25\text{V}$	-7.5	7.5		140			160		200	
$V_i=-7.5\text{V}$	-7.5	7.5		140			160		200			
ΔR_{ON}	ON resistance variations between switches of the same package		-2.5	2.5				30			Ω	
			-5	5				15				
			-7.5	7.5				10				
I_{OFF}	Input-to-output off-state leakage current ($X_0 \sim X_7$ -COMMON)	$V_{IO}=10\text{V}, V_{OI}=0\text{V}$	0	10					125		nA	
		$V_{IO}=0\text{V}, V_{OI}=10\text{V}$	0	10					-125			
		$V_{IO}=18\text{V}, V_{OI}=0\text{V}$	0	18		250			250			1000
		$V_{IO}=0\text{V}, V_{OI}=18\text{V}$	0	18		-250			-250			-1000
I_{DD}	Quiescent supply current	$V_i=V_{DD}, V_{SS}$	0	5		20			20		150	
			0	10		40			40		300	
			0	15		80			80		600	
I_{IH}	"H" input current (A~C, INH)	$V_{IH}=18\text{V}$	0	18		0.3			0.3		1.0	μA
I_{IL}	"L" input current (A~C, INH)	$V_{IL}=0\text{V}$	0	18		-0.3			-0.3		-1.0	μA

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8-CHANNEL ANALOG MULTIPLEXER/DEMULTIPLEXER

SWITCHING CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{SS}=0\text{V}$)

Symbol	Parameter	Tset conditions	Limits			Unit		
			$V_{EE}(\text{V})$	$V_{DD}(\text{V})$	Min		Typ	Max
$f_{\text{max}(I/O)}$	Maximum transfer frequency	$R_L=10\text{k}\Omega$ $C_L=15\text{pF}$ Test circuit 2	-5	5		25	MHz	
t_{PLH}	"L-H" and "H-L" output propagation time (A, B, C- $X_0\sim X_7$, COMMON)	$R_L=10\text{k}\Omega$ $C_L=50\text{pF}$ Test circuit 3	0	5			1000	ns
			0	10			500	
			0	15			400	
			-5	5			700	
			-7.5	7.5			500	
t_{PHL}	"L-H" and "H-L" output propagation time (A, B, C- $X_0\sim X_7$, COMMON)	$R_L=10\text{k}\Omega$ $C_L=50\text{pF}$ Test circuit 3	0	5			1000	ns
			0	10			500	
			0	15			400	
			-5	5			700	
			-7.5	7.5			500	
t_{PLH}	"L-H" and "H-L" output propagation time (INHIBIT- $X_0\sim X_7$, COMMON)	$R_L=10\text{k}\Omega$ $C_L=50\text{pF}$ Test circuit 4	0	5			1400	ns
			0	10			700	
			0	15			500	
			-5	5			900	
			-7.5	7.5			500	
t_{PHL}	"L-H" and "H-L" output propagation time (INHIBIT- $X_0\sim X_7$, COMMON)	$R_L=10\text{k}\Omega$ $C_L=50\text{pF}$ Test circuit 4	0	5			1400	ns
			0	10			700	
			0	15			500	
			-5	5			900	
			-7.5	7.5			500	
t_{PLH}	"L-H" and "H-L" output propagation time ($X_0\sim X_7$ /COMMON-COMMON/ $X_0\sim X_7$)	$R_L=10\text{k}\Omega$ $C_L=50\text{pF}$ Test circuit 5	0	5			45	ns
			0	10			30	
			0	15			20	
t_{PHL}	"L-H" and "H-L" output propagation time ($X_0\sim X_7$ /COMMON-COMMON/ $X_0\sim X_7$)	$R_L=10\text{k}\Omega$ $C_L=50\text{pF}$ Test circuit 5	0	5			45	ns
			0	10			30	
			0	15			20	
—	Sine-wave distortion	$R_L=10\text{k}\Omega$ $f_i=1\text{kHz}$ Test circuit 2	-5	5		0.1	%	
—	Feedthrough (switch off)	$R_L=1\text{k}\Omega$ Test circuit 6	-5	5		500	kHz	
—	Crosstalk (A, B, C, INHIBIT- $X_0\sim X_7$, COMMON)	$R_L=1\text{k}\Omega$ $R_L=10\text{k}\Omega$ $C_L=15\text{pF}$ Test circuit 7	0	5			200	mV
			0	10			300	
			0	15			400	
C_i	Input capacitance	Control and inhibit inputs					7.5	pF
		Channel and common inputs				10		

TEST CIRCUITS ($V_{SS}=0\text{V}$, capacitance C_L includes stray wiring capacitance and probe input capacitance)

1 ON resistance (R_{ON})

$R_{ON} = 10 \times \frac{(V_1 - V_0)}{V_0} (\text{k}\Omega)$

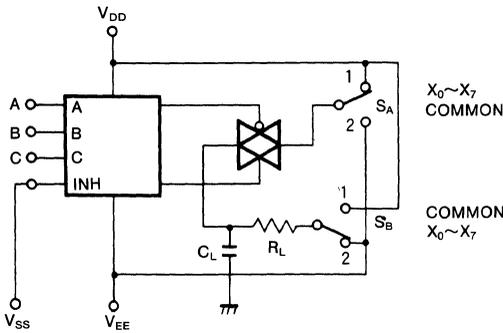
Refer to the function table for conditions of control inputs A, B, and C.

2 Maximum transfer frequency ($f_{\text{max}(I/O)}$) Sine-wave distortion

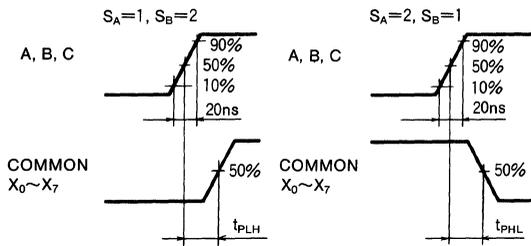
$f_{\text{max}(I/O)}$ is taken as that frequency f_i at which, using a sine-wave input of $2.5V_{p-p}$, $20 \log_{10}(V_0/V_1) = -3\text{dB}$. Refer to the function table for conditions of control inputs A, B, and C.

8-CHANNEL ANALOG MULTIPLEXER/DEMULPLEXER

3 "L-H" and "H-L" output propagation time
 (A, B, C-X₀-X₇, COMMON)

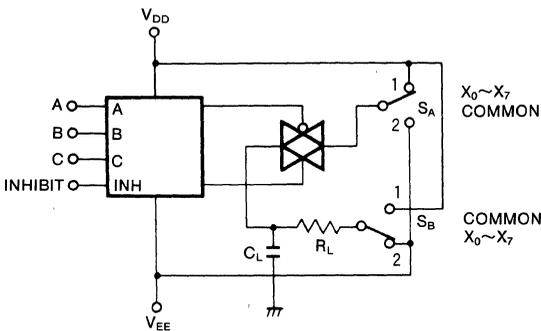


TIMING DIAGRAM

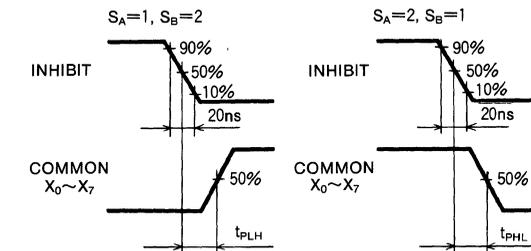


Refer to the function table for conditions of control inputs A, B, and C

4 "L-H" and "H-L" output propagation time
 (INHIBIT-X₀-X₇, COMMON)

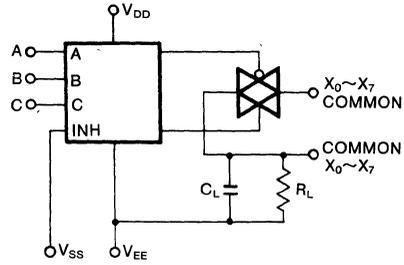


TIMING DIAGRAM

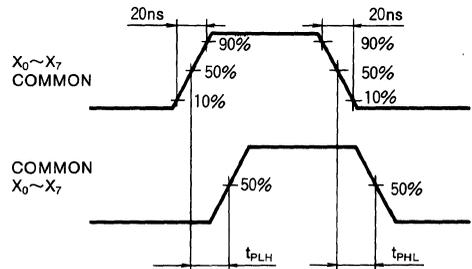


Refer to the function table for conditions of control inputs A, B, and C.

5 "L-H" and "H-L" output propagation time
 (X₀-X₇/COMMON-COMMON/X₀-X₇)

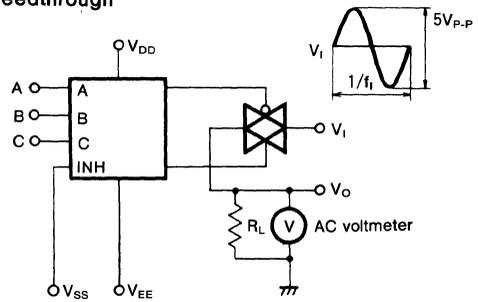


TIMING DIAGRAM



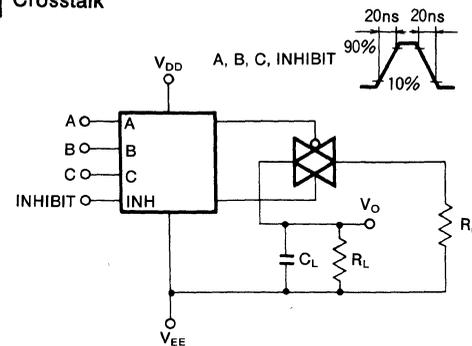
Refer to the function table for conditions of control inputs A, B, and C

6 Feedthrough



The feedthrough is taken as that frequency f_i at which, using a sine-wave input of $2.5V_{P-P}$, $20 \log_{10}(V_o/V_i) = -50\text{dB}$. Refer to the function table for conditions of control inputs A, B, and C.

7 Crosstalk



Refer to the function table for conditions of control inputs A, B, and C.

M4052BP M4052BFP

DUAL 4-CHANNEL ANALOG MULTIPLEXER/DEMULTIPLEXER

DESCRIPTION

The M4052BP is a semiconductor integrated circuit consisting of two multiplexer/demultiplexers which use 2-bit digital inputs to perform selection of four analog switches.

FEATURES

- Low ON resistance: 50Ω typ. ($V_{DD}=15V$)
- High OFF resistance: $10^9\Omega$ or greater (typ)
- Small differences in ON resistance between each switch in the package: 10Ω typ. ($V_{DD}=7.5V$, $V_{SS}=-7.5V$)
- Linearized transfer characteristics: 0.07% distortion (typ)
- Signals with amplitude greater than the logic level amplitude of the control inputs may be switched.
- Provided with an inhibit input

APPLICATION

General purpose, for use in industrial and consumer digital equipment.

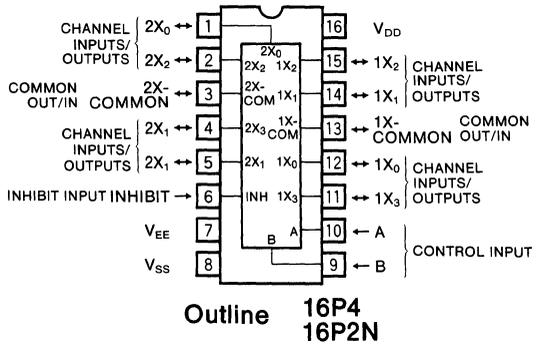
FUNCTIONAL DESCRIPTION

When a 2-bit binary input signal is applied to the control inputs (A and B), the channel number corresponding to the binary value input (X_0 through X_3) is set at low impedance with respect to the corresponding (X-COMMON). All other channels remain at high impedance.

In this operation, if the (INHIBIT) input is held high, all channels (X_0 through X_3) will be put in the high-impedance state, regardless of the state of the other inputs.

It is possible to switch an analog signal of amplitude $V_{DD}-V_{EE}$ if this is greater than the logic level span $V_{DD}-V_{SS}$ for inputs (A and B)

PIN COFIGRATION (TOP VIEW)



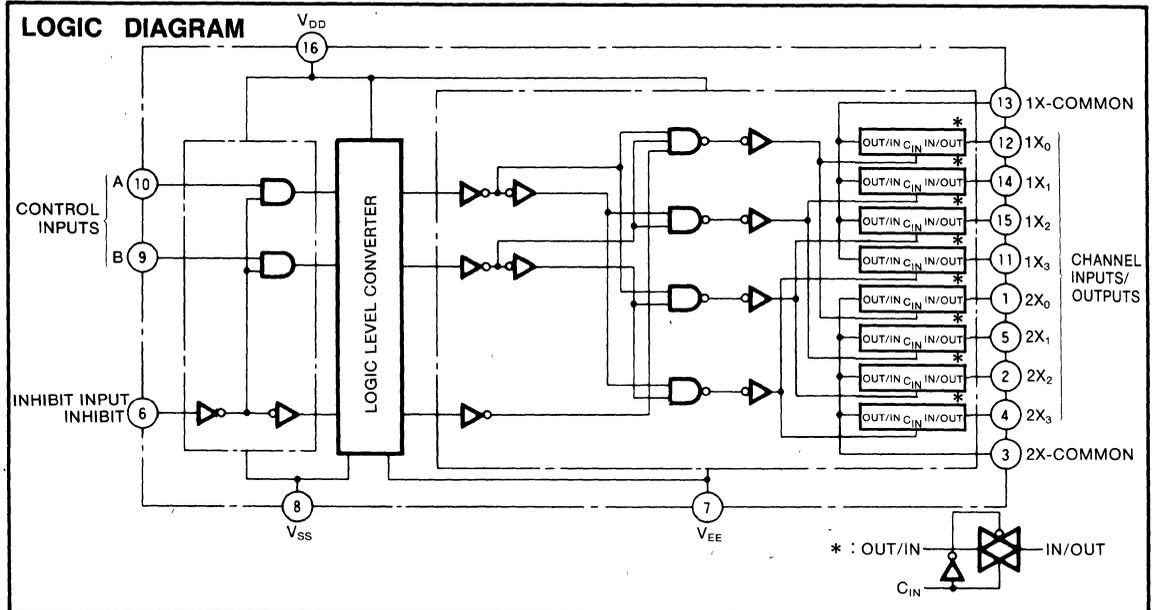
FUNCTION TABLE (Note 1)

Inhibit input	Control inputs		Channel INPUT/OUTPUT to COMMON switch selection			
	B	A	X_0	X_1	X_2	X_3
INHIBIT						
L	L	L	ON	OFF	OFF	OFF
L	L	H	OFF	ON	OFF	OFF
L	H	L	OFF	OFF	ON	OFF
L	H	H	OFF	OFF	OFF	ON
H	X	X	OFF	OFF	OFF	OFF

Note 1 : X : Irrelevant

ON : Low impedance between X_n and X-COMMON ($n=0\sim3$)

OFF : High impedance between X_n and X-COMMON ($n=0\sim3$)



M4052BP

M4052BFP

DUAL 4-CHANNEL ANALOG MULTIPLEXER/DEMULPLEXER

ABSOLUTE MAXIMUM RATINGS (T_a=-40~+85°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{DD} -V _{SS}	Supply voltage		-0.5~20	V
V _{DD} -V _{EE}			-0.5~20	V
V _I	Input voltage	Control and inhibit inputs	V _{SS} -0.5~V _{DD} +0.5	V
		Channel and common inputs	V _{EE} -0.5~V _{DD} +0.5	V
V _{IO}	Input-to-output voltage		±0.5	V
I _I	Input current	Control and inhibit inputs	±10	mA
I _O	Output current	Switch-off	±10	mA
V _O	Output voltage	Channel and common outputs	V _{EE} -0.5~V _{DD} +0.5	V
T _{opr}	Operating temperature range		-40~+85	°C
T _{stg}	Storage temperature range		-65~+150	°C

RECOMMENDED OPERATING CONDITING CONDITIONS (T_a=-40~+85°C, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V _{DD} -V _{SS}	Supply voltage	3		18	V
V _{DD} -V _{EE}		3		18	V
V _I	Input voltage	Control and inhibit inputs	V _{SS}	V _{DD}	V
		Channel and common inputs	V _{EE}	V _{DD}	
V _O	Output voltage	V _{EE}		V _{DD}	V

ELECTRICAL CHARACTERISTICS (V_{SS}=0V)

Symbol	Parameter	Test conditions		Limits						Unit			
				-40°C		25°C		85°C					
				V _{EE} (V)	V _{DD} (V)	Min	Max	Min	Typ		Max	Min	Max
V _{IH}	"H" input voltage (A, B, INHIBIT)	Input-to-output current=10μA	0	5	3.5		3.5				3.5	V	
			0	10	7.0		7.0				7.0		
			0	15	11.0		11.0				11.0		
V _{IL}	"L" input current (A, B, INHIBIT)	Input-to-output current=10μA	0	5		1.5				1.5	1.5	V	
			0	10		3.0				3.0	3.0		
			0	15		4.0				4.0	4.0		
R _{ON}	ON resistance	Test circuit 1	V _I =5V	0	5		500				600	800	Ω
			V _I =2.5V	0	5		850				950	1300	
			V _I =0.25V	0	5		500				600	800	
			V _I =10V	0	10		210				250	300	
			V _I =5V	0	10		210				250	300	
			V _I =0.25V	0	10		210				250	300	
			V _I =15V	0	15		140				160	200	
			V _I =7.5V	0	15		140				160	200	
			V _I =0.25V	0	15		140				160	200	
			V _I =5V	-5	5		210				250	300	
			V _I =±0.25V	-5	5		210				250	300	
			V _I =-5V	-5	5		210				250	300	
			V _I =7.5V	-7.5	7.5		140				160	200	
V _I =±0.25V	-7.5	7.5		140				160	200				
V _I =-7.5V	-7.5	7.5		140				160	200				
ΔR _{ON}	ON resistance variations between switches of the same package		-2.5	2.5				30				Ω	
			-5	5				15					
			-7.5	7.5				10					
I _{OFF}	Input-to-output off-state leakage current (X ₀ ~X ₃ -X-COMMON)		V _{I/O} =10V, V _{O/I} =0V	0	10					125		nA	
			V _{I/O} =0V, V _{O/I} =10V	0	10					-125			
			V _{I/O} =18V, V _{O/I} =0V	0	18		250				250		1000
			V _{I/O} =0V, V _{O/I} =18V	0	18		-250				-250		-1000
I _{DD}	Quiescent supply current	V _I =V _{DD} , V _{SS}	0	5		20				20	150	μA	
			0	10		40				40	300		
			0	15		80				80	600		
I _{IH}	"H" input current (A, B, INH)	V _{IH} =18V	0	18		0.3			0.3		1.0	μA	
I _{IL}	"L" input current (A, B, INH)	V _{IL} =0V	0	18		-0.3			-0.3		-1.0	μA	

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DUAL 4-CHANNEL ANALOG MULTIPLEXER/DEMULTIPLEXER

SWITCHING CHARACTERISTICS (T_a=25°C, V_{SS}=0V)

Symbol	Parameter	Tset conditions	Limits			Unit		
			V _{EE} (V)	V _{DD} (V)	Min		Typ	Max
f _{max(I/O)}	Maximum transfer frequency	R _L =10kΩ C _L =15pF Test circuit 2	-5	5		25		MHz
t _{PLH}	"L-H" and "H-L" output propagation time (A, B-X ₀ ~X ₃ , X-COMMON)	R _L =10kΩ C _L =50pF Test circuit 3	0	5			1000	ns
			0	10			500	
			0	15			400	
			-5	5			700	
			-7.5	7.5			500	
t _{PHL}	"L-H" and "H-L" output propagation time (A, B-X ₀ ~X ₃ , X-COMMON)	R _L =10kΩ C _L =50pF Test circuit 3	0	5			1000	ns
			0	10			500	
			0	15			400	
			-5	5			700	
			-7.5	7.5			500	
t _{PLH}	"L-H" and "H-L" output propagation time (INHIBIT-X ₀ ~X ₃ , X-COMMON)	R _L =10kΩ C _L =50pF Test circuit 4	0	5			1400	ns
			0	10			700	
			0	15			500	
			-5	5			900	
			-7.5	7.5			500	
t _{PHL}	"L-H" and "H-L" output propagation time (INHIBIT-X ₀ ~X ₃ , X-COMMON)	R _L =10kΩ C _L =50pF Test circuit 4	0	5			1400	ns
			0	10			700	
			0	15			500	
			-5	5			900	
			-7.5	7.5			500	
t _{PLH}	"L-H" and "H-L" output propagation time (X ₀ ~X ₃ /X-COMMON-X-COMMON/X ₀ ~X ₃)	R _L =10kΩ C _L =50pF Test circuit 5	0	5			45	ns
			0	10			30	
			0	15			20	
t _{PHL}	"L-H" and "H-L" output propagation time (X ₀ ~X ₃ /X-COMMON-X-COMMON/X ₀ ~X ₃)	R _L =10kΩ C _L =50pF Test circuit 5	0	5			45	ns
			0	10			30	
			0	15			20	
—	Sine-wave distortion	R _L =10kΩ f _i =1kHz Test circuit 2	-5	5		0.1		%
—	Feedthrough (switch off)	R _L =1kΩ Test circuit 6	-5	5		500		kHz
—	Crosstalk (A, B, INHIBIT-X ₀ ~X ₃ , X-COMMON)	R _L =1kΩ	0	5		200		mV
		R _L =10kΩ C _L =15pF Test circuit 7	0	10		300		
C _I	Input capacitance	Control and inhibit inputs					7.5	pF
		Channel and common inputs				10		

TEST CIRCUITS (V_{SS}=0V, capacitance C_L includes stray wiring capacitance and probe input capacitance)

1 ON resistance (R_{ON})

$$R_{ON} = 10 \times \frac{(V_I - V_O)}{V_O} \text{ (k}\Omega\text{)}$$

Refer to the function table for conditions of control inputs A and B

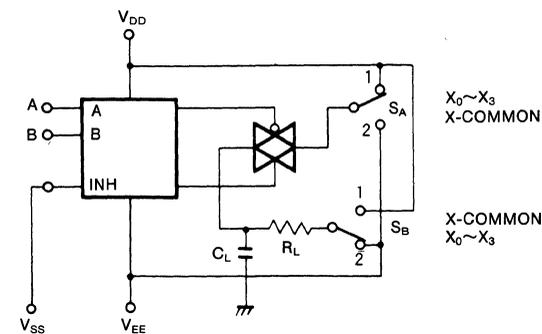
2 Maximum transfer frequency (f_{max(I/O)}) Sine-wave distortion

f_{max(I/O)} is taken as that frequency f_i at which, using a sine-wave input of 2.5V_{P-P}, 20 log₁₀(V_O/V_I) = -3dB Refer to the function table for conditions of control inputs A and B.

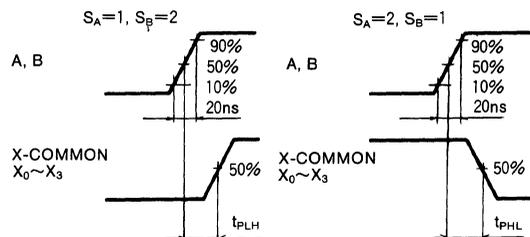
M4052BP M4052BFP

DUAL 4-CHANNEL ANALOG MULTIPLEXER/DEMULTIPLEXER

3 "L-H" and "H-L" output propagation time (A, B-X₀-X₇, COMMON)

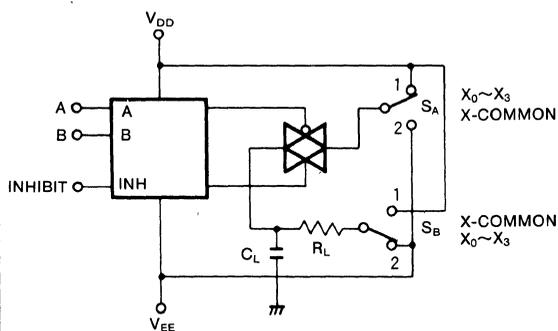


TIMING DIAGRAM

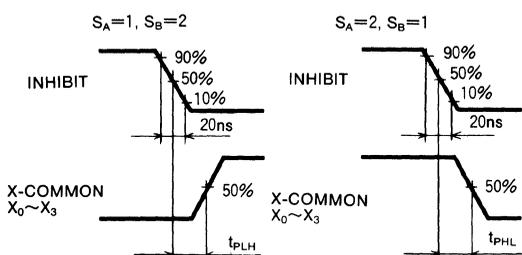


Refer to the function table for conditions of control inputs A and B.

4 "L-H" and "H-L" output propagation time (INHIBIT-X₀-X₃, X-COMMON)

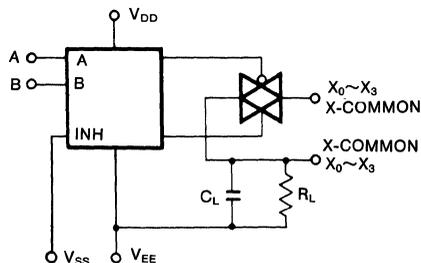


TIMING DIAGRAM

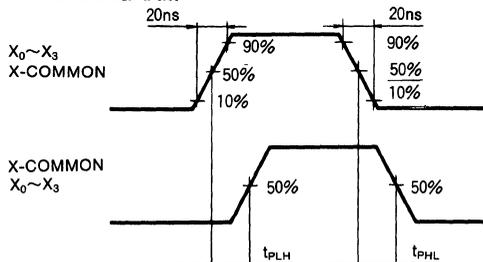


Refer to the function table for conditions of control inputs A and B.

5 "L-H" and "H-L" output propagation time (X₀-X₃/X-COMMON-X-COMMON/X₀-X₃)

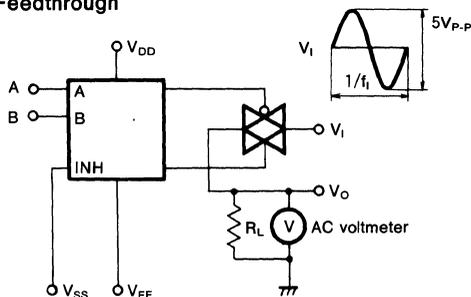


TIMING DIAGRAM



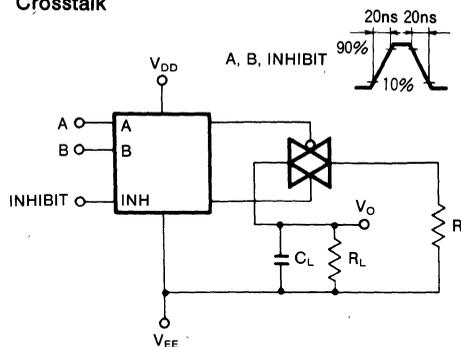
Refer to the function table for conditions of control inputs A and B.

6 Feedthrough



The feedthrough is taken as that frequency f_i at which, using a sine-wave input of $2.5V_{P-P}$, $20 \log_{10}(V_o/V_i) = -50\text{dB}$. Refer to the function table for conditions of control inputs A and B.

7 Crosstalk



Refer to the function table for conditions of control inputs A and B.

M4053BP M4053BFP

TRIPLE 2-CHANNEL ANALOG MULTIPLEXER/DEMULTIPLEXER

DESCRIPTION

The M4053BP is a semiconductor integrated circuit consisting of three multiplexer/demultiplexers which use 1-bit digital inputs to perform selection of two analog switches.

FEATURES

- Low ON resistance: 50Ω typ. ($V_{DD}=15V$)
- High OFF resistance: $10^9\Omega$ or greater (typ)
- Small differences in ON resistance between each switch in the package: 10Ω typ. ($V_{DD}=7.5V$, $V_{SS}=-7.5V$)
- Linearized transfer characteristics: 0.07% distortion (typ)
- Signals with amplitude greater than the logic level amplitude of the control inputs may be switched.
- Provided with an inhibit input

APPLICATION

General purpose, for use in industrial and consumer digital equipment.

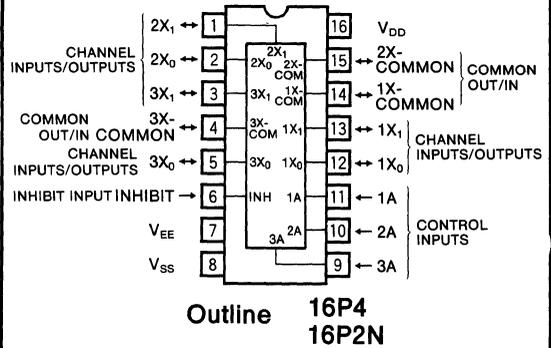
FUNCTIONAL DESCRIPTION

When a 1-bit binary input signal is applied to the control inputs (A), the channel numbers corresponding to the binary value input (X_0 , X_1) are set to low impedance with respect to the corresponding (X-COMMON). All other channels remain at high impedance.

In this operation, if the (INHIBIT) input is held high, all channels (X_0 , X_1) will be put in the high-impedance state, regardless of the state of the other inputs.

It is possible to switch an analog signal of amplitude $V_{DD}-V_{EE}$ if this is greater than the logic level span $V_{DD}-V_{SS}$ for inputs (A).

PIN COFIGRATION (TOP VIEW)



FUNCTION TABLE (Note 1)

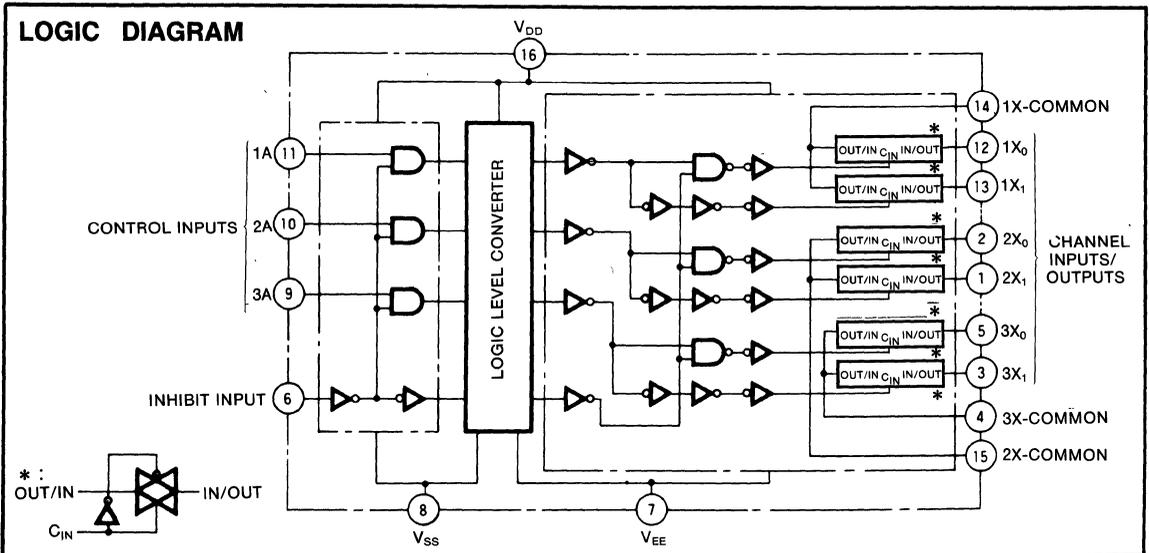
Inhibit input	Control inputs	Channel INPUT/OUTPUT to COMMON switch selection	
INHIBIT	A	X_0	X_1
L	L	ON	OFF
L	H	OFF	ON
H	X	FF	FF

Note 1 : X : Irrelevant

ON : Low impedance between X_n and X-COMMON ($n=0,1$)

OFF : High impedance between X_n and X-COMMON ($n=0,1$)

LOGIC DIAGRAM



M4053BP

M4053BFP

TRIPLE 2-CHANNEL ANALOG MULTIPLEXER/DEMULPLEXER

ABSOLUTE MAXIMUM RATINGS ($T_a = -40 \sim +85^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
$V_{DD}-V_{SS}$	Supply voltage		$-0.5 \sim 20$	V
$V_{DD}-V_{EE}$			$-0.5 \sim 20$	V
V_i	Input voltage	Control and inhibit inputs	$V_{SS} - 0.5 \sim V_{DD} + 0.5$	V
		Channel and common inputs	$V_{EE} - 0.5 \sim V_{DD} + 0.5$	
V_{IO}	Input-to-output voltage		± 0.5	V
I_i	Input current	Control and inhibit inputs	± 10	mA
I_o	Output current	Switch-off	± 10	mA
V_o	Output voltage	Channel and common outputs	$V_{EE} - 0.5 \sim V_{DD} + 0.5$	V
T_{opr}	Operating temperature range		$-40 \sim +85$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-65 \sim +150$	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITING CONDITIONS ($T_a = -40 \sim +85^\circ\text{C}$, unless otherwise noted)

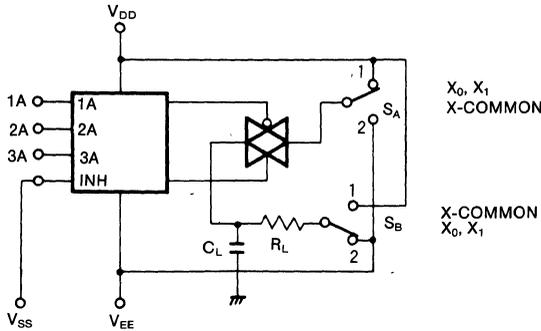
Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
$V_{DD}-V_{SS}$	Supply voltage	3		18	V
$V_{DD}-V_{EE}$		3		18	
V_i	Input voltage	V_{SS}		V_{DD}	V
	Control and inhibit inputs	V_{EE}		V_{DD}	
V_o	Output voltage	V_{EE}		V_{DD}	V

ELECTRICAL CHARACTERISTICS ($V_{SS}=0\text{V}$)

Symbol	Parameter	Test conditions	Limits								Unit	
					-40 $^\circ\text{C}$		25 $^\circ\text{C}$		85 $^\circ\text{C}$			
			$V_{EE}(\text{V})$	$V_{DD}(\text{V})$	Min	Max	Min	Typ	Max	Min		Max
V_{IH}	"H" input voltage (A, B, C, INHIBIT)	Input-to-output current= $10\mu\text{A}$	0	5	3.5		3.5			3.5		V
			0	10	7.0		7.0			7.0		
			0	15	11.0		11.0			11.0		
V_{IL}	"L" input current (A, B, C, INHIBIT)	Input-to-output current= $10\mu\text{A}$	0	5		1.5			1.5		1.5	V
			0	10		3.0			3.0		3.0	
			0	15		4.0			4.0		4.0	
R_{ON}	ON resistance	$V_i=5\text{V}$	0	5		500			600		800	Ω
		$V_i=2.5\text{V}$	0	5		850			950		1300	
		$V_i=0.25\text{V}$	0	5		500			600		800	
		$V_i=10\text{V}$	0	10		210			250		300	
		$V_i=5\text{V}$	0	10		210			250		300	
		$V_i=0.25\text{V}$	0	10		210			250		300	
		$V_i=15\text{V}$	0	15		140			160		200	
		$V_i=7.5\text{V}$	0	15		140			160		200	
		$V_i=0.25\text{V}$	0	15		140			160		200	
		$V_i=5\text{V}$	-5	5		210			250		300	
		$V_i=\pm 0.25\text{V}$	-5	5		210			250		300	
		$V_i=-5\text{V}$	-5	5		210			250		300	
		$V_i=7.5\text{V}$	-7.5	7.5		140			160		200	
$V_i=\pm 0.25\text{V}$	-7.5	7.5		140			160		200			
$V_i=-7.5\text{V}$	-7.5	7.5		140			160		200			
ΔR_{ON}	ON resistance variations between switches of the same package		-2.5	2.5				30			Ω	
			-5	5				15				
			-7.5	7.5				10				
I_{OFF}	Input-to-output off-state leakage current ($X_0 \sim X_1$ -X-COMMON)	$V_{i/o}=10\text{V}, V_{o/i}=0\text{V}$	0	10					125		nA	
		$V_{i/o}=0\text{V}, V_{o/i}=10\text{V}$	0	10					-125			
		$V_{i/o}=18\text{V}, V_{o/i}=0\text{V}$	0	18		250			250			1000
		$V_{i/o}=0\text{V}, V_{o/i}=18\text{V}$	0	18		-250			-250			-1000
I_{DD}	Quiescent supply current	$V_i=V_{DD}, V_{SS}$	0	5		20			20		150	
			0	10		40			40		300	
			0	15		80			80		600	
I_{IH}	"H" input current (A, INH)	$V_{IH}=18\text{V}$	0	18		0.3			0.3		1.0	μA
I_{IL}	"L" input current (A, INH)	$V_{iL}=0\text{V}$	0	18		-0.3			-0.3		-1.0	μA

TRIPLE 2-CHANNEL ANALOG MULTIPLEXER/DEMULTIPLEXER

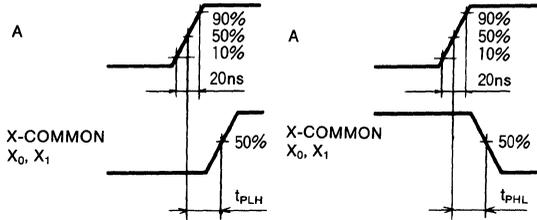
3 "L-H" and "H-L" output propagation time
(A-X₀, X₁, X-COMMON)



TIMING DIAGRAM

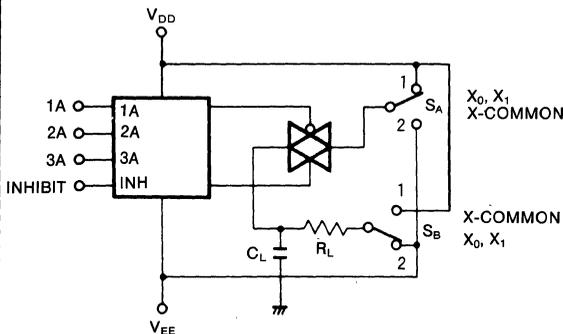
S_A=1, S_B=2

S_A=2, S_B=1



Refer to the function table for conditions of control input A.

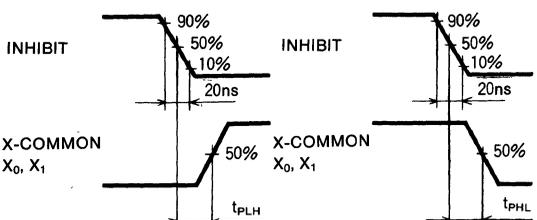
4 "L-H" and "H-L" output propagation time
(INHIBIT-X₀, X₁, X-COMMON)



TIMING DIAGRAM

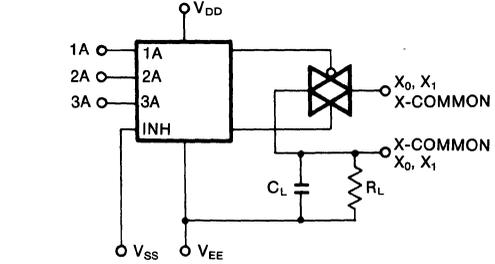
S_A=1, S_B=2

S_A=2, S_B=1

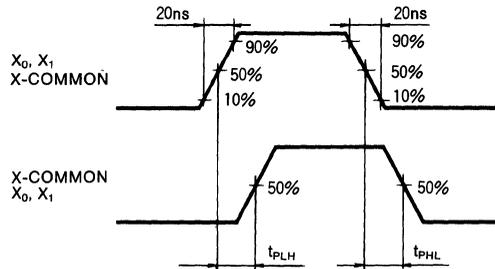


Refer to the function table for conditions of control input A.

5 "L-H" and "H-L" output propagation time
(X₀, X₁/X-COMMON-X-COMMON/X₀, X₁)

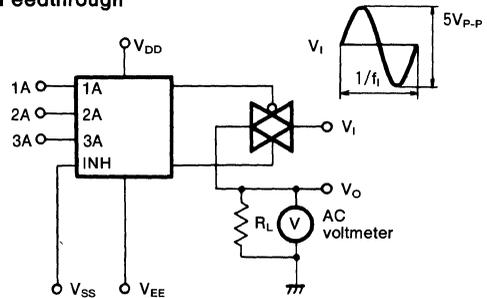


TIMING DIAGRAM



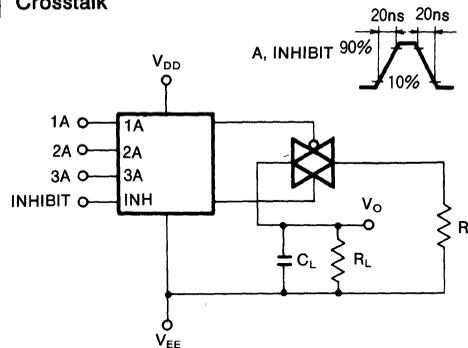
Refer to the function table for conditions of control input A.

6 Feedthrough



The feedthrough is taken as that frequency f_i at which, using a sine-wave input of $2.5V_{P-P}$, $20 \log_{10}(V_O/V_i) = -50\text{dB}$. Refer to the function table for conditions of control input A.

7 Crosstalk



Refer to the function table for conditions of control input A.

M4066BP M4066BFP

QUADRUPLE BILATERAL SWITCH

DESCRIPTION

The M4066BP is a semiconductor integrated circuit consisting of four independent bilateral analog switches.

FEATURES

- Low ON resistance: 50Ω typ. ($V_{DD}=15V$)
- High OFF resistance: $10^9\Omega$ or greater (typ)
- Small differences in ON resistance between each switch in the package: 10Ω typ. ($V_{DD}=15V$)
- Linearized transfer characteristics: 0.07% distortion (typ)
- Wide operating voltage range: $V_{DD}=3\sim 18V$
- Wide operating temperature range: $T_a=-40\sim +85^\circ C$

APPLICATION

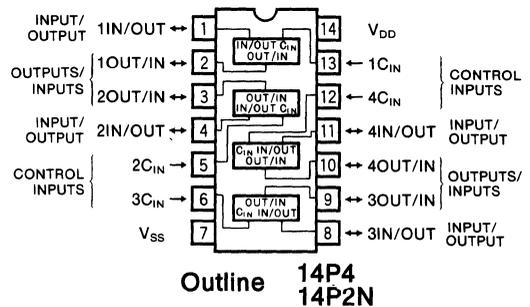
General purpose, for use in industrial and consumer digital equipment.

FUNCTIONAL DESCRIPTION

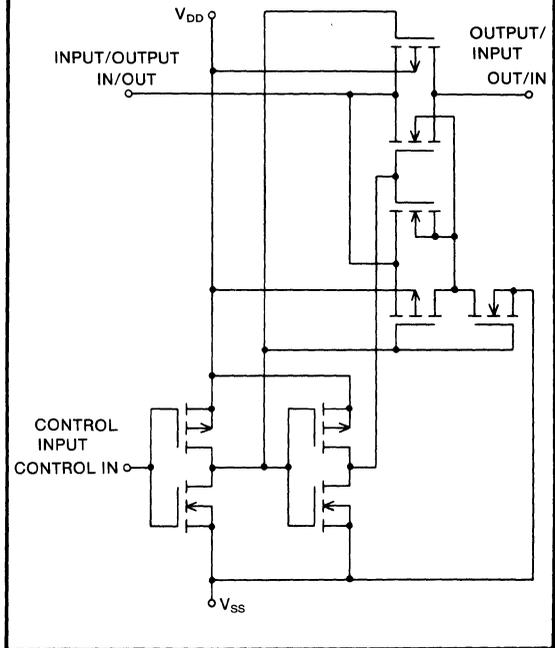
The control input (C_{IN}) can be used to change the input-to-output impedance (IN/OUT-OUT/IN) of the switches. When (C_{IN}) is made high, the input-to-output switch impedance is low and when set to low, this impedance is high. While this device is compatible with the M4016BP, the lower ON resistance and better transfer characteristics allow a larger input voltage range.

Input C_{IN}	Input/output and output/input resistance ($V_{DD}=10V, 15V$)
H	$0.5\sim 3 \times 10^2 \Omega$
L	$> 10^9 \Omega$ (typ)

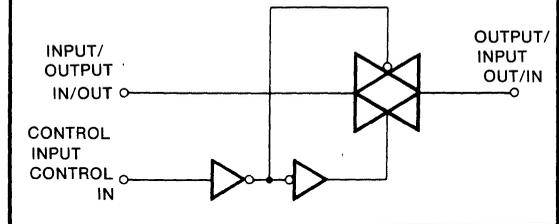
PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC (EACH SWITCH)



LOGIC DIAGRAM (EACH SWITCH)



M4066BP
M4066BFP

QUADRUPLE BILATERAL SWITCH

ABSOLUTE MAXIMUM RATINGS ($T_a = -40 \sim +85^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{DD}	Supply voltage		$V_{SS} - 0.5 \sim V_{SS} + 20$	V
V_I	Input voltage		$V_{SS} - 0.5 \sim V_{DD} + 0.5$	V
V_{IO}	Input-to-output voltage		± 0.5	V
I_I	Input current	Control inputs	± 10	mA
I_O	Output current	Switch-off	± 10	mA
T_{opr}	Operating temperature range		$-40 \sim +85$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-65 \sim +150$	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a = -40 \sim +85^\circ\text{C}$, $V_{SS} = 0\text{V}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{DD}	Supply voltage	3		18	V
V_I	Input voltage	0		V_{DD}	V

ELECTRICAL CHARACTERISTICS

Symbol	Parameter	Test conditions	Limits								Unit	
					-40 $^\circ\text{C}$		25 $^\circ\text{C}$			85 $^\circ\text{C}$		
			$V_{SS}(\text{V})$	$V_{DD}(\text{V})$	Min	Max	Min	Typ	Max	Min		Max
V_{IH}	"H" input voltage (C_{IN})	Input-to-output current = 10 μA	0	5	3.5		3.5			3.5		V
			0	10	7.0		7.0			7.0		
			0	15	11.0		11.0			11.0		
V_{IL}	"L" input current (C_{IN})	Input-to-output current = 10 μA	0	5		1.5			1.5		1.5	V
			0	10		2.0			2.0		2.0	
			0	15		2.5			2.5		2.5	
R_{ON}	ON resistance	$V_I = 5\text{V}$	0	5		500			600		800	Ω
		$V_I = 2.5\text{V}$	0	5		850			950		1300	
		$V_I = 0.25\text{V}$	0	5		500			600		800	
		$V_I = 10\text{V}$	0	10		210			250		300	
		$V_I = 5\text{V}$	0	10		210			250		300	
		$V_I = 0.25\text{V}$	0	10		210			250		300	
		$V_I = 15\text{V}$	0	15		140			160		200	
		$V_I = 7.5\text{V}$	0	15		140			160		200	
		$V_I = 0.25\text{V}$	0	15		140			160		200	
		$V_I = 5\text{V}$	-5	5		210			250		300	
		$V_I = \pm 0.25\text{V}$	-5	5		210			250		300	
		$V_I = -5\text{V}$	-5	5		210			250		300	
		$V_I = 7.5\text{V}$	-7.5	7.5		140			160		200	
$V_I = \pm 0.25\text{V}$	-7.5	7.5		140			160		200			
$V_I = -7.5\text{V}$	-7.5	7.5		140			160		200			
ΔR_{ON}	ON resistance variations between switches of the same package		-2.5	2.5				30			Ω	
			-5	5				15				
			-7.5	7.5				10				
I_{OFF}	Input/output off-state leakage current	$V_{IO} = 10\text{V}$, $V_{OI} = 0\text{V}$	0	10					125		nA	
		$V_{IO} = 0\text{V}$, $V_{OI} = 10\text{V}$	0	10					-125			
		$V_{IO} = 18\text{V}$, $V_{OI} = 0\text{V}$	0	18		250			250			1000
		$V_{IO} = 0\text{V}$, $V_{OI} = 18\text{V}$	0	18		-250			-250			-1000
I_{DD}	Quiescent supply current	$V_I(C_{IN}) = V_{DD}$, V_{SS}	0	5		1			1		7.5	μA
			0	10		2			2		15	
			0	15		4			4		30	
I_{IH}	"H" input current (C_{IN})	$V_I = 18\text{V}$	0	18		0.3			0.3		1.0	μA
I_{IL}	"L" input current (C_{IN})	$V_{IL} = 0\text{V}$	0	18		-0.3			-0.3		-1.0	μA

M4066BP
M4066BFP

QUADRUPLE BILATERAL SWITCH

SWITCHING CHARACTERISTICS ($T_a=25^\circ\text{C}$)

Symbol	Parameter	Test conditions	Limits			Unit			
			V_{SS} (V)	V_{DD} (V)	Min		Typ	Max	
$f_{\text{max}}(I/O)$	Maximum transfer frequency	$R_L=10\text{k}\Omega$ $C_L=15\text{pF}$ Test circuit 2	-5	5		25		MHz	
$f_{\text{max}}(C_{IN})$	Maximum control frequency	$R_L=300\Omega$ $C_L=15\text{pF}$ Test circuit 3	0	5		6		MHz	
			0	10		18			
			0	15		22			
t_{PLH}	"L-H" and "H-L" output propagation time (IN/OUT-OUT/IN)	$R_L=10\text{k}\Omega$ $C_L=50\text{pF}$ Test circuit 4	0	5			45	ns	
				0	10				30
				0	15				20
t_{PHL}			0	5			45	ns	
				0	10				30
				0	15				20
t_{PLH}	"L-H" and "H-L" output propagation time (CONTROL IN-OUT/IN)	$R_L=10\text{k}\Omega$ $C_L=50\text{pF}$ Test circuit 5	0	5			200	ns	
				0	10				70
				0	15				60
t_{PHL}			0	5			200	ns	
				0	10				70
				0	15				60
—	Sine-wave distortion	$R_L=10\text{k}\Omega$ $f_i=1\text{kHz}$ Test circuit 2	-5	5		0.07		%	
—	Feedthrough (switch off)	$R_L=1\text{k}\Omega$ Test circuit 6	-5	5		500		kHz	
—	Crosstalk (CONTROL IN-OUT/IN)	$R_L=1\text{k}\Omega$ $R_L=10\text{k}\Omega$ $C_L=15\text{pF}$ Test circuit 7	0	5		200		mV	
				0	10		300		
				0	15		400		
C_i	Input capacitance	Control input					7.5	pF	
		Switch input/output				10			

TEST CIRCUITS

1 ON resistance (R_{ON})

$$R_{ON} = 10 \times \frac{(V_I - V_O)}{V_O} \text{ [k}\Omega\text{]}$$

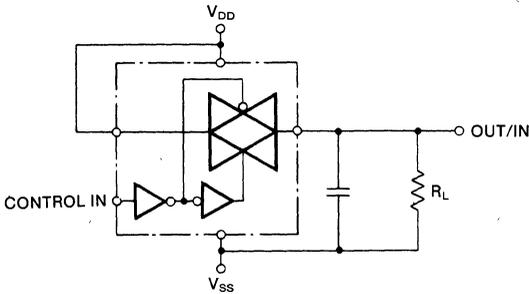
2 Maximum transfer frequency ($f_{\text{max}}(I/O)$)
Sine-wave distortion

$f_{\text{max}}(I/O)$ is taken as that frequency f_i at which, using a sine-wave input of $2.5V_{P-P}$, $20 \log_{10}(V_O/V_I) = -3\text{dB}$

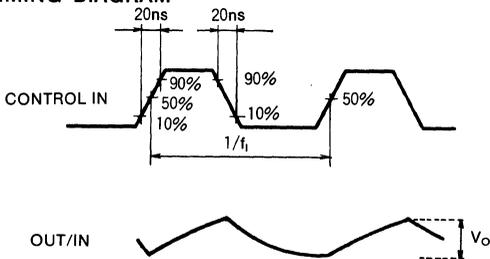
M4066BP
M4066BFP

QUADRUPLE BILATERAL SWITCH

3 Maximum control frequency ($f_{max}(C_{IN})$)

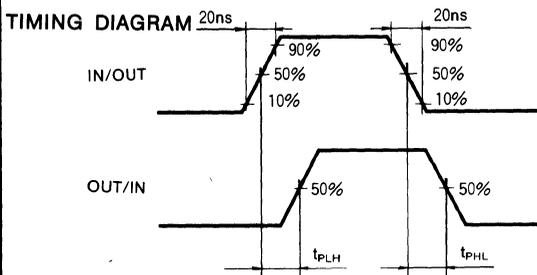
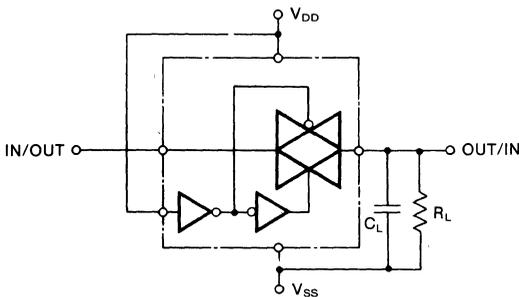


TIMING DIAGRAM

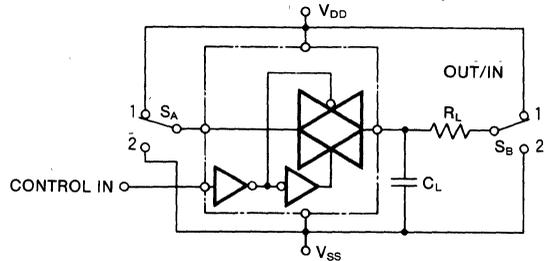


$f_{max}(C_{IN})$ is taken as that frequency f_i at which the output amplitude (V_o) is 1/2 that at 1kHz

4 "L-H" and "H-L" output propagation time (IN/OUT-OUT/IN)



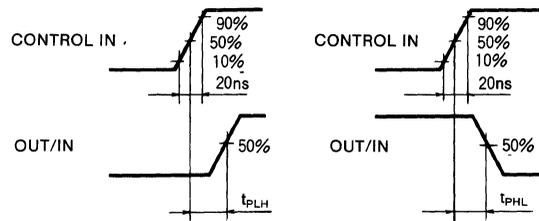
5 "L-H" and "H-L" output propagation time (CONTROL IN-OUT/IN)



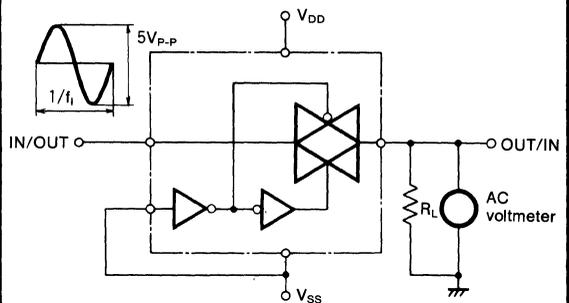
$S_A=1, S_B=2$

$S_A=2, S_B=1$

TIMING DIAGRAM

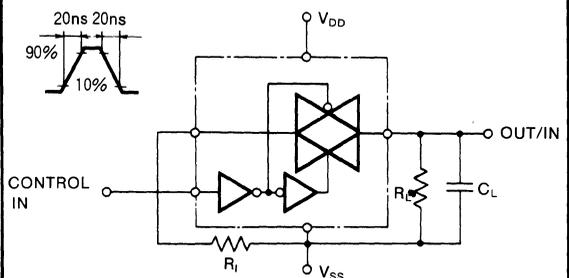


6 Feedthrough



The feedthrough is taken as that frequency f_i at which, using a sine-wave input of $2.5V_{p-p}$, $20 \log_{10}(V_o/V_i) = -50\text{dB}$.

7 Crosstalk



M51320P

ANALOG SWITCH

DESCRIPTION

The M51320P is a semiconductor integrated circuit containing an analog switch designed for use in a video system. It contains two audio switches and one video switch. One of the audio switches has two inputs and the other, three. The video switch has two inputs. Each switch can be independently controlled. In addition, the video switch contains an amplifier with a gain of about 6.7dB.

FEATURES

- Video and stereo sound switches in one package
- Wide frequency range (video switch) DC~10MHz
- High separation (video) Crosstalk 55dB(typ.) @5MHz

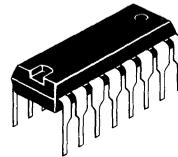
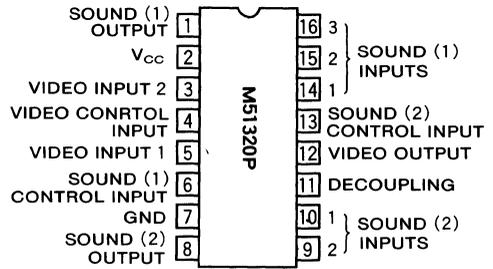
APPLICATION

Video equipment.

RECOMMENDED OPERATING CONDITIONS

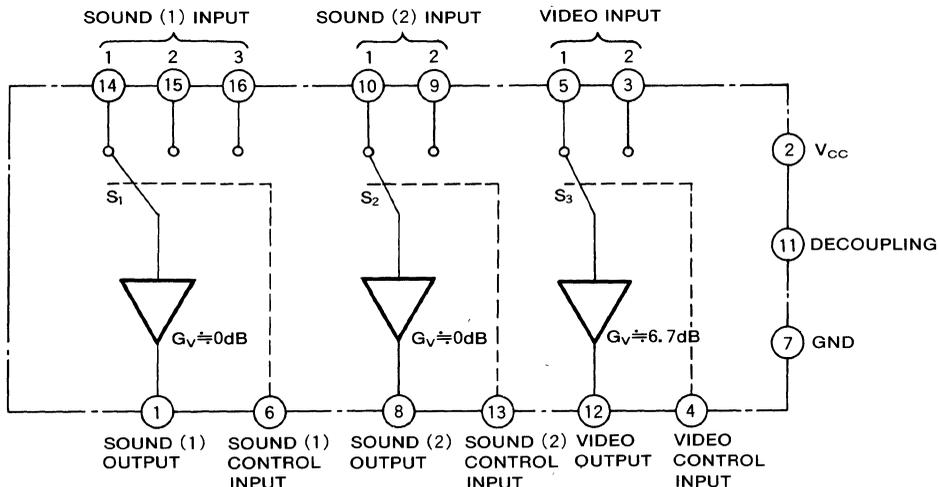
Supply voltage range 5~14V

PIN CONFIGURATION (TOP VIEW)



16-pin molded plastic DIP

BLOCK DIAGRAM



ANALOG SWITCH

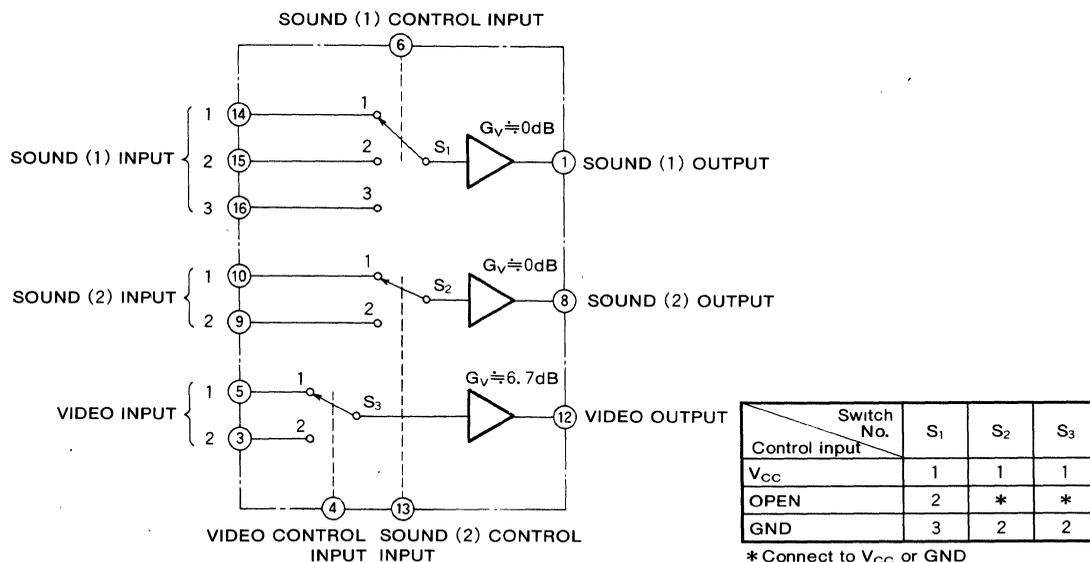
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		14	V
V_{IS}	Input signal voltage	Sound	6	V
		Video	6	
V_{IC}	Input control voltage		V_{CC}	V
P_d	Power dissipation		1.25	W
K_θ	Thermal derating		12.5	mW/ $^\circ\text{C}$
T_{opr}	Operating temperature range		-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-40~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

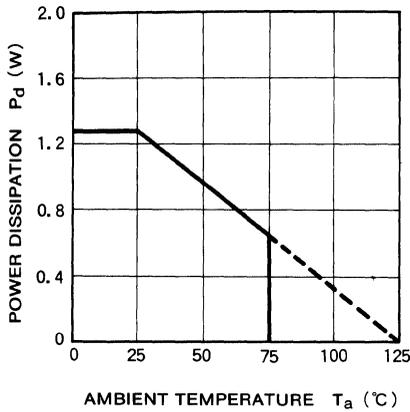
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current			40	48	mA
$V_{I,DC}$	Input bias voltage		3.8	4.2	4.6	V
$V_{O,DC}$	Output bias voltage	Video	5.0	5.6	6.2	V
		Sound	3.0	3.6	4.2	
V_{OP}	Output DC offset voltage			15	100	mV
$V_{IC,H}$	Control-pin threshold voltage	Sound (1) (Pin ⑥ 3-state input)	7.0	8.0	9.0	V
$V_{IC,L}$		Sound (1) (Pin ⑥ 3-state input)	3.0	4.0	5.0	
V_{IC}		Sound (2) and video (Pins ④, ⑬)	1.7	2.1	2.5	
G_V	Voltage gain	Video, $f=1\text{MHz}$	5.7	6.7	7.7	dB
		Sound, $f=1\text{kHz}$	-0.5	-0.1		
THD	Total harmonic distortion	Sound, $f=1\text{kHz}$, $V_O=1\text{Vrms}$		0.02	0.2	%
V_N	Output noise voltage	Sound, $R_g=600\Omega$, $BW=15\text{kHz}$		3	50	μVrms
		Video, $R_g=75\Omega$, $BW=10\text{MHz}$		0.5	1.0	mVrms
CT	Crosstalk	Sound, $f=1\text{kHz}$	65	80		dB
		Video, $f=5\text{MHz}$	45	55		

SWITCH MODE

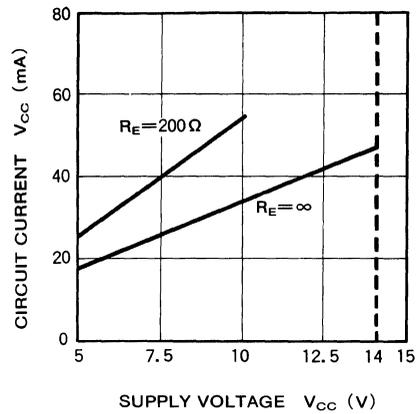


TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

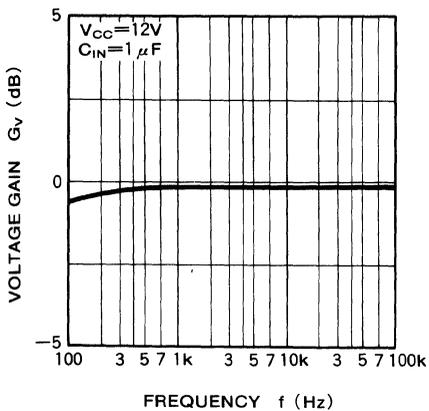
THERMAL DERATING (MAXIMUM RATING)



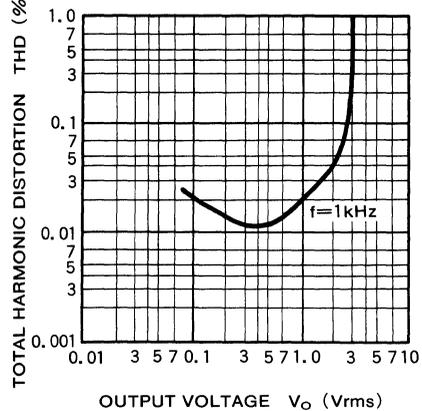
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



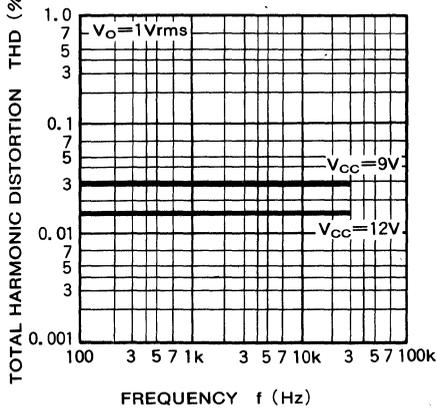
VOLTAGE GAIN VS. FREQUENCY (SOUND)



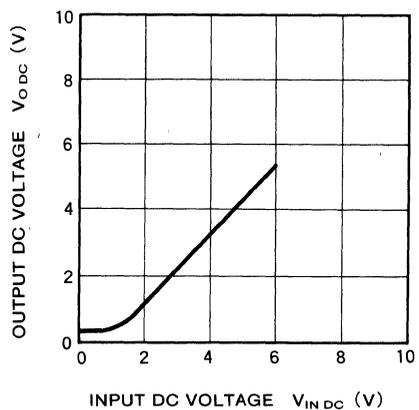
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (SOUND)



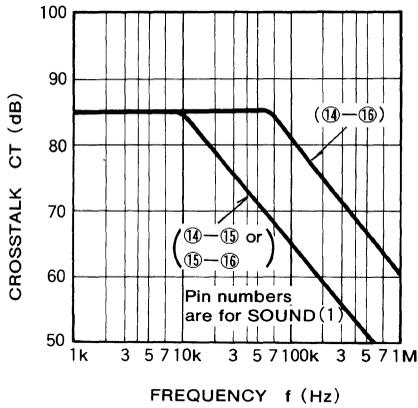
TOTAL HARMONIC DISTORTION VS. FREQUENCY (SOUND)



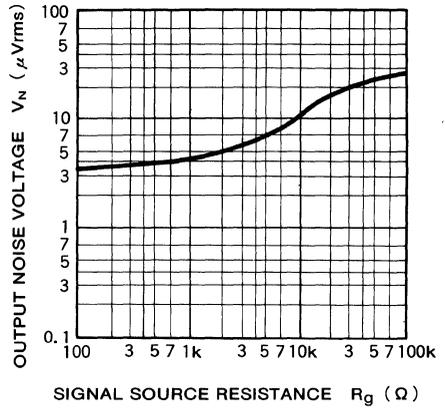
OUTPUT DC VOLTAGE VS. INPUT DC VOLTAGE (SOUND)



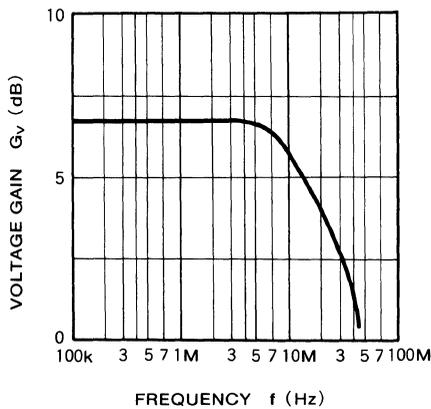
CROSSTALK VS. FREQUENCY (SOUND)



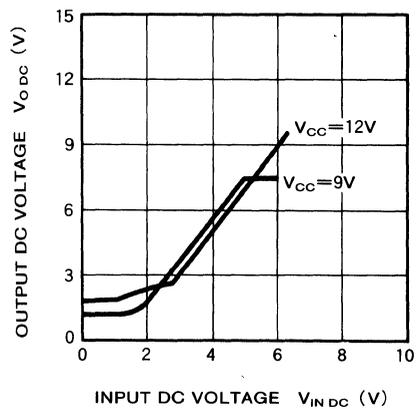
OUTPUT NOISE VOLTAGE VS. SIGNAL SOURCE RESISTANCE (SOUND)



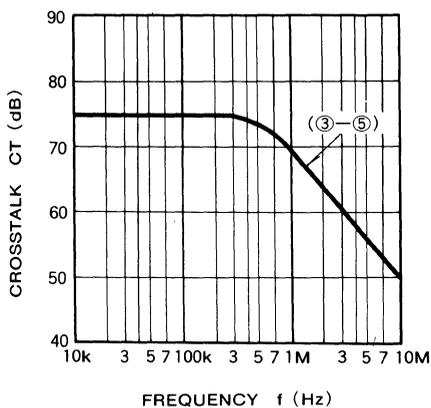
VOLTAGE GAIN VS. FREQUENCY (VIDEO)



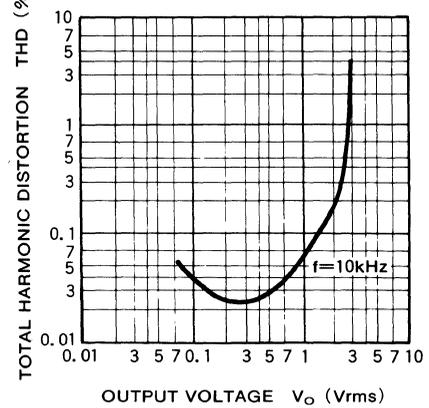
OUTPUT DC VOLTAGE VS. INPUT DC VOLTAGE (VIDEO)



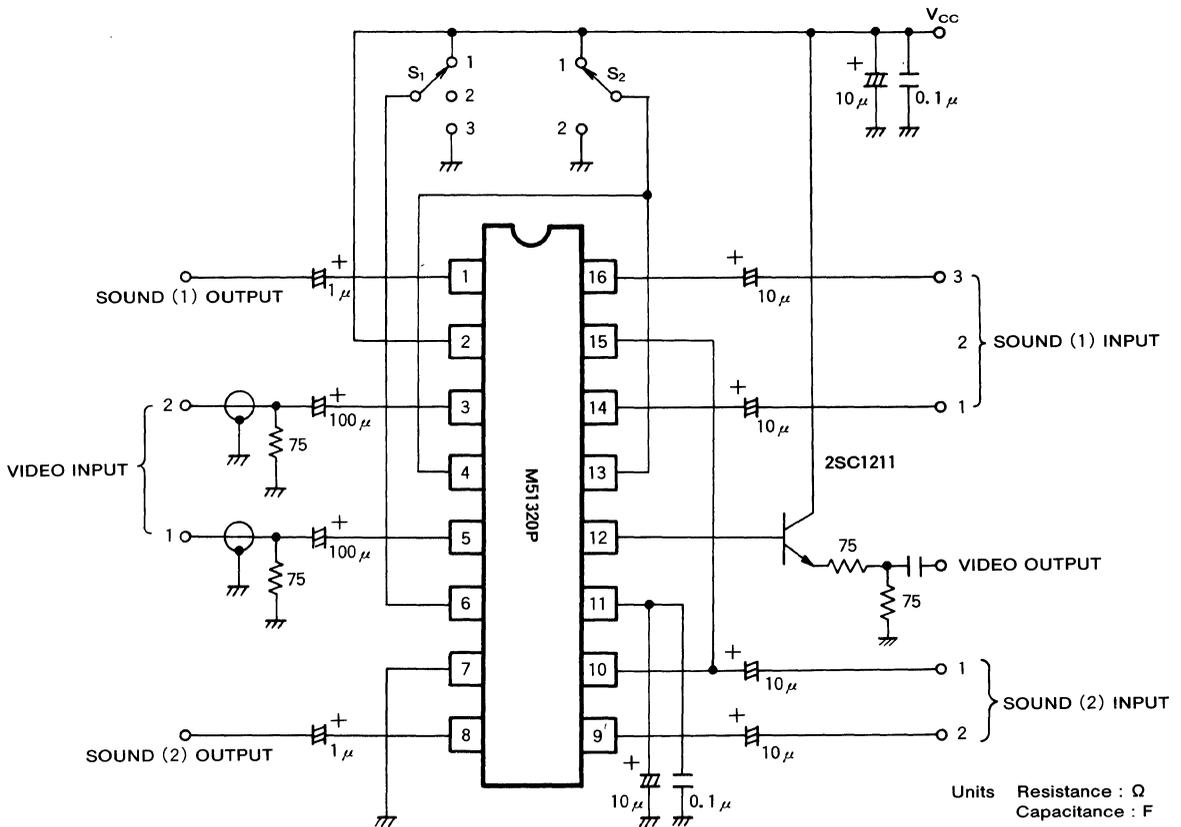
CROSSTALK VS. FREQUENCY (VIDEO)



TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (VIDEO)



APPLICATION EXAMPLE

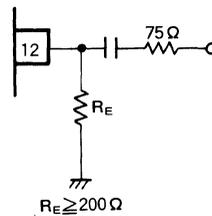


PRECAUTIONS FOR USE

When the video output is sent through a 75 Ω output and the supply voltage of 12V is used, and external transistor buffer such as the one shown in the above diagram should be connected to maintain a balance with the power dissipation of the package.

Should 9V be the sole supply voltage, a direct drive arrangement can be used as in the diagram at right by connecting resistor R_E between the output pin and ground.

Since an emitter-follower output is used in the video and sound outputs, when the external wiring is long or a capacitive load is connected, a resistor with a value of several tens of ohms should be connected at a position near the output pin.



MITSUBISHI LINEAR ICs M51321P

ANALOG SWITCH

DESCRIPTION

The M51321P is a semiconductor integrated circuit containing an analog switch designed for use in a video system. It contains two audio switches and one video switch. Each switch has three inputs and can be independently controlled. In addition, the video switch contains an amplifier with a gain of about 6.7dB.

FEATURES

- Video and stereo sound switches in one package
- Wide frequency range (video switch) DC~10MHz
- High separation (video) ... Crosstalk 55dB(typ.) (@5MHz)

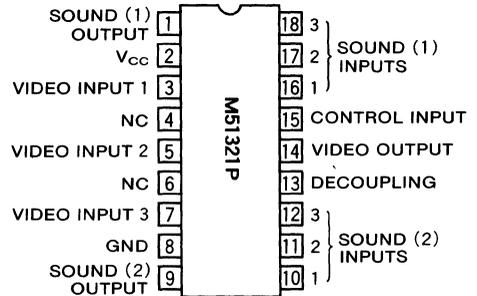
APPLICATION

Video equipment.

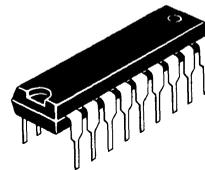
RECOMMENDED OPERATING CONDITIONS

Supply voltage range 5~14V

PIN CONFIGURATION (TOP VIEW)

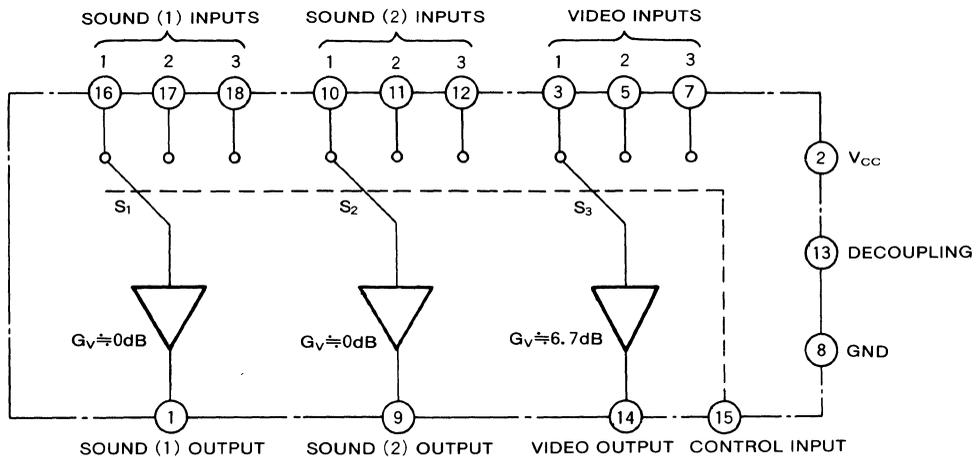


NC : NO CONNECTION



18-pin molded plastic DIP

BLOCK DIAGRAM



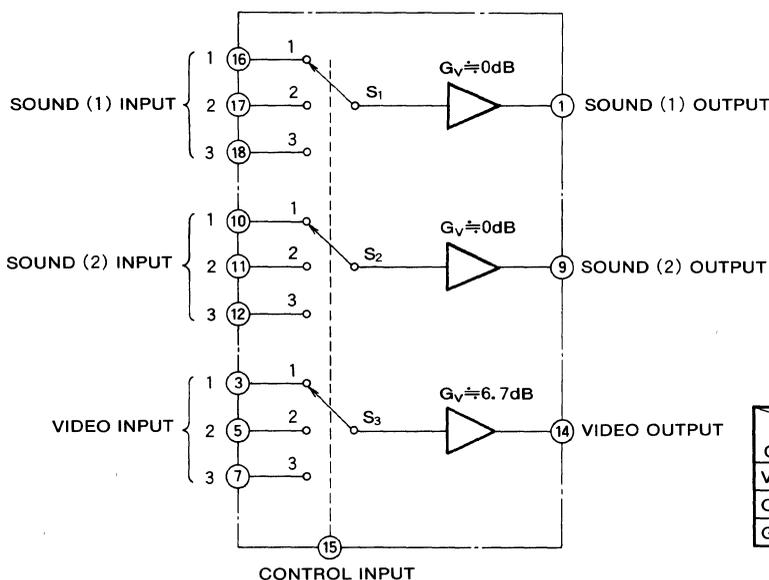
ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		14	V
V_{IS}	Input signal voltage	Sound	6	V
		Video	6	
V_{IC}	Input control voltage		V_{CC}	V
P_d	Power dissipation		1.25	W
K_{θ}	Thermal derating		12.5	mW/ $^{\circ}\text{C}$
T_{opr}	Operating temperature range		$-20\sim+75$	$^{\circ}\text{C}$
T_{stg}	Storage temperature range		$-40\sim+125$	$^{\circ}\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current			39	48	mA
V_{IDC}	Input bias voltage		3.8	4.2	4.6	V
V_{ODC}	Output bias voltage	Video	5.0	5.6	6.2	V
		Sound	3.0	3.6	4.2	
V_{OP}	Output DC offset voltage			15	100	mV
$V_{IC H}$	Control-pin threshold voltage		7.0	8.0	9.0	V
$V_{IC L}$			3.0	4.0	5.0	V
G_V	Voltage gain	Video, $f=1\text{MHz}$	5.7	6.7	7.7	dB
		Sound, $f=1\text{kHz}$	-0.5	-0.1		
THD	Total harmonic distortion	Sound, $f=1\text{kHz}$, $V_O=1\text{Vrms}$		0.02	0.2	%
V_N	Output noise voltage	Sound, $R_g=600\Omega$, $BW=15\text{kHz}$		3	50	μVrms
		Video, $R_g=75\Omega$, $BW=10\text{MHz}$		0.5	1.0	mVrms
CT	Crosstalk	Sound, $f=1\text{kHz}$	65	80		dB
		Video, $f=5\text{MHz}$	45	55		

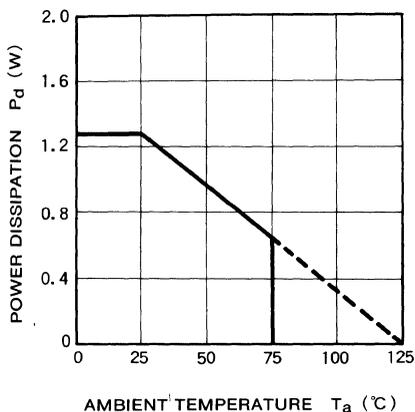
SWITCH MODE



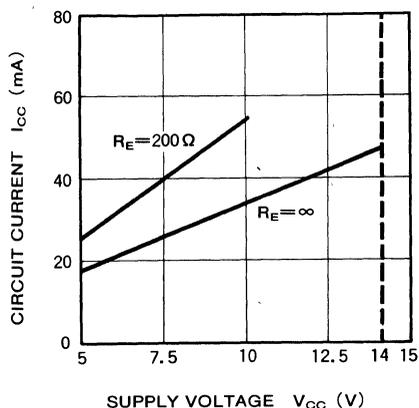
Switch No.	S_1	S_2	S_3
Control input			
V_{CC}	1	1	1
OPEN	2	2	2
GND	3	3	3

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

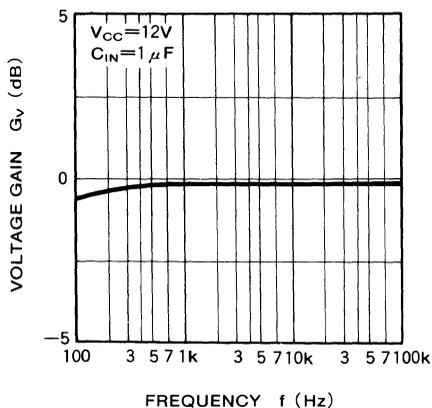
THERMAL DERATING (MAXIMUM RATING)



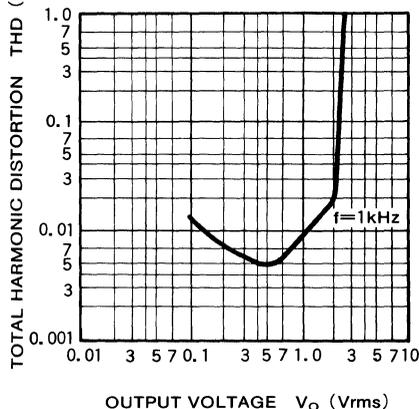
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



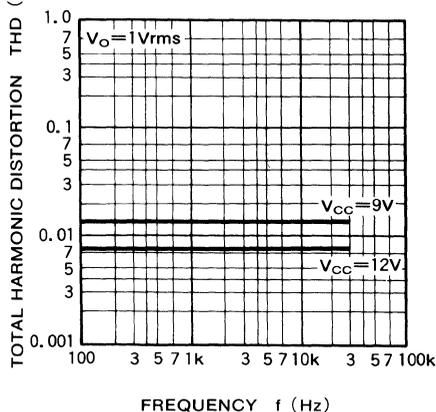
VOLTAGE GAIN VS. FREQUENCY (SOUND)



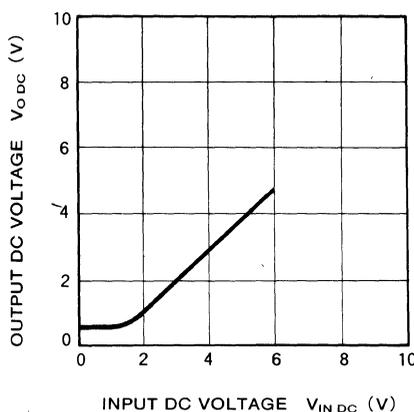
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (SOUND)



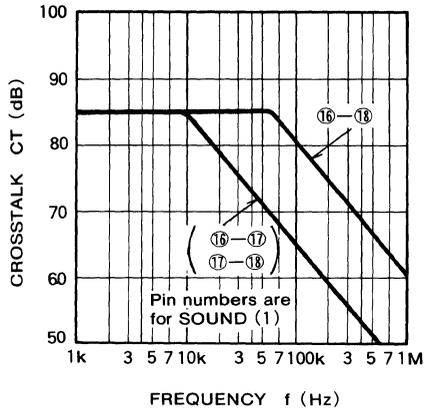
TOTAL HARMONIC DISTORTION VS. FREQUENCY (SOUND)



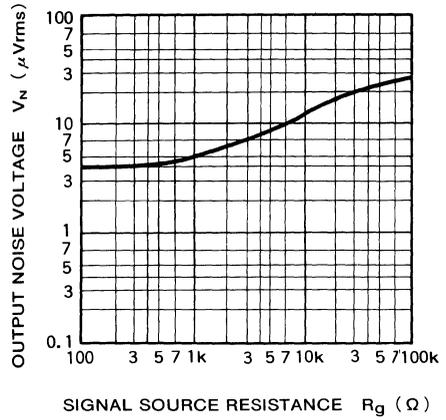
OUTPUT DC VOLTAGE VS. INPUT DC VOLTAGE (SOUND)



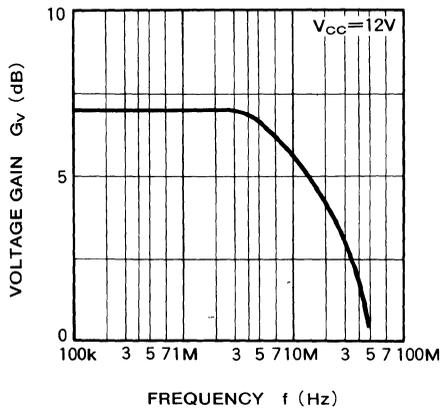
CROSSTALK VS. FREQUENCY (SOUND)



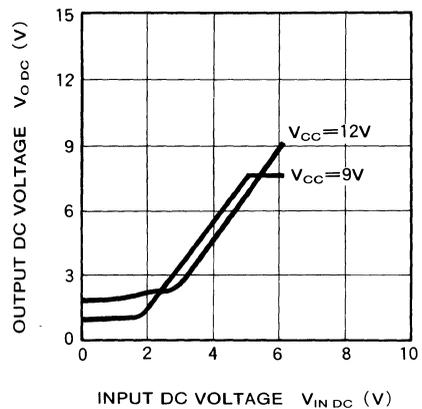
OUTPUT NOISE VOLTAGE VS. SIGNAL SOURCE RESISTANCE (SOUND)



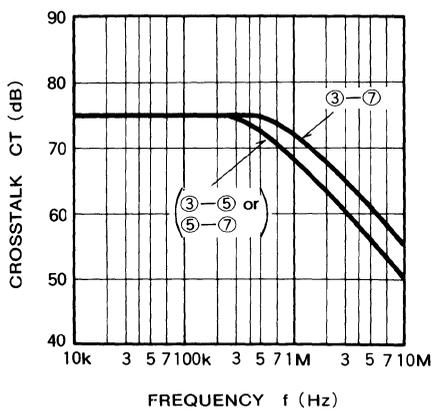
VOLTAGE GAIN VS. FREQUENCY (VIDEO)



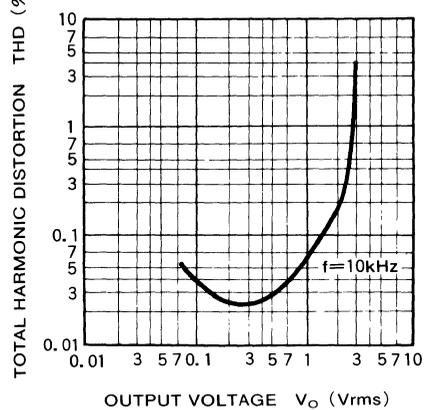
OUTPUT DC VOLTAGE VS. INPUT DC VOLTAGE (VIDEO)



CROSSTALK VS. FREQUENCY (VIDEO)



TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (VIDEO)



M51326P

ANALOG SWITCH

DESCRIPTION

The M51326P is a semiconductor integrated circuit containing an analog switch designed for use in a video system. It contains two audio switches and one video switch. One of the audio switches has two inputs and the other, three. The video switch has two inputs. Each switch can be independently controlled.

FEATURES

- Video and stereo sound switches in one package
- Wide frequency range (video switch) DC~10MHz
- High separation (video) Crosstalk 55dB(typ.) @5MHz

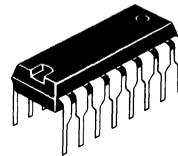
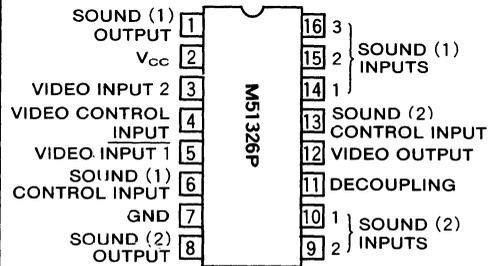
APPLICATION

Video equipment.

RECOMMENDED OPERATING CONDITIONS

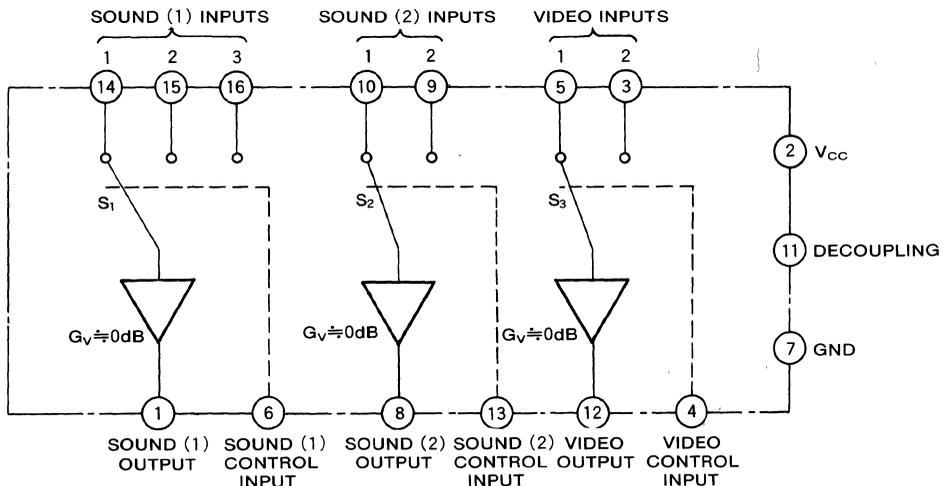
Supply voltage range 5~14V

PIN CONFIGURATION (TOP VIEW)



16-pin molded plastic DIP

BLOCK DIAGRAM



ANALOG SWITCH

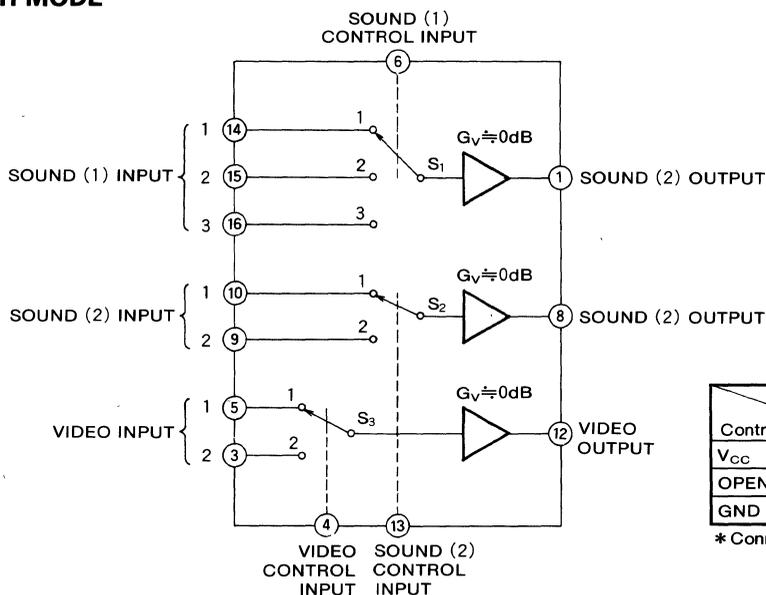
ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		14	V
V_{IS}	Input signal voltage		6	V
V_{IC}	Input control voltage		V_{CC}	V
P_d	Power dissipation		1.25	W
K_{θ}	Thermal derating		12.5	mW/ $^{\circ}\text{C}$
T_{opr}	Operating temperature range		-20~+75	$^{\circ}\text{C}$
T_{stg}	Storage temperature range		-40~+125	$^{\circ}\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current			28	36	mA
$V_{I\ DC}$	Input bias voltage		3.8	4.2	4.6	V
$V_{O\ DC}$	Output bias voltage		3.0	3.6	4.2	V
V_{OP}	Output DC offset voltage			15	100	mV
$V_{IC\ H}$	Control-pin threshold voltage	Sound (1) (pin ⑥ 3-state input)	7.0	8.0	9.0	V
$V_{IC\ L}$		Sound (1) (pin ⑥ 3-state input)	3.0	4.0	5.0	V
V_{IC}		Sound (2) and video (pins ④, ⑬)	1.7	2.1	2.5	V
G_V	Voltage gain	$f=1\text{kHz}$	-0.5	-0.1		dB
THD	Total harmonic distortion	Sound, $f=1\text{kHz}$, $V_o=1\text{Vrms}$		0.02	0.2	%
V_N	Output noise voltage	Sound, $R_g=600\ \Omega$, $BW=15\text{kHz}$		3	50	μVrms
		Video, $R_g=75\ \Omega$, $BW=10\text{MHz}$		0.5	1.0	mVrms
CT	Crosstalk	Sound, $f=1\text{kHz}$	65	80		dB
		Video, $f=5\text{MHz}$	45	55		

SWITCH MODE

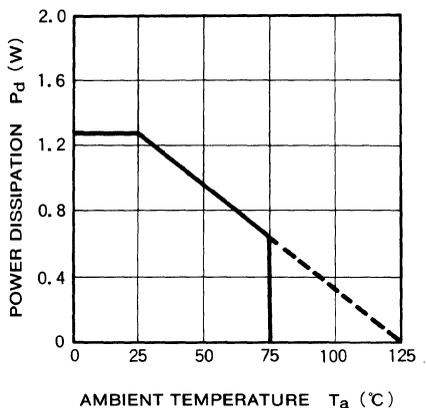


Control input	Switch No.		
	S ₁	S ₂	S ₃
V_{CC}	1	1	1
OPEN	2	*	*
GND	3	2	2

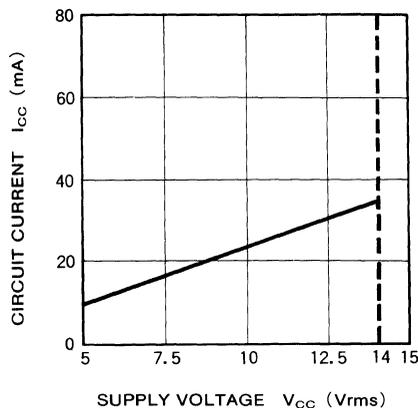
* Connect to V_{CC} or GND

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

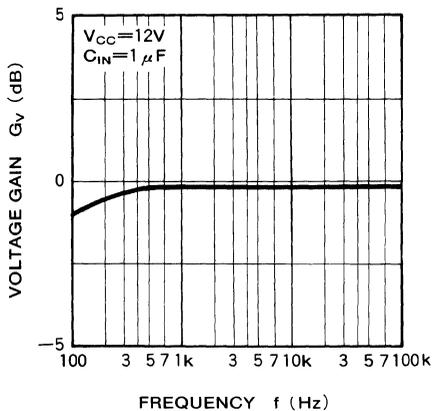
THERMAL DERATING (MAXIMUM RATING)



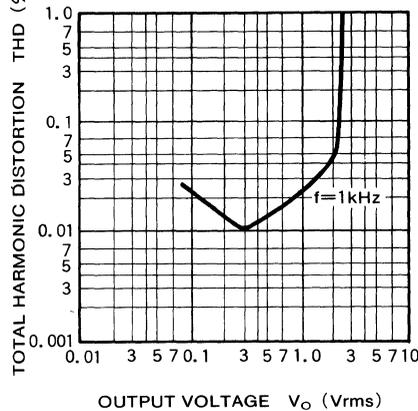
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



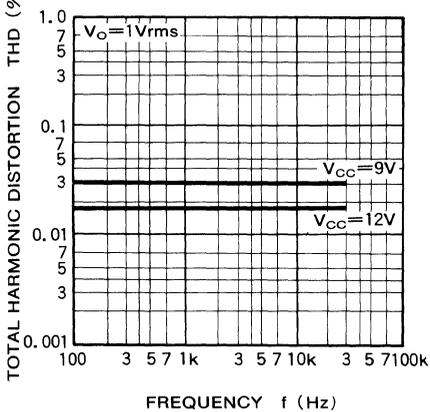
VOLTAGE GAIN VS. FREQUENCY (SOUND)



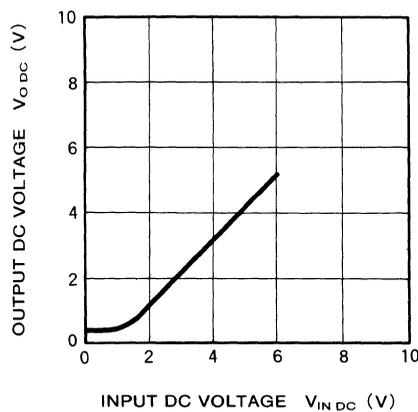
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (SOUND)



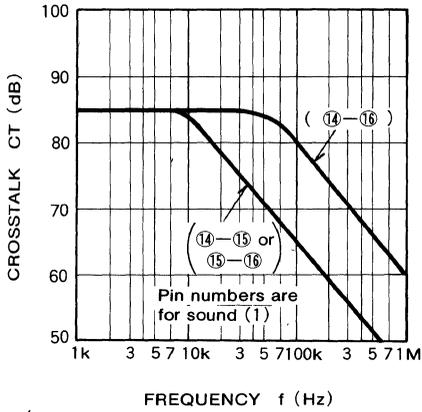
TOTAL HARMONIC DISTORTION VS. FREQUENCY (SOUND)



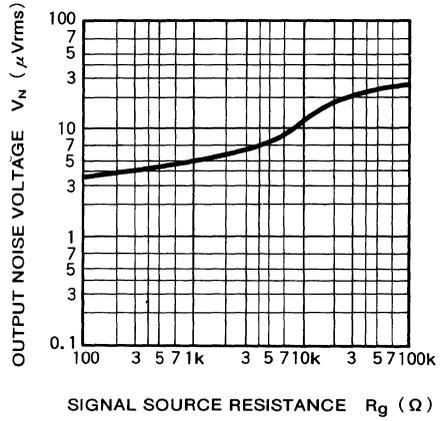
OUTPUT DC VOLTAGE VS. INPUT DC VOLTAGE (SOUND)



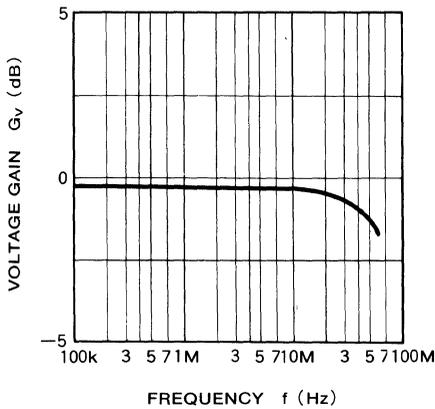
CROSSTALK VS. FREQUENCY (SOUND)



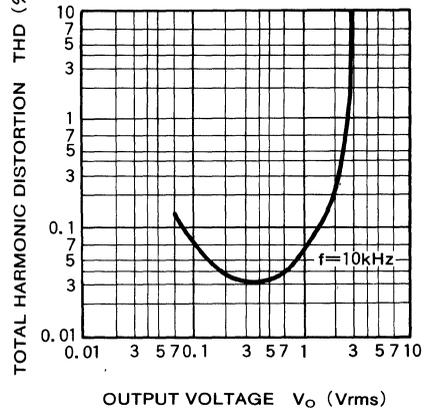
OUTPUT NOISE VOLTAGE VS SIGNAL SOURCE RESISTANCE (SOUND)



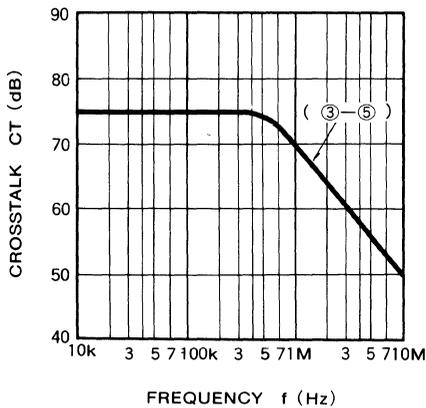
VOLTAGE GAIN VS. FREQUENCY (VIDEO)



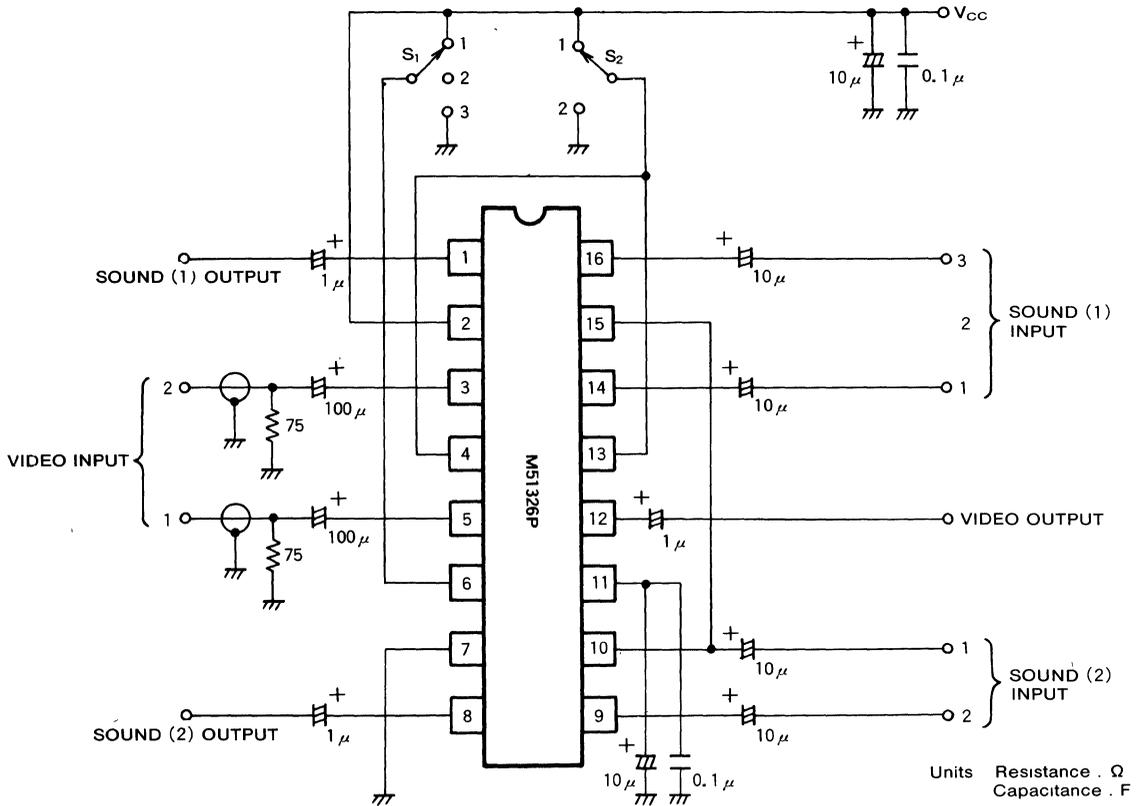
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (VIDEO)



CROSSTALK VS. FREQUENCY (VIDEO)



APPLICATION EXAMPLE



PRECAUTIONS FOR USE

Since an emitter-follower output is used in the video and sound outputs, when the external wiring is long or a capacitive load is connected, a resistor with a value of several tens of ohms should be connected at a position near the output pin.

MITSUBISHI LINEAR ICs M51327P

ANALOG SWITCH

DESCRIPTION

The M51327P is a semiconductor integrated circuit containing an analog switch designed for use in a video system. It contains two audio switches and one video switch. Each switch has three inputs and can be independently controlled.

FEATURES

- Video and stereo sound switches in one package
- Wide frequency range (video switch) DC~10MHz
- High separation (video) Crosstalk 55dB(typ.) @5MHz

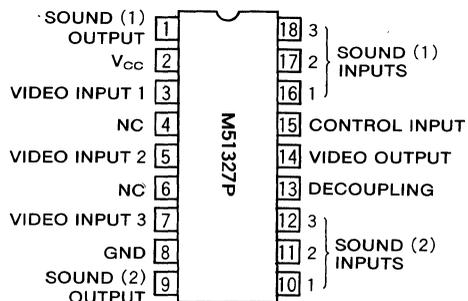
APPLICATION

Video equipment.

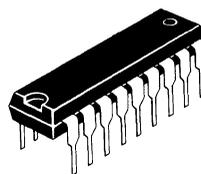
RECOMMENDED OPERATING CONDITIONS

Supply voltage range 5~14V

PIN CONFIGURATION (TOP VIEW)

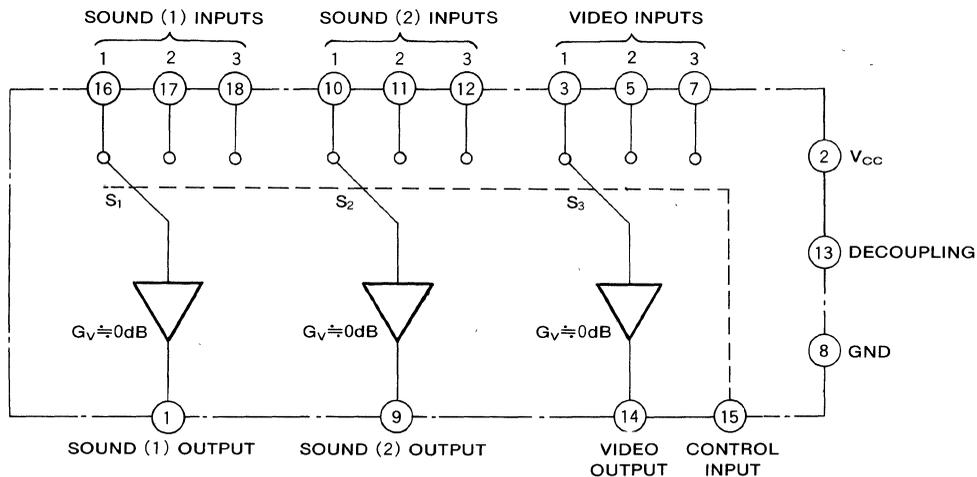


NC NO CONNECTION



18-pin molded plastic DIP

BLOCK DIAGRAM



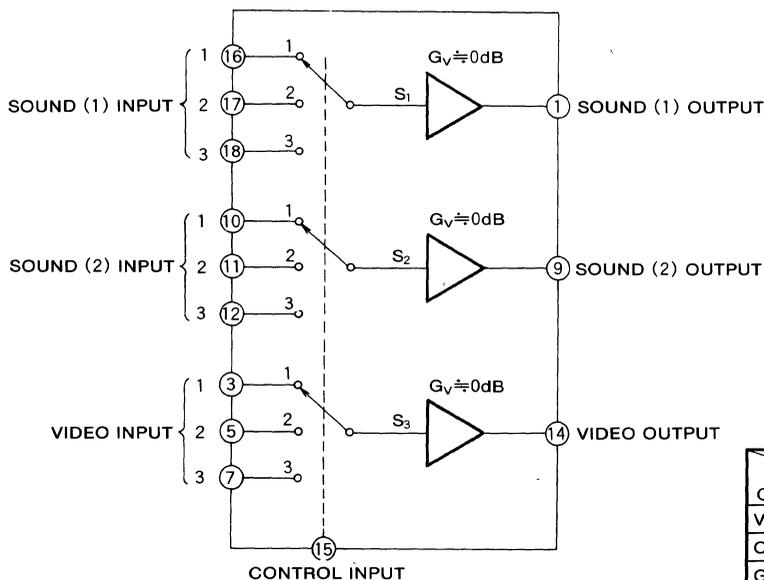
ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		14	V
V_{IS}	Input signal voltage		6	V
V_{IC}	Input control voltage		V_{CC}	V
P_d	Power dissipation		1.25	W
K_{θ}	Thermal derating		12.5	mW/ $^{\circ}\text{C}$
T_{opr}	Operating temperature range		-20~+75	$^{\circ}\text{C}$
T_{stg}	Storage temperature range		-40~+125	$^{\circ}\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current			28	36	mA
$V_{I\ DC}$	Input bias voltage		3.8	4.2	4.6	V
$V_{O\ DC}$	Output bias voltage		3.0	3.6	4.2	V
V_{OP}	Output DC offset voltage			15	100	mV
$V_{IC\ H}$	Control-pin threshold voltage		7.0	8.0	9.0	V
$V_{IC\ L}$			3.0	4.0	5.0	V
G_V	Voltage gain	Sound, $f=1\text{kHz}$	-0.5	-0.1		dB
THD	Total harmonic distortion	Sound, $f=1\text{kHz}$, $V_O=1\text{Vrms}$		0.02	0.2	%
V_N	Output noise voltage	Sound, $R_g=600\Omega$, $BW=15\text{kHz}$		3	50	μVrms
		Video, $R_g=75\Omega$, $BW=10\text{MHz}$		0.5	1.0	mVrms
CT	Crosstalk	Sound, $f=1\text{kHz}$	65	80		dB
		Video, $f=5\text{MHz}$	45	55		

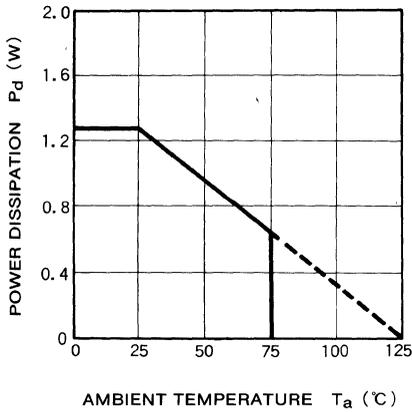
SWITCH MODE



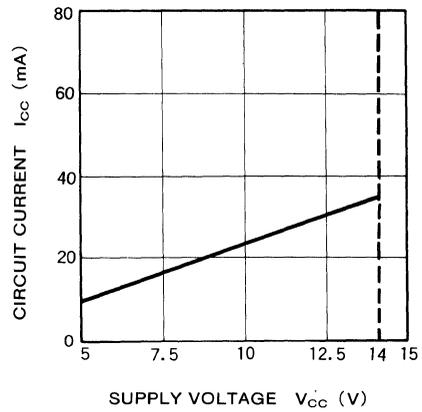
Control input \ Switch No.	S ₁	S ₂	S ₃
V_{CC}	1	1	1
OPEN	2	2	2
GND	3	3	3

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

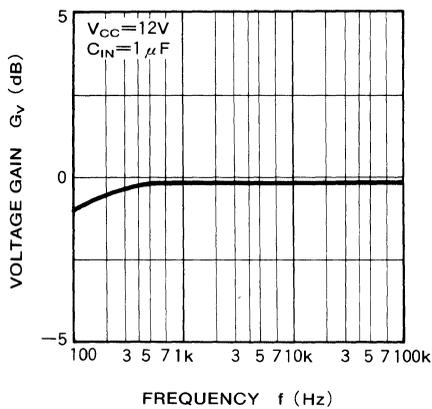
THERMAL DERATING (MAXIMUM RATING)



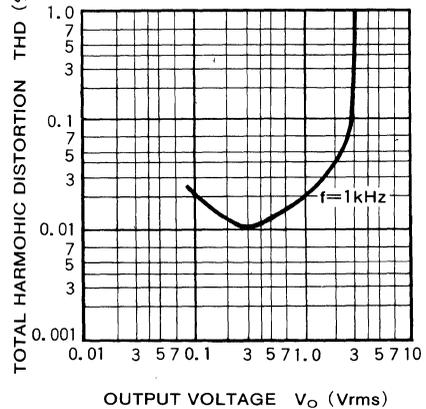
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



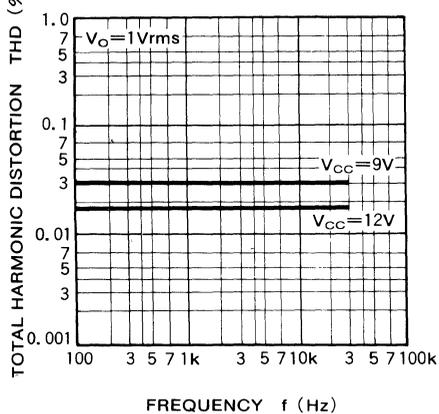
VOLTAGE GAIN VS. FREQUENCY (SOUND)



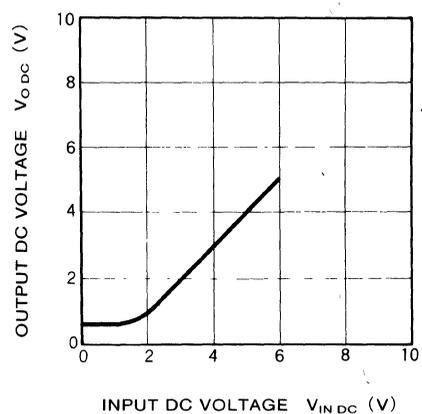
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (SOUND)



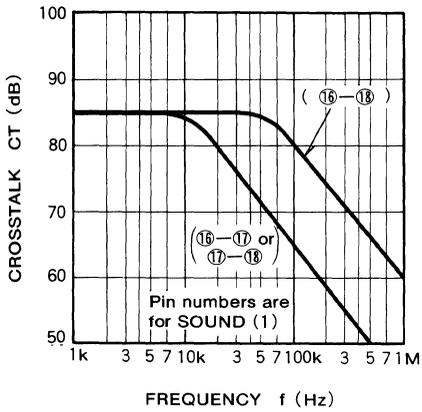
TOTAL HARMONIC DISTORTION VS. FREQUENCY (SOUND)



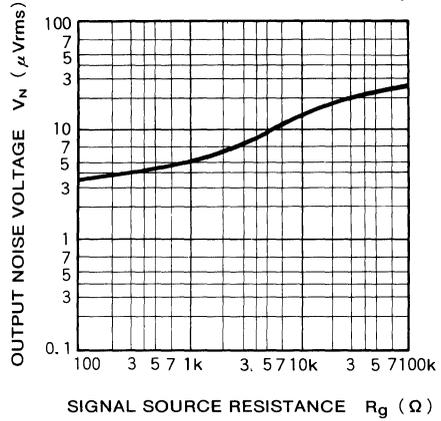
OUTPUT DC VOLTAGE VS. INPUT DC VOLTAGE (SOUND)



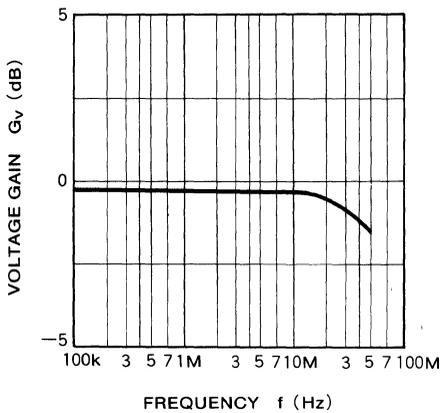
CROSSTALK VS. FREQUENCY (SOUND)



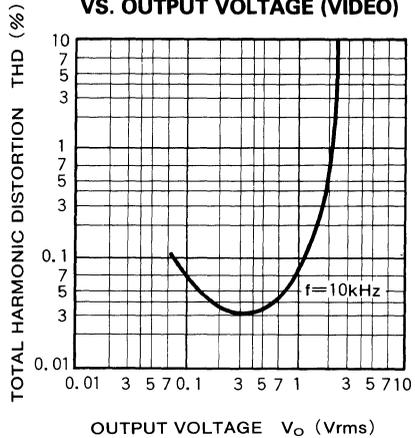
OUTPUT NOISE VOLTAGE VS. SIGNAL SOURCE RESISTANCE (SOUND)



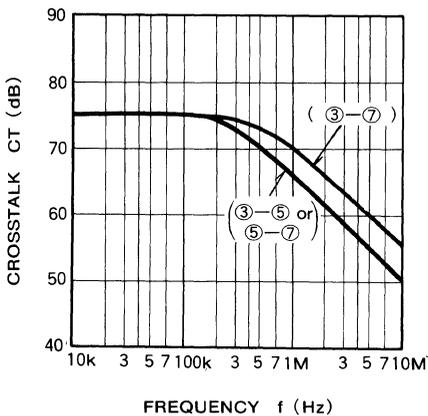
VOLTAGE GAIN VS. FREQUENCY (VIDEO)



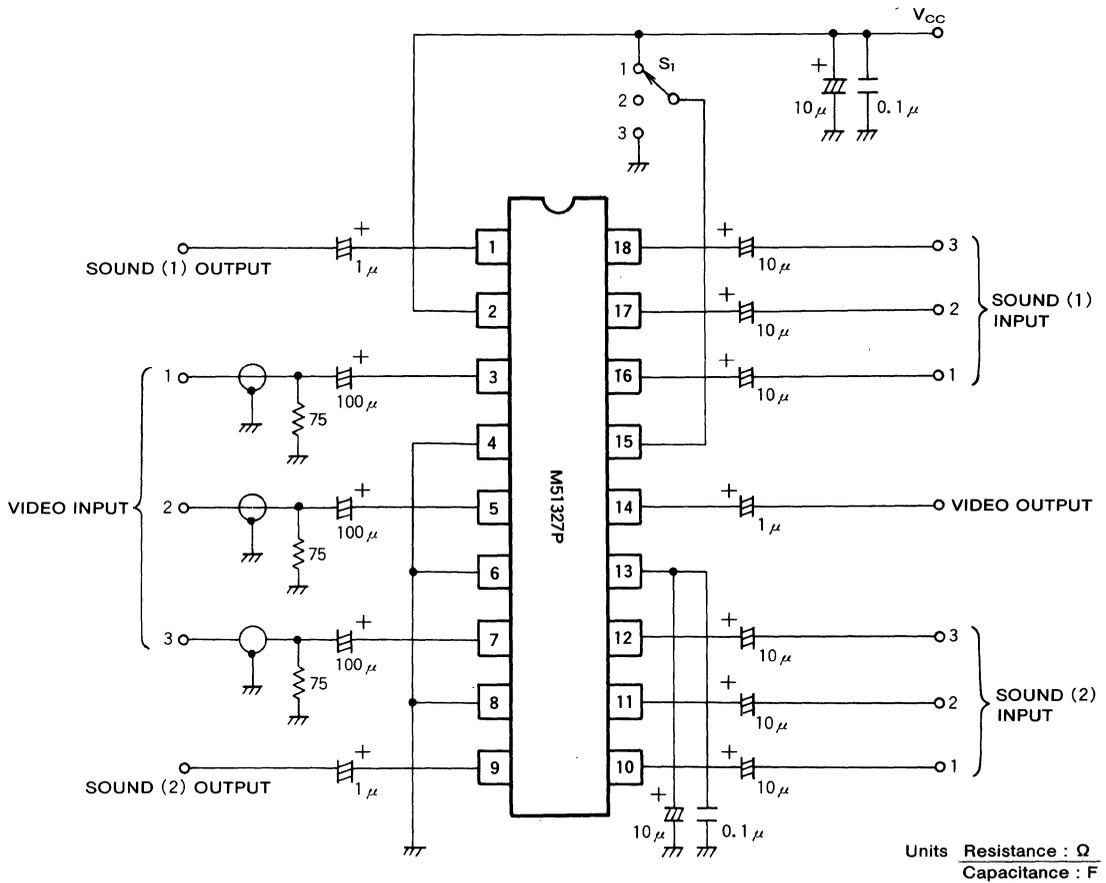
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (VIDEO)



CROSSTALK VS. FREQUENCY (VIDEO)



APPLICATION EXAMPLE



PRECAUTIONS FOR USE

Since an emitter-follower output is used in the video and sound outputs, when the external wiring is long or a capacitive load is connected, a resistor with a value of several tens of ohms should be connected at a position near the output pin.

MITSUBISHI LINEAR ICs

M51329P

ANALOG SWITCH

DESCRIPTION

The M51329P is a semiconductor integrated circuit containing an analog switch designed for use in a video system. It contains two audio switches and one video switch. Each switch has three inputs and can be independently controlled in each of the audio and the video sections. In addition, the video switch contains an amplifier with a gain of about 6.7dB.

FEATURES

- Video and stereo sound switches in one package
- Wide frequency range (video switch) DC~10MHz
- High separation (video) Crosstalk 55dB(typ.) @5MHz

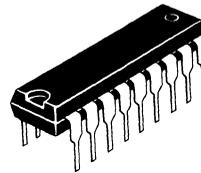
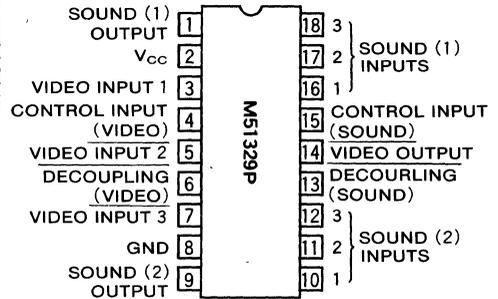
APPLICATION

Video equipment.

RECOMMENDED OPERATING CONDITIONS

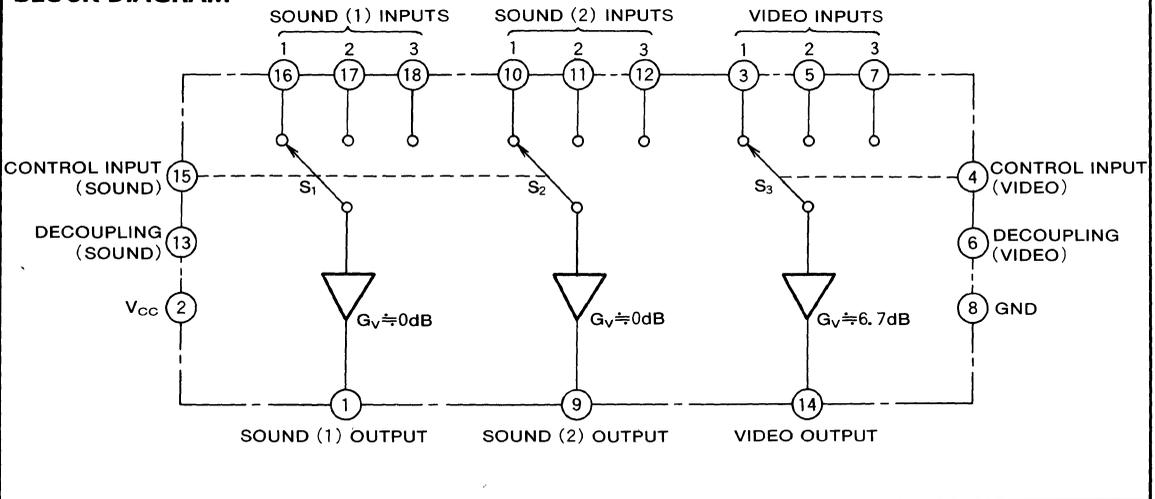
Supply voltage range 5~14V

PIN CONFIGURATION (TOP VIEW)



18-pin molded plastic DIP

BLOCK DIAGRAM



ANALOG SWITCH

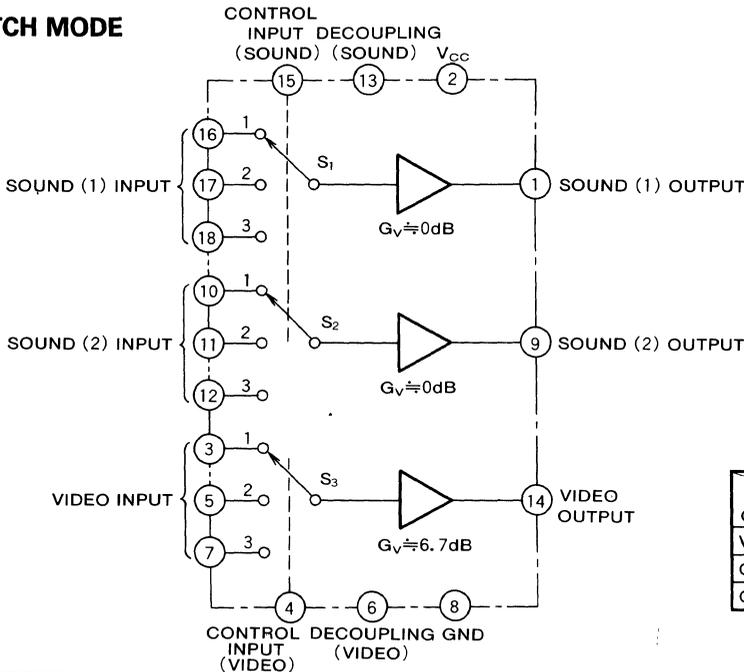
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		14	V
V_{IS}	Input signal voltage	Sound	6	V
		Video	6	
V_{IC}	Input control voltage		V_{CC}	V
P_d	Power dissipation		1.25	W
K_θ	Thermal derating		12.5	mW/ $^\circ\text{C}$
T_{opr}	Operating temperature range		$-20\sim+75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-40\sim+125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current			44	56	mA
V_{IDC}	Input bias voltage		3.8	4.2	4.6	V
V_{ODC}	Output bias voltage	Video	5.0	5.6	6.2	V
		Sound	3.0	3.6	4.2	
V_{OP}	Output DC offset voltage			15	100	mV
$V_{IC H}$	Control-pin threshold voltage		7.0	8.0	9.0	V
			3.0	4.0	5.0	V
G_V	Voltage gain	Video, $f=1\text{MHz}$	5.7	6.7	7.7	dB
		Sound, $f=1\text{kHz}$	-0.5	-0.1		
R_i	Input resistance	Sound, $f=1\text{kHz}$		22		k Ω
		Video, $f=5\text{MHz}$		11		
C_i	Input capacitance	Video, $f=5\text{MHz}$		4		pF
THD	Total harmonic distortion	Sound, $f=1\text{kHz}$, $V_o=1\text{Vrms}$		0.02	0.2	%
V_N	Output noise voltage	Video, $R_g=75\Omega$, $BW=10\text{MHz}$		0.5	1.0	mVrms
		Sound, $R_g=600\Omega$, $BW=15\text{kHz}$		3	50	μVrms
CT	Crosstalk	Sound, $f=1\text{kHz}$	65	80		dB
		Video, $f=5\text{MHz}$	45	55		

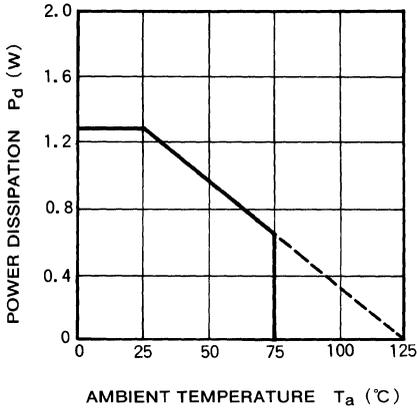
SWITCH MODE



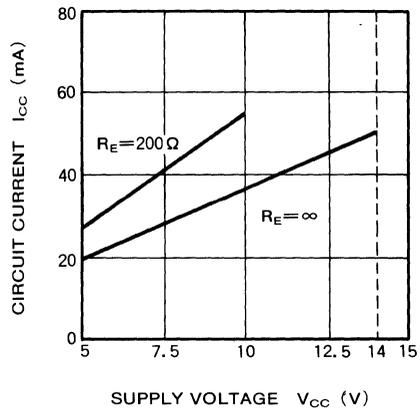
Control input	Switch No.		
	S ₁	S ₂	S ₃
V_{CC}	1	1	1
OPEN	2	2	2
GND	3	3	3

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

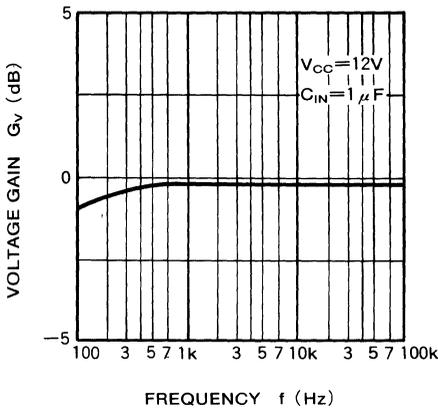
THERMAL DERATING (MAXIMUM RATING)



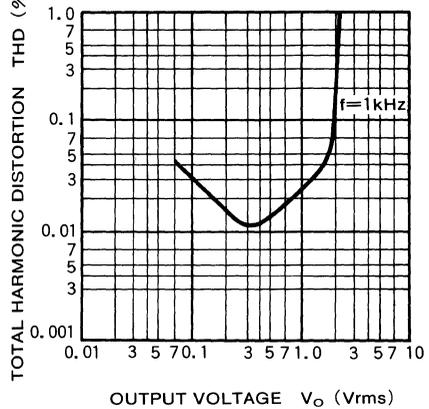
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



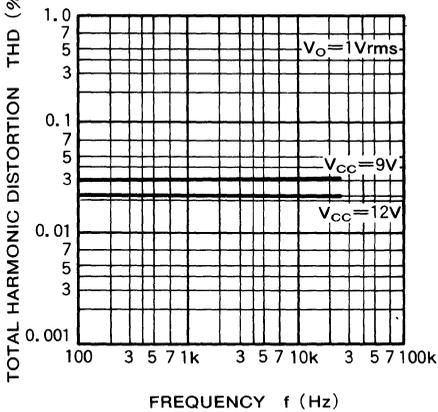
VOLTAGE GAIN VS. FREQUENCY (SOUND)



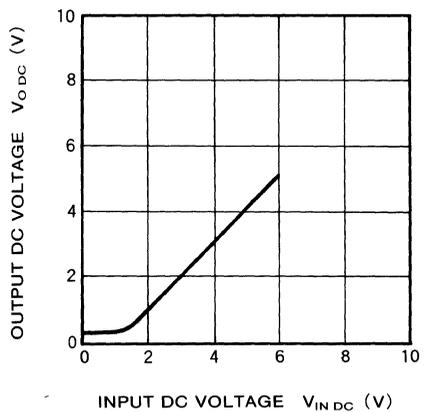
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (SOUND)



TOTAL HARMONIC DISTORTION VS. FREQUENCY (SOUND)

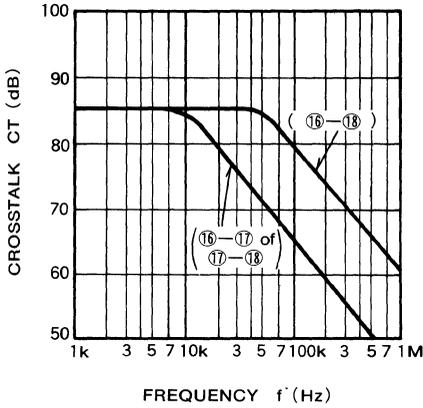


OUTPUT DC VOLTAGE VS. INPUT DC VOLTAGE (SOUND)

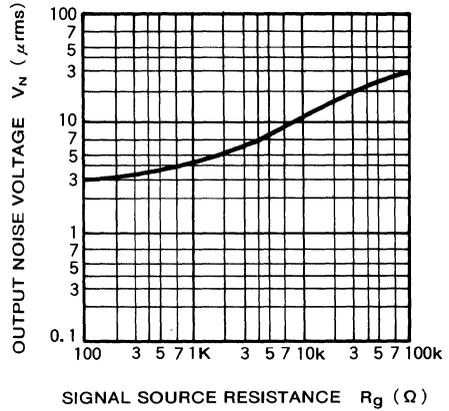


ANALOG SWITCH

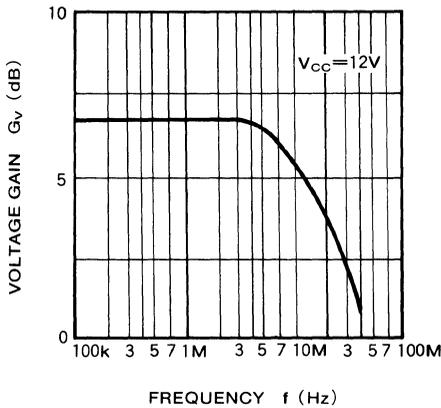
CROSSTALK VS. FREQUENCY (SOUND)



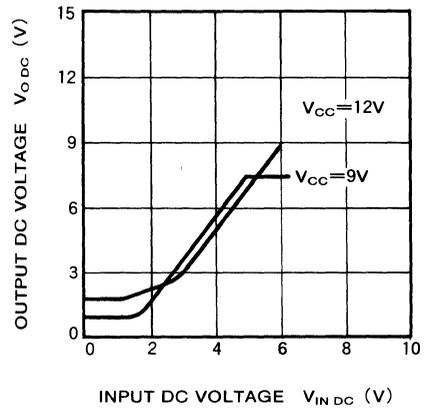
OUTPUT NOISE VOLTAGE VS. SIGNAL SOURCE RESISTANCE (SOUND)



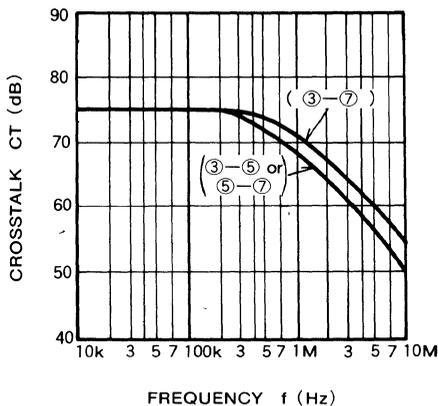
VOLTAGE GAIN VS. FREQUENCY (VIDEO)



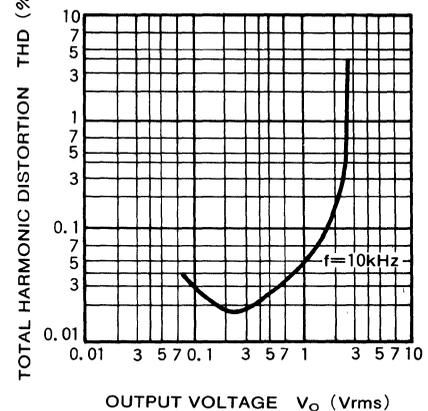
OUTPUT DC VOLTAGE VS. INPUT DC VOLTAGE (VIDEO)



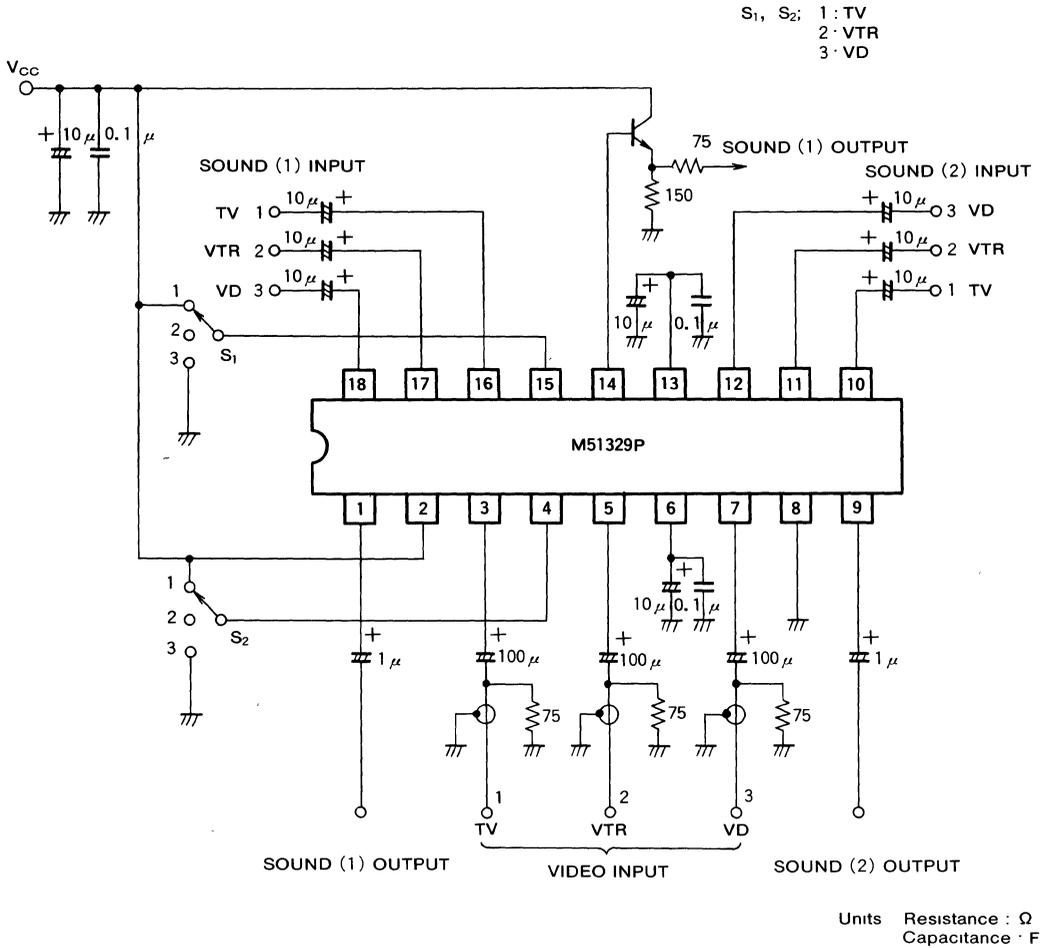
CROSSTALK VS. FREQUENCY (VIDEO)



TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (VIDEO)



APPLICATION EXAMPLE



PRECAUTIONS FOR USE

Since an emitter-follower output is used in the video and sound outputs, when the external wiring is long or a capacitive load is connected, a resistor with a value of several tens of ohms should be connected at a position near the output pin.

M51330P

ANALOG SWITCH

DESCRIPTION

The M51330P is a semiconductor integrated circuit containing an analog switch designed for use in a video system. It contains two audio switches and one video switch. Each switch has three inputs and can be independently controlled in each of the audio and the video sections.

FEATURES

- Video and stereo sound switches in one package
- Wide frequency range (video switch) DC~10MHz
- High separation (video) Crosstalk 55dB(typ.) @5MHz

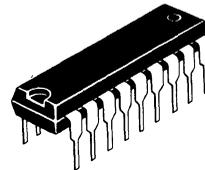
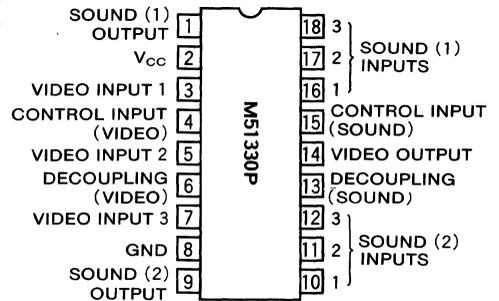
APPLICATION

Video equipment.

RECOMMENDED OPERATING CONDITIONS

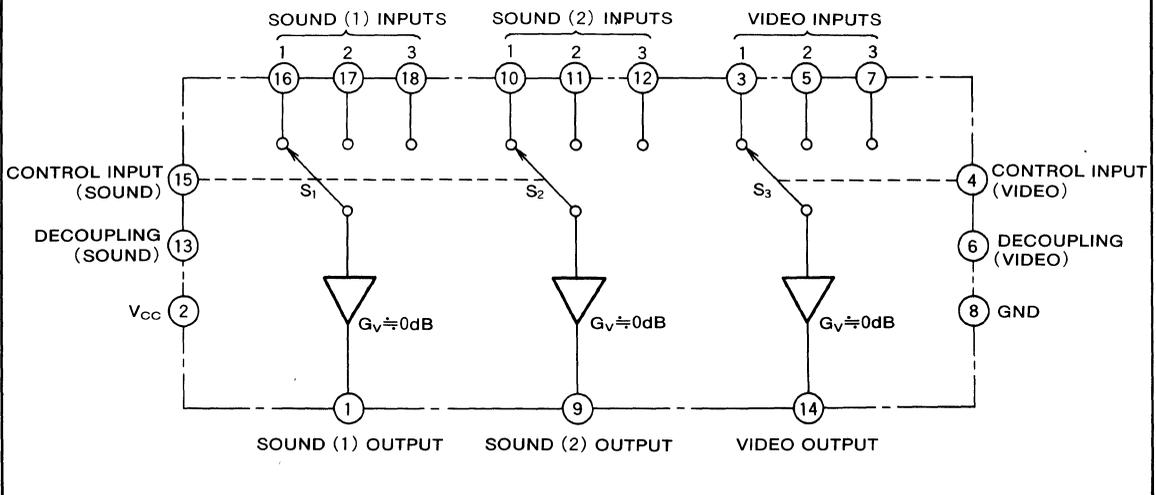
Supply voltage range 5~14V

PIN CONFIGURATION (TOP VIEW)



18-pin molded plastic DIP

BLOCK DIAGRAM



ANALOG SWITCH

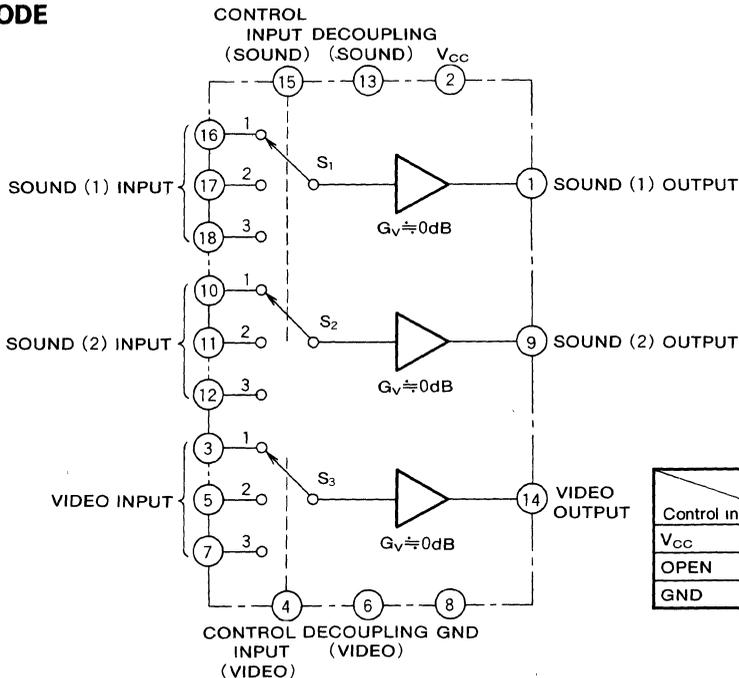
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		14	V
V_{IS}	Input signal voltage	Sound	6	V
		Video	6	
V_{IC}	Input control voltage		V_{CC}	V
P_d	Power dissipation		1.25	W
K_θ	Thermal derating		12.5	mW/ $^\circ\text{C}$
T_{opr}	Operating temperature range		$-20\sim+75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-40\sim+125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current			36	48	mA
$V_{I\ DC}$	Input bias voltage		3.8	4.2	4.6	V
$V_{O\ DC}$	Output bias voltage		3.0	3.6	4.2	V
V_{OP}	Output DC offset voltage			15	100	mV
$V_{IC\ H}$ $V_{IC\ L}$	Control-pin threshold voltage		7.0	8.0	9.0	V
			3.0	4.0	5.0	V
G_V	Voltage gain	Video, $f=1\text{MHz}$	-0.5	-0.1		dB
		Sound, $f=1\text{kHz}$	-0.5	-0.1		
R_i	Input resistance	Sound, $f=1\text{kHz}$		22		k Ω
		Video, $f=5\text{MHz}$		11		
C_i	Input capacitance	Video, $f=5\text{MHz}$		4		pF
THD	Total harmonic distortion	Sound, $f=1\text{kHz}$, $V_o=1\text{Vrms}$		0.02	0.2	%
V_N	Output noise voltage	Video, $R_g=75\Omega$, $BW=10\text{MHz}$		0.5	1.0	mVrms
		Sound, $R_g=600\Omega$, $BW=15\text{kHz}$		3	50	μVrms
CT	Crosstalk	Sound, $f=1\text{kHz}$	65	80		dB
		Video, $f=5\text{MHz}$	45	55		

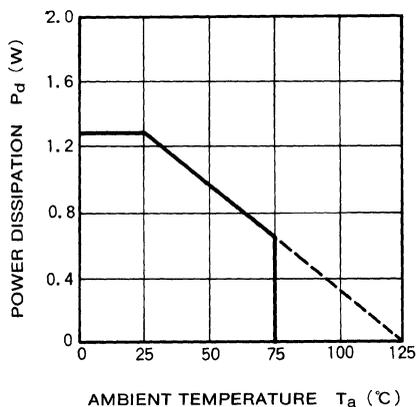
SWITCH MODE



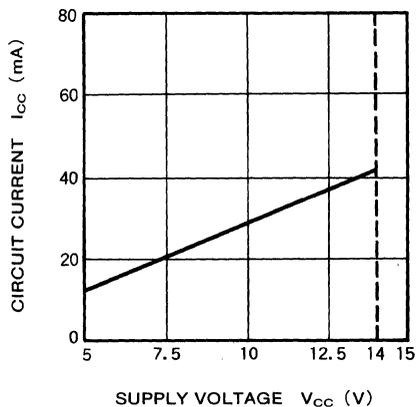
Control input	Switch No.		
	S_1	S_2	S_3
V_{CC}	1	1	
OPEN	2	2	
GND	3	3	

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

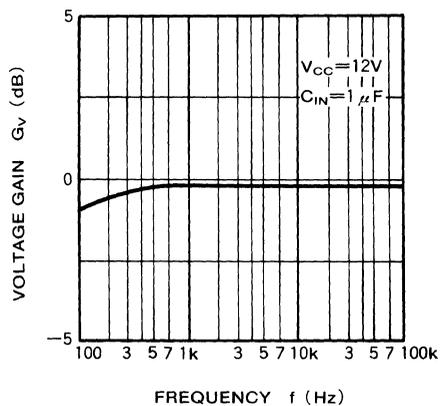
THERMAL DERATING (MAXIMUM RATING)



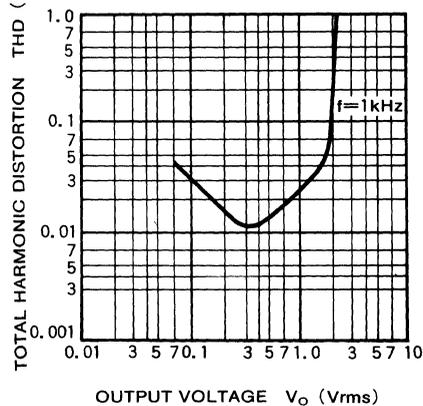
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



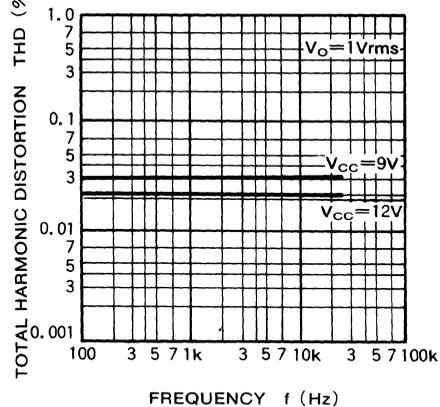
VOLTAGE GAIN VS. FREQUENCY (SOUND)



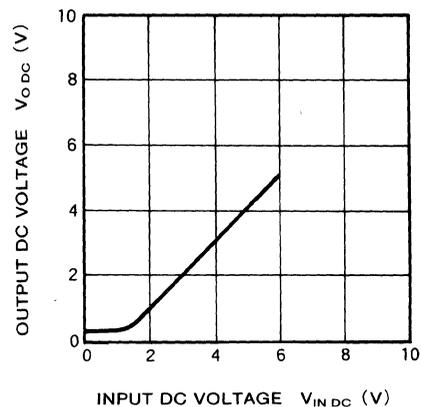
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (SOUND)



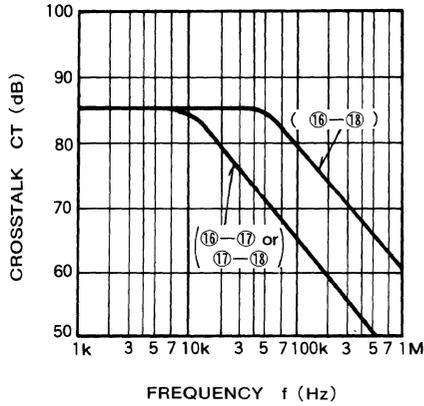
TOTAL HARMONIC DISTORTION VS. FREQUENCY (SOUND)



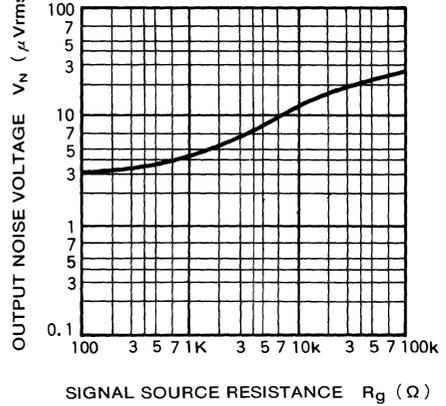
OUTPUT DC VOLTAGE VS. INPUT DC VOLTAGE (SOUND)



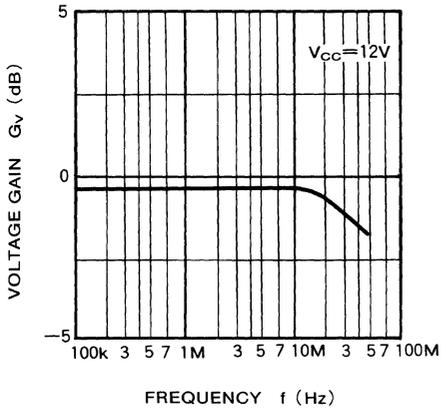
CROSSTALK VS. FREQUENCY (SOUND)



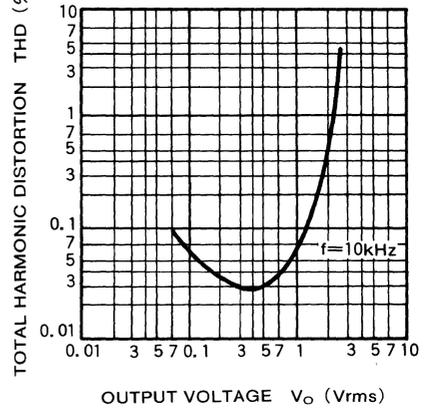
OUTPUT NOISE VOLTAGE VS. SIGNAL SOURCE RESISTANCE (SOUND)



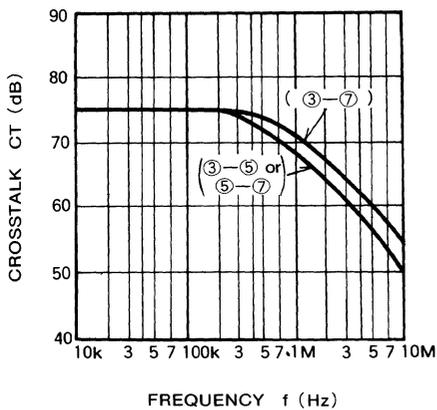
VOLTAGE GAIN VS. FREQUENCY (VIDEO)



TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE (VIDEO)



CROSSTALK VS. FREQUENCY (VIDEO)



MITSUBISHI LINEAR ICs

M51551P,FP

ANALOG SWITCH

DESCRIPTION

The M51551P,FP is a semiconductor integrated circuit containing two electronic switches, each with two inputs and a single output.

Channel input is simultaneously selectable for supply DC voltage to the control pin. The device also contains an output circuit designed for use in driving an LED or other indicator to display which channel is operating.

M51551FP is housed in a 14-pin plastic flat package, while the housing for M51551P is housed in a 14-pin molded plastic dual in-line (DIL) package.

FEATURES

- Dual channel 2-input 1-output analogue signal switch.
- Mode switched for supply DC voltage.
- Built-in output circuit for use in driving a mode indicator.
- Bipolar construction assures low distortion
..... 0.006% (typ) ($V_i=1.0V_{rms}$)
- Operation possible with either dual or single polarity power supplies. (Control pin operates by switching between GND and supply.)

APPLICATION

Audio selector, stereo radio cassette tape recorders, car-stereos.

RECOMMENDED OPERATING CONDITIONS

Dual polarity power supplies :

Supply voltage range $\pm 6 \sim \pm 10V$

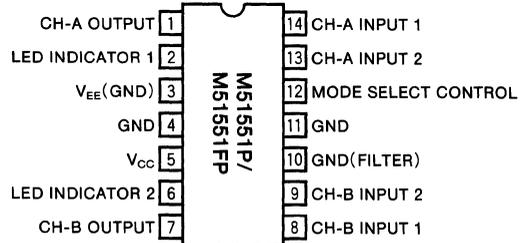
Rated supply voltage $\pm 8V$

Single polarity power supplies :

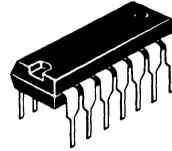
Supply voltage range $6 \sim 20V$

Rated supply voltage $15V$

PIN CONFIGURATION (TOP VIEW)



ITEMS IN PARENTHESES APPLICABLE TO SINGLE POLARITY POWER SUPPLY.

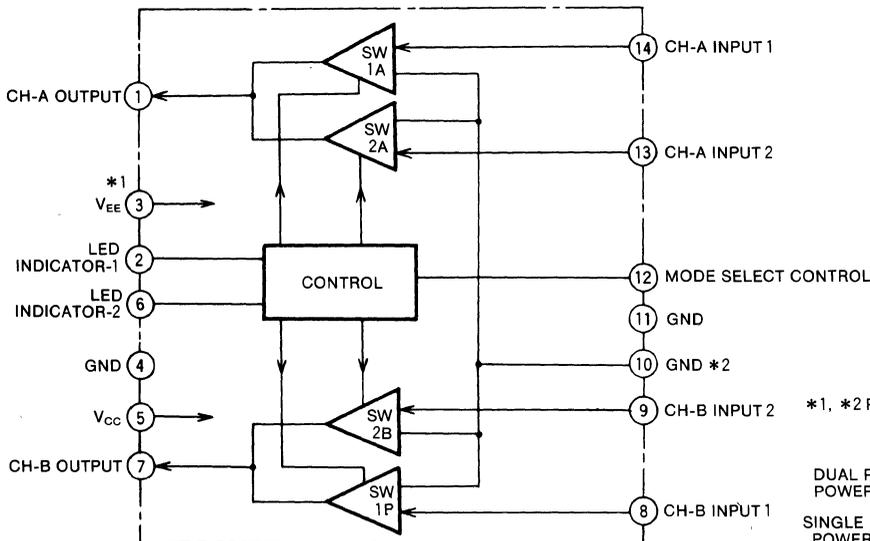


14-pin molded plastic DIP



14-pin molded plastic FLAT

BLOCK DIAGRAM



*1, *2 PIN DESCRIPTION

	PIN 3	PIN 10
DUAL POLARITY POWER SUPPLY	V_{EE}	GND
SINGLE POLARITY POWER SUPPLY	GND	FILTER

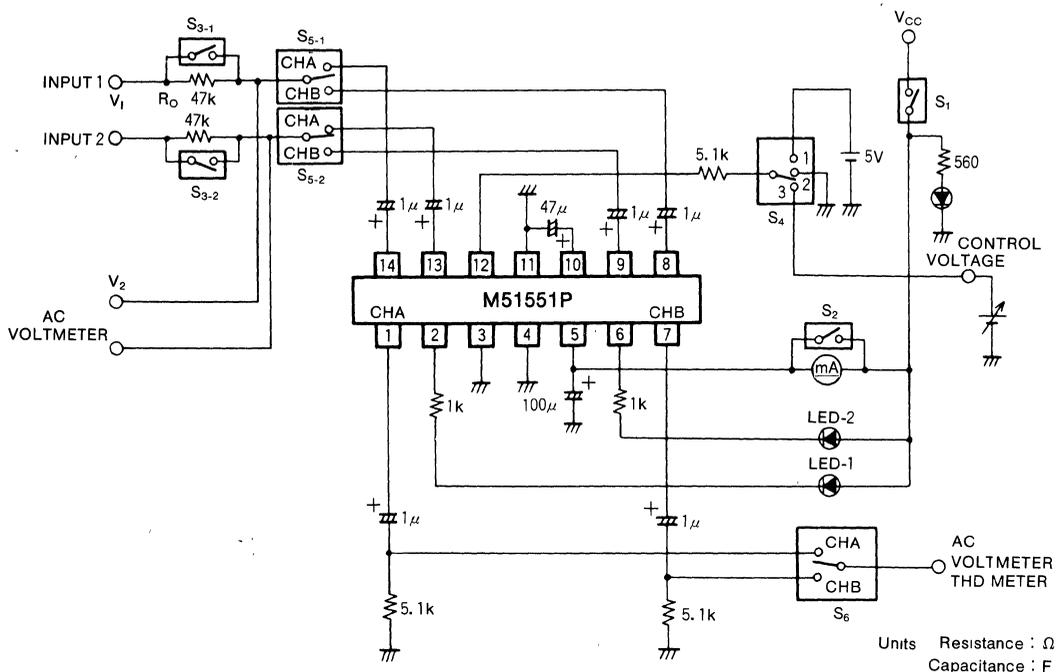
ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage	Quiescent, between pins 5 and 3.	24	V
I_{CC}	Circuit current	Except pins 2 and 6	30	mA
$I_{CC(2)}, \textcircled{2}$	LED drive current	Pin 2, Pin 6	40	mA
P_d	Power dissipation		620	mW
K_{θ}	Thermal derating	$T_a \geq 25^{\circ}\text{C}$	6.2	mW/ $^{\circ}\text{C}$
T_{opr}	Operating temperature		$-20 \sim +75$	$^{\circ}\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^{\circ}\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, $V_{CC}=15\text{V}$, $V_i=1.5\text{Vrms}$, $f=1\text{kHz}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	Except pins 2 and 6		12	16	mA
V_{CNTL1}	Select control voltage 1	Control voltage when input 1 is selected	2.0			V
V_{CNTL2}	Select control voltage 2	Control voltage when input 2 is selected.			0.7	V
R_i	Input resistance	$V_i=1.5\text{Vrms}$, $f=1\text{kHz}$, $R_o=47\text{k}\Omega$	30	47		$\text{k}\Omega$
G_v	Voltage gain	$V_i=1.5\text{Vrms}$, $f=1\text{kHz}$	-1	0	1	dB
$V_{O(max)}$	Maximum output voltage	The output voltage at THD=1%	4.0	4.5		Vrms
THD	Total harmonic distortion	$V_o=1.0\text{Vrms}$		0.006	0.017	%
N_o	Output noise level	BPF(20Hz~20kHz), connect pin 1 to GND		5.5	10	μVrms
C.T.	Crosstalk	The ratio of signal leakage between input 1 and input 2, $f=1\text{kHz}$	52	58		dB
C.L.	Channel leakage	The ratio of signal leakage between CH-A and CH-B, $f=1\text{kHz}$	77	83		dB
C.B.	Channel balance	Output level ratio between CH-A and CH-B	-0.5	0	0.5	dB

TEST CIRCUIT (Using single polarity power supply)



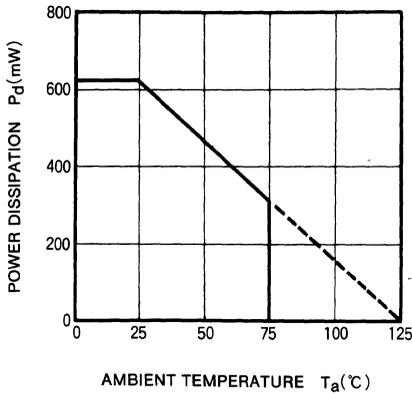
Note : When a dual polarity power supply is used, pin 3 is the V_{EE} pin (Ground pin 3 for signal polarity supply)

TEST METHODS

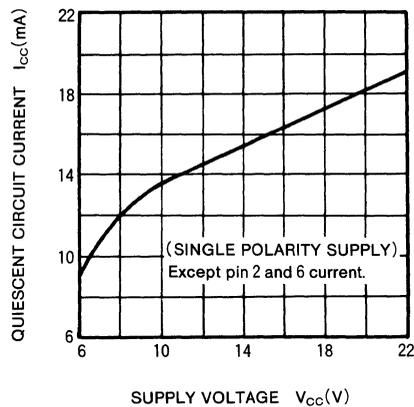
Symbol	Switch conditions							Test method
	S ₁	S ₂	S ₃	S ₄	S ₅₋₁	S ₅₋₂	S ₆	
I _{CC}	ON	OFF	ON	1/2				Read ammeter with V _{CC} =15V.
V _{CNTL1}	ON	ON	ON	3				Measure the voltage applied to pin 12 at the point the signal applied to input 1 appears at output.
V _{CNTL2}	ON	ON	ON	3				Measure the voltage applied to pin 12 at the point the signal applied to input 2 appears at output.
R _i	ON	ON	OFF	1/2	CHA CHB	CHA CHB		$R_i = \frac{V_2 \times R_0}{V_1 - V_2}$ At V ₁ =1.5Vrms, R ₀ =47kΩ measure V ₂ relative to input 1 and input 2.
G _v	ON	ON	ON	1/2	CHA CHB	CHA CHB	CHA CHB	G _v =20 log $\frac{V_o}{V_i}$ (dB) V ₁ =1.5Vrms, V _o =voltage output
V _{o(max)}	ON	ON	ON	1/2	CHA CHB	CHA CHB	CHA CHB	Measure output voltage for THD=1% at output.
THD	ON	ON	ON	1/2	CHA CHB	CHA CHB	CHA CHB	Measure output distortion for V ₂ =1.0Vrms
N _o	ON	ON	ON	1/2			CHA CHB	Input terminal AC ground, BPF 20Hz~20kHz
C.T.	ON	ON	ON	1→2	CHA CHB		CHA CHB	C.T.=20 log $\left(\frac{V_o(S_2 \rightarrow 1)}{V_o(S_4 \rightarrow 2)} \right)$ (dB)
C.L.	ON	ON	ON	1	CHA	CHA	CHA CHB	C.L.=20 log $\left(\frac{V_o(CHA)}{V_o(CHB)} \right)$ (dB)
C.B.	ON	ON	ON	1	CHA CHB	CHA CHB	CHA CHB	C.B.=20 log $\left(\frac{V_o(CHA)}{V_o(CHB)} \right)$ (dB)

TYPICAL CHARACTERISTICS (T_a=25°C, V_{CC}=15V, unless otherwise noted)

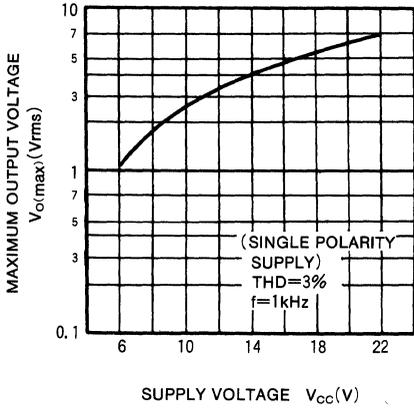
THERMAL DERATING
(MAXIMUM RATING)



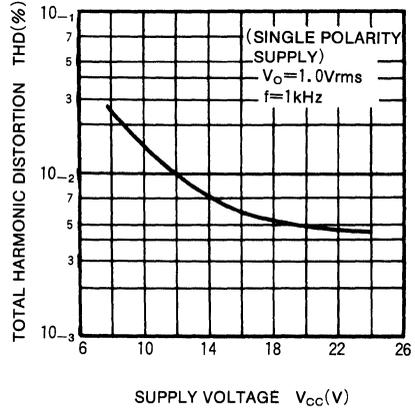
QUIESCENT CIRCUIT CURRENT
VS SUPPLY VOLTAGE



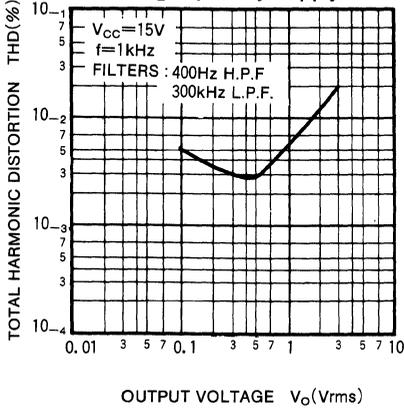
MAXIMUM OUTPUT VOLTAGE VS SUPPLY VOLTAGE



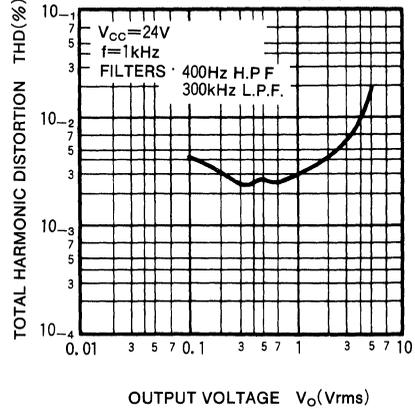
TOTAL HARMONIC DISTORTION VS SUPPLY VOLTAGE



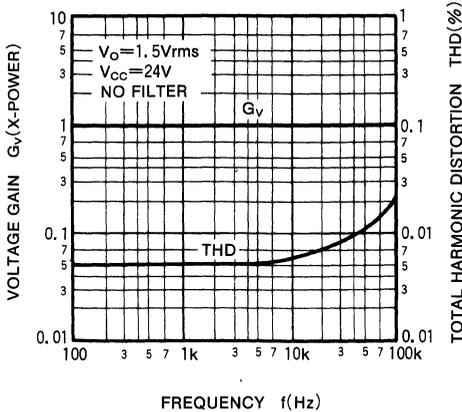
TOTAL HARMONIC DISTORTION VS OUTPUT VOLTAGE (V_{CC}=15V)
(Single polarity supply)



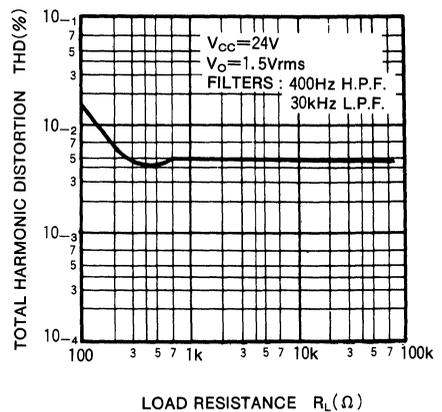
TOTAL HARMONIC DISTORTION VS OUTPUT VOLTAGE (V_{CC}=24V)
(Single polarity supply)



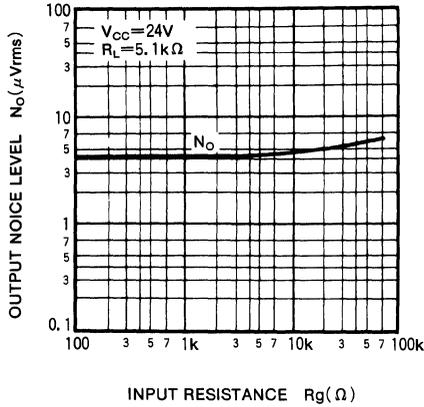
FREQUENCY RESPONSE
(Single polarity supply)



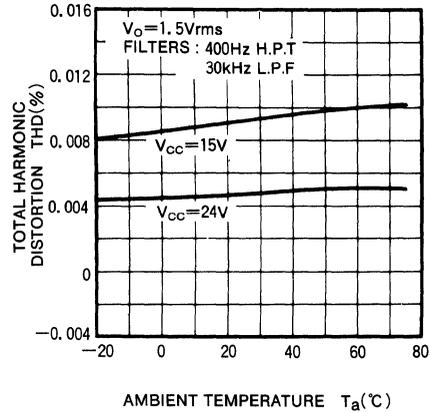
TOTAL HARMONIC DISTORTION VS LOAD RESISTANCE (Single polarity supply)



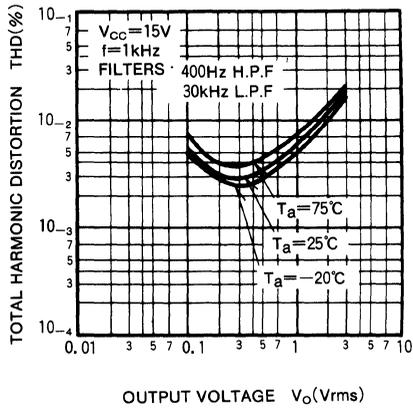
OUTPUT NOISE LEVEL VS INPUT RESISTANCE (Single polarity supply)



TOTAL HARMONIC DISTORTION VS AMBIENT TEMPERATURE (Single polarity supply)

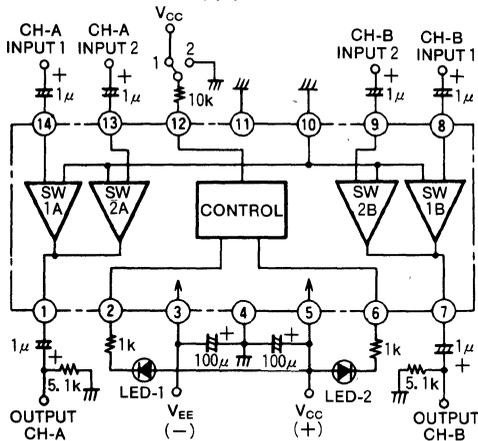


TEMPERATURE CHARACTERISTICS (Single polarity supply)



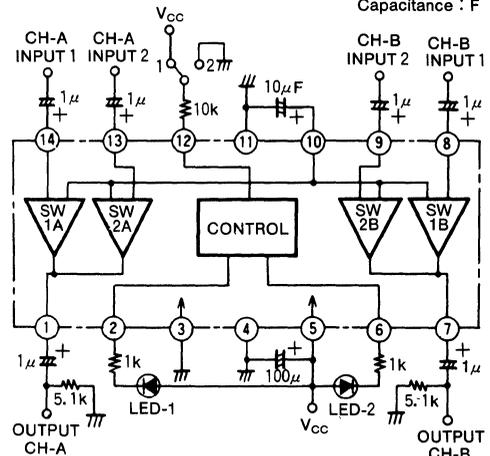
APPLICATION EXAMPLES

(Dual polarity power supply)



(Single polarity power supply)

Unit Resistance : Ω
Capacitance : F



LEVEL INDICATOR

QUICK REFERENCE TABLE OF LEVEL INDICATOR

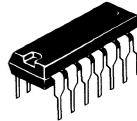
Type No.	Function						Features	Maximum ratings			Package outline
	Max No of LEDs driven	Display format	Display threshold	Output format	Ref voltage	Detection amplifier		Supply voltage (V)	Load current	Allowable power consumption (mW)	
M51901P	12	Dot display, 1 or 2 LEDs	Linear	Open collector 	Externally set 5.0~7.5V	—	<ul style="list-style-type: none"> ● 23 operation modes ● Blanking mode included 	10.2~16.5	30mAmax	650	16-pin DIP
M51903L	5	Bar display, linear LEDs	Linear	Open collector 	Internal 1.25V	—	<ul style="list-style-type: none"> ● Wide supply voltage range ● Brightness of uppermost LED varies linearly with the input voltage 	4~18	15mAmax	450	8-pin SIP
M51910P	9	Dot display, only 1 LEDs	Linear	Open collector 	Externally set 0~3.5V	—	<ul style="list-style-type: none"> ● Built-in timer ● 2 LED-drive modes, so 2 inputs can be displayed 	8~18	20mAmax	1400	22-pin DIP
M51906P	6	Bar display	Acoustic threshold (dB)	Open collector 	Internal 1.25V	Internal (gain set externally)	<ul style="list-style-type: none"> ● Wide supply voltage range ● Detection amp built-in Cut-off frequency.....500kHz typ Offset voltage.....2mV typ 	4~15	30mAmax	1500	14-pin DIP
M51907P	8	Bar display	Acoustic threshold (dB)	Constant current cascade connection 	Internal 1.25V (Ref voltage terminal included)	Internal (gain set externally)	<ul style="list-style-type: none"> ● Wide supply voltage range ● Output cascade connection reduces power consumption by half 	4~15	25mAmax	1600	16-pin DIP
M51909P	8	Bar display	Linear	Constant current cascade connection 	Internal 1.25V (Ref voltage terminal included)	Internal (gain set externally)	<ul style="list-style-type: none"> ● Detection amp built-in ● Abundant applications circuits 	4~15	25mAmax	1600	
M51911L	6	Bar display	Acoustic threshold (dB)	Constant current cascade connection 	Internal 1.25V	Internal (gain 1/7dB)	<ul style="list-style-type: none"> ● Wide supply voltage range ● Few external attachments required ● Output cascade connection reduces power consumption by half 	4~15	13mA typ internally decided	1100	10-pin SIP
M51912L	6	Bar display	Linear	Constant current cascade connection 	Internal 1.25V	Internal (gain 1/7dB)	<ul style="list-style-type: none"> ● Detection amp built-in 	4~15	13mA typ internally decided	1100	



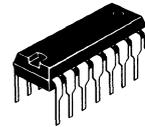
8-pin SIP



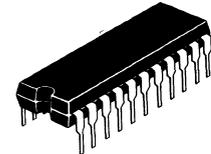
10-pin SIP



14-pin DIP



16-pin DIP



22-pin DIP

M51901P

12-POINT/23-MODE LED DRIVER

DESCRIPTION

The M51901P is a semiconductor integrated circuit consisting of a driver circuit capable of driving 12 LEDs in 23 modes.

When a DC voltage is applied to the input pin the LED driving outputs are activated either 1 or 2 at a time to provide 23 LED drive modes in accordance with the applied voltage level. In addition, a blanking function is available when the reference voltage is made a low level.

The M51901P consists of 12 differential amplifiers and the associated ladder circuit as well as a blanking circuit.

FEATURES

- 12 LEDs may be driven in accordance with the level of a DC voltage applied to the input, using a built-in A-D conversion capability.
- 23 operating modes are provided
- Built-in blanking function
- The reference voltage level may be freely selected

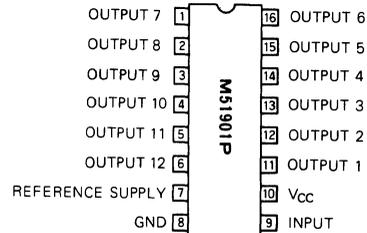
APPLICATION

23-mode drivers for 12 LEDs, simplified A-D converters

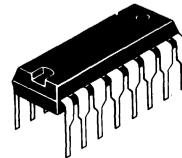
RECOMMENDED OPERATING CONDITIONS

Supply voltage range	10.2~16.5V
Rated supply voltage	13.2V
Reference voltage range	5.0~7.5V
Input voltage range	0~9.2V

PIN CONFIGURATION (TOP VIEW)

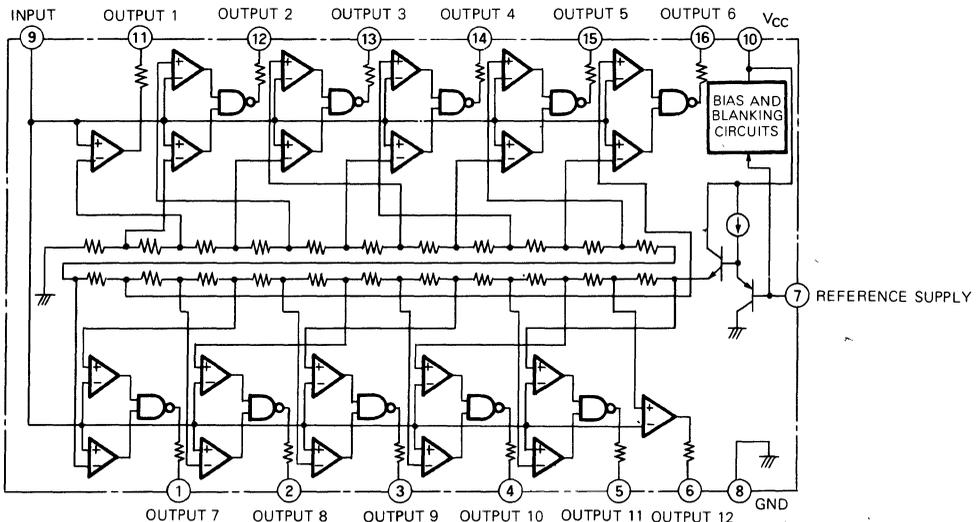


Outline 16P4



16-pin molded plastic DIP

BLOCK DIAGRAM



12-POINT/23-MODE LED DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

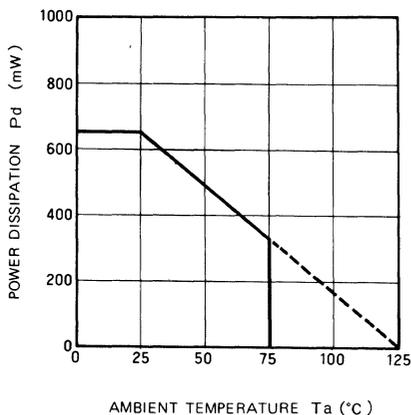
Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		18	V
I _O	Output current		30	mA
P _d	Power dissipation		650	mW
K _θ	Derating	T _a ≥ 25°C	6 5	mW/°C
T _{opg}	Operating temperature		-20 ~ +75	°C
T _{stg}	Storage temperature		-40 ~ +125	°C

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, V_{CC}=±13.2V, V_{REF}=7.20V, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I _{CC}	Circuit current	V _{IN} =0V, outputs open		2	5	mA
I _I	Input current	V _{IN} =9.2V			100	μA
V _O	Output voltage (pins ①-⑫)	R _L =620 Ω	4 2	5.5	6 8	V
V _{BL}	Blanking voltage	V _{IN} =9.2V, I _O =100μA			0 8	V
I _⑦	Pin ⑦ output current	V _{IN} =0V			15	μA
V _②	Output 2 on-state central input voltage	I _O ≥ 1mA		1 99		V
V _③	Output 3 on-state central input voltage	I _O ≥ 1mA		2 51		V
V _④	Output 4 on-state central input voltage	I _O ≥ 1mA		3 03		V
V _⑤	Output 5 on-state central input voltage	I _O ≥ 1mA		3 55		V
V _⑥	Output 6 on-state central input voltage	I _O ≥ 1mA		4 07		V
V _⑦	Output 7 on-state central input voltage	I _O ≥ 1mA		4 59		V
V _⑧	Output 8 on-state central input voltage	I _O ≥ 1mA		5 11		V
V _⑨	Output 9 on-state central input voltage	I _O ≥ 1mA		5 63		V
V _⑩	Output 10 on-state central input voltage	I _O ≥ 1mA		6 15		V
V _⑪	Output 11 on-state central input voltage	I _O ≥ 1mA		6 67		V

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

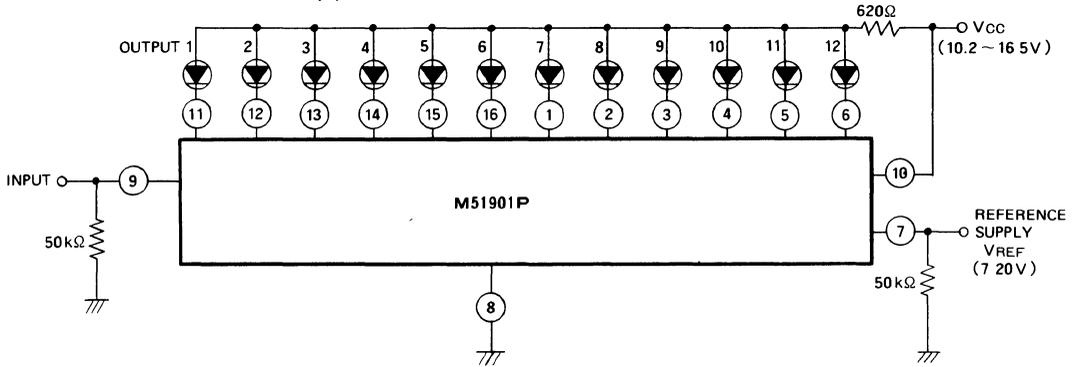
**THERMAL DERATING
(MAXIMUM RATING)**



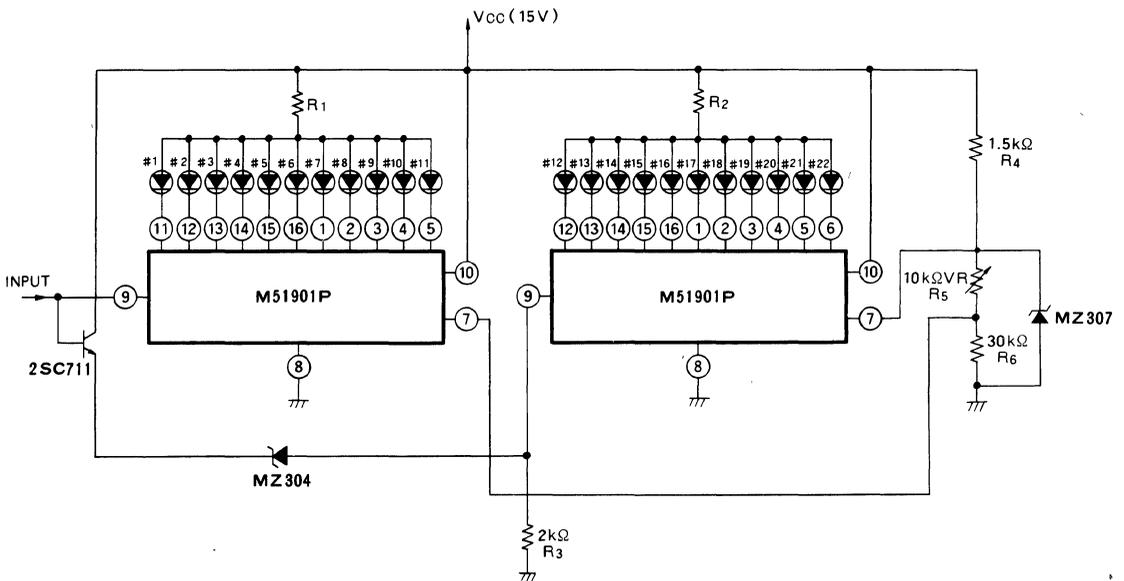
12-POINT/23-MODE LED DRIVER

APPLICATION EXAMPLES

(1) 23-MODE DRIVER FOR 12 LEDS



(2) 22-LED DRIVER (USING CASCADE CONNECTION)



Note R_5 is chosen such that the lower drive level limit for LED #12 is just 0.24V higher than higher drive level limit for LED #10

M51903L

LED LINEAR LEVEL INDICATOR

DESCRIPTION

The M51903L is a semiconductor integrated circuit consisting of a circuit designed for use in level meters. It is capable of driving 5 LEDs to create a bar-type display. In accordance with the input level, the uppermost LED brightness varies to form a linear indicator, making this device ideal for use in signal meters and VU meters. A low-voltage reference power supply is built in, so that the only external components required are LEDs, resistors and capacitors.

FEATURES

- Bar-type display of input level using 5 LEDs
- The uppermost LED brightness varies linearly with respect to the input level resulting in a high-resolution display with no radiation.
- By changing the external resistance values, the LED brightness can be adjusted $I_O=15\text{mA}(\text{max.})$
- Operates over a wide range of supply voltages $V_{CC}=4\sim 18\text{V}$
- Built-in reference supply
- High input impedance $I_{IN}=100\text{nA}(\text{typ.})$

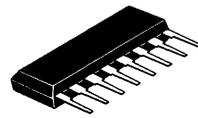
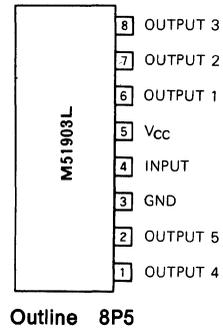
APPLICATION

Signal meters, VU meters, tuning meters, and other general display applications

RECOMMENDED OPERATING CONDITIONS

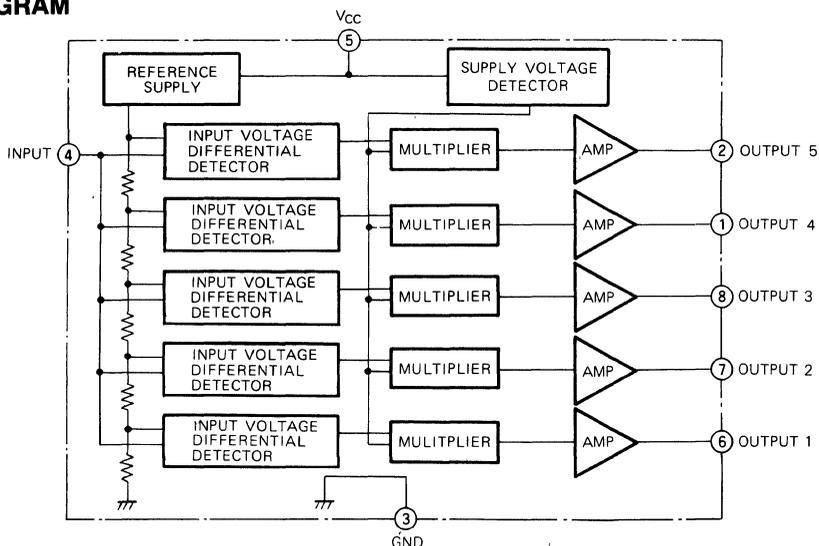
- Supply voltage range 4~18V
- Rated supply voltage 10V

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic SIP

BLOCK DIAGRAM



LED LINEAR LEVEL INDICATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
V_{IN}	Input voltage		6	V
BV_O	Output breakdown voltage		18	V
I_O	Output current		15 (per pin)	mA
P_{dF}	Power dissipation	With the M51903L soldered to a printed circuit board (copper-clad area 4.5 x 5.5cm, thickness 35 μ , board thickness 2.0mm)	550	mW
$K_{\theta F}$	Derating	$T_a \geq 25^\circ\text{C}$	5.5	mW/°C
T_{opg}	Operating temperature		-20 ~ +75	°C
T_{stg}	Storage temperature		-40 ~ +125	°C

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=\pm 10\text{V}$, unless otherwise noted)

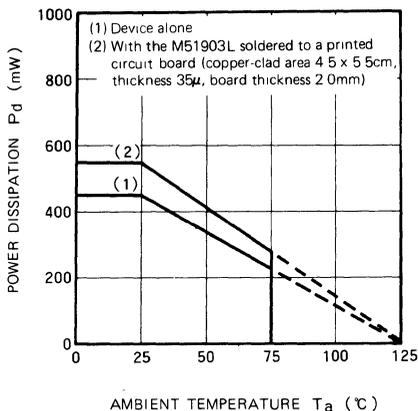
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage range		4	10	18	V
V_{INF}	Fullscale input voltage			1320		mV
V_{step}	Step voltage			210		mV
I_{IN}	Input current	$V_{IN}=0\text{V}$ (Note 1)		0.1	1.0	μA
I_{CC}	Circuit current	$V_{IN}=0\text{V}$		5	8	mA
$V_{IT\textcircled{6}}$	Output 1 LED drive voltage	$R_L = 1.5\text{k}\Omega$ $I_L = 100\mu\text{A}$ Using red GaAlAs LEDs	170	230	300	mV
$V_{IT\textcircled{7}}$	Output 2 LED drive voltage		380	450	530	mV
$V_{IT\textcircled{8}}$	Output 3 LED drive voltage		580	660	730	mV
$V_{IT\textcircled{9}}$	Output 4 LED drive voltage		780	860	940	mV
$V_{IT\textcircled{2}}$	Output 5 LED drive voltage		980	1070	1180	mV

Note 1 Current flowing from pin ④ is taken as positive current

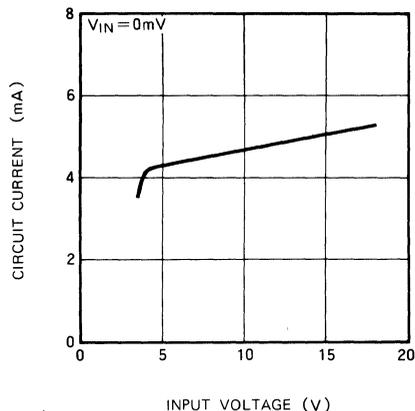
TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=10\text{V}$ unless otherwise noted)

(For the following typical characteristics, R_L in the application example (1) is 1.5k Ω and red GaAlAs LEDs are used for measurements)

**THERMAL DERATING
(MAXIMUM RATING)**

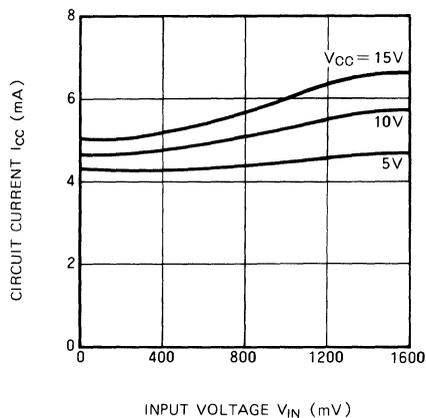


**CIRCUIT CURRENT VS
SUPPLY VOLTAGE**

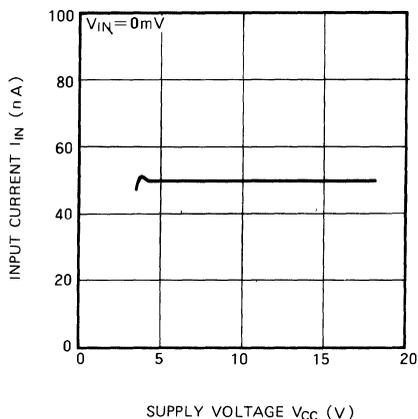


LED LINEAR LEVEL INDICATOR

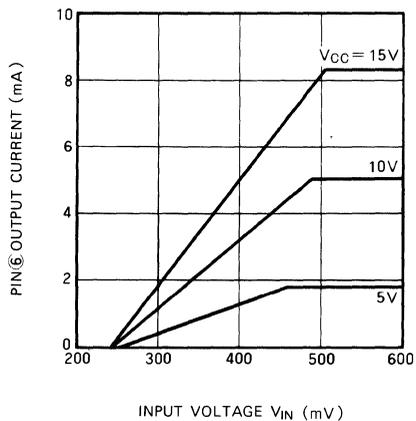
CIRCUIT CURRENT VS INPUT VOLTAGE



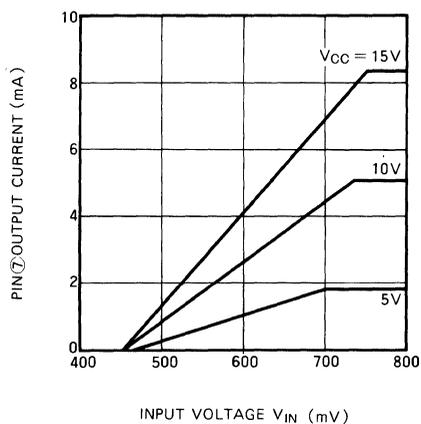
INPUT CURRENT VS SUPPLY VOLTAGE



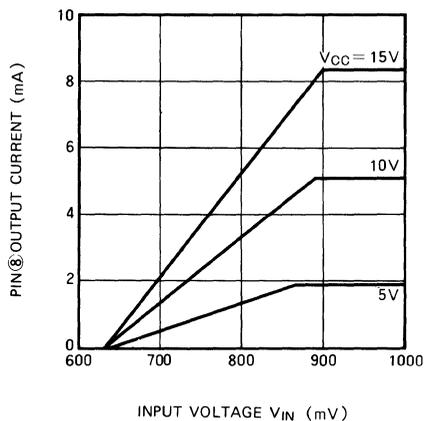
PIN ⑥ OUTPUT CURRENT VS INPUT VOLTAGE



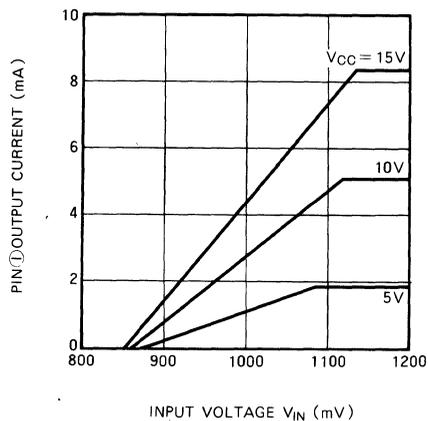
PIN ⑦ OUTPUT CURRENT VS INPUT VOLTAGE



PIN ⑧ OUTPUT CURRENT VS INPUT VOLTAGE

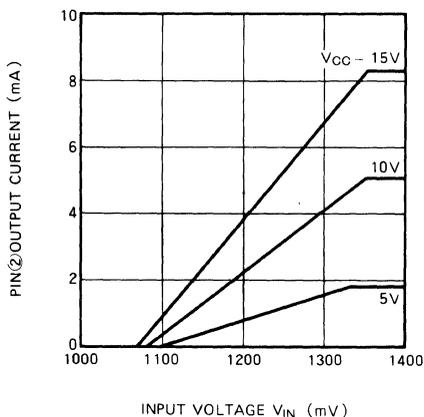


PIN ① OUTPUT CURRENT VS INPUT VOLTAGE

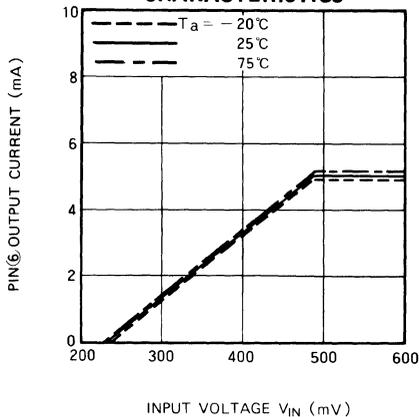


LED LINEAR LEVEL INDICATOR

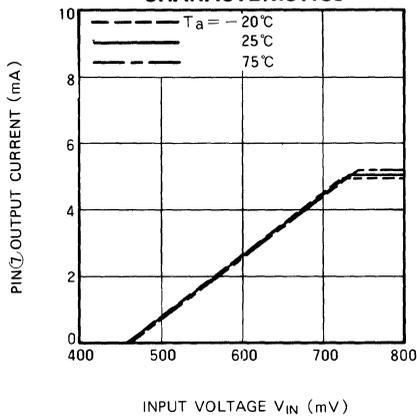
PIN ② OUTPUT CURRENT VS INPUT VOLTAGE



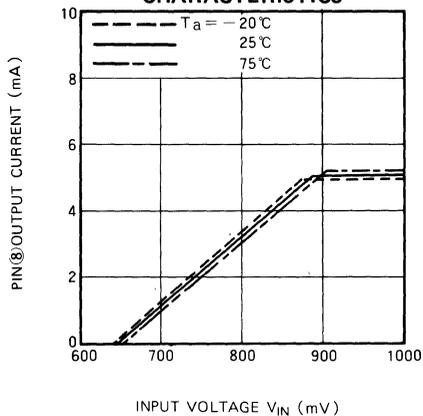
PIN ⑥ OUTPUT CURRENT VS INPUT VOLTAGE TEMPERATURE CHARACTERISTICS



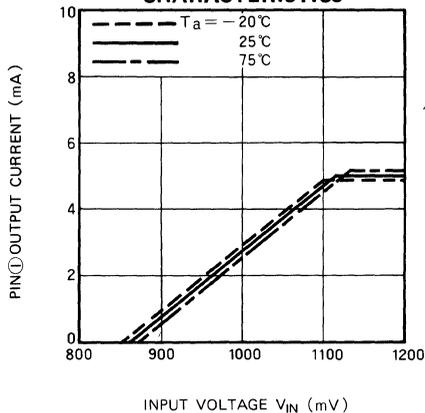
PIN ⑦ OUTPUT CURRENT VS INPUT VOLTAGE TEMPERATURE CHARACTERISTICS



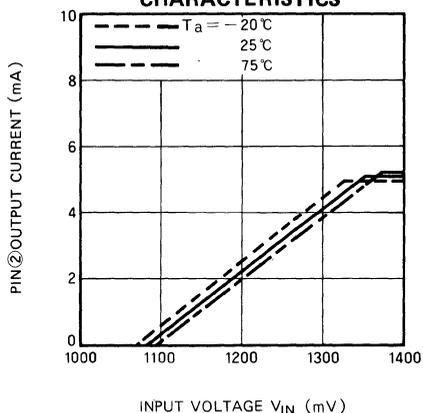
PIN ⑧ OUTPUT CURRENT VS INPUT VOLTAGE TEMPERATURE CHARACTERISTICS



PIN ① OUTPUT CURRENT VS INPUT VOLTAGE TEMPERATURE CHARACTERISTICS



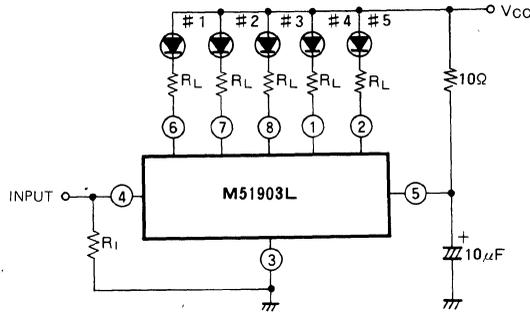
PIN ② OUTPUT CURRENT VS INPUT VOLTAGE TEMPERATURE CHARACTERISTICS



LED LINEAR LEVEL INDICATOR

APPLICATION EXAMPLES

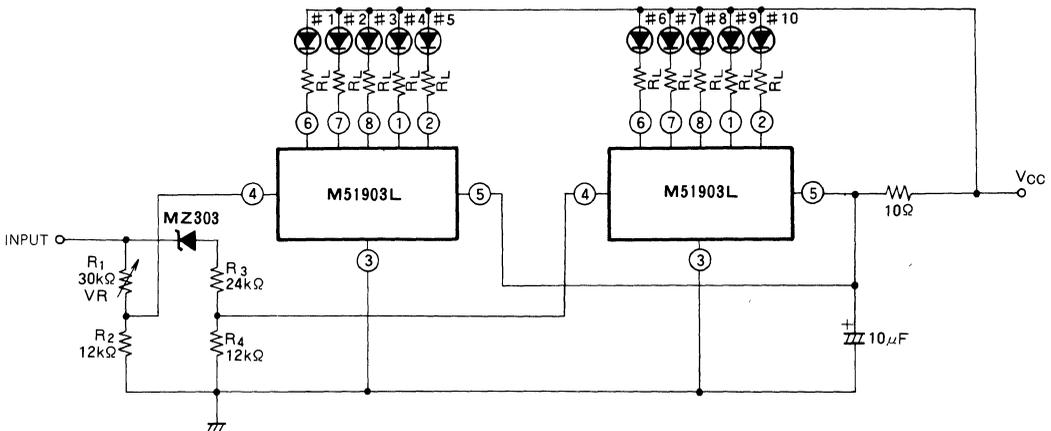
(1) M51903L used alone



Note 1 The value of R_L is chosen to suit the LED devices to be used. The maximum LED current $\approx (V_{CC} - \text{LED forward voltage drop} - 1.0) / R_L \leq 15\text{mA}$

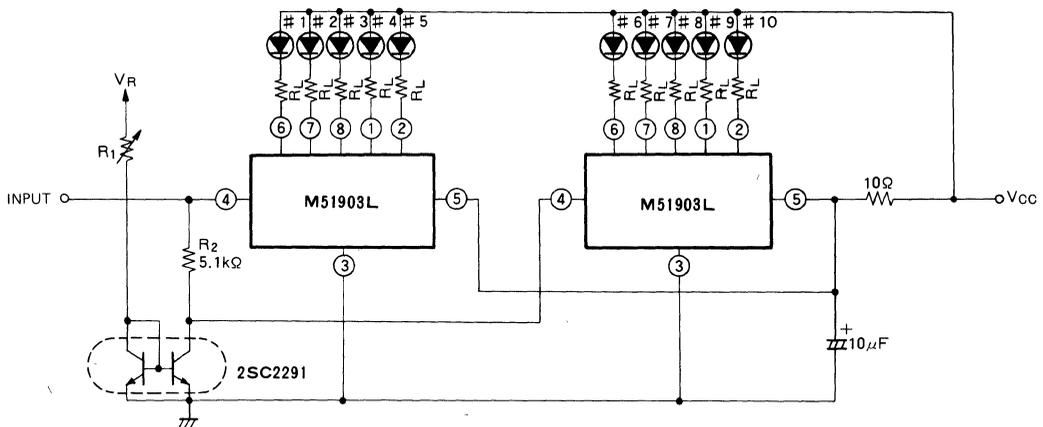
(2) M51903L used in cascade connection

(i) Circuit for a maximum input voltage of 7V



Note 1 Resistance R_1 is selected so that the turn-on voltage for LED #6 is approximately 630mV higher than that of LED #5

(3) Circuit for a maximum input voltage of 2.3V



Note 1 Resistance R_1 selected so that the turn-on voltage for LED #6 is approximately 210mV higher than that of LED #5

M51910P

8-POINT/2-INPUT LED DRIVER

DESCRIPTION

The M51910P is an IC capable of driving a LED input level display according to the LED display mode. It can also be used as a comparator because it can issue a control signal after comparing input values. The M51910P has a built-in timer for control use.

There are two input pins for the display. One input is for the continuous display mode, and the other is for the on-off display mode, so that two inputs can be displayed on the same LED.

FEATURES

- Drives two display modes, displaying two inputs
- Reference voltage can be freely selected0.5~3.5V
- Built-in regulated power supply
..... $V_S=4.8V$, $I_{omax}=10mA$
- Built-in timersetting range is 5min.

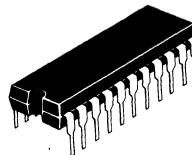
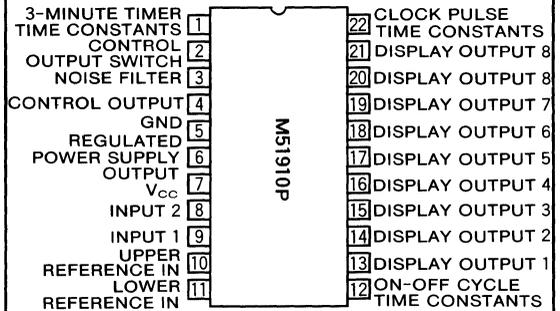
APPLICATION

Control equipment level indicator, thermometer for temperature adjustment devices.

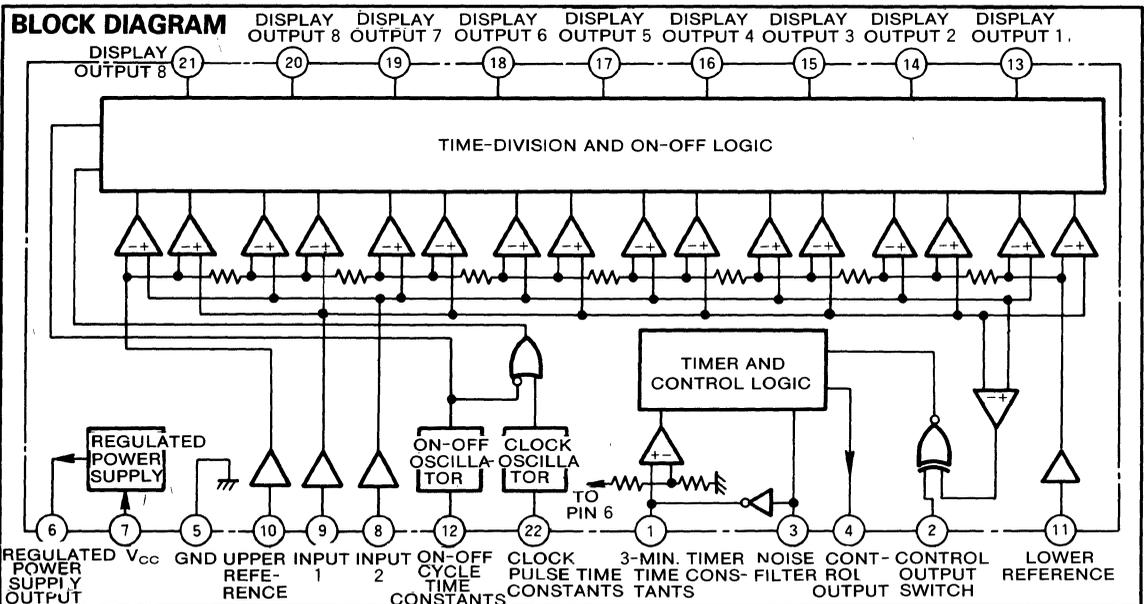
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 8~18V
- Rated supply voltage12V

PIN CONFIGURATION (TOP VIEW)



22-pin molded plastic DIP



8-POINT/2-INPUT LED DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
I_{ODIS}	Display output current		*30	mA
BV_{ODIS}	Display output dielectric strength		18	V
$I_{④}$	Pin ④ input current		50	mA
$BV_{④}$	Pin ④ dielectric strength		30	V
$I_{⑥}$	Pin ⑥ output current		-10	mA
P_d	Power dissipation		1400	mW
K_θ	Thermal derating		1.4	mW/°C
T_{opr}	Operating temperature		-20~+75	°C
T_{stg}	Storage temperature		-40~+125	°C

* The peak current may be as high as 40mA when the clock frequency is over 30Hz.

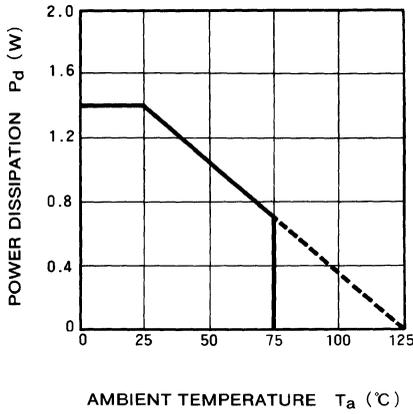
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	Pin ⑦ input current		10	20	mA
V_S	Regulated power voltage	Pin ⑥ voltage	4.0	4.8	5.4	V
$I_{⑧in}$	Pin ⑧ output current			1.7	10	μA
$I_{⑨in}$	Pin ⑨ output current			1.7	10	μA
$I_{⑩in}$	Pin ⑩ output current			1.7	10	μA
$I_{⑪in}$	Pin ⑪ output current			2.7	15	μA
$V_{⑧on2}$	Pin ⑧ threshold voltage output 2	$V_{⑧}=2.2\text{V}$ $V_{⑩}=2.8\text{V}$ Pin ⑧ voltage increases from low level and all outputs go ON.	2.180	2.22	2.250	V
$V_{⑧on3}$	Pin ⑧ threshold voltage output 3		2.265	2.30	2.335	V
$V_{⑧on4}$	Pin ⑧ threshold voltage output 4		2.345	2.38	2.415	V
$V_{⑧on5}$	Pin ⑧ threshold voltage output 5		2.425	2.46	2.495	V
$V_{⑧on6}$	Pin ⑧ threshold voltage output 6		2.505	2.54	2.575	V
$V_{⑧on7}$	Pin ⑧ threshold voltage output 7		2.585	2.62	2.655	V
$V_{⑧on8}$	Pin ⑧ threshold voltage output 8		2.670	2.70	2.740	V
$V_{⑧on9}$	Pin ⑧ threshold voltage output 9		2.750	2.79	2.820	V
$V_{⑨on2}$	Pin ⑨ threshold voltage output 2		$V_{⑨}=2.2\text{V}$ $V_{⑩}=2.8\text{V}$ Pin ⑨ voltage increases from low level and all outputs go ON	2.180	2.22	2.250
$V_{⑨on3}$	Pin ⑨ threshold voltage output 3	2.265		2.38	2.335	V
$V_{⑨on4}$	Pin ⑨ threshold voltage output 4	2.345		2.38	2.415	V
$V_{⑨on5}$	Pin ⑨ threshold voltage output 5	2.425		2.46	2.495	V
$V_{⑨on6}$	Pin ⑨ threshold voltage output 6	2.505		2.54	2.575	V
$V_{⑨on7}$	Pin ⑨ threshold voltage output 7	2.585		2.62	2.655	V
$V_{⑨on8}$	Pin ⑨ threshold voltage output 8	2.670		2.70	2.740	V
$V_{⑨on9}$	Pin ⑨ threshold voltage output 9	2.750		2.79	2.820	V
$V_{⑧TH}$	Pin ⑧ display output hysteresis	Difference between input levels when output is ON and OFF		3	5	7
$V_{⑨TH}$	Pin ⑨ display output hysteresis					
$\Delta V_{⑧}$	Pin ⑧ output i to i+1 step voltage	Difference between input levels when neighboring outputs are ON	70	82	95	mV
$\Delta V_{⑨}$	Pin ⑨ output i to i+1 step voltage					
$\Delta V_{⑧⑨}$	Threshold voltage difference for same output between pin ⑧ and pin ⑨		-12	0	12	mV
F_{CL}	Oscillator frequency	1/2 pin ⑫ oscillator frequency		90		Hz
F_{ONOF}	On-off oscillator frequency	1/2 pin ⑫ oscillator frequency	2			Hz
V_{CTH}	Error detector threshold voltage	$V_{⑧}=2,500\text{V}$ and $V_{⑨}$ changes	2.490	2.505	2.520	V
Δ_{CT}	Error detector hysteresis	Pin ④ inversion voltage and hysteresis	5	15	25	mV
T	Timer operating time	1.5M Ω resistance attached to pin ① 100 Ω F capacity connected to pin ①		200		sec

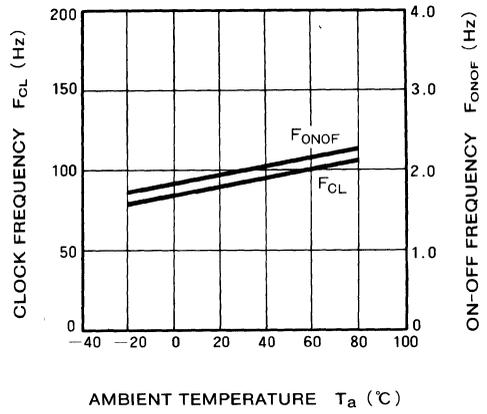
8-POINT/2-INPUT LED DRIVER

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

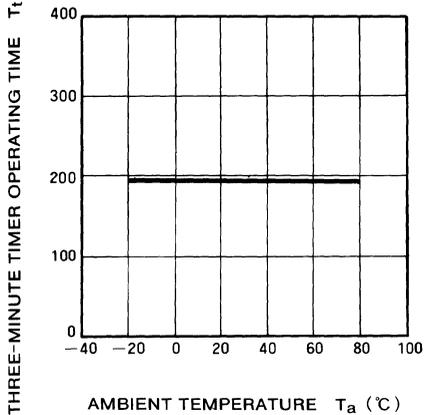
THERMAL DERATING (MAXIMUM RATING)



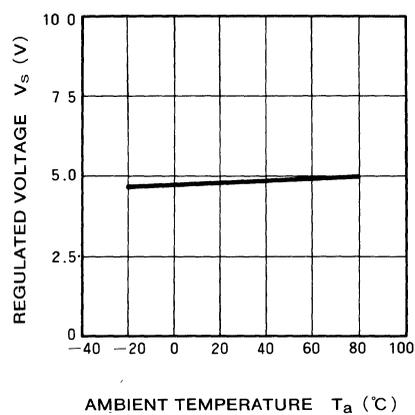
ON-OFF, CLOCK FREQUENCIES VS. AMBIENT TEMPERATURE



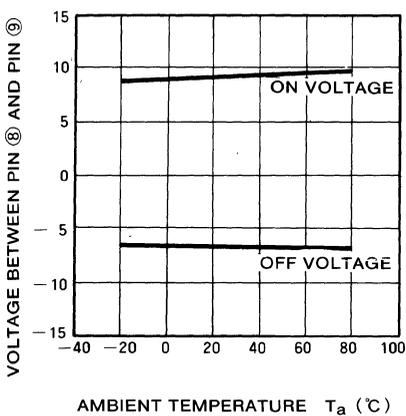
THREE-MINUTE TIMER OPERATING TIME VS. AMBIENT TEMPERATURE



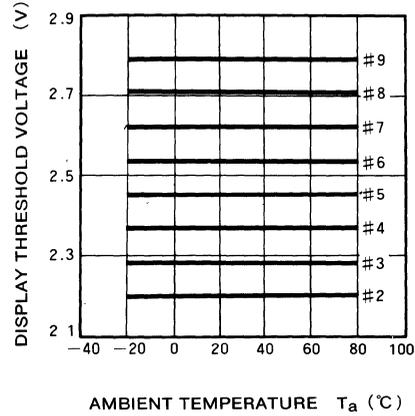
REGULATED VOLTAGE VS. AMBIENT TEMPERATURE



ERROR AMPLIFIER ON-OFF VOLTAGE VS. AMBIENT TEMPERATURE

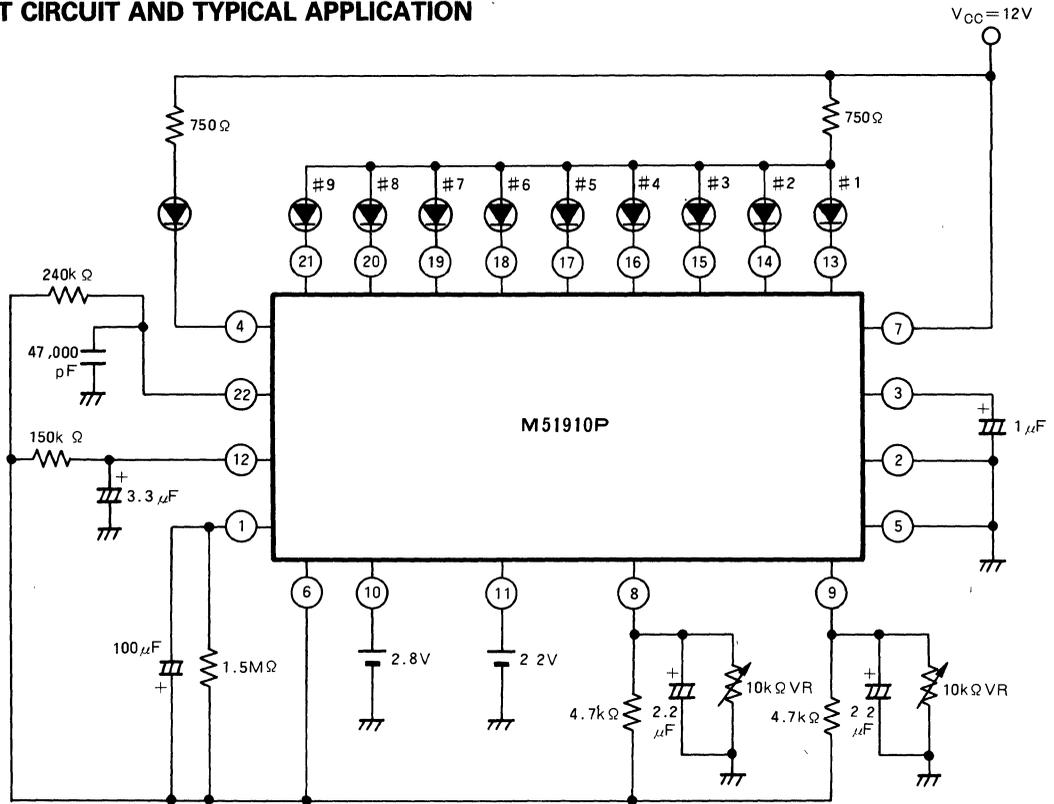


DISPLAY THRESHOLD VOLTAGE VS. AMBIENT TEMPERATURE



8-POINT/2-INPUT LED DRIVER

TEST CIRCUIT AND TYPICAL APPLICATION



M51906P**6-STEP BAR TYPE LED LEVEL INDICATOR****DESCRIPTION**

The M51906P is a semiconductor integrated circuit designed for LED level meters. It functions as a bar-type display for inputs of up to 6 LEDs. With its built-in advanced half-wave rectification operational amplifier, it accepts direct input of either AC or DC signals.

A logarithmic scale of 3, 0, -3, -7, -12, and -18dB display levels is provided, making the M51906P ideal for VU meters applications.

FEATURES

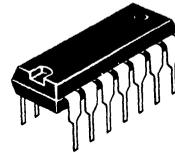
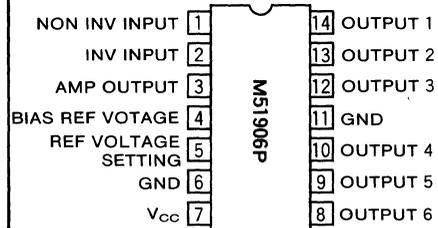
- Advanced half-wave rectification OP amp built in
Cut-off frequency 500kHz(typ.)
Offset voltage 2mV(typ.)
- High output drive current 30mA(max)
- Wide supply voltage range 4V~15V
- Small circuit current when in no-signal condition
..... 0.9mA(typ.)
- Open-collector-type output, so power dissipation is low and low-voltage operation is possible
- Input amp gain is varied by external resistance
- Shifting the LED turn-on voltage is easy with the reference voltage setting pin

APPLICATION

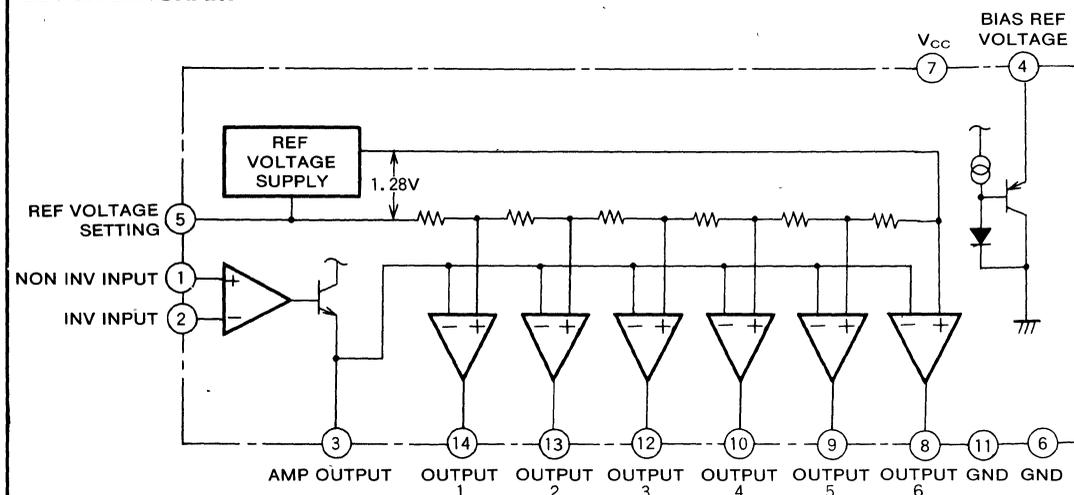
Signal meters, VU meters, and tuning meters.

RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 4~15V
Rated power supply $9V \pm 10\%$

PIN CONFIGURATION (TOP VIEW)

14-pin molded plastic DIP

BLOCK DIAGRAM

6-STEP BAR TYPE LED LEVEL INDICATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		16	V
BV_O	Output voltage		16	V
I_O	Output current		30	mA
V_{IN}	Input voltage	Input terminal to GND	$-3 \sim V_{CC} - 0.8$	V
$ V_{①} - V_{②} $	Input differential voltage	Pin ① to pin ②	5	V
$V_{⑤}$	Pin ⑤ voltage	Pin ⑤ to GND	4	V
$I_{④}$	Pin ④ input current		1.5	mA
$I_{③}$	Pin ③ output current		-1	mA
P_d	Power dissipation		1500	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	12	mW/ $^\circ\text{C}$
T_{opr}	Operating temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^\circ\text{C}$
$V_{③}$	Pin ③ voltage	Pin ③ to GND	6	V

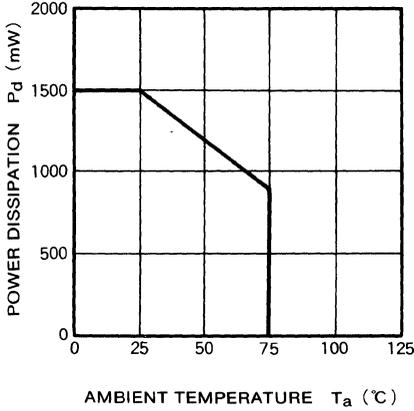
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=9\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4.0		15.0	V
I_{CC1}	Circuit current	$V_{①} - V_{⑤} = 0\text{V}$ (when all LEDs are off)		0.9	2	mA
I_{CC2}	Circuit current	$V_{①} - V_{⑤} = 2\text{V}$ (when all LEDs are on)		10	20	mA
V_{IO}	Input amp offset	Input voltage $V_{①} = 2\text{V}$		2	10	mV
I_{IB}	Input amp bias current	Input voltage $V_{①} = 2\text{V}$		50	250	nA
V_{ref}	Reference voltage	Pin ④ voltage	1.15	1.35	1.55	V
$I_{⑤}$	Pin ⑤ output current	$V_{⑤} = 0\text{V}$	-600	-400	-260	μA
V_{SAT}	Output saturation voltage	$I_{sink} = 30\text{mA}$		0.3	1	V
V_{th1}	Output 1 threshold voltage	Amp gain=1 Pin ⑤ voltage is taken as reference	91	114	144	mV
V_{th2}	Output 2 threshold voltage		-20	-18	-16	dB
V_{th3}	Output 3 threshold voltage		181	228	287	mV
V_{th4}	Output 4 threshold voltage		-14	-12	-10	dB
V_{th5}	Output 5 threshold voltage		341	405	481	mV
V_{th6}	Output 6 threshold voltage		-8.5	-7	-5.5	dB
			572	641	720	mV
			-4	-3	-2	dB
			807	906	1017	mV
			-1	0	+1	dB
			1141	1280	1436	mV
			+2	+3	+4	dB
I_{OL}	Output leak current	$V_{OUT} = V_{CC}$			1	μA

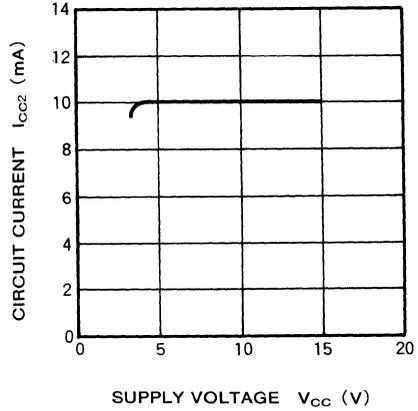
6-STEP BAR TYPE LED LEVEL INDICATOR

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

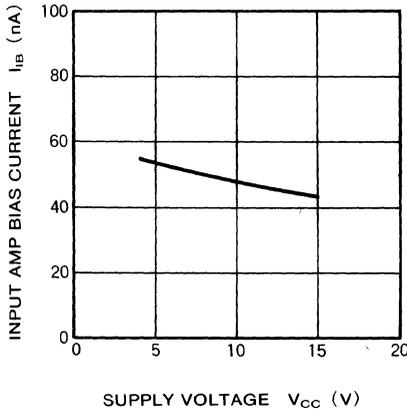
THERMAL DERATING (MAXIMUM RATING)



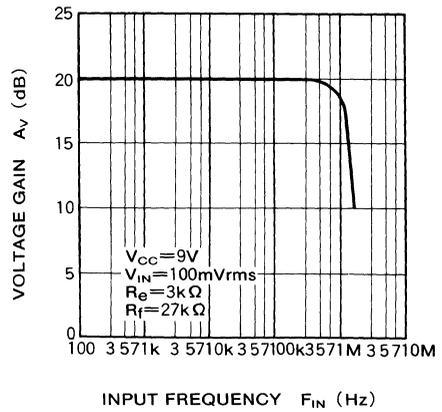
CIRCUIT CURRENT VS. SUPPLY VOLTAGE



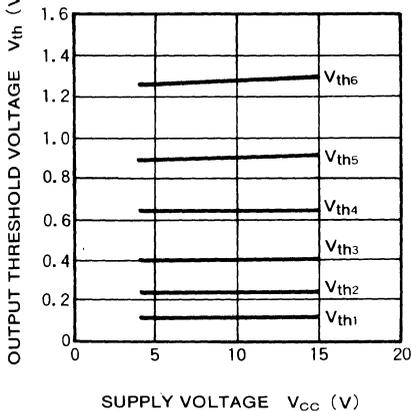
INPUT AMP BIAS CURRENT VS. SUPPLY VOLTAGE



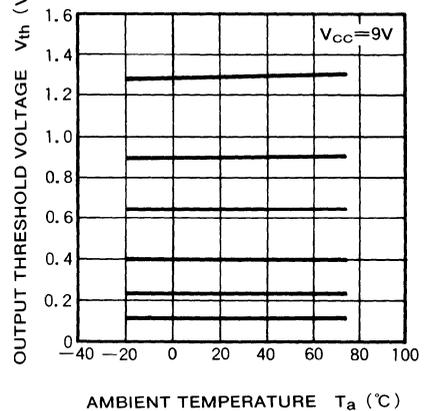
VOLTAGE GAIN VS. INPUT FREQUENCY



OUTPUT THRESHOLD VOLTAGE VS. SUPPLY VOLTAGE



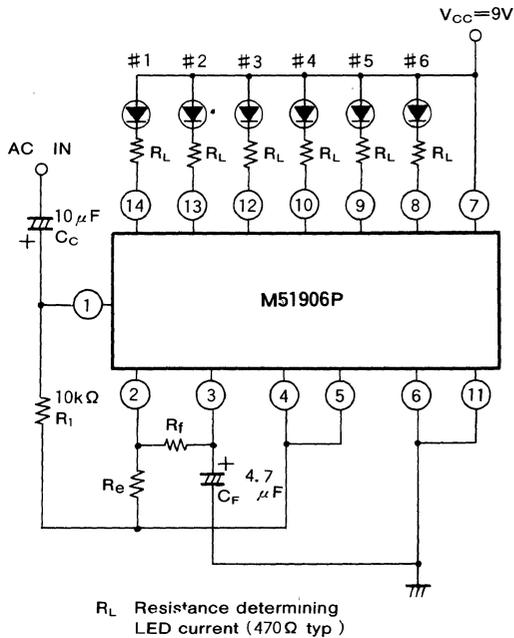
OUTPUT THRESHOLD VOLTAGE VS. AMBIENT TEMPERATURE



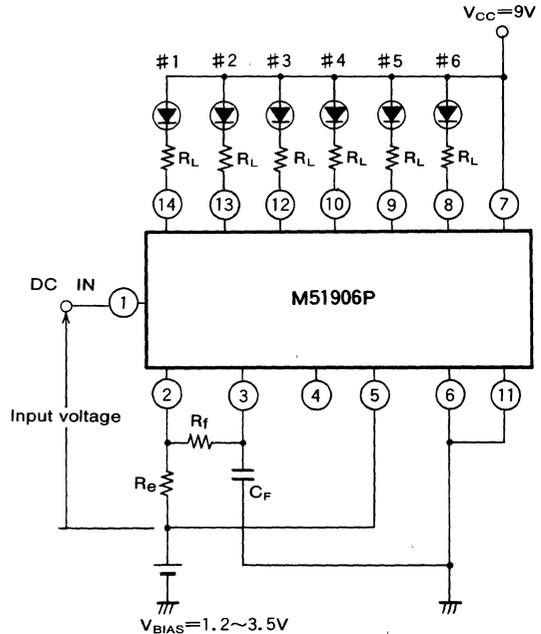
6-STEP BAR TYPE LED LEVEL INDICATOR

APPLICATION EXAMPLES

(1) AC INPUT



(2) DC INPUT



OPERATION

The AC signal applied through the coupling capacitor C_C is rectified and amplified mainly at reference Pin ④, and then output at Pin ③. The voltage at Pin ③ is compared with the reference voltage at Pin ⑤ (internal reference voltage 1.28V divided on a logarithmic scale), and the LED is driven according to the comparator output.

Note 1. Amp gain

$$\frac{R_e + R_f}{R_e} \quad (R_e + R_f \sim 30k\Omega)$$

The peak value is output at Pin ③, so a DC voltage of approximately 1.4 times the rms value is output.

Note 2. Select the value of C_F according to the recovery time.

Recovery time: $C_F \times (R_e + R_f)$

Attack time: $C_F \times 430\Omega$

Note 3. LED current

$$\frac{V_{CC} - V_{FLED} - V_{SAT}}{R_L}$$

V_{FLED} : LED forward voltage drop

V_{SAT} : IC output saturation voltage

Note 4. Pin ④ voltage is approximately 1.35V at normal temperature (25°C), the temperature coefficient is $-4mV/^\circ C$.

Note 5. If less than 6 LEDs are used, leave the unused output terminals open.

Note 6. When using DC input, leave Pin ④ open and connect the external reference voltage of 1.2~3.5V to Pin ⑤

CAUTION

1. If the power supply connection positive and negative terminals are reversed, the M51906P will lose a lot of current and this can cause serious damage to the device.
2. The output is an open collector, so load resistances must be connected in series with the LEDs.

M51907P

8-STEP BAR TYPE LED LEVEL INDICATOR

DESCRIPTION

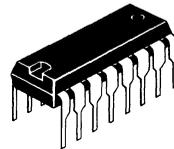
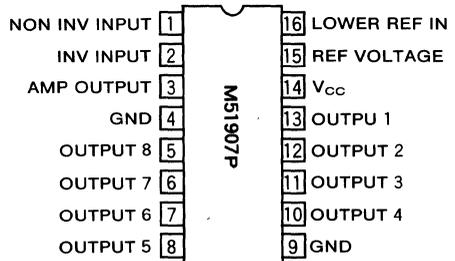
The M51907P is a semiconductor integrated circuit designed for LED level meters. It drives a bar-type input-level display for up to 8 LEDs. With its built-in advanced half-wave rectification operational amplifier, the M51907P accepts direct input of either AC or DC signals.

A logarithmic scale of 5, 2, 0, -2, -5, -8, -13, and -18dB display levels is provided, making the M51907P ideal for signal meter applications.

FEATURES

- Advanced half-wave rectification OP amp built in
Cut-off frequency 200kHz (typ.)
Offset voltage 2mV (typ.)
- Output current can be adjusted with a single resistor
..... 2~25mA
- Wide supply voltage range 4~15V
- LEDs are grouped in two's by cascade connection, so when all LEDs are driven in parallel, the current required is reduced by half
- Parallel shifting of the LED turn-on voltage is easy with the lower ref in pin
- Advanced half-wave OP amp gain is varied by external resistance
- Reference voltage terminal for full-scale coordination, making it easy to use the M51907P and the companion M51909P in cascade connection to drive more than 8 LEDs.

PIN CONFIGURATION (TOP VIEW)



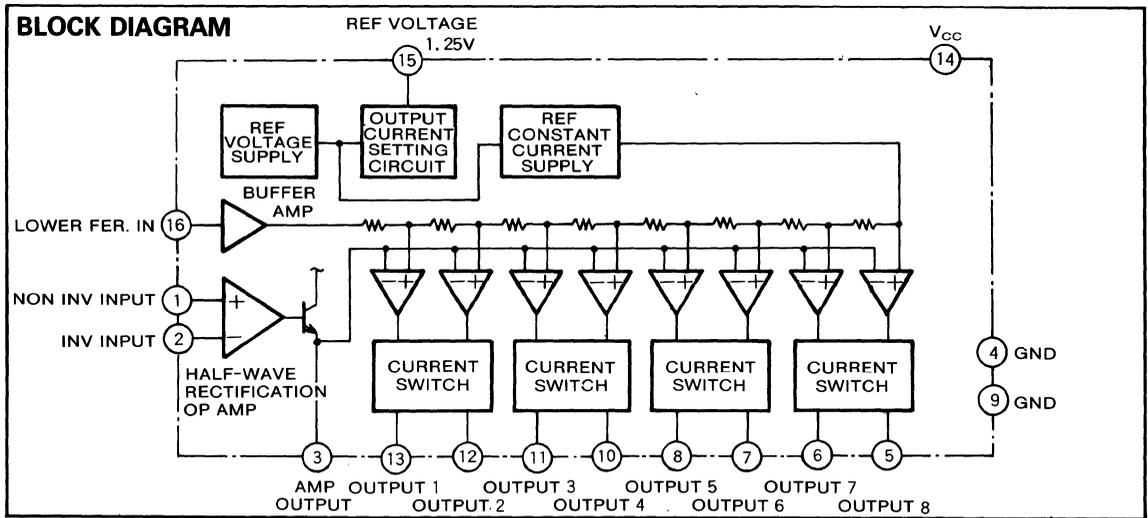
16-pin molded plastic DIP

APPLICATION

Signal meters, VU meters, and tuning meters.

RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 4~15V
- Rated power supply 9V±10%



8-STEP BAR TYPE LED LEVEL INDICATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

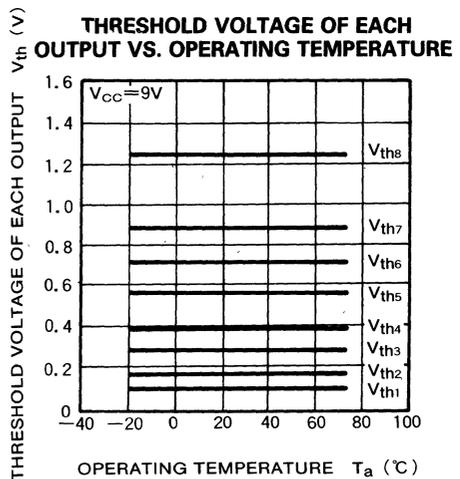
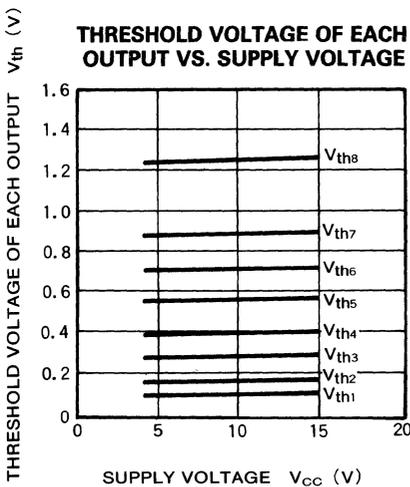
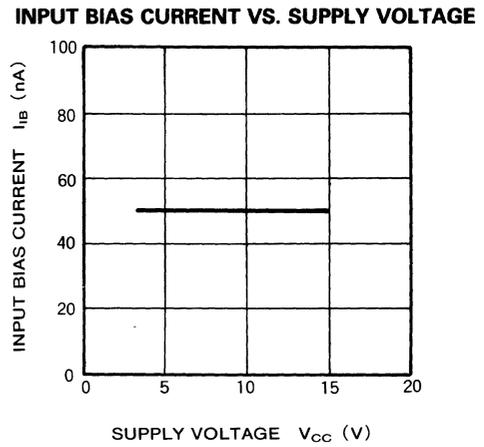
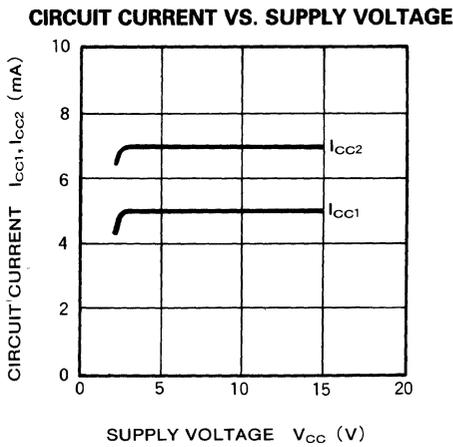
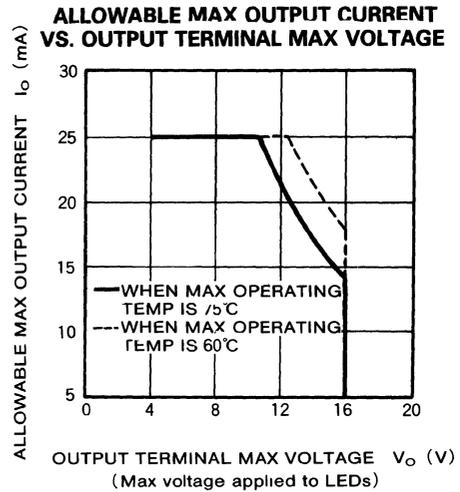
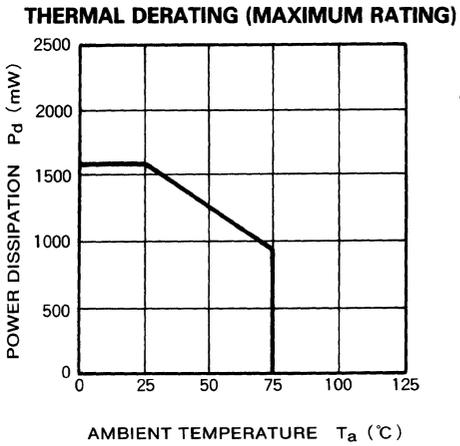
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		16	V
BV_O	Output voltage		16	V
I_O	Output current		25	mA
V_{IN}	Input voltage	Input pin to GND	$-3 \sim V_{CC}$	V
$ V_{①}-V_{②} $	Input differential voltage	Pin ① to pin ②	5	V
$V_{⑫}$	Pin ⑫ voltage	Pin ⑫ to GND	V_{CC}	V
$I_{⑫}$	Pin ⑫ input current		500	μA
$I_{③}$	Pin ③ output current	Static value	-1	mA
P_d	Power dissipation		1600	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	12.8	mW/ $^\circ\text{C}$
T_{opr}	Operating temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=9\text{V}$, unless otherwise noted) R_{ad} : Pin ⑫ adjustable resistor

Symbol	Parameter	Test conditions	Limits			Unit	
			Min	Typ	Max		
V_{CC}	Supply voltage		4.0		15.0	V	
I_{CC1}	Circuit current	$V_{①}=0\text{V}$ (when all LEDs are off) $R_{ad}=2.8\text{k}\Omega$		5.0	8.0	mA	
I_{CC2}	Circuit current	$V_{①}=2\text{V}$ (when all LEDs are on) $R_{ad}=2.8\text{k}\Omega$		7.0	11.2	mA	
V_{IO}	Input offset voltage	Input voltage $V_{①}=1\text{V}$		2	10	mV	
I_{IB}	Input bias current	Input voltage $V_{①}=0\text{V}$	-300	-50		nA	
V_{IN}	Input voltage range		0		$V_{CC}-2$	V	
V_{ref}	Reference voltage	$R_{ad}=6.4\text{k}\Omega$	1.125	1.250	1.375	V	
$V_{⑫}$	Pin ⑫ set voltage range		-0.2		$V_{CC}-3.5$	V	
$I_{⑫}$	Pin ⑫ output current		-2000	-50		nA	
V_{th1}	Output 1 threshold voltage	Amp gain=1 Pin ⑫ voltage is taken as reference	70	89	111	mV _{DC}	
V_{th2}	Output 2 threshold voltage		-20	-18	-16	dB	
V_{th3}	Output 3 threshold voltage		125	157	198	mV _{DC}	
V_{th4}	Output 4 threshold voltage		-15	-13	-11	dB	
V_{th5}	Output 5 threshold voltage		235	280	333	mV _{DC}	
V_{th6}	Output 6 threshold voltage		-9.5	-8	-6.5	dB	
V_{th7}	Output 7 threshold voltage		352	395	443	mV _{DC}	
V_{th8}	Output 8 threshold voltage		-6	-5	-4	dB	
I_{OL}	Output leak current					1	μA
I_O	Output current		$R_{ad}=6.4\text{k}\Omega$	9.6	12	14.4	mA
I_O'	Output current		$R_{ad}=2.8\text{k}\Omega$	20	25	30	mA
V_{SAT}	Output saturation voltage		$R_{ad}=2.8\text{k}\Omega$ $I_O=12.5\text{mA}$			500	mV

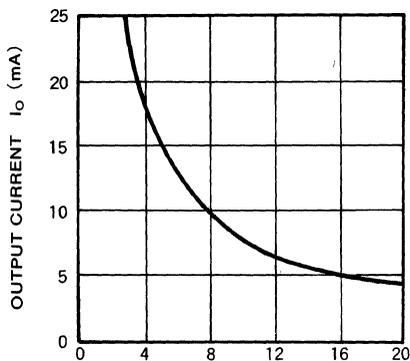
8-STEP BAR TYPE LED LEVEL INDICATOR

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)



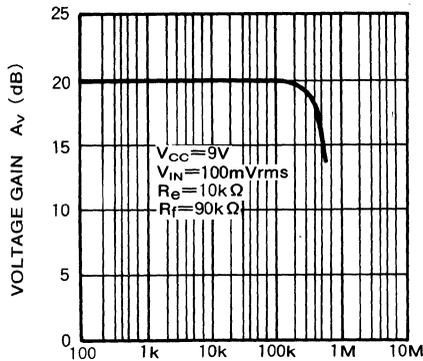
8-STEP BAR TYPE LED LEVEL INDICATOR

**OUTPUT CURRENT VS.
 PIN 15 ADJUSTABLE RESISTOR**



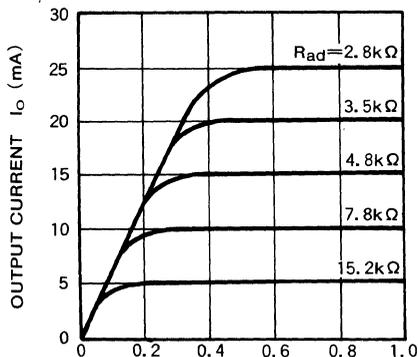
PIN 15 ADJUSTABLE RESISTOR R_{ad} (k Ω)

**HALF-WAVE RECTIFIER OP AMP
 VOLTAGE GAIN VS. INPUT FREQUENCY**



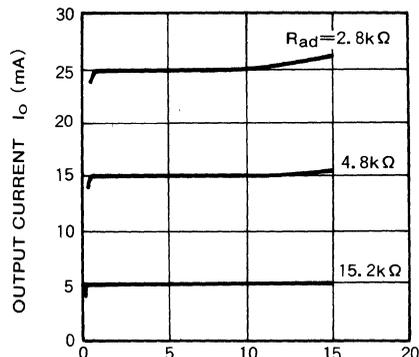
INPUT FREQUENCY F_{IN} (Hz)

**OUTPUT CURRENT VS.
 OUTPUT VOLTAGE (1)**



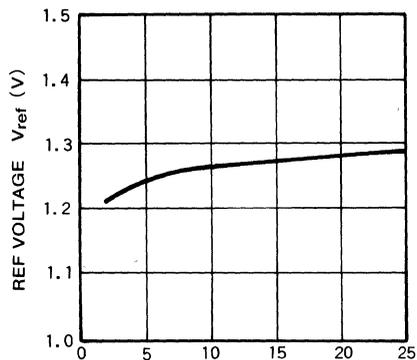
OUTPUT VOLTAGE V_o (V)

**OUTPUT CURRENT VS.
 OUTPUT VOLTAGE (2)**



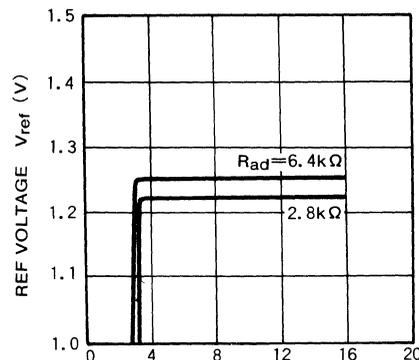
OUTPUT VOLTAGE V_o (V)

**REF VOLTAGE VS. PIN 15
 ADJUSTABLE RESISTOR**



PIN 15 ADJUSTABLE RESISTOR R_{ad} (k Ω)

REF VOLTAGE VS. SUPPLY VOLTAGE

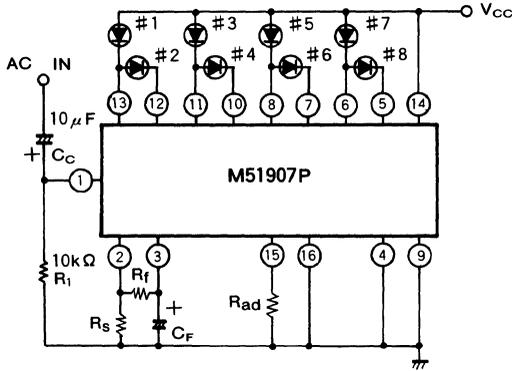


SUPPLY VOLTAGE V_{cc} (V)

8-STEP BAR TYPE LED LEVEL INDICATOR

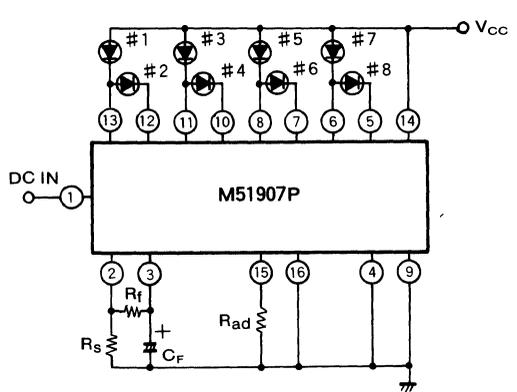
APPLICATION EXAMPLES

(1) AC INPUT



Rad: Resistance determining LED current
 (See OUTPUT CURRENT VS. PIN 15 ADJUSTABLE RESISTANCE graph)

(2) DC INPUT



OPERATION

The AC signal applied through the coupling capacitor C_C is rectified and amplified mainly by the GND voltage and output at Pin 3. The voltage at Pin 3 is compared with the reference voltage at Pin 16 (internal reference voltage 1.25V divided on a logarithmic scale), and the resultant comparator output drives the LEDs.

Note 1. Amp gain

$$\frac{R_S + R_f}{R_S}$$

The peak value is output at Pin 3, so when this is a sine wave, a DC voltage of approximately 1.4 times the rms value is output.

Note 2. Select the value of C_F according to the recovery time.

Recovery time: $C_F \times (R_S + R_f)$

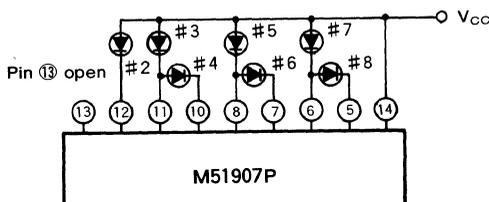
Attack time: $C_F \times 460 \Omega$

Note 3. LED current

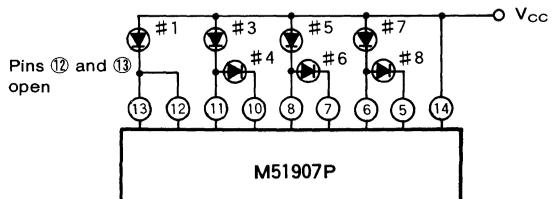
Pin 15 adjustable resistor determines the LED current. (See the OUTPUT CURRENT VS. PIN 15 ADJUSTABLE RESISTOR graph)

Note 4. When less than 8 LEDs are used, proceed as follows.

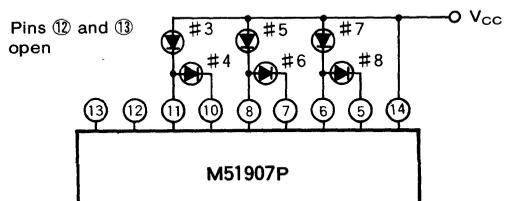
1. When #1 LED is not used:



2. When #2 LED is not used:

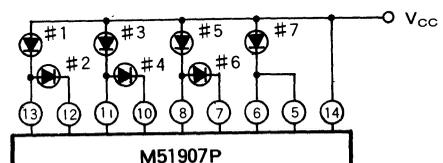


3. When #1 and #2 LEDs are not used:



4. When any of LEDs #3 to #8 are not used:

When an even-numbered LED is not used, follow the procedure given for LED #2, and when an odd-numbered LED is not used, follow that for LED #3.

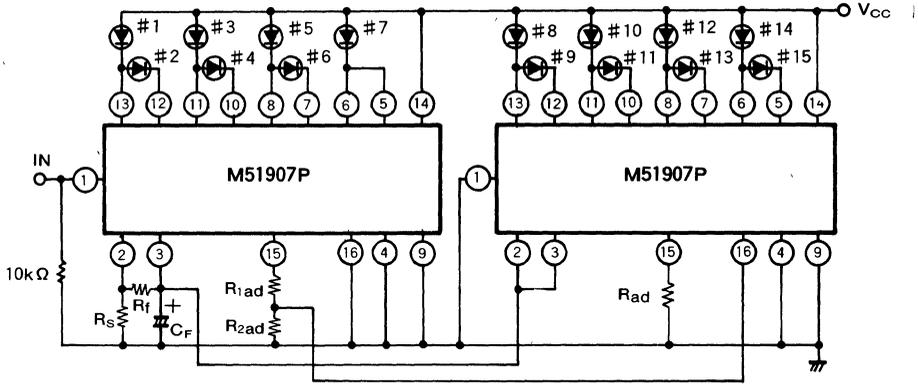


8-STEP BAR TYPE LED LEVEL INDICATOR

(3) Cascade connection

It is easy to create a 16-step indicator by using two M51907Ps, but the drive threshold value (dB) is irregular, so there is a tendency to unnatural deviation in the drive threshold value at the connection. The following

example illustrates how this problem is avoided. Fifteen LEDs can be driven using this configuration with the M51907P and the M51909P.

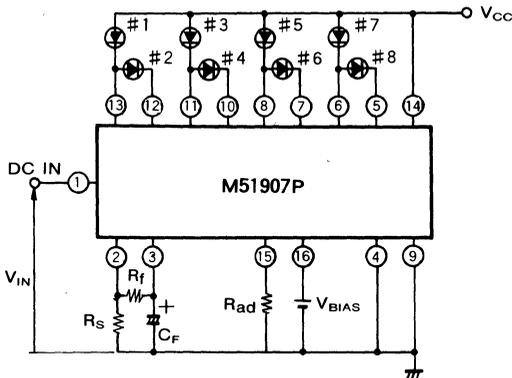


Note 1. To equalize the brightness levels of the M51907P and the M51909P, select R_{1ad} and R_{2ad} so that $R_{ad} = R_{1ad} + R_{2ad}$.

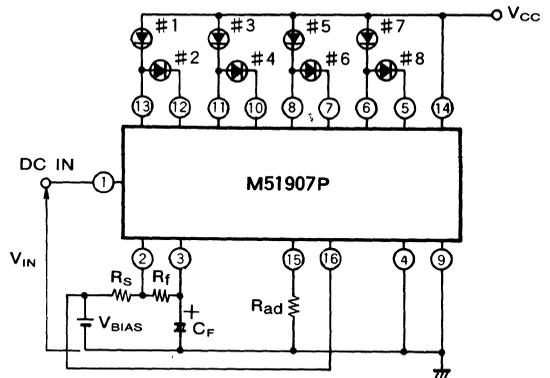
Note 2. When values for R_{1ad} and R_{2ad} are selected so that $R_{1ad}:R_{2ad} = 1:2.53$, the threshold values are as follows.

#1	#2	#3	#4	#5	#6	#7	#8	#9	#10	#11	#12	#13	#14	#15
-24.6dB	-19.7dB	-14.7dB	-11.7dB	-8.7dB	-6.7dB	-4.7dB	-3.2dB	-2.0dB	-0.9dB	0dB	0.86dB	1.63dB	2.34dB	3dB

(4) Fixed reference bias DC input



Display is given by the formula $V_{IN} \times \frac{R_S + R_f}{R_S} - V_{BIAS}$



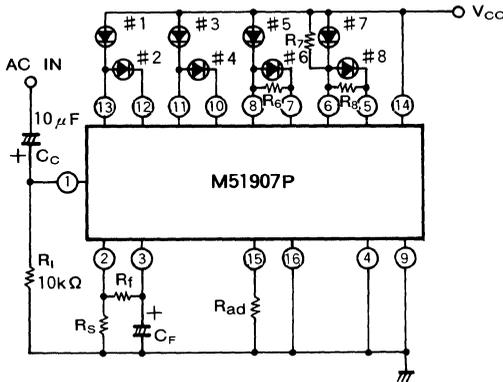
Display is given by the formula $(V_{IN} - V_{BIAS}) \times \frac{R_S + R_f}{R_S}$

8-STEP BAR TYPE LED LEVEL INDICATOR

(5) Different color LEDs (when different drive currents are desired for the LEDs)

Select Pin ⑮ adjustable resistor for connection to the LED that requires the maximum drive current. Then connect resistors in parallel to the remaining LEDs with reduced current.

For example, to reduce the current for LEDs #6, #7, and #8, use the following arrangement.



The currents flowing through each of the LEDs in the above circuit configuration are given by the following formulas.

For LEDs #1~#5

$$I_{LEDA} \sim 60 \times \frac{V_{⑮}}{R_{ad}}$$

For LED #6

$$I_{LEDB} \sim 60 \times \frac{V_{⑮}}{R_{ad}} - \frac{V_{F6}}{R_6} = I_{LEDA} - \frac{V_{F6}}{R_6}$$

For LED #7

$$I_{LEDC} \sim 60 \times \frac{V_{⑮}}{R_{ad}} - \frac{V_{F7}}{R_7} = I_{LEDA} - \frac{V_{F7}}{R_7}$$

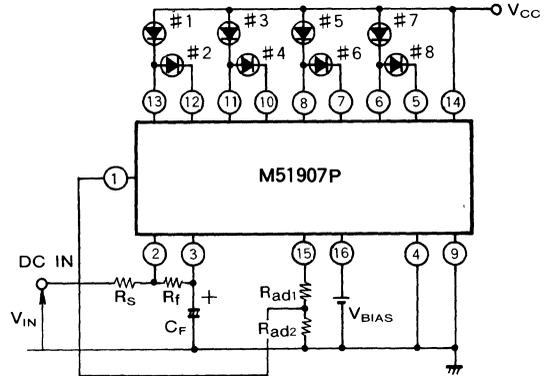
For LED #8

$$I_{LEDD} \sim 60 \times \frac{V_{⑮}}{R_{ad}} - \frac{V_{F8}}{R_8} = I_{LEDA} - \frac{V_{F8}}{R_8}$$

$V_{⑮}$, Pin ⑮ voltage, is 1.25V and V_{F6} , V_{F7} , and V_{F8} are the forward voltage drops for LEDs #6, #7, and #8 respectively.

The M51907P is designed so that the temperature-dependency of the voltage at Pin ⑮ is very low. The LED forward voltage drop, only -2 to $-2.5\text{mV}/^\circ\text{C}$, also has a relatively low temperature dependence. These design features make it possible to accurately set the current flowing through each LED.

(6) All LEDs light at minimum-voltage input signal, and the LEDs are extinguished as the input voltage increases



Display follows the formula:

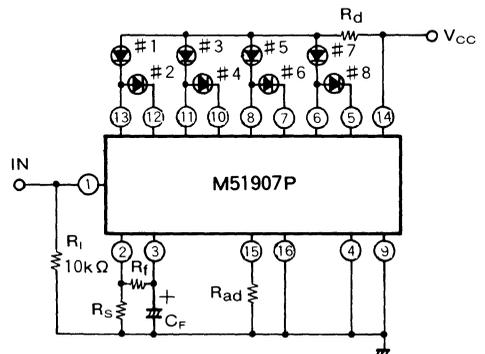
$$\frac{1.25(V) \times R_{ad2}}{R_{ad1} + R_{ad2}} \times \left(1 + \frac{R_f}{R_s}\right) - V_{IN} \times \frac{R_f}{R_s} - V_{BIAS}$$

Note LED drive voltage is determined by $R_{ad1} + R_{ad2}$

(7) High supply voltage produces high output currents

The relation between the allowable output current and the maximum voltage at the output terminal is determined by the thermal derating curve. (See the ALLOWABLE OUTPUT CURRENT VS. OUTPUT TERMINAL MAXIMUM VOLTAGE graph)

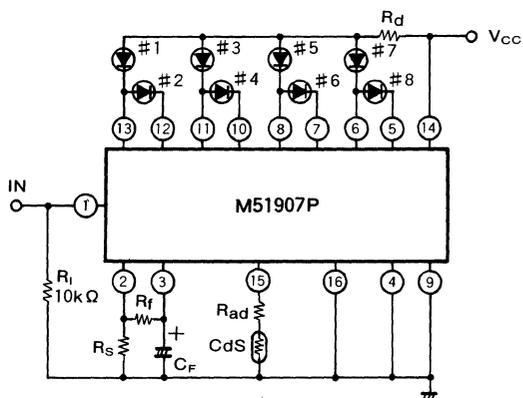
Therefore, if the user wants a high supply voltage to produce high output currents, the following circuit configuration can be used, but the output terminal voltage must be low



A realistic design involves taking the required output current (LED current I_{LED}) as the allowable maximum output current, and then determining the corresponding output terminal maximum voltage V_{Omax} from the ALLOWABLE OUTPUT CURRENT VS. OUTPUT TERMINAL MAX VOLTAGE graph. Then if the maximum value for the supply voltage is V_{CCmax} , the value of R_d has to be determined to satisfy the relation $4 \cdot I_{LED} \cdot R_d > V_{CCmax} - V_{Omax}$.

8-STEP BAR TYPE LED LEVEL INDICATOR

(8) Adjusting LED brightness with respect to the surrounding light



The LED current is determined by $R_{ad} + R_{cds}$.

MITSUBISHI LINEAR ICs
M51909P

8-STEP BAR TYPE LED LEVEL INDICATOR

DESCRIPTION

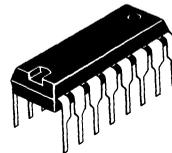
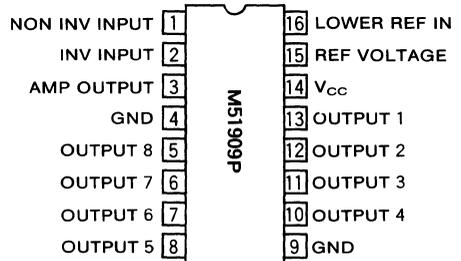
The M51909P is a semiconductor integrated circuit designed for LED level meters. It drives a bar-type input-level display for up to 8 LEDs. With its built-in advanced half-wave rectification operational amplifier, the M51909P accepts direct input of either AC or DC signals.

A linear scale of display levels is provided, making the M51909P especially suitable for signal meter applications.

FEATURES

- Advanced half-wave rectification OP amp built in
 Cut-off frequency 200kHz(typ.)
 Offset voltage 2mV(typ.)
- Output current can be adjusted with a single resistor
 2~25mA
- Wide supply voltage range 4~15V
- LEDs are grouped in two's by cascade connection, so when all LEDs are driven in parallel, the current required is reduced by half
- Parallel shifting of the LED turn-on voltage is easy with the lower ref in pin
- Advanced half-wave OP amp gain is varied by external resistance
- Reference voltage terminal for full-scale coordination, making it easy to use the M51907P and the M51909P in cascade connection to drive more than 8 LEDs

PIN CONFIGURATION (TOP VIEW)



16-pin molded plastic DIP

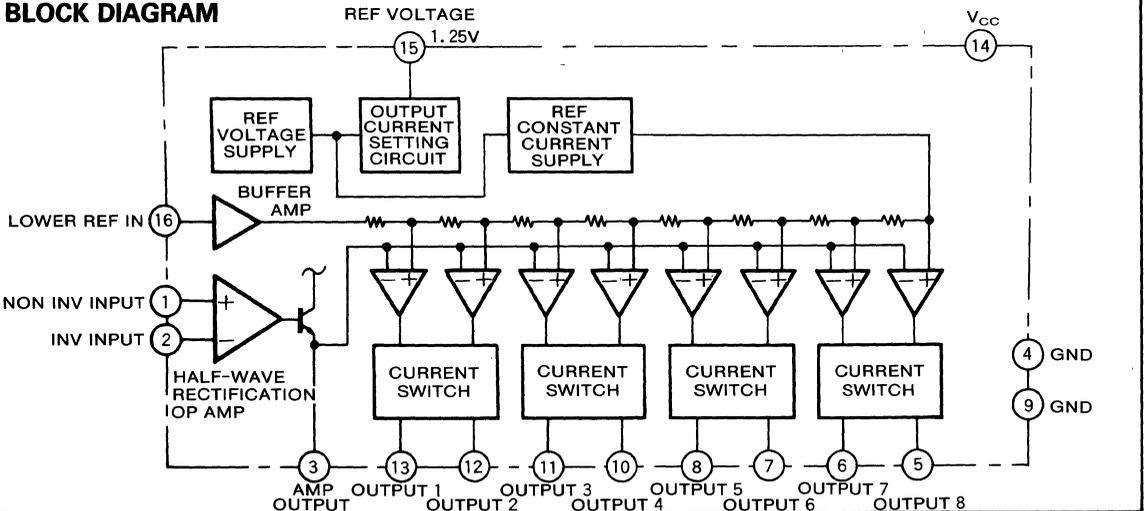
APPLICATION

Signal meters, VU meters, and tuning meters.

RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 4~15V
- Rated power supply 9V±10%

BLOCK DIAGRAM



8-STEP BAR TYPE LED LEVEL INDICATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

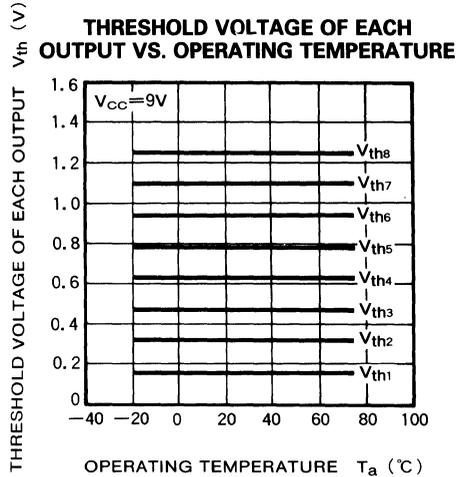
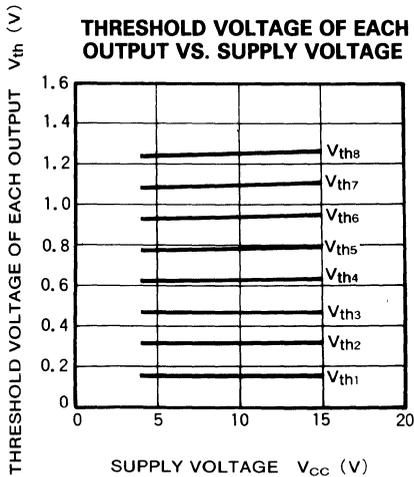
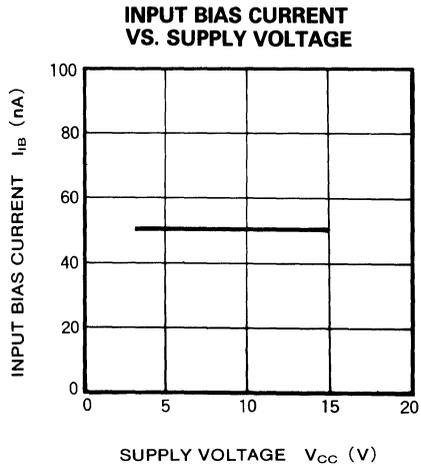
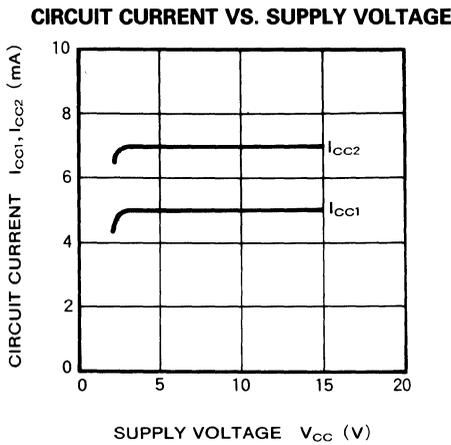
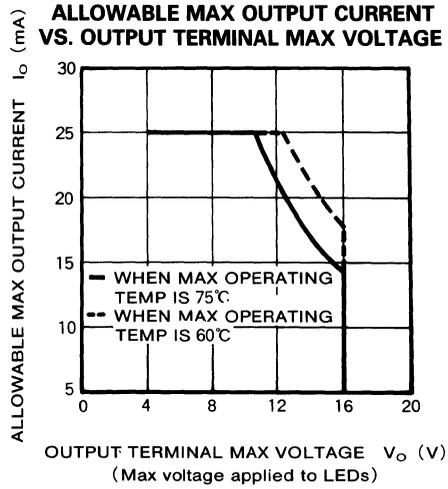
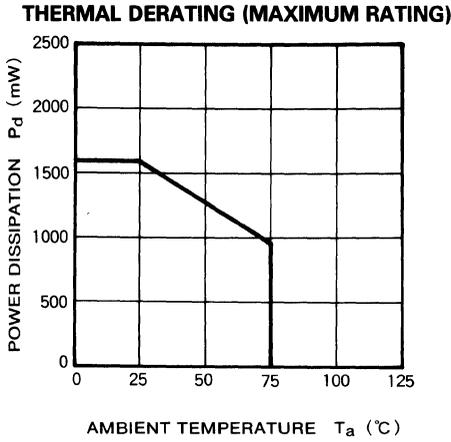
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		16	V
BV_O	Output voltage		16	V
I_O	Output current		25	mA
V_{IN}	Input voltage	Input pin to GND	$-3\sim V_{CC}$	V
$ V_{(1)}-V_{(2)} $	Input differential voltage	Pin ① to pin ②	5	V
$V_{(8)}$	Pin ⑧ voltage	Pin ⑧ to GND	V_{CC}	V
$I_{(8)}$	Pin ⑧ input current		-500	μA
$I_{(3)}$	Pin ③ output current	Static value	-1	mA
P_d	Power dissipation		1600	mW
K_θ	Thermal derating	$T_a\geq 25^\circ\text{C}$	12.8	mW/ $^\circ\text{C}$
T_{opr}	Operating temperature		$-20\sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature		$-40\sim +125$	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=9\text{V}$, unless otherwise noted) R_{ad} Pin ⑧ adjustable resistor

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4.0		15.0	V
I_{CC1}	Circuit current	$V_{(1)}=0\text{V}$ (when all LEDs are off) $R_{ad}=2.8\text{k}\Omega$		5.0	8.0	mA
I_{CC2}	Circuit current	$V_{(1)}=2\text{V}$ (when all LEDs are on) $R_{ad}=2.8\text{k}\Omega$		7.0	11.2	mA
V_{IO}	Input offset voltage	Input voltage $V_{(1)}=1\text{V}$		2	10	mV
I_{IB}	Input bias current	Input voltage $V_{(1)}=0\text{V}$	-300	-50		nA
V_{IN}	Input voltage range		0		$V_{CC}-2$	V
V_{ref}	Reference voltage	$R_{ad}=6.4\text{k}\Omega$	1.125	1.250	1.375	V
$V_{(8)}$	Pin ⑧ set voltage		-0.2		$V_{CC}-3.5$	V
$I_{(8)}$	Pin ⑧ output current		-2000	-50		nA
V_{th1}	Output 1 threshold voltage	Amp gain=1 Pin ⑧ voltage is taken	136	156	177	mV
V_{th2}	Output 2 threshold voltage		276	313	349	mV
V_{th3}	Output 3 threshold voltage		417	469	521	mV
V_{th4}	Output 4 threshold voltage		558	625	693	mV
V_{th5}	Output 5 threshold voltage		698	781	864	mV
V_{th6}	Output 6 threshold voltage		839	938	1036	mV
V_{th7}	Output 7 threshold voltage		979	1094	1208	mV
V_{th8}	Output 8 threshold voltage		1120	1250	1380	mV
I_{OL}	Output leak current				1	μA
I_O	Output current	$R_{ad}=6.4\text{k}\Omega$	9.6	12	14.4	mA
I_O'	Output current	$R_{ad}=2.8\text{k}\Omega$	20	25	30	mA
V_{SAT}	Output saturation voltage	$R_{ad}=2.8\text{k}\Omega$ $I_O=12.5\text{mA}$			500	mV

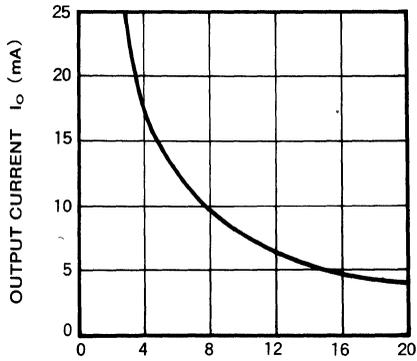
8-STEP BAR TYPE LED LEVEL INDICATOR

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)



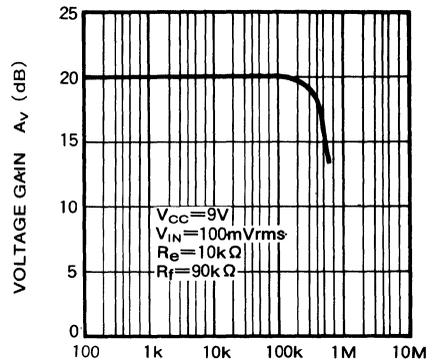
8-STEP BAR TYPE LED LEVEL INDICATOR

**OUTPUT CURRENT VS.
 PIN 15 ADJUSTABLE RESISTOR**



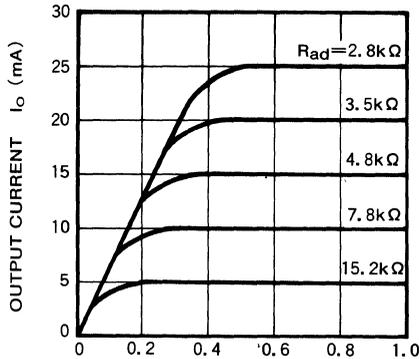
PIN 15 ADJUSTABLE RESISTOR R_{ad} (k Ω)

**HALF-WAVE RECTIFIER OF AMP
 VOLTAGE GAIN VS. INPUT FREQUENCY**



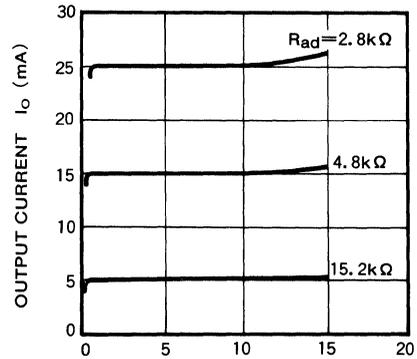
INPUT FREQUENCY F_{IN} (Hz)

**OUTPUT CURRENT VS.
 OUTPUT VOLTAGE (1)**



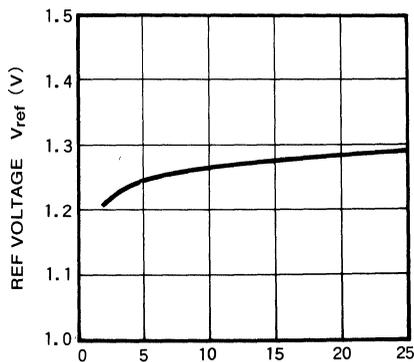
OUTPUT VOLTAGE V_o (V)

**OUTPUT CURRENT VS.
 OUTPUT VOLTAGE (2)**



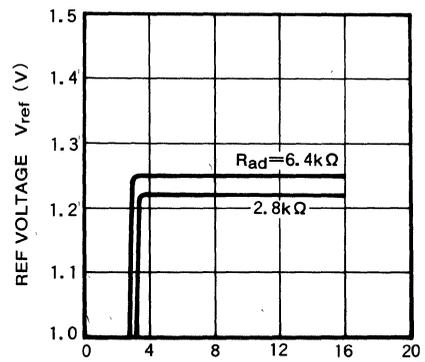
OUTPUT VOLTAGE V_o (V)

**REF VOLTAGE VS. PIN 15
 ADJUSTABLE RESISTOR**



PIN 15 ADJUSTABLE RESISTOR R_{ad} (k Ω)

REF VOLTAGE VS. SUPPLY VOLTAGE

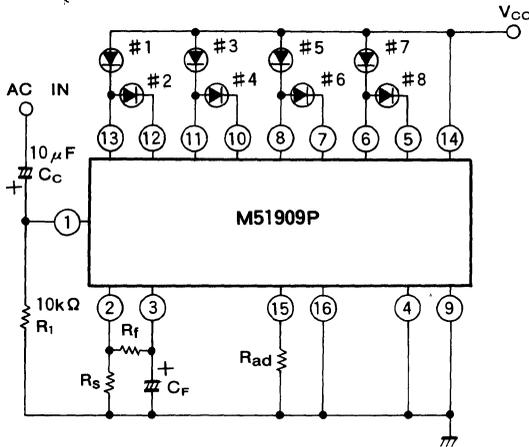


SUPPLY VOLTAGE V_{CC} (V)

8-STEP BAR TYPE LED LEVEL INDICATOR

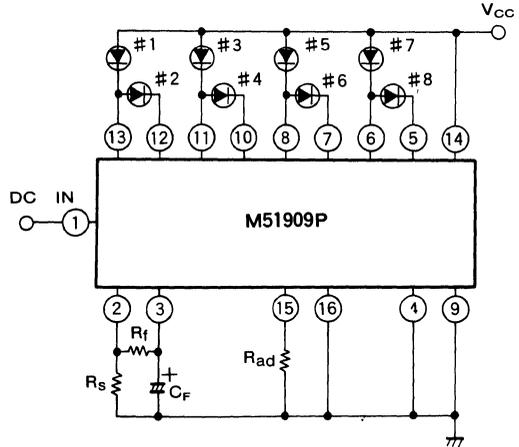
APPLICATION EXAMPLES

(1) AC INPUT



Rad: Resistance determining LED current
(See OUTPUT CURRENT VS. PIN 15 ADJUSTABLE RESISTOR graph)

(2) DC INPUT



OPERATION

The AC signal applied through the coupling capacitor C_C is rectified and amplified mainly with the GND voltage and output at Pin 3. The voltage at Pin 3 compared with the reference voltage at Pin 15 (internal reference voltage 1.25V divided on a logarithmic scale), and the resultant comparator output drives the LEDs.

Note 1. Amp gain

$$\frac{R_S + R_f}{R_S}$$

The peak value is output at Pin 3, so when this is a sine wave, a DC voltage of approximately 1.4 times the rms value is output.

Note 2. Select the value of C_F according to the recovery time.

Recovery time: $C_F \times (R_S + R_f)$

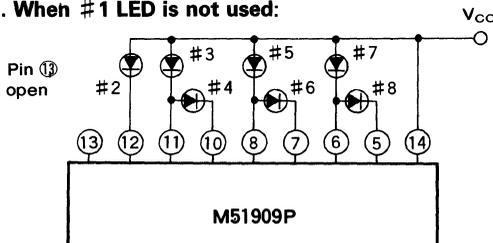
Attack time: $C_F \times 460\Omega$

Note 3. LED current

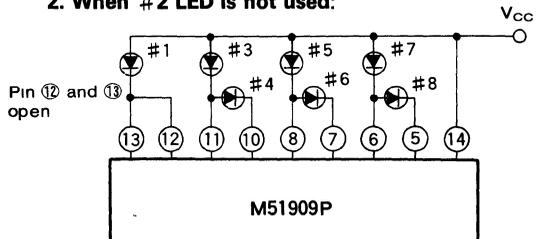
Pin 15 adjustable resistor determines the LED current. (See the OUTPUT CURRENT VS. PIN 15 ADJUSTABLE RESISTOR graph)

Note 4. When less than 8 LEDs are used, proceed as follows.

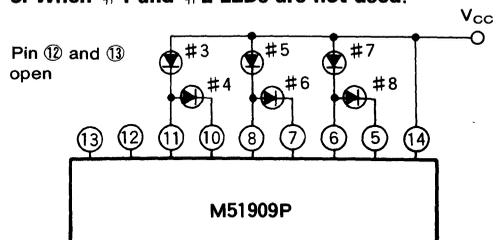
1. When #1 LED is not used:



2. When #2 LED is not used:

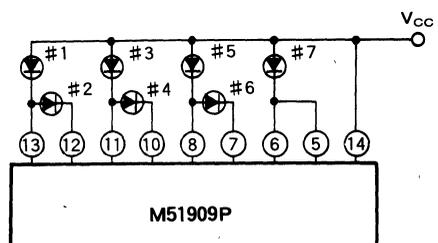


3. When #1 and #2 LEDs are not used:



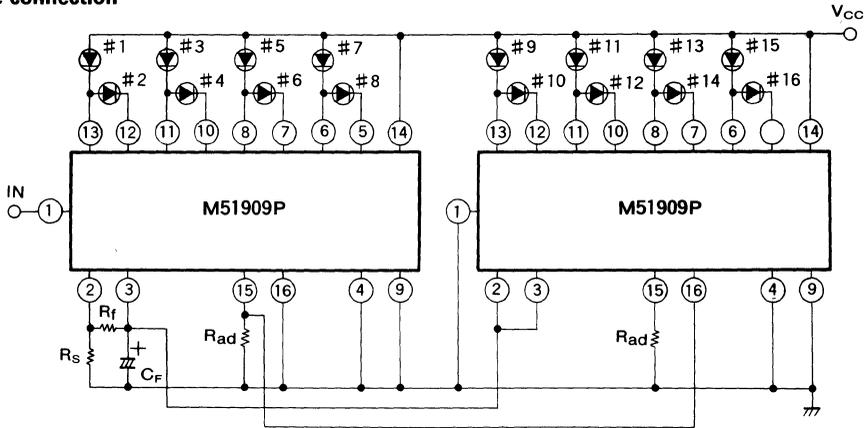
4. When any of LEDs #3 to #8 are not used:

When an even-numbered LED is not used, follow the procedure given for LED #2, and when an odd-numbered LED is not used, follow that for LED #3.

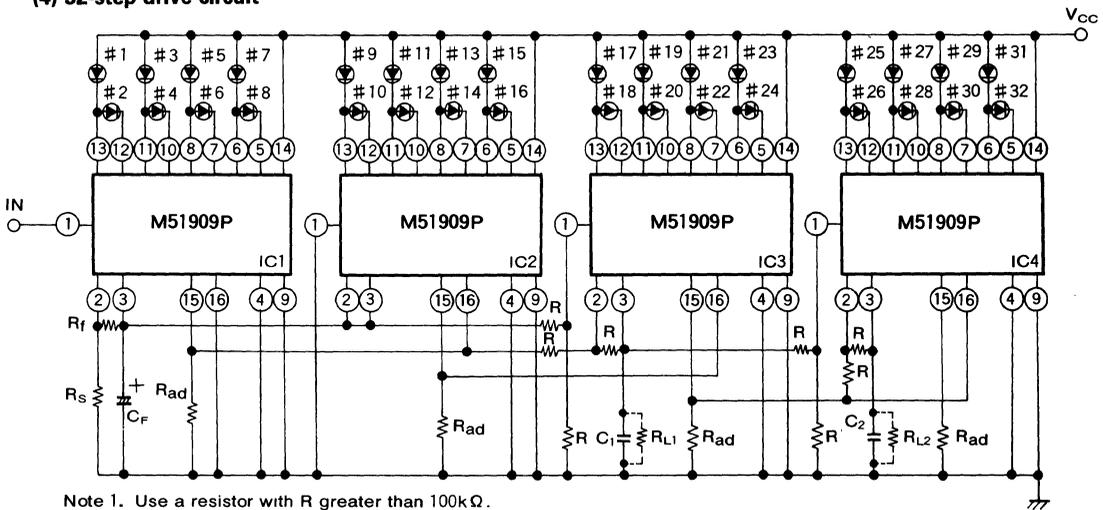


8-STEP BAR TYPE LED LEVEL INDICATOR

(3) Cascade connection



(4) 32-step drive circuit



Note 1. Use a resistor with R greater than 100kΩ.

Note 2. The circuit above works best when V_{CC} is greater than 6V.

Note 3. Adjust the LED drive current by adjusting R_{ad}

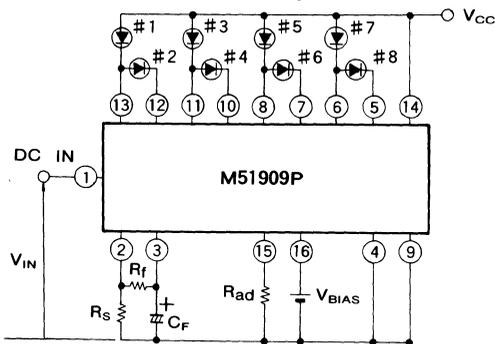
Note 4. Select C_F for the desired recovery time.

Time constants ~ C_F(R_F+10k)

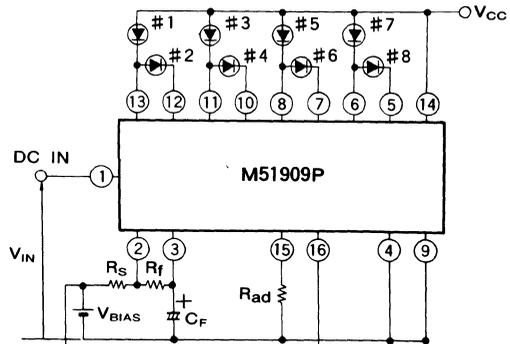
Note 5. C₁ and C₂ stop oscillation.

The recovery time must be extended so that the following relation holds between charge resistors R_{L1} and R_{L2} for C₁ and C₂.

(5) Fixed reference bias DC input



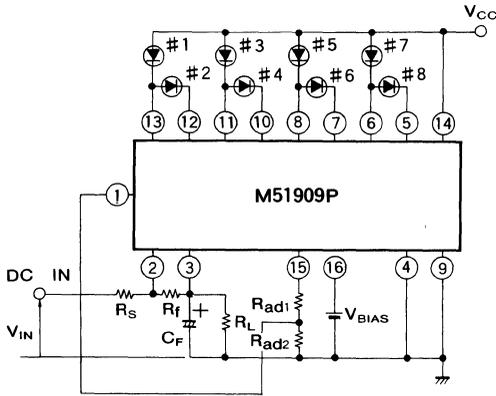
Display is given by the formula $V_{IN} \times \frac{R_S + R_f}{R_S} - V_{BIAS}$



Display is given by the formula $(V_{IN} - V_{BIAS}) \times \frac{R_S + R_f}{R_S}$

8-STEP BAR TYPE LED LEVEL INDICATOR

(6) All LEDs light at minimum-voltage input signal, and the LEDs are extinguished as the input voltage increases



For LED #6

$$I_{LEDB} \sim 60 \times \frac{V_{15}}{R_{ad}} - \frac{V_{F6}}{R_6} = I_{LEDA} - \frac{V_{F6}}{R_6}$$

For LED #7

$$I_{LEDC} \sim 60 \times \frac{V_{15}}{R_{ad}} - \frac{V_{F7}}{R_7} = I_{LEDA} - \frac{V_{F7}}{R_7}$$

For LED #8

$$I_{LEDC} \sim 60 \times \frac{V_{15}}{R_{ad}} - \frac{V_{F8}}{R_8} = I_{LEDA} - \frac{V_{F8}}{R_8}$$

V_{15} , Pin 15 voltage, is 1.25V and V_{F6} , V_{F7} , and V_{F8} are the forward voltage drops for LEDs #6, #7, and #8 respectively.

The M51907P is designed so that the temperature-dependency of the voltage at Pin 15 is very low. The LED forward voltage drop, only -2 to $-2.5\text{mV}/^\circ\text{C}$, also has a relatively low temperature dependence. These design features make it possible to accurately set the current flowing through each LED.

Display follows the formula:

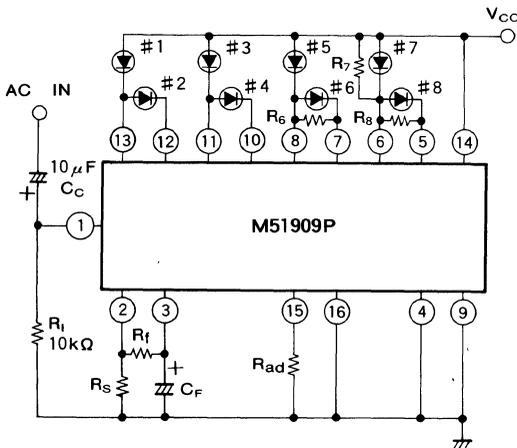
$$\frac{1.25(V) \times R_{ad2}}{R_{ad1} + R_{ad2}} \times \left(1 + \frac{R_f}{R_s}\right) - V_{IN} \times \frac{R_f}{R_s} - V_{BIAS}$$

Note LED drive voltage is determined by $R_{ad1} + R_{ad2}$

(7) Different color LEDs (when different drive currents are desired for the LEDs)

Select Pin 15 adjustable resistor for connection to the LED that requires the maximum drive current. Then connect resistors in parallel to the remaining LEDs using reduced current.

For example, to reduce the current for LEDs #6, #7, and #8, use the following arrangement.



The currents flowing through each of the LEDs in the above circuit configuration are given by the following formulas.

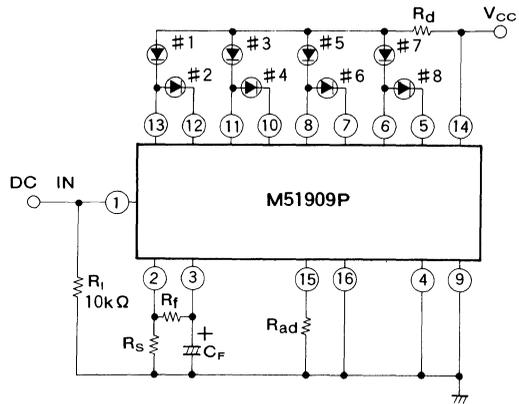
For LEDs #1~#5

$$I_{LEDA} \sim 60 \times \frac{V_{15}}{R_{ad}}$$

(8) High supply voltage produces high output currents

The relation between the allowable output current and the maximum voltage at the output terminal is determined by the thermal derating curve. (See the ALLOWABLE OUTPUT CURRENT VS. OUTPUT TERMINAL MAXIMUM VOLTAGE graph)

Therefore, if the user wants a high supply voltage to produce high output currents, the following circuit configuration can be used, but the output terminal voltage must be low.

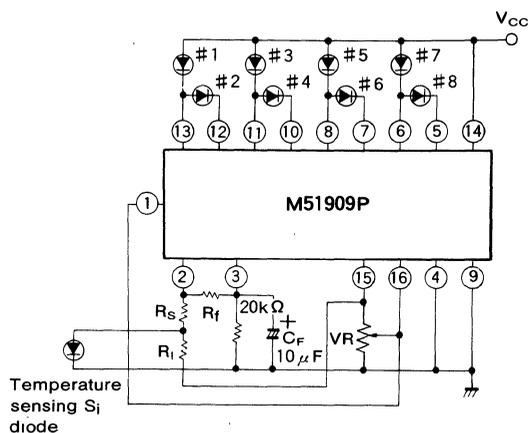


A realistic design involves taking the required output current (LED current: I_{LED}) as the allowable maximum output current, and then determining the corresponding output terminal maximum voltage V_{Omax} from the ALLOWABLE OUTPUT CURRENT VS. OUTPUT TERMINAL MAX VOLTAGE graph. Then if the maximum value for the supply voltage is V_{CCmax} , the value of R_d has to be determined to satisfy the relation $4 \cdot I_{LED} \cdot R_d > V_{CCmax} - V_{Omax}$.

8-STEP BAR TYPE LED LEVEL INDICATOR

(9) Thermometer

This application example makes use of the S_i diode's temperature sensing characteristics. A thermistor can be used in exactly the same way. Also, a CdS can be used to create a lux meter.

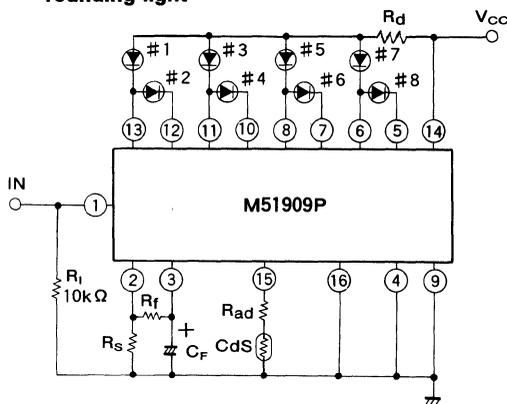


- Note 1. The values of VR and R_1 connected to Pin ⑮ determine the LED brightness
 If V_f is the temperature sensing diode forward voltage drop then the following relation holds

$$I_{LED} \sim 60 \times \left(\frac{V_{⑮}}{R_{VR}} + \frac{V_{⑮} - V_f}{R_1} \right)$$

- Note 2. The VR setting determines the range of temperature desired on the display.
- Note 3. The values of R_s and R_f determine how many degrees rise in temperature constitutes one display step. The diode has a temperature coefficient of approximately $-2\text{mV}/^\circ\text{C}$. To advance display one step in one direction requires around 156mV. For example, if one step is to be set at 1°C , then $R_f/R_s \approx 78$.

(10) Adjusting LED brightness with respect to the surrounding light



The LED current is determined by $R_{ad} + R_{cds}$.

MITSUBISHI LINEAR ICs
M51911L

6-STEP BAR TYPE LED LEVEL INDICATOR

DESCRIPTION

The M51911L is a semiconductor integrated circuit designed for LED level meters. It drives a bar-type input-level display for up to 6 LEDs. With its built-in advanced half-wave rectification operational amplifier, the M51911L accepts direct input of either AC or DC signals. Output is provided for the LEDs grouped in two's by cascade connection, drastically reducing power consumption.

The logarithmic scale of 3, 0 -3, -7, -12, and -18dB display levels makes this device ideal for VU meter applications.

FEATURES

- Advanced half-wave rectification OP amp built in
 Cut-off frequency 200kHz(typ.)
 Offset voltage 2mV(typ.)
- Output current is determined by internal circuits, reducing the number of attachments 13mA(typ.)
- The internal reference voltage threshold value for the supply voltage is only slightly affected by temperature 1.25V(typ.)
- Wide supply voltage range 4~15V
- Advanced half-wave OP amp gain is set internally 17dB(typ.)

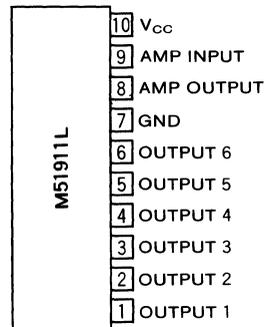
APPLICATION

Signal meters, VU meters, and tuning meters, and other general display applications.

REMONNECED OPERATING CONDITIONS

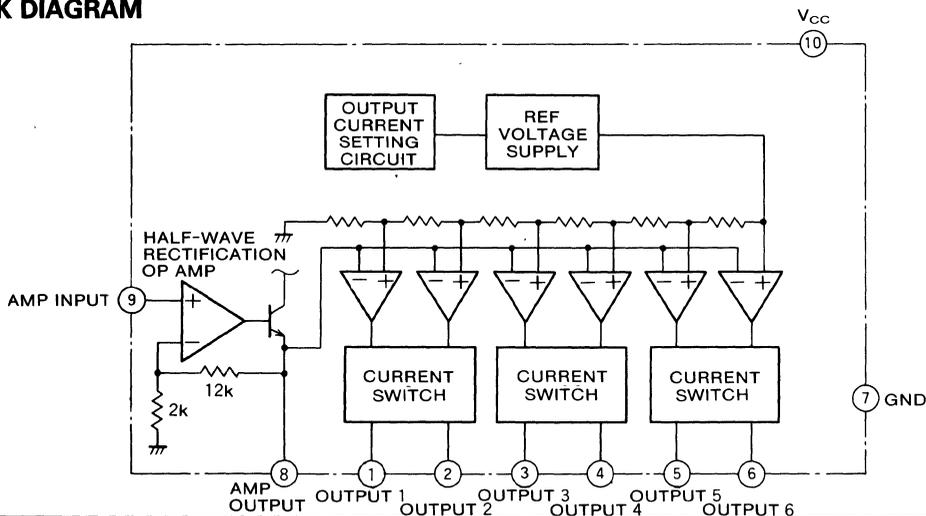
- Supply voltage range 4~15V
- Rated power supply 9V±10%

PIN CONFIGURATION (TOP VIEW)



10-pin molded plastic SIP

BLOCK DIAGRAM



6-STEP BAR TYPE LED LEVEL INDICATOR

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

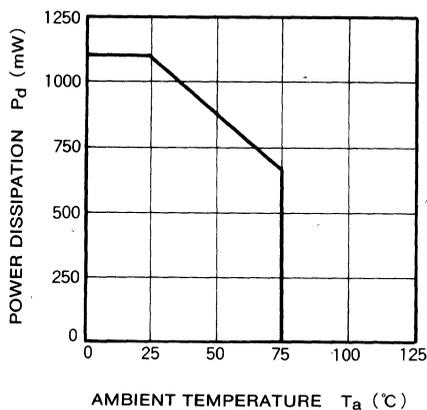
Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		16	V
BV _O	Output voltage		16	V
V _{IN}	Input voltage	Pin ⑨—GND	-2~V _{CC}	V
I _⑧	Pin ⑧ output current		-1	mA
V _⑧	Pin ⑧ output voltage	Pin ⑧—GND	6	V
P _d	Power dissipation		1100	mW
K _θ	Thermal derating	T _a ≥25°C	8.8	mW/°C
T _{opr}	Operating temperature		-20~+75	°C
T _{stg}	Storage temperature		-40~+125	°C

ELECTRICAL CHARACTERISTICS (T_a=25°C, V_{CC}=9V, unless otherwise noted)

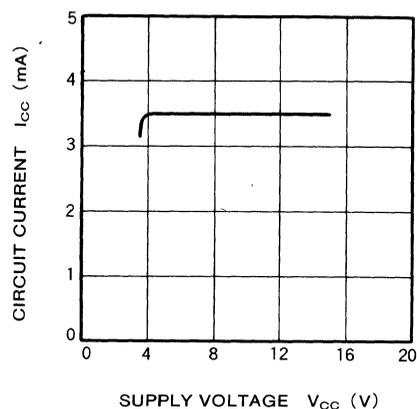
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage		4.0		15.0	V
I _{CC1}	Circuit current	V _⑧ =0V (when all LEDs are off)		3.5	5.6	mA
I _{CC2}	Circuit current	V _⑧ =200mV (when all LEDs are on)		4.0	6.4	mA
I _{IB}	Input bias current	V _⑧ =0V	-500	-150		nA
A _v	Input amp gain			17		dB
V _{th1}	Output 1 threshold voltage	Amp gain=17dB Pin ⑨—GND	6.3	11.2	16.8	mVrms
			-23	-18	-14.5	dB
V _{th2}	Output 2 threshold voltage		16.8	22.5	28.3	mVrms
			-14.5	-12	-10	dB
V _{th3}	Output 3 threshold voltage		33.7	40	47.6	mVrms
			-8.5	-7	-5.5	dB
V _{th4}	Output 4 threshold voltage		56.6	63.5	71.3	mVrms
			-4	-3	-2	dB
V _{th5}	Output 5 threshold voltage		79.9	89.7	100.6	mVrms
			-1	0	+1	dB
V _{th6}	Output 6 threshold voltage		112.9	126.6	142.2	mVrms
			+2	+3	+4	dB
I _O	Output current		8	13	20	mA
I _{OL}	Output leak current	Output voltage=V _{CC}			1	μA
R _⑧	Pin ⑧ adjustable resistor	V _⑧ =0.3V	9	14	19	kΩ

TYPICAL CHARACTERISTICS (T_a=25°C, unless otherwise noted)

THERMAL DERATING (MAXIMUM RATING)

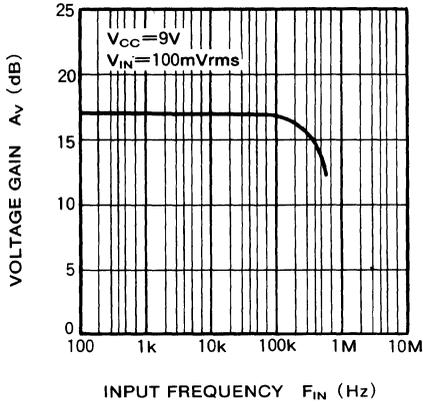


CIRCUIT CURRENT VS. SUPPLY VOLTAGE

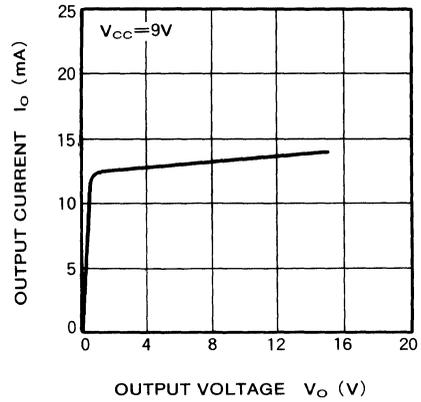


6-STEP BAR TYPE LED LEVEL INDICATOR

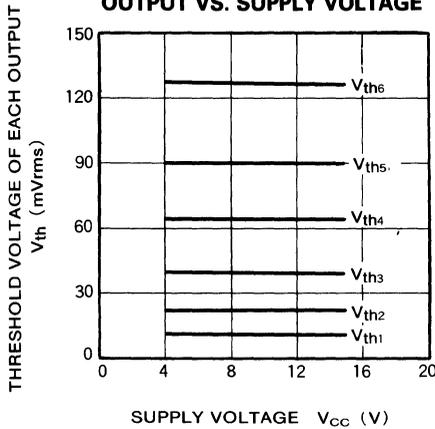
**HALF-WAVE RECTIFIER OP AMP
 VOLTAGE GAIN VS. INPUT FREQUENCY**



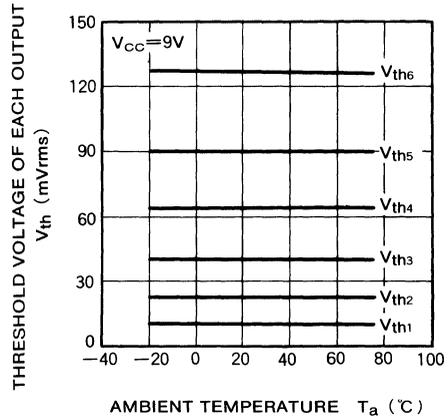
OUTPUT CURRENT VS. OUTPUT VOLTAGE



**THRESHOLD VOLTAGE OF EACH
 OUTPUT VS. SUPPLY VOLTAGE**

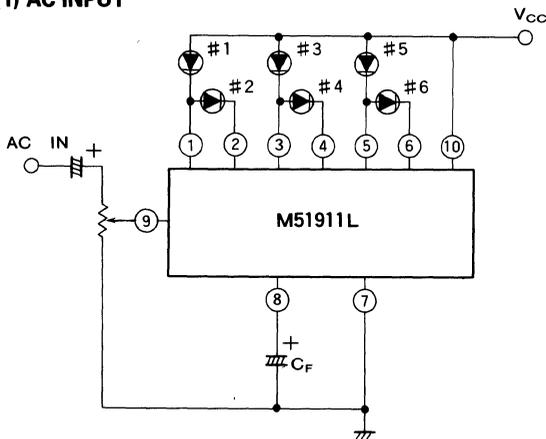


**THRESHOLD VOLTAGE OF EACH
 OUTPUT VS. AMBIENT TEMPERATURE**

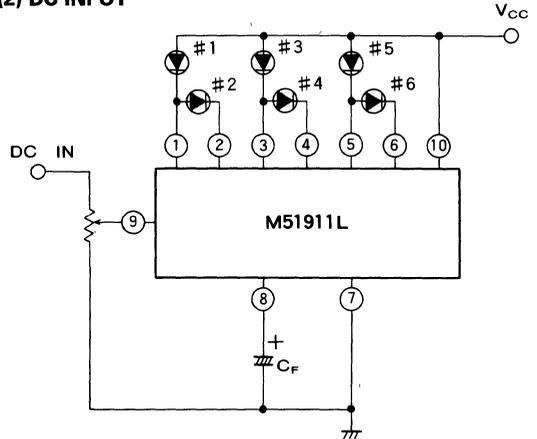


APPLICATION EXAMPLES

(1) AC INPUT



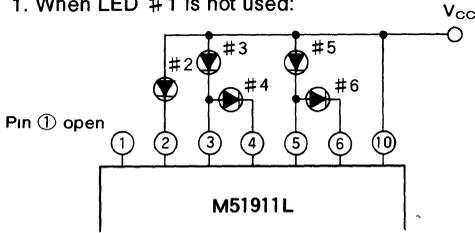
(2) DC INPUT



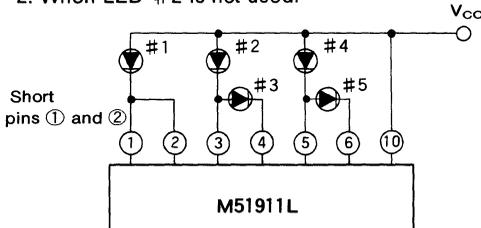
6-STEP BAR TYPE LED LEVEL INDICATOR

(3) Less than six LEDs are required

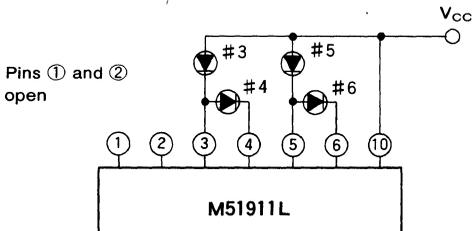
1. When LED #1 is not used:



2. When LED #2 is not used:

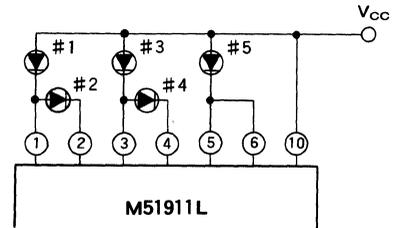


3. When #1 and #2 LEDs are not used:



4. When any of LEDs #3 to #6 are not used:

When an even-numbered LED is not used, follow the procedure given for LED #2, and when an odd-numbered LED is not used, follow that for LED #3.



PRECAUTIONS

1. Output current determined by internal circuit 13mA
2. Amp gain is set by internal resistor 17dB
3. Recovery time: $C_F \times 14k\Omega$
4. Attack time: $C_F \times 430\Omega$

M51912L

6-STEP BAR TYPE LED LEVEL INDICATOR

DESCRIPTION

The M51912L is a semiconductor integrated circuit designed for LED level meters. It drives a bar-type input-level display for up to 6 LEDs. With its built-in advanced half-wave rectification operational amplifier, the M51912L accepts direct input of either AC or DC signals. Output is provided for the LEDs grouped in two's by cascade connection, drastically reducing power consumption.

The M51912L provides linear-scale display levels, just right for signal meter applications.

FEATURES

- Advanced half-wave rectification OP amp built in
Cut-off frequency 200kHz(typ.)
Offset voltage 2mV(typ.)
- Output current is set internally, reducing the number of attachments 13mA(typ.)
- The internal reference voltage threshold value for the supply voltage is only slightly affected by temperature 1.25V(typ.)
- Wide supply voltage range 4~15V
- Advanced half-wave OP amp gain is set by internal resistor 17dB(typ.)

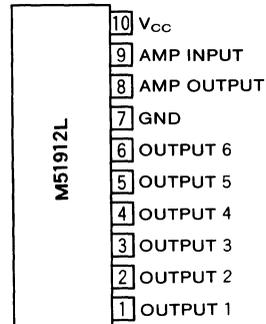
APPLICATION

Signal meters, VU meters, and tuning meters, and other general display applications.

RECOMMENDED OPERATING CONDITIONS

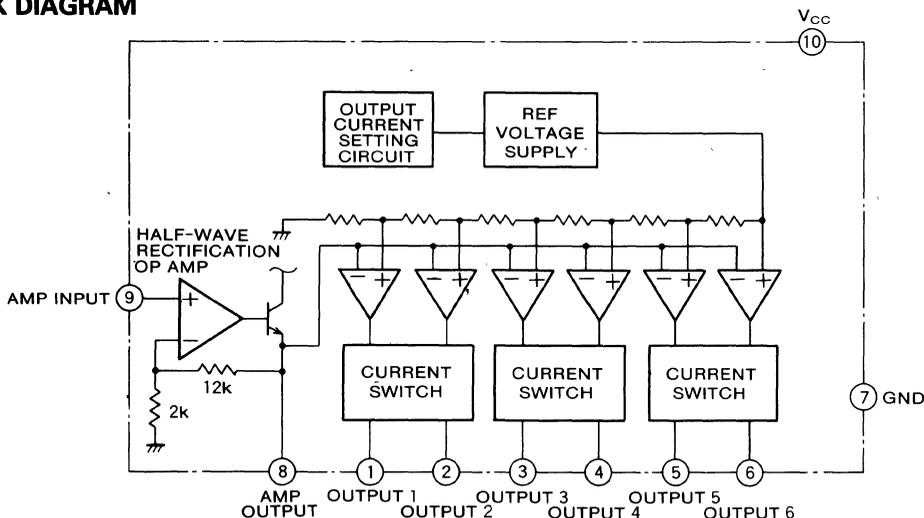
- Supply voltage range 4~15V
- Rated power supply $9V \pm 10\%$

PIN CONFIGURATION (TOP VIEW)



10-pin molded plastic SIP

BLOCK DIAGRAM



6-STEP BAR TYPE LED LEVEL INDICATOR

ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		16	V
BV_O	Output voltage		16	V
V_{IN}	Input voltage	Pin ⑨—GND	$-2 \sim V_{CC}$	V
$I_{\text{⑧}}$	Pin ⑧ output current		-1	mA
$V_{\text{⑧}}$	Pin ⑧ output voltage	Pin ⑧—GND	6	V
P_d	Power dissipation		1100	mW
K_{θ}	Thermal derating	$T_a \geq 25^{\circ}\text{C}$	8.8	mW/ $^{\circ}\text{C}$
T_{opr}	Operating temperature		$-20 \sim +75$	$^{\circ}\text{C}$
T_{stg}	Storage temperature		$-40 \sim +125$	$^{\circ}\text{C}$

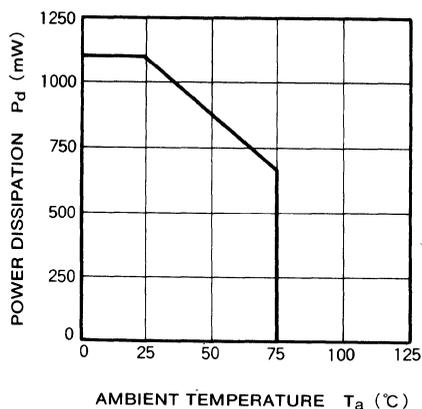
ELECTRICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, $V_{CC}=9\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4.0		15.0	V
I_{CC1}	Circuit current	$V_{\text{⑧}}=0\text{V}$ (when all LEDs are off)		3.5	5.6	mA
I_{CC2}	Circuit current	$V_{\text{⑧}}=200\text{mV}$ (when all LEDs are on)		4.0	6.4	mA
I_{IB}	Input bias current	$V_{\text{⑧}}=0\text{V}$	-500	-150		nA
A_v	Input amp gain			17		dB
V_{th1}	Output 1 threshold voltage	Amp gain=17dB Pin ⑨—GND	22	29	36	mV
V_{th2}	Output 2 threshold voltage		49	59	69	mV
V_{th3}	Output 3 threshold voltage		75	89	*103	mV
V_{th4}	Output 4 threshold voltage		*102	119	*136	mV
V_{th5}	Output 5 threshold voltage		*128	149	*170	mV
V_{th6}	Output 6 threshold voltage		*153	179	205	mV
I_O	Output current		8	13	20	mA
I_{OL}	Output leak current	Output voltage= V_{CC}			1	μA
$R_{\text{⑧}}$	Pin ⑧ adjustable resistor		9	14	19	k Ω

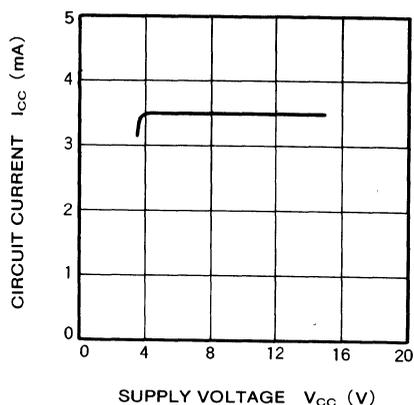
*: Over-wrap for the IC package is not provided.

TYPICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, unless otherwise noted)

THERMAL DERATING (MAXIMUM RATING)

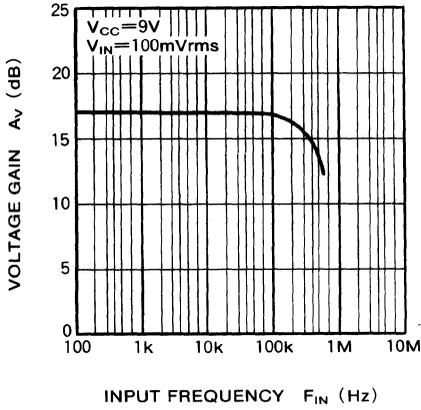


CIRCUIT CURRENT VS. SUPPLY VOLTAGE

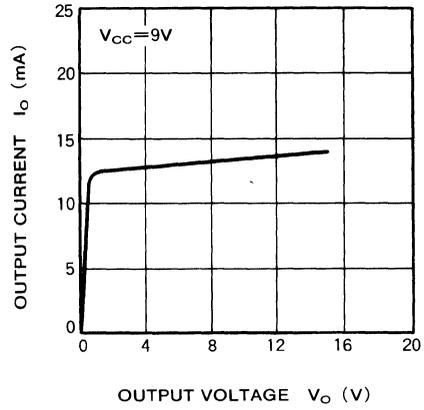


6-STEP BAR TYPE LED LEVEL INDICATOR

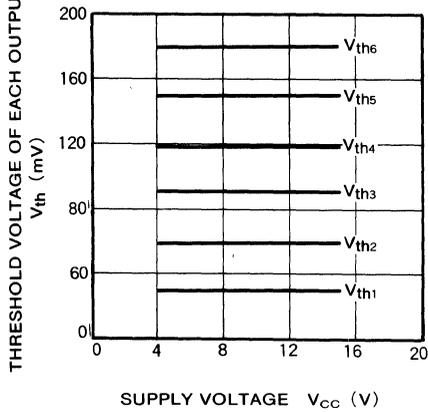
**HALF-WAVE RECTIFIER OP AMP
 VOLTAGE GAIN VS. INPUT FREQUENCY**



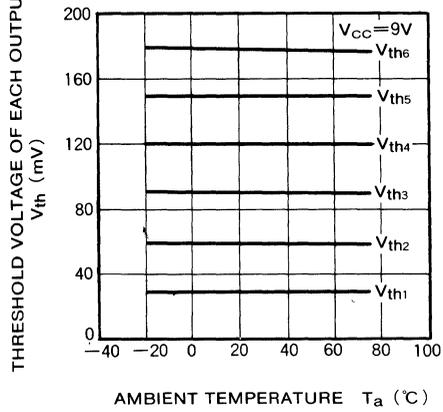
OUTPUT CURRENT VS. OUTPUT VOLTAGE



**THRESHOLD VOLTAGE OF EACH
 OUTPUT VS. SUPPLY VOLTAGE**

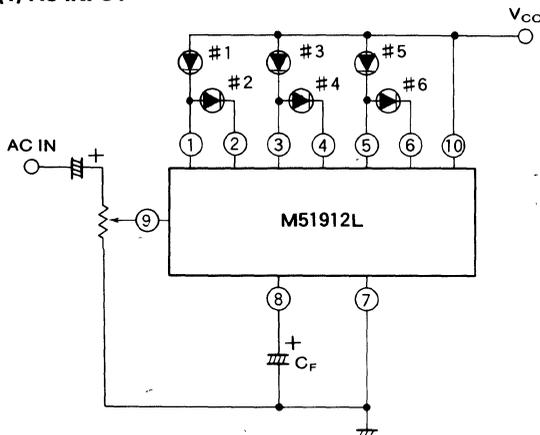


**THRESHOLD VOLTAGE OF EACH
 OUTPUT VS. AMBIENT TEMPERATURE**

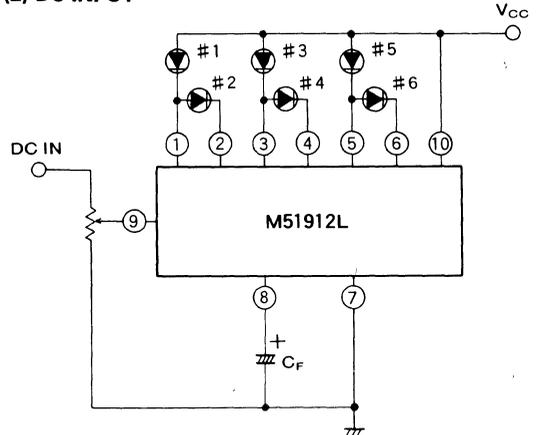


APPLICATION EXAMPLES

(1) AC INPUT



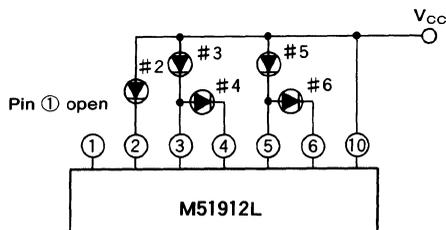
(2) DC INPUT



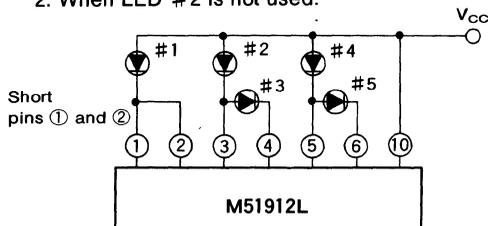
6-STEP BAR TYPE LED LEVEL INDICATOR

(3) Less than six LEDs are required

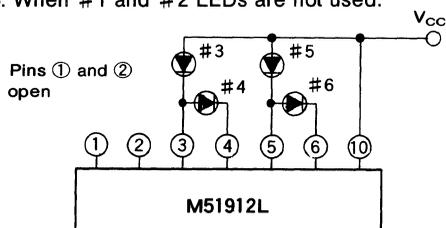
1. When LED #1 is not used:



2. When LED #2 is not used:

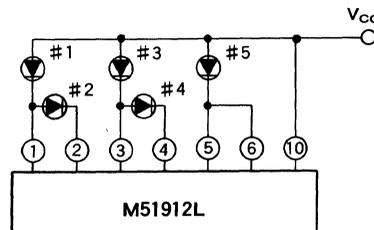


3. When #1 and #2 LEDs are not used:



4. When any of LEDs #3 to #6 are not used:

When an even-numbered LED is not used, follow the procedure given for LED #2, and when an odd-numbered LED is not used, follow that for LED #3.



PRECAUTIONS

1. Output current is determined by internal circuit
 13mA
2. Amp gain is set by internal resistor 17dB
3. Recovery time: $C_F \times 14k\Omega$
4. Attack time: $C_F \times 430\Omega$

TRANSISTOR ARRAY

MITSUBISHI GENERAL-PURPOSE INTEGRATED CIRCUITS INDEX BY FUNCTION OF TRANSISTOR ARRAYS

QUICK REFERENCE TABLE OF TRANSISTOR ARRAY SERIES

No.	Input/output type		No. of circuits	Type No.	Output current (mA)				Withstand voltage (V)				Output type		CD	Remarks			
	Input	Output			1.5A	500~300	Less than 150	80	50	40	20	Darlington	Low saturation						
1	"H"	active	Current sink	M54512L			○						○						
				M54532L	○					○				○					
				M54594P*	○			■					■		■		Sprague ULN2065B		
				M54595P*	○			■					■		■		Sprague ULN2067B		
				M54516P, 521P		○								○					
				M54529AP, 529P		○								○		○			
				M54533P, 534P, 539P, 578P		○								○		○	○		
				M54527P			○					○				○			
				M54571P		○						○				○	○		
				M54514AP, 515P, 577P, 577FP				50						○		○			Output withstand voltage Up to V _{CC} for the 535P, 536P
				M54537P, 535P, 536P		○							○						
				M54517P, 519P		○						○							
				M54528P			○						○				○		Sprague ULN2000A series
				M54523P, 524P, 525P, 526P,		○							○				○		
				M54530P, 531P		○							○				○		
				M54513P			○						○				○		
				M54538P		○								○			○		
				M54522P									○				○		
				M54585P*		■								■		■		■	Sprague ULN2803A
M54584P*		■									■				8-unit version of the 537P				
M54590P, 591P, 592P, 593P**		■			■						■		■		Sprague ULN2823A, 2821A ULN2822A, 2824A				
2	"H"	Source	8	M54562P, 563P		○				○			○		○	Sprague ULN2981A, 2982A			
				M54564P		○					○						Changed from DC to R of the 562P		
				M54597P, 598P*		■			■				■		■		Sprague ULN2983A, 2984A		
3	"L"	Sink	4	M54567P	○					○				○		Toshiba TD62308AP			
				M54596P*	■				■			■		■		Toshiba TD62308BP			
				M54661P	■				■				■		■		Tokyo Sanyo LB1205		
			7	M54576P, 576FP			50						○		○				
				M54566P		○						○						Toshiba TD62305AP	
				M54565P			○						○			○			
M54583P*		○						○						8-unit version of the 566P					
4	"L"	Source	4	M54568L			30					○		○					
				M54560P, 561P			○					○			○				
				M54580P			○					○							
			8	M54581P		○						○				○		Sprague UDN2580A	
				M54586P*		■						■			■			Changed from CD to R of the 581P	
M54660P*		■				■				■		■		Sprague UDN2580A-1					

★: New products, ★★: Under development, ○: Current product, ■: New and under-development product

M54512L

4-UNIT 50mA TRANSISTOR ARRAY

DESCRIPTION

The M54512L, 4-channel sink driver, consists of four NPN transistors, and designed for use in medium-current switching applications.

FEATURES

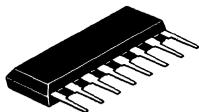
- Output breakdown voltage to 20V
- 50mA output sink current capability
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

LED or incandescent display driver

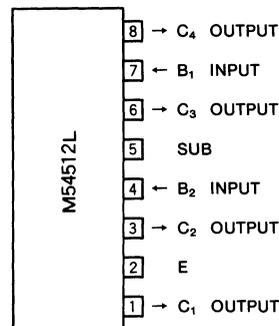
FUNCTION

The M54512L is comprised of four NPN transistors with a $10\text{k}\Omega$ series input resistor, connected to form dual 2-parallel output drivers. All emitters of transistors are connected together to pin 2. The substrate is connected to pin 5 and pin 5 must be tied to the most negative point in the external circuit. The drivers are capable of sinking 50mA and will withstand 20V in the OFF state.



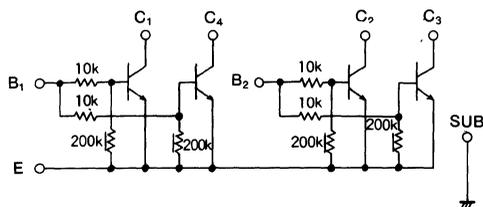
8-pin molded plastic SIP

PIN CONFIGURATION (TOP VIEW)



Outline 8P5

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = 25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5~+20	V
V_{EBO}	Emitter-base sustaining voltage		4	V
I_C	Collector current	Transistor ON	50	mA
V_I	Input voltage		20	V
P_d	Power dissipation	$T_a = 75^\circ\text{C}$	500	mW
T_{opr}	Operating ambient temperature range		-10~+75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a = 25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_C	Output voltage	0		18	V
I_C	Collector current per channel	0		20	mA
V_{IH}	"H" Input voltage	11		18	V
V_{IL}	"L" Input voltage	0		0.2	V

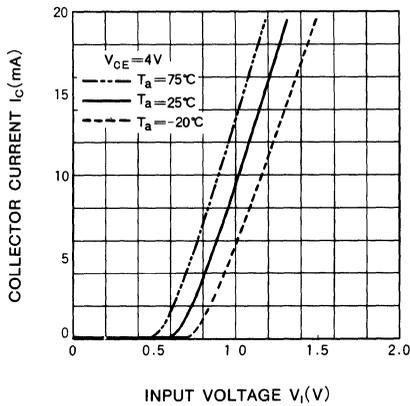
4-UNIT 50mA TRANSISTOR ARRAY

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

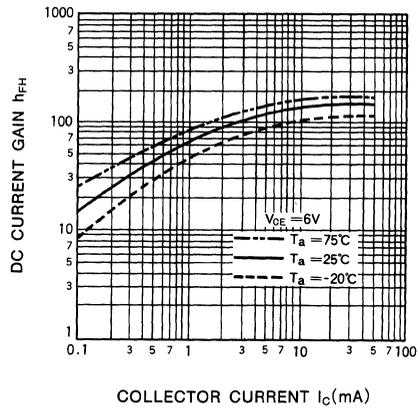
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{C(\text{leak})}$	Output leakage current	$V_{CE} = 20\text{V}$			20	μA
$V_{CE(\text{sat})}$	Output saturation voltage	$I_B = 2\text{mA}$ $I_C = 10\text{mA}$ $I_C = 20\text{mA}$		0.02 0.04	0.1 0.2	V
BV_{EBO}	Emitter-base sustaining voltage	$I_{EBO} = 150\mu\text{A}$	4			V
V_I	Input voltage	$I_B = 2\text{mA}$	4	11	18	V
h_{FE}	DC forward current gain	$V_{CE} = 6\text{V}$, $I_C = 20\text{mA}$, $T_a = 25^\circ\text{C}$	60	150		—

TYPICAL CHARACTERISTICS

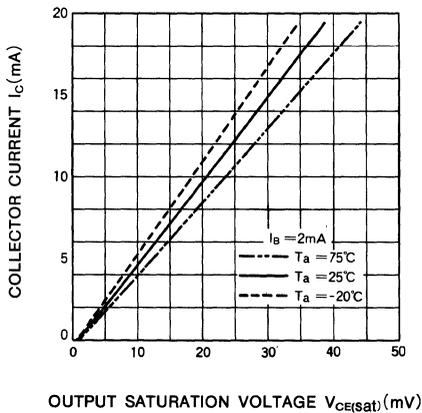
OUTPUT CURRENT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



OUTPUT CHARACTERISTICS



M54532P

4-UNIT 1.5A DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54532P, 4-channel sink driver, consists of 8 NPN transistors connected to form high current gain driver pairs.

FEATURES

- High output sustaining voltage to 50V
- High output sink current to 1.5A
- Integral diodes for transient suppression
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

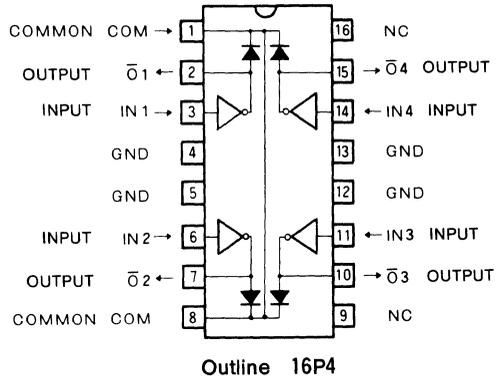
APPLICATION

Relay and printer driver, Display driver

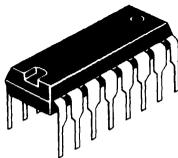
FUNCTION

The M54532P is comprised of eight NPN darlington driver pairs with 340Ω series input resistors. Each output has a diode for inductive load transient suppression and the cathodes of the diodes are connected to pin 8. The outputs are capable of sinking 1.5A and will withstand 50V in the OFF state.

PIN CONFIGURATION (TOP VIEW)

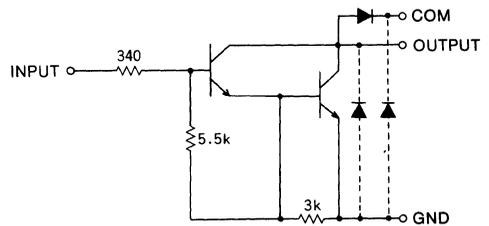


NC : NO CONNECTION



16-pin molded plastic DIP

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +50	V
I_C	Collector current	Transistor ON	1.5	A
V_I	Input voltage		10	V
$I_{F(D)}$	Clamp diode forward current		1.5	A
			1.25	A
$V_{R(D)}$	Clamp diode reverse voltage		50	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.92	W
T_{OPR}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{STG}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

4-UNIT 1.5A DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

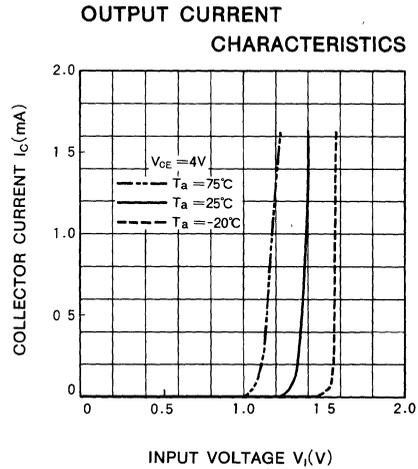
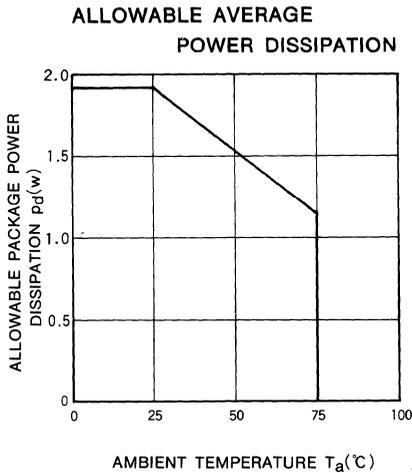
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		50	V
I_C	Collector current per channel	Percent duty cycle less than 4%	0		1.25	A
		Percent duty cycle less than 18%	0		700	mA
V_{IH}	"H" Input voltage	$I_C = 1.25\text{A}$	3		6	V
V_{IL}	"L" Input voltage	$I_{O(leak)} = 50\mu\text{A}$	0		0.4	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CEO} = 100\mu\text{A}$	50			V
$V_{CE(sat)}$	Output saturation voltage	$I_I = 2\text{mA}$	$I_C = 1.25\text{A}$	1.3	2.2	V
			$I_C = 700\text{mA}$	1.1	1.7	
I_I	Input current	$V_I = 3\text{V}$		5	8.5	mA
V_F	Clamp diode forward voltage	$I_F = 1.25\text{A}$		1.6	2.3	V
V_R	Clamp diode reverse voltage	$I_R = 100\mu\text{A}$	50			V
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 1\text{A}, T_a = 25^\circ\text{C}$	800	7000		—

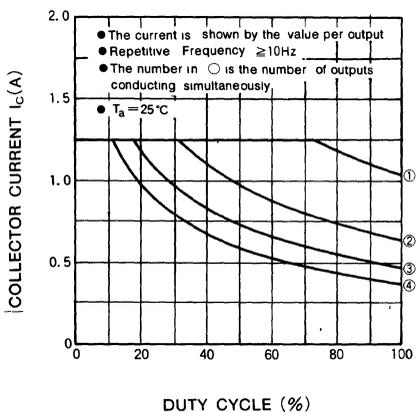
* : A typical value is at $T_a = 25^\circ\text{C}$

TYPICAL CHARACTERISTICS

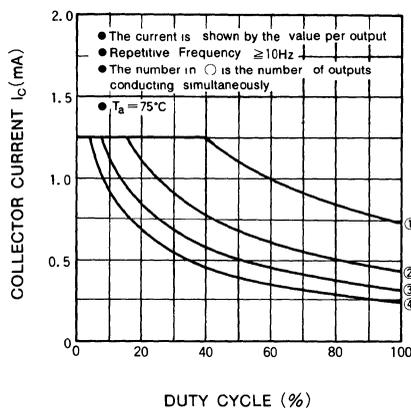


4-UNIT 1.5A DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

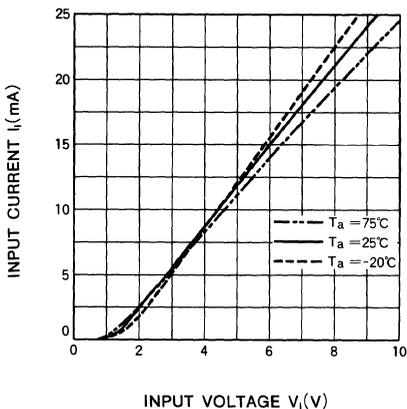
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



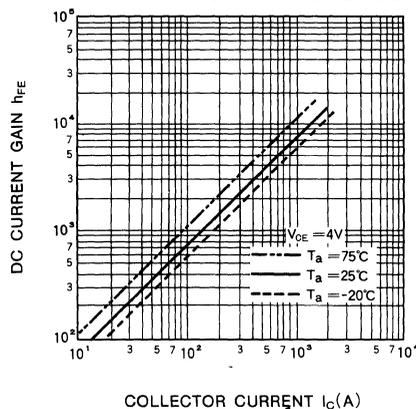
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



INPUT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54516P

5-UNIT 500mA DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54516P, 5-channel sink driver, consists of 10 NPN transistors connected to form five high current gain driver pairs.

FEATURES

- Output sustaining voltage to 25 V
- High output sink current to 500mA
- PMOS Compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

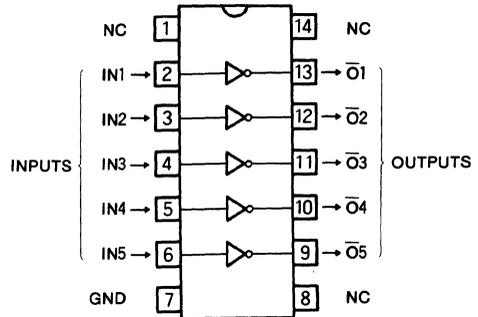
APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics.

FUNCTION

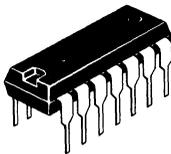
The M54516P is comprised of five NPN darlington driver pairs with $20\text{k}\Omega$ series input resistors. All emitter and the substrate are connected together to pin 7. The output are capable of sinking 500mA and will withstand 25V in the OFF state.

PIN CONFIGURATION (TOP VIEW)



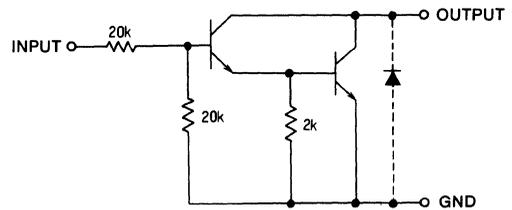
Outline 14P4

NC : NO CONNECTION



14-pin molded plastic DIP

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +25$	V
I_C	Collector current	Transistor ON	500	mA
V_i	Input voltage		25	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

5-UNIT 500mA DARLINGTON TRANSISTOR ARRAY

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

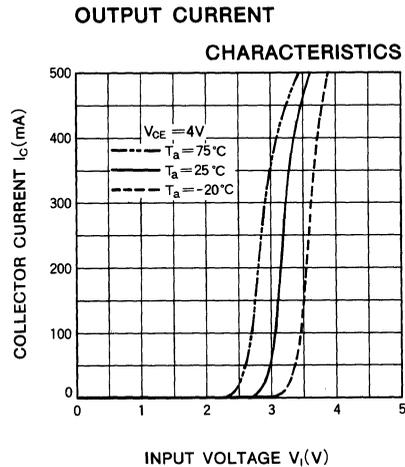
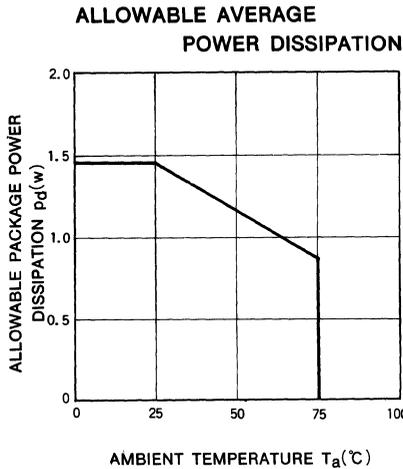
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		25	V
I_C	Collector current per channel	Percent duty cycle less than 10%	0		400	mA
		Percent duty cycle less than 55%	0		200	
V_{IH}	"H" Input voltage	$I_C = 400\text{mA}$	8		20	V
		$I_C = 200\text{mA}$	5		20	
V_{IL}	"L" Input voltage	$I_{oleak} = 50\mu\text{A}$	0		0.5	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CEO} = 100\mu\text{A}$	25			V
$V_{CE(sat)}$	Output saturation voltage	$V_I = 8\text{V}, I_C = 400\text{mA}$		1.15	2.2	V
		$V_I = 5\text{V}, I_C = 200\text{mA}$		0.95	1.4	
I_I	Input current	$V_I = 17\text{V}$		0.8	1.8	mA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 400\text{mA}, T_a = 25^\circ\text{C}$	1000	4000		—

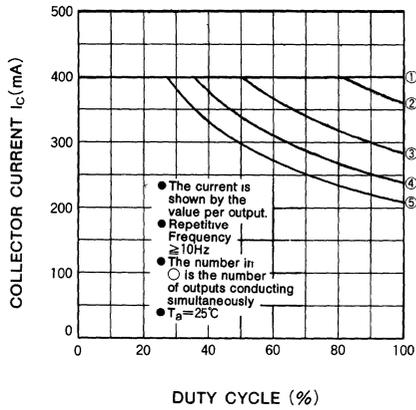
* : A typical value is at $T_a = 25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

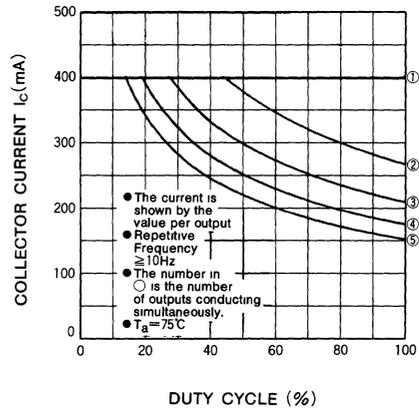


5-UNIT 500mA DARLINGTON TRANSISTOR ARRAY

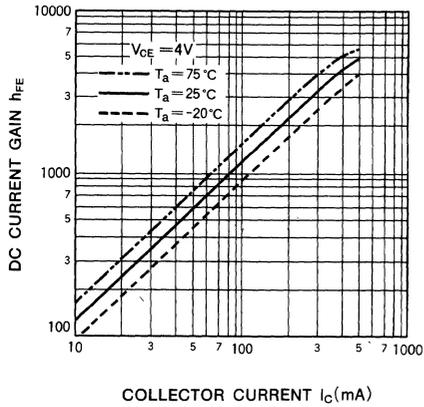
**ALLOWABLE COLLECTOR CURRENT
AS A FUNCTION OF DUTY CYCLE**



**ALLOWABLE COLLECTOR CURRENT
AS A FUNCTION OF DUTY CYCLE**



**DC CURRENT GAIN
CHARACTERISTICS**



M54521P

5-UNIT 500mA DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54521P, 5-channel sink driver, consists of 10 NPN transistors connected to form high current gain driver pairs.

FEATURES

- Output sustaining voltage to 30V
- High output sink current to 500mA
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

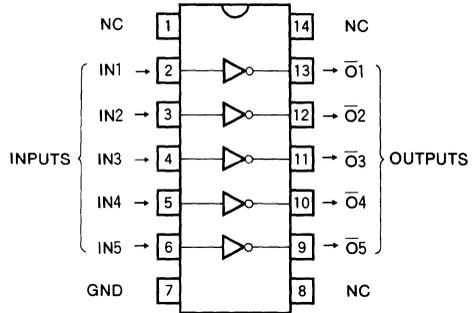
APPLICATION

Relay and printer drivers, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

FUNCTION

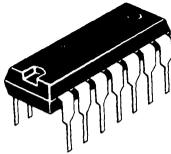
The M54521P is comprised of five NPN darlington driver pairs. All emitters and the substrate are connected together to pin 7. The output are capable of sinking 500mA and will withstand 30V in the OFF state.

PIN CONFIGURATION (TOP VIEW)



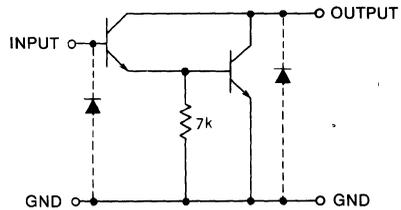
Outline 14P4

NC : NO CONNECTION



14-pin molded plastic DIP

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +30	V
I_C	Collector current	Transistor ON	500	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

5-UNIT 500mA DARLINGTON TRANSISTOR ARRAY

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

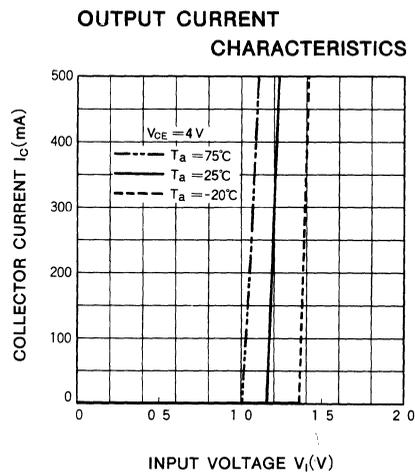
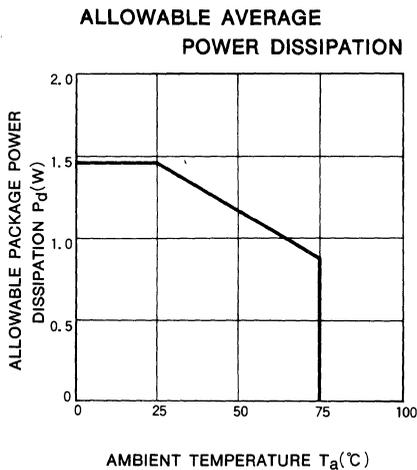
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		30	V
I_C	Collector current per channel	Percent duty cycle less than 10%	0		400	mA
		Percent duty cycle less than 55%	0		200	
I_{IH}	"H" Input current	$I_C = 200\text{mA}$	1		5	mA
		$I_C = 400\text{mA}$	2		5	
I_{IL}	"L" Input current			0		μA

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CEO} = 100\mu\text{A}$	30			V
$V_{CE(sat)}$	Output saturation voltage	$V_i = 2\text{mA}, I_C = 400\text{mA}$		1.0	2.4	V
		$V_i = 1\text{mA}, I_C = 200\text{mA}$		0.8	1.6	
V_i	Input voltage	$I_i = 1\text{mA}$	0.6	1.35	1.7	V

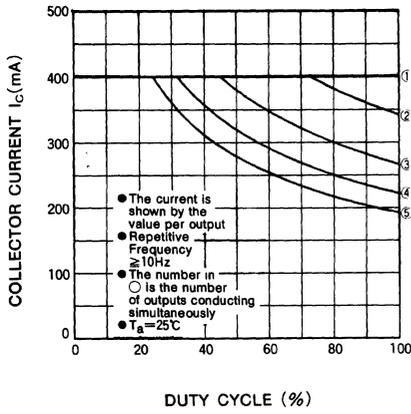
* : A typical value is at $T_a = 25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

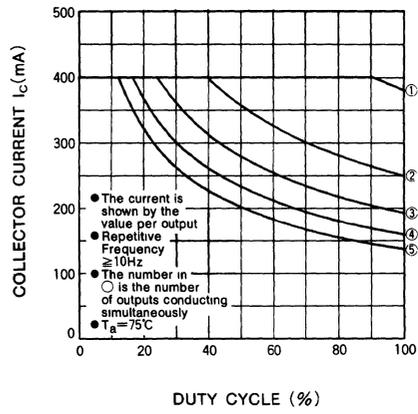


5-UNIT 500mA DARLINGTON TRANSISTOR ARRAY

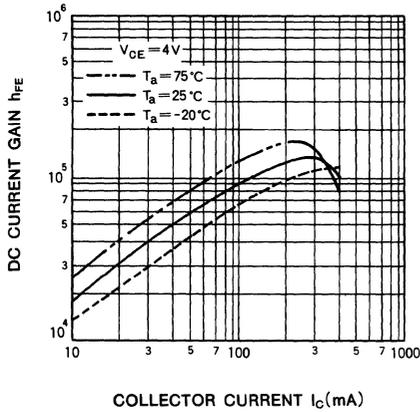
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54529AP

5-UNIT 320mA TRANSISTOR ARRAY WITH STROBE

DESCRIPTION

The M54529AP, 5-channel sink driver, consists of 10 NPN transistors connected to form high current gain driver pairs.

FEATURES

- Output sustaining voltage to 20V
- High output sink current to 320mA
- CMOS compatible input with strobe control
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

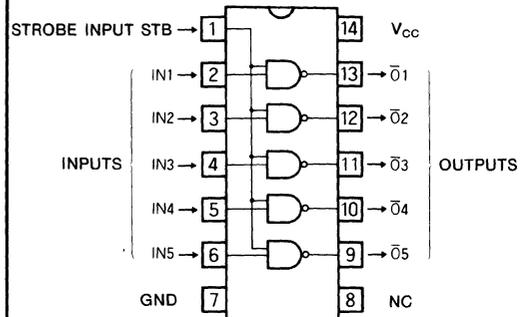
APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

FUNCTION

The M54529AP uses a predriver stage. Each input has a diode and $30k\Omega$ resistor in series to have a wide input voltage range from -25V to $+20\text{V}$. All input can be controlled simultaneously by a strobe input at pin 1. The power supply of the predrivers is connected to pin 14. All emitters and the substrate are connected together to pin 7. The outputs are capable of sinking 320mA and will withstand 20V in the OFF state.

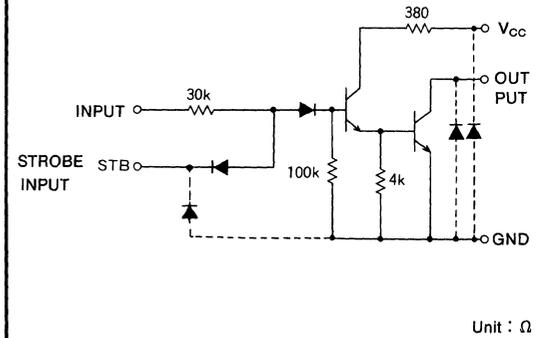
PIN CONFIGURATION (TOP VIEW)



Outline 14P4

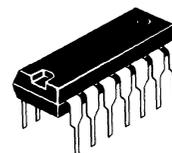
NC : NO CONNECTION

CIRCUIT SCHEMATIC



FUNCTIONAL TABLE

IN	STB	OUT
L	L	H
H	L	H
L	H	H
H	H	L



14-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +20$	V
I_C	Collector current	Transistor ON	320	mA
V_I	Input voltage		$-20 \sim +20$	V
$V_{I(STB)}$	Strobe input voltage		20	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

5-UNIT 320mA TRANSISTOR ARRAY WITH STROBE

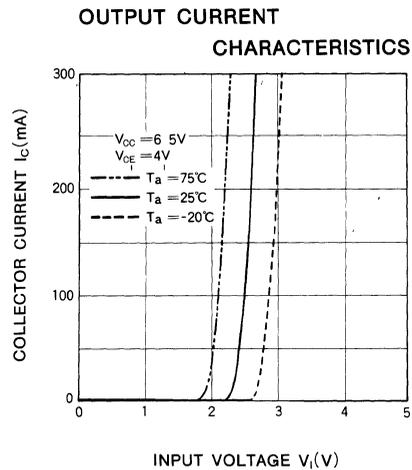
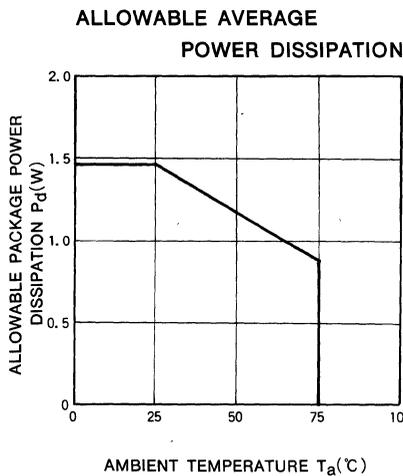
RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		3	5	8	V
V_O	Output voltage		0		20	V
I_C	Collector current per channel	Percent duty cycle less than 33%, $V_{CC}=6.5\text{V}$	0		300	mA
		Percent duty cycle less than 80%, $V_{CC}=6.5\text{V}$	0		150	mA
V_{IH}	"H" Input voltage	$I_C=150\text{mA}$	3.5		15	V
		$I_C=300\text{mA}$	5		15	V
V_{IL}	"L" Input voltage	$I_{O(Leak)}=50\mu\text{A}$	0		1	V
$V_{IH(STB)}$	"H" Input voltage (strobe input)		2.4		15	V
$V_{IL(STB)}$	"L" Input voltage (strobe input)		0		0.2	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

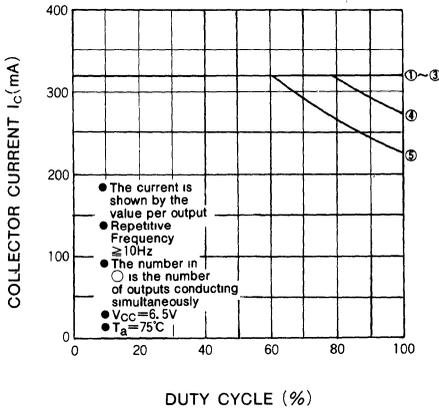
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$V_{CC}=8\text{V}$, $V_I=8\text{V}$, $V_{I(STB)}=0.2\text{V}$ $I_{CEO}=100\mu\text{A}$	20			V
$V_{CE(sat)}$	Output saturation voltage	$V_{I(STB)}=2.4\text{V}$		0.35	0.85	V
		$V_{CC}=6.5\text{V}$, $V_I=5\text{V}$, $I_C=250\text{mA}$ $V_{CC}=3\text{V}$, $V_I=3.5\text{V}$, $I_C=150\text{mA}$		0.2	0.6	
I_I	Input current	$V_{CC}=5\text{V}$, $V_I=3.5\text{V}$, $V_{I(STB)}=2.4\text{V}$		20	120	μA
I_R	Input leakage current	$V_{CC}=8\text{V}$, $V_I=-20\text{V}$			-20	μA
$I_{I(STB)}$	Strobe input current	$V_{CC}=5\text{V}$, $V_I=5\text{V}$ all input $V_{I(STB)}=0.2\text{V}$		-0.8	-1.5	mA
$I_{R(STB)}$	Strobe input leakage current	$V_{CC}=8\text{V}$, $V_I=0\text{V}$, $V_{I(STB)}=20\text{V}$			10	μA
I_{CC}	Supply current	$V_{CC}=8\text{V}$, $V_I=5\text{V}$ all input $V_{I(STB)}=2.4\text{V}$		95	170	mA
h_{FE}	DC forward current gain	$V_{CE}=4\text{V}$, $V_{CC}=6.5\text{V}$, $I_C=300\text{mA}$, $T_a=25^\circ\text{C}$	1000	18000		—

TYPICAL CHARACTERISTICS

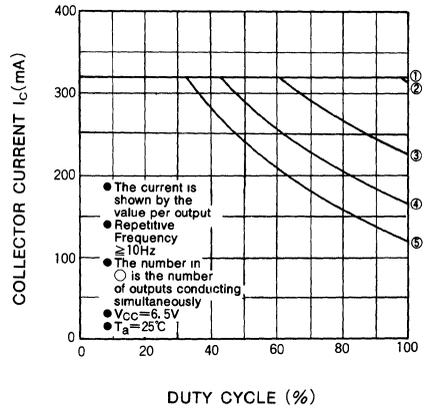


5-UNIT 320mA TRANSISTOR ARRAY WITH STROBE

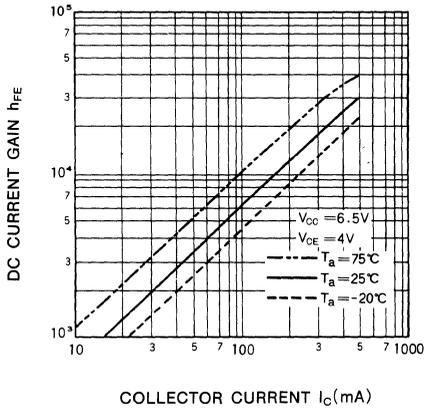
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54529P

5-UNIT 320mA TRANSISTOR ARRAY WITH STROBE

DESCRIPTION

The M54529P, 5-channel sink driver, consists of 10 NPN transistors connected to form high current gain driver pairs.

FEATURES

- Output sustaining voltage to 20V
- High output sink current to 320mA
- PMOS Compatible input with strobe control
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

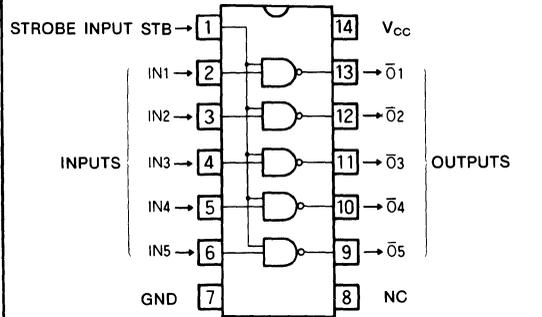
APPLICATION

Relay and printer driver, LED and incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

FUNCTION

The M54529P uses a predriver stage. Each input has a diode and $20\text{k}\Omega$ resistor in series to have a wide input voltage range from -25V to $+20\text{V}$. All input can be controlled simultaneously by a strobe input at pin 1. The power supply of the predrivers is connected to pin 14. All emitters and the substrate are connected together to pin 7. The outputs are capable of sinking 320mA and will withstand 20V in the OFF state.

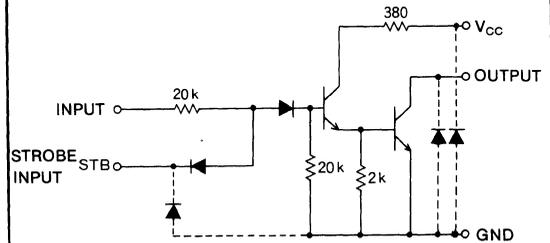
PIN CONFIGURATION (TOP VIEW)



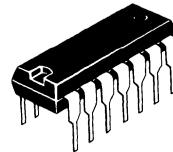
Outline 14P4

NC : NO CONNECTION

CIRCUIT SCHEMATIC



Unit : Ω



14-pin molded plastic DIP

FUNCTIONAL TABLE

IN	STB	OUT
L	L	H
H	L	H
L	H	H
H	H	L

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +20$	V
I_C	Collector current	Transistor ON	320	mA
V_i	Input voltage		$-25 \sim +20$	V
$V_{i(STB)}$	Strobe input voltage		20	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

5-UNIT 320mA TRANSISTOR ARRAY WITH STROBE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit	
		Min	Typ	Max		
V_{CC}	Supply voltage	3		8	V	
V_O	Output voltage	0		20	V	
I_C	Collector current per channel	Percent duty cycle less than 33%, $V_{CC}=6.5\text{V}$	0		300	mA
		Percent duty cycle less than 80%, $V_{CC}=6.5\text{V}$	0		150	
V_{IH}	"H" Input voltage	$I_C=300\text{mA}$	7		15	V
		$I_C=150\text{mA}$	6		15	
V_{IL}	"L" Input voltage	$I_{O(leak)}=50\mu\text{A}$	0		1	V
$V_{IH(STB)}$	"H" Input voltage (strobe input)		2.4		15	V
$V_{IL(STB)}$	"L" Input voltage (strobe input)		0		0.2	V

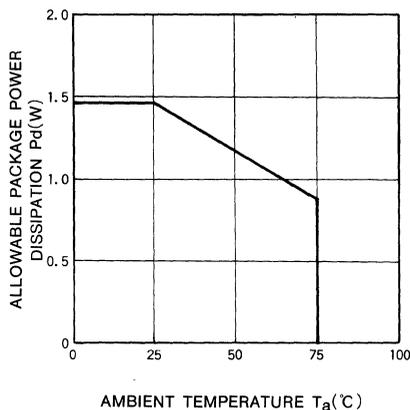
ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$V_{CC}=8\text{V}$, $V_I=7\text{V}$, $V_{I(STB)}=0.2\text{V}$ $I_{CEO}=100\mu\text{A}$	20			V
$V_{CE(sat)}$	Output saturation voltage	$V_I=7\text{V}$		0.5	0.85	V
		$V_{I(STB)}=2.4\text{V}$	$V_{CC}=6.5\text{V}$, $I_C=250\text{mA}$ $V_{CC}=3\text{V}$, $I_C=120\text{mA}$	0.3	0.5	
I_I	Input current	$V_{CC}=8\text{V}$, $V_I=18\text{V}$, $V_{I(STB)}=2.4\text{V}$		0.9	1.8	mA
I_R	Input leakage current	$V_{CC}=8\text{V}$, $V_I=-25\text{V}$		0	-20	μA
$I_{I(STB)}$	Strobe input current	$V_{CC}=8\text{V}$, $V_I=7\text{V}$ all input $V_{I(STB)}=0.2\text{V}$		-4		mA
$I_{R(STB)}$	Strobe input leakage current	$V_{CC}=8\text{V}$, $V_I=0\text{V}$, $V_{I(STB)}=20\text{V}$		0	10	μA
I_{CC}	Supply current	$V_{CC}=8\text{V}$, $V_I=7\text{V}$ all input $V_{I(STB)}=2.4\text{V}$		95	170	mA
h_{FE}	DC forward current gain	$V_{CE}=4\text{V}$, $V_{CC}=6.5\text{V}$, $I_C=300\text{mA}$, $T_a=25^\circ\text{C}$	1000	3000		—

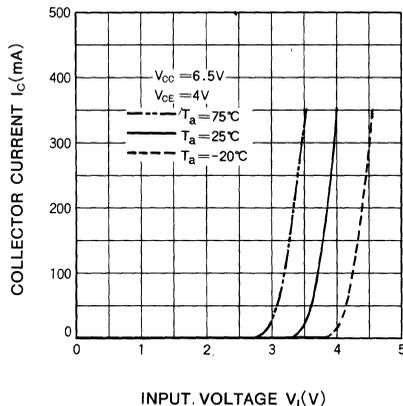
* : All typical values are at $T_a=25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION

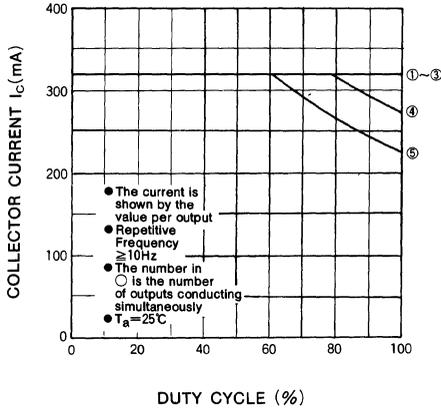


OUTPUT CURRENT CHARACTERISTICS

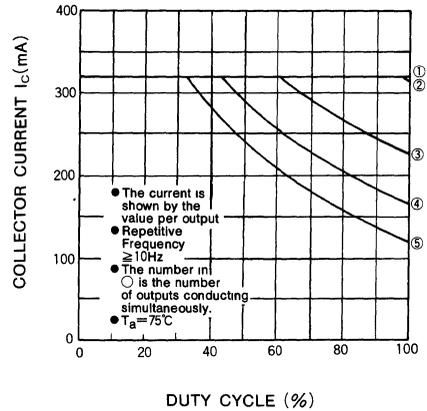


5-UNIT 320mA TRANSISTOR ARRAY WITH STROBE

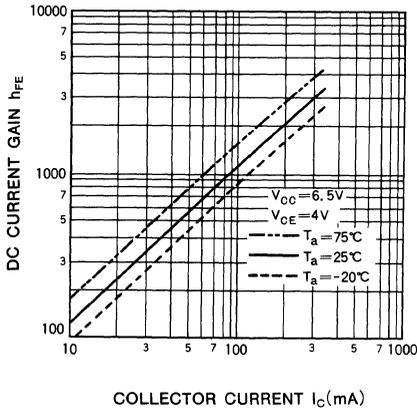
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54533P

6-UNIT 320mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

DESCRIPTION

The M54533P, 6-channel sink driver, consists of 12 NPN transistors to form high current gain driver pairs.

FEATURES

- Output breakdown voltage to 20V
- High output sink current to 320mA
- Integral diode for transient suppression
- Strobe control input
- Wide input voltage range from -25V to +20V
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED or incandescent display digit driver

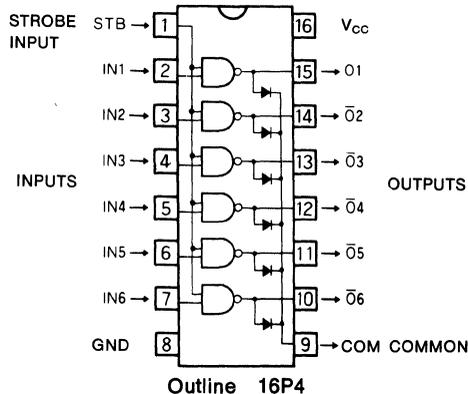
FUNCTION

The M54533P uses a predriver stage. Each input has a diode and 20kΩ resistor in series to allow a negative voltage input. All input can be controlled simultaneously by a strobe input at pin 1.

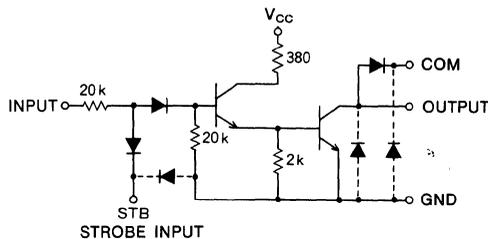
The power supply of the predrivers is connected to pin 16. All emitters and the substrate are connected together to pin 8. Each output has an integral diode for inductive load transient suppression and the cathodes of the diodes are connected to pin 9.

The outputs are capable of sinking 320mA and will withstand 20V in the OFF state.

PIN CONFIGURATION (TOP VIEW)



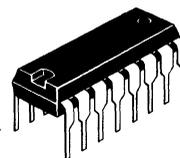
CIRCUIT SCHEMATIC



Unit: Ω

FUNCTIONAL TABLE

iN	STB	OUT
L	L	H
H	L	H
L	H	H
H	H	L



16-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +20	V
I_C	Collector current	Transistor ON	320	mA
V_I	Input voltage		-25 ~ +20	V
$V_{I(STB)}$	Strobe input voltage		20	V
$V_{R(D)}$	Clamp diode reverse voltage		20	V
$I_{F(D)}$	Clamp diode forward current		320	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{Opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{Stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

6-UNIT 320mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		3		8	V
V_O	Output voltage		0		20	V
I_C	Collector current per channel	Percent duty cycle less than 25%, $V_{CC}=6.5\text{V}$	0		300	mA
		Percent duty cycle less than 65%, $V_{CC}=6.5\text{V}$	0		150	
V_{IH}	"H" Input voltage	$I_C=300\text{mA}$	7		18	V
		$I_C=150\text{mA}$	5		18	
V_{IL}	"L" Input voltage	$I_{O(leak)}=50\mu\text{A}$	0		1	V
$V_{IH(STB)}$	"H" Input voltage (strobe input)		2.4		18	V
$V_{IL(STB)}$	"L" Input voltage (strobe input)		0		0.2	V

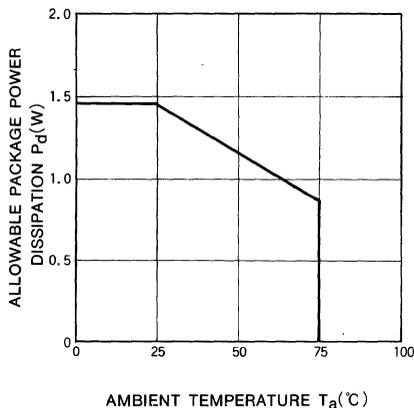
ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$V_{CC}=8\text{V}$, $V_I=18\text{V}$, $V_{I(STB)}=0.2\text{V}$ $I_{CEO}=100\mu\text{A}$	20			V
$V_{CE(sat)}$	Output saturation voltage	$V_I=7\text{V}$		0.5	0.85	V
		$V_{I(STB)}=2.4\text{V}$	$V_{CC}=6.5\text{V}$, $I_C=250\text{mA}$ $V_{CC}=3\text{V}$, $I_C=120\text{mA}$	0.3	0.5	
I_I	Input current	$V_{CC}=8\text{V}$, $V_I=18\text{V}$, $V_{I(STB)}=2.4\text{V}$		0.8	1.8	mA
I_R	Input leakage current	$V_{CC}=8\text{V}$, $V_I=-25\text{V}$			-20	μA
$I_{I(STB)}$	Strobe input current	$V_{CC}=8\text{V}$, $V_I=18\text{V}$ (all input), $V_{I(STB)}=0.2\text{V}$		-4		mA
$I_{R(STB)}$	Strobe input leakage current	$V_{CC}=8\text{V}$, $V_I=0\text{V}$, $V_{I(STB)}=20\text{V}$			20	μA
$V_{F(D)}$	Clamp diode forward voltage	$I_{F(D)}=320\text{mA}$		1.4	2.4	V
$V_{R(D)}$	Clamp diode reverse voltage	$I_{R(D)}=100\mu\text{A}$	20	40		V
I_{CC}	Supply current	$V_{CC}=8\text{V}$, $V_I=7\text{V}$ (all input)		120	200	mA
		$V_{I(STB)}=2.4\text{V}$				
h_{FE}	DC forward current gain	$V_{CE}=4\text{V}$, $V_{CC}=6.5\text{V}$, $I_C=300\text{mA}$, $T_a=25^\circ\text{C}$	1000	3000		—

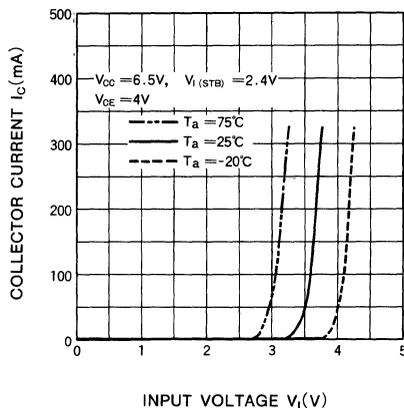
6-UNIT 320mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

TYPICAL CHARACTERISTICS

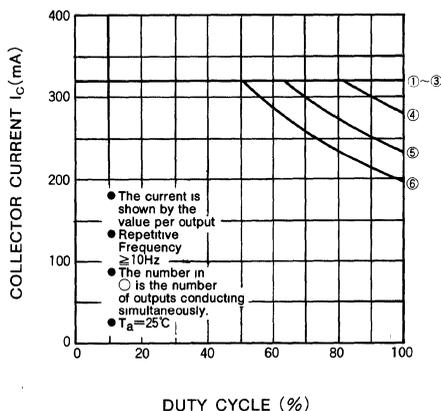
ALLOWABLE AVERAGE POWER DISSIPATION



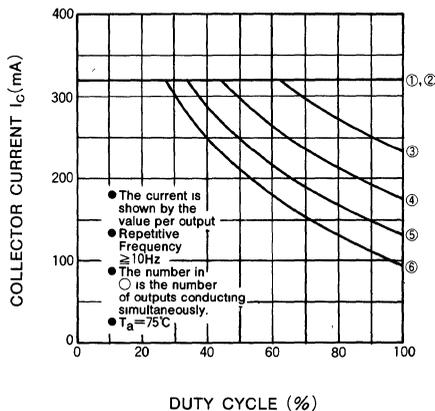
OUTPUT CURRENT CHARACTERISTICS



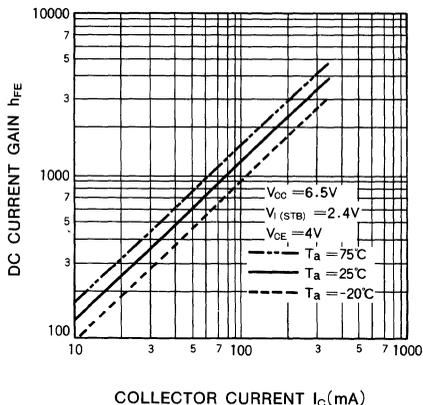
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54534P

6-UNIT 320mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

DESCRIPTION

The M54534P, 6-channel sink driver, consists of 12 NPN transistors connected to form high current gain driver pairs.

FEATURES

- Output breakdown voltage to 20V
- High output sink current to 320mA
- Integral diodes for transient suppression
- Strobe control input
- Wide input voltage range from $-25V$ to $+20V$
- Wide operating temperature range ($T_a = -20 \sim +75^\circ C$)

APPLICATIONS

Relay and printer driver, LED or incandescent display digit driver

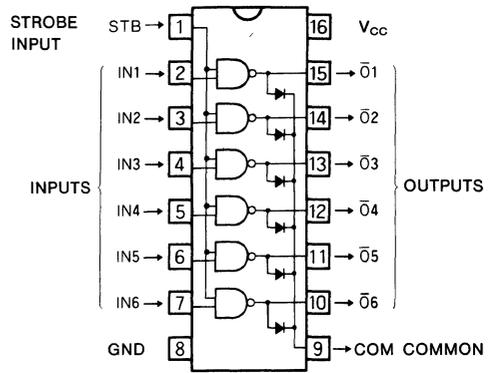
FUNCTION

The M54534P uses a predriver stage. Each input has a diode and 1.6kΩ resistor in series to allow a negative voltage input. All input can be controlled simultaneously by a strobe input at pin 1.

The power supply of the predrivers is connected to pin 16. Each output has an integral diode for inductive load transient suppression and the cathodes of the diodes are connected to pin 9. All emitters and the substrate are connected together to pin 8.

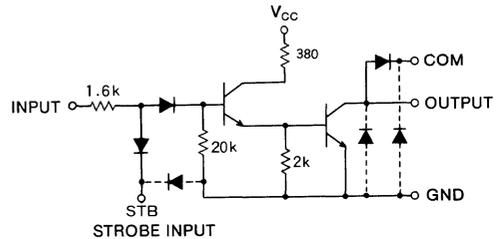
The outputs are capable of sinking 320mA and will withstand 20V in the OFF state.

PIN CONFIGURATION (TOP VIEW)



Outline 16P4

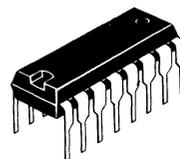
CIRCUIT SCHEMATIC



Unit : Ω

FUNCTIONAL TABLE

IN	STB	OUT
L	L	H
H	L	H
L	H	H
H	H	L



16-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ C$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +20$	V
I_C	Collector current	Transistor ON	320	mA
V_I	Input voltage		$-25 \sim +20$	V
$V_{I(STB)}$	Strobe input voltage		20	V
$V_{R(D)}$	Clamp diode reverse voltage		20	V
$I_{F(D)}$	Clamp diode forward current		320	mA
P_d	Power dissipation	$T_a = 25^\circ C$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ C$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ C$

6-UNIT 320mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		3		8	V
V_O	Output voltage		0		20	V
I_C	Collector current per channel	Percent duty cycle less than 25%, $V_{CC}=6.5\text{V}$	0		300	mA
		Percent duty cycle less than 65%, $V_{CC}=6.5\text{V}$	0		150	
V_{IH}	"H" Input voltage	$I_C=300\text{mA}$	3.2		18	V
V_{IL}	"L" Input voltage	$I_{O(leak)}=50\mu\text{A}$	0		0.7	V
$V_{IH(STB)}$	"H" Input voltage (strobe input)		2.4		18	V
$V_{IL(STB)}$	"L" Input voltage (strobe input)		0		0.2	V

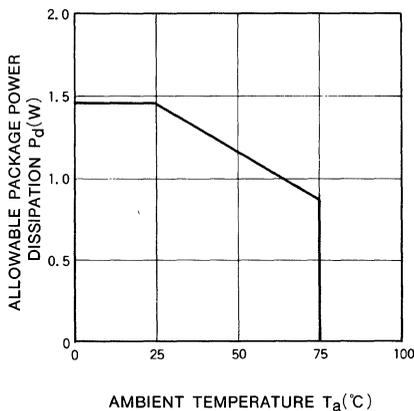
ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$V_{CC}=8\text{V}$, $V_I=3.2\text{V}$, $V_{I(STB)}=0.2\text{V}$ $I_{CEO}=100\mu\text{A}$	20			V
$V_{CE(sat)}$	Output saturation voltage	$V_I=3\text{V}$		0.5	0.85	V
		$V_{I(STB)}=2.4\text{V}$	$V_{CC}=6.5\text{V}$, $I_C=250\text{mA}$ $V_{CC}=3\text{V}$, $I_C=120\text{mA}$	0.3	0.5	
I_I	Input current	$V_{CC}=8\text{V}$, $V_I=3.2\text{V}$, $V_{I(STB)}=2.4\text{V}$			1.4	mA
I_R	Input leakage current	$V_{CC}=8\text{V}$, $V_I=-25\text{V}$			-20	μA
$I_{I(STB)}$	Strobe input current	$V_{CC}=8\text{V}$, $V_I=3.2\text{V}$ (all input), $V_{I(STB)}=0.2\text{V}$		-7.9		mA
$I_{R(STB)}$	Strobe input leakage current	$V_{CC}=8\text{V}$, $V_I=0\text{V}$, $V_{I(STB)}=20\text{V}$			20	μA
$V_{F(D)}$	Clamp diode forward voltage	$I_{F(D)}=320\text{mA}$		1.4	2.4	V
$V_{R(D)}$	Clamp diode reverse voltage	$I_{R(D)}=100\mu\text{A}$	20	40		V
I_{CC}	Supply current	$V_{CC}=8\text{V}$, $V_I=3.2\text{V}$ (all input) $V_{I(STB)}=2.4\text{V}$		120	200	mA
h_{FE}	DC forward current gain	$V_{CE}=4\text{V}$, $V_{CC}=6.5\text{V}$, $I_C=300\text{mA}$, $T_a=25^\circ\text{C}$	1000	3000		—

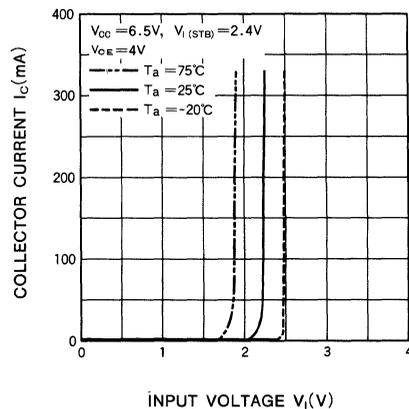
* : All typical values are at $T_a=25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION

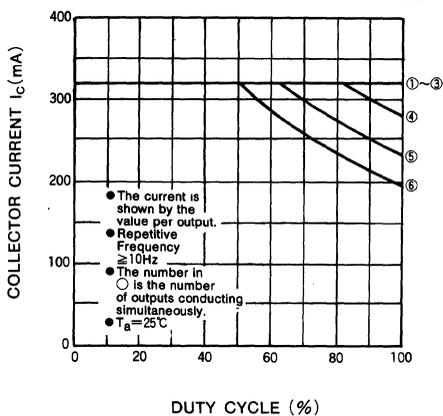


OUTPUT CURRENT CHARACTERISTICS

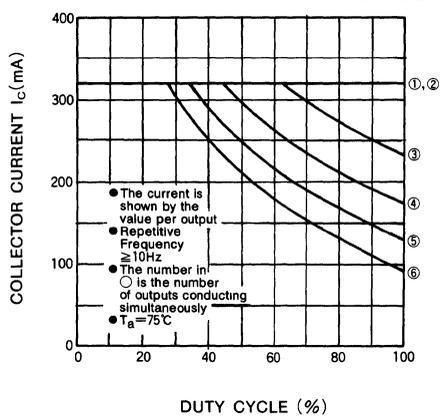


6-UNIT 320mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

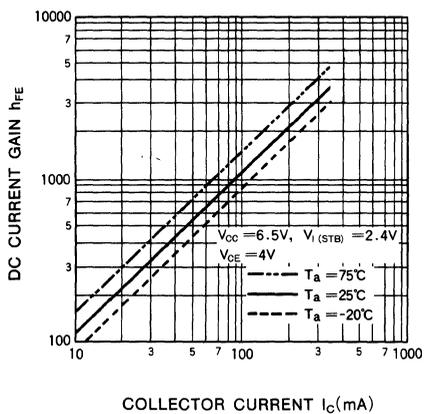
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54539P

6-UNIT 700mA TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54539P, 6-channel sink driver, consists of 12 NPN transistors connected to form high current gain driver pairs.

FEATURES

- Output breakdown voltage to 20V
- High output sink current to 700mA
- Integral diodes for transient suppression
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

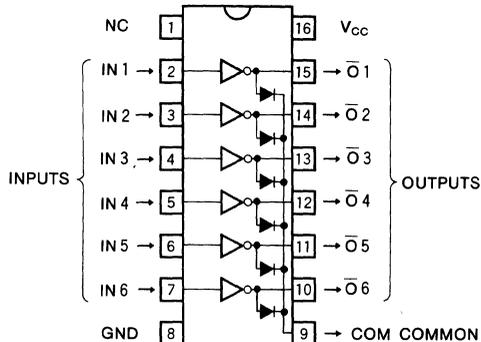
APPLICATIONS

Relay and solenoid driver, LED or incandescent display driver, Thermal head driver

FUNCTION

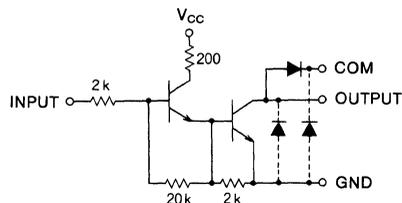
The M54539P uses a predriver stage with $2k\Omega$ series input resistor. The power supply of the predrivers is connected to pin 16. Each output has an integral diode for inductive load transient suppression and the cathodes of the diodes are connected to pin 9. All emitters and the substrate are connected together to pin 8. The outputs are capable of sinking 700mA and will withstand 20V in the OFF state.

PIN CONFIGURATION (TOP VIEW)



Outline 16P4 NC : NO CONNECTION

CIRCUIT SCHEMATIC



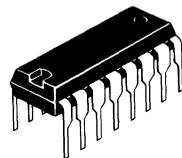
Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +20	V
I_C	Collector current	Transistor ON	700	mA
V_i	Input voltage		10	V
V_R	Clamp diode reverse voltage		20	V
I_F	Clamp diode forward current	Pulse width $\leq 35\text{ms}$, Percent duty cycle $\leq 5\%$	700	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	350	mA
T_{opr}	Operating ambient temperature range		1.47	W
T_{stg}	Storage temperature range		-20 ~ +75	$^\circ\text{C}$
			-55 ~ +125	$^\circ\text{C}$

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit	
		Min	Typ	Max		
V_{CC}	Supply voltage	3	5	7	V	
V_O	Output voltage	0		20	V	
I_C	Collector current per channel	The three outputs conducting simultaneously Percent duty cycle less than 20%	0		700	mA
		The three outputs conducting simultaneously Percent duty cycle less than 90%	0		200	
V_{IH}	"H" Input voltage	$I_C = 450\text{mA}$	3	6	V	
V_{IL}	"L" Input voltage	$I_{o(\text{leak})} = 50\mu\text{A}$	0	0.3	V	



16-pin molded plastic DIP

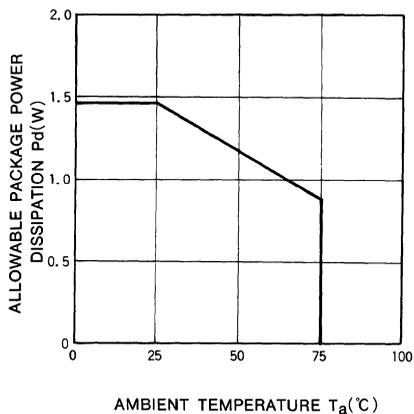
6-UNIT 700mA TRANSISTOR ARRAY WITH CLAMP DIODE

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

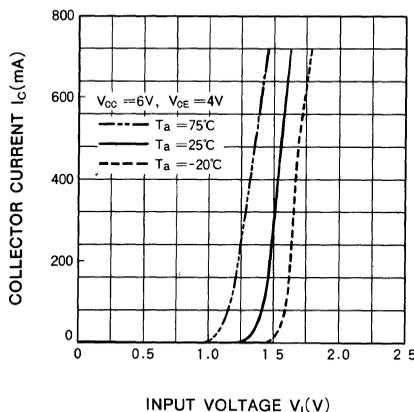
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$V_{CC}=7V, I_{CEO}=100\mu A$	20			V
$V_{CE(sat)}$	Output saturation voltage	$V_{CC}=5V$		0.46	0.8	V
		$V_I=3V$	$I_C=450mA$	0.2	0.45	
I_I	Input current	$V_{CC}=7V, V_I=3.2V$		0.75	1.4	mA
V_R	Clamp diode reverse voltage	$I_R=100\mu A$	20			V
V_F	Clamp diode forward voltage	$I_F=350mA$		1.5	2.7	V
I_{CC}	Supply current	$V_{CC}=7V, V_I=3.2V$ (all input)		190	300	mA
h_{FE}	DC forward current gain	$V_{CE}=4V, V_{CC}=6V, I_C=300mA, T_a=25^\circ\text{C}$	3000	8000		—

TYPICAL CHARACTERISTICS

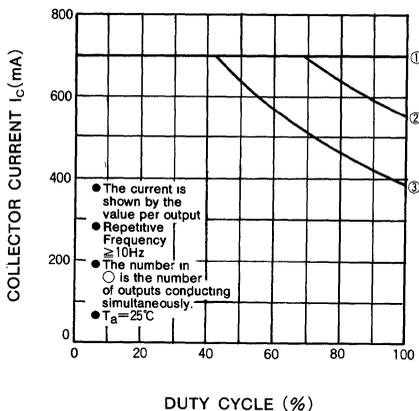
ALLOWABLE AVERAGE POWER DISSIPATION



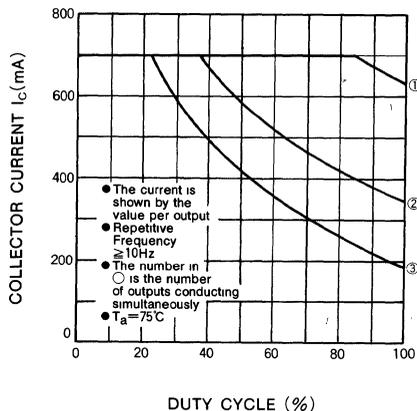
OUTPUT CURRENT CHARACTERISTICS



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE

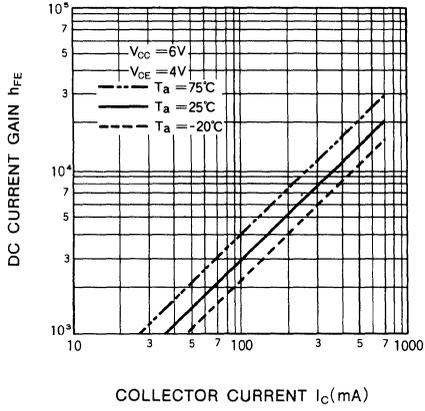


ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



6-UNIT 700mA TRANSISTOR ARRAY WITH CLAMP DIODE

**DC CURRENT GAIN
CHARACTERISTICS**



M54578P

6-UNIT 700mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

DESCRIPTION

The M54578P, 6-channel sink driver, consists of 12 NPN transistors connected to form high current gain driver pairs.

FEATURES

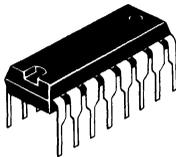
- 20V breakdown
- High output sink current to 700mA
- PMOS Compatible
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATIONS

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

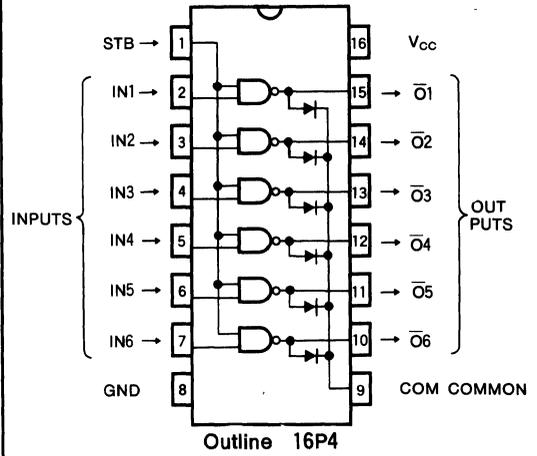
FUNCTION

The M54578P uses a predriver stage. Each input has a diode and 2 k Ω resistor in series to allow a negative voltage input. All input can be controlled simultaneously by a strobe input at pin 1. The power supply of the predrivers is connected to pin 16. Each output has an integral diode for inductive load transient suppression and the cathodes of the diodes are connected to pin 9. All emitters and the substrate are connected together to pin 8. The outputs are capable of sinking 700mA and will withstand 20V in the OFF state.

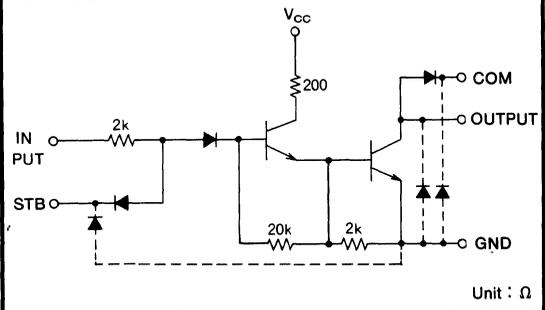


16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +20	V
I_C	Collector current	Transistor ON	700	mA
V_I	Input voltage		-25 ~ +20	V
$V_{I(STB)}$	Strobo input voltage		20	V
$V_{R(D)}$	Clamp diode reverse voltage		20	V
$I_{F(D)}$	Clamp diode forward current	Pulse width $\leq 35\text{ms}$, Duty cycle $\leq 5\%$	700	mA
			350	●
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

6-UNIT 700mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

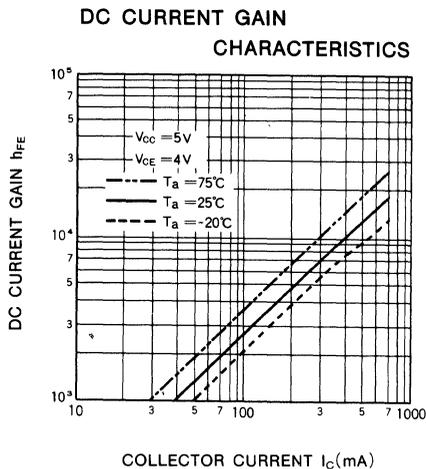
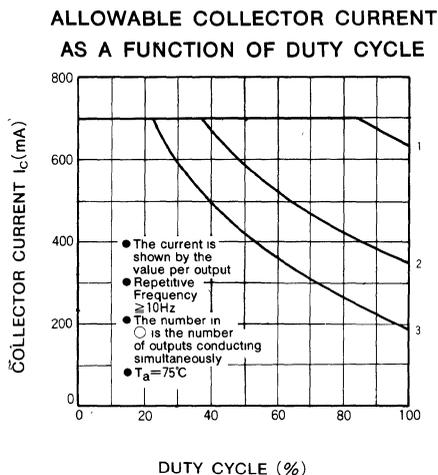
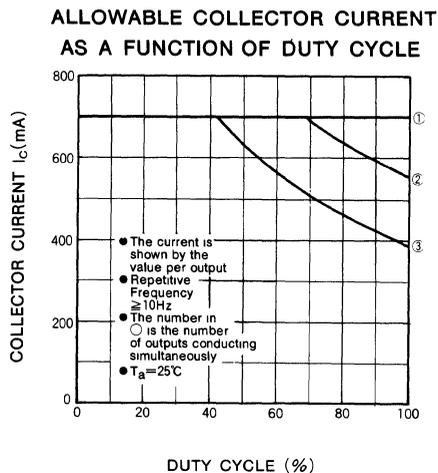
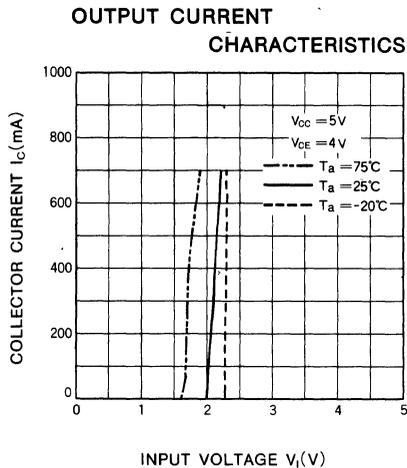
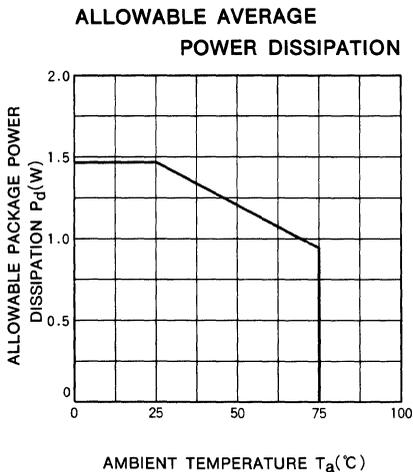
Symbol	Parameter	Limits			Unit	
		Min	Typ	Max		
V_{CC}	Supply voltage	3	5	8	V	
V_O	Output voltage	0		20	V	
I_C	Collector current per channel	The three outputs conducting simultaneously Percent duty cycle less than 20%	0		700	mA
		The three outputs conducting simultaneously Percent duty cycle less than 90%	0		200	
$V_{IH(STB)}$	"H" Input voltage (strobe input)	2.4		V_{CC}	V	
$V_{IL(STB)}$	"L" Input voltage (strobe input)	0		0.2	V	
V_{IH}	"H" Input voltage	$I_C = 450\text{mA}$, $V_{CC} = 5\text{V}$	3.5		V_{CC}	V
		$I_C = 700\text{mA}$, $V_{CC} = 6\text{V}$	5		V_{CC}	
V_{IL}	"L" Input voltage	$I_{O(Leak)} = 50\mu\text{A}$	0	0.8	V	

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$V_{CC} = 7\text{V}$, $V_{I(STB)} = 0.4\text{V}$ $V_I = 3.5\text{V}$, $I_{CEO} = 100\mu\text{A}$	20			V
$V_{CE(sat)}$	Collector-emitter saturation voltage	$V_{CC} = 5\text{V}$		0.5	0.8	V
		$V_I = 3.5\text{V}$	$I_C = 450\text{mA}$ $I_C = 200\text{mA}$	0.23	0.45	
I_I	Input current	$V_{CC} = 7\text{V}$, $V_I = 3.5\text{V}$ $V_{I(STB)} = 2.4\text{V}$		0.6	1.4	mA
I_R	Input leakage current	$V_{CC} = 7\text{V}$, $V_R = -25\text{V}$			-20	μA
$I_{I(STB)}$	Strobe input current	$V_{CC} = 7\text{V}$, $V_{I(STB)} = 0.4\text{V}$ $V_I = 3.5\text{V}$ (all input)		-7.2	-10.7	mA
$I_{R(STB)}$	Strobe input leakage current	$V_{CC} = 7\text{V}$, $V_I = 0\text{V}$, $V_{I(STB)} = 20\text{V}$			20	μA
$V_{F(D)}$	Clamp diode forward current	$I_{F(D)} = 600\text{mA}$		1.6	5	V
$V_{R(D)}$	Clamp diode reverse voltage	$I_{R(D)} = 100\mu\text{A}$	20			V
I_{CC}	Supply current	$V_{CC} = 8\text{V}$, $V_{I(STB)} = 2.4\text{V}$ $V_I = 3.5\text{V}$ (all input)		220	320	mA
h_{FE}	DC forward current gain	$V_{CC} = 5\text{V}$ $V_{CE} = 4\text{V}$, $I_C = 450\text{mA}$, $T_a = 25^\circ\text{C}$	2000	10000		—

6-UNIT 700mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

TYPICAL CHARACTERISTICS



M54527P

6-UNIT 150mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54527P, 6-channel sink driver, consists of 12 NPN transistors connected to form high current gain driver pairs.

FEATURES

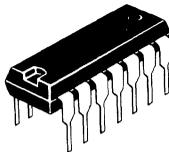
- High output sustaining voltage to 40V
- Output sink current to 150mA
- PMOS compatible input
- Integral diode for transient suppression
- Wide input voltage range from $-40V$ to $+40V$
- Wide operating temperature range ($T_a = -20 \sim +75^\circ C$)

APPLICATIONS

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

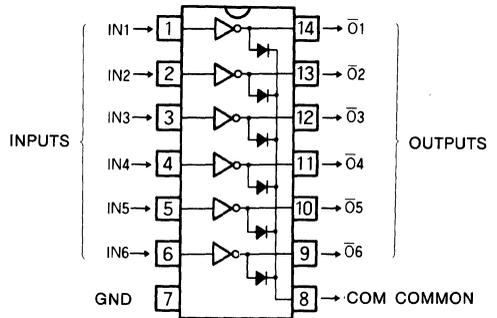
FUNCTION

The M54527P is comprised of six darlington driver pairs. Each input has a diode and $20k\Omega$ resistor in series to allow a negative voltage input. Between pin 8 and each output, there are integral diodes for inductive load transient suppression. All emitters and the substrate are connected together to pin 7. The outputs are capable of sinking 150mA and will withstand 40V in the OFF state.



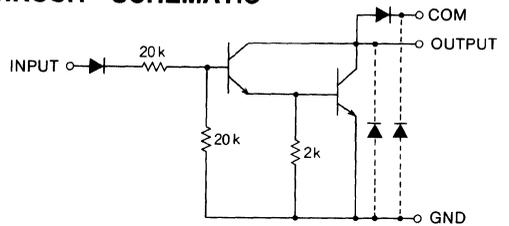
14-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



Outline 14P4

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ C$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +40$	V
I_C	Collector current	Transistor ON	150	mA
V_I	Input voltage		$-40 \sim +40$	V
$I_{F(D)}$	Clamp diode forward current		150	mA
$V_{R(D)}$	Clamp diode reverse voltage		40	V
P_d	Power dissipation	$T_a = 25^\circ C$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ C$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ C$

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ C$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		40	V
I_C	Collector current per channel	Percent duty cycle less than 50%	0		150	mA
V_{IH}	"H" Input voltage	$I_C = 150mA$	7		35	V
V_{IL}	"L" Input voltage	$I_{OL(peak)} = 50\mu A$	0		1	V

6-UNIT 150mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

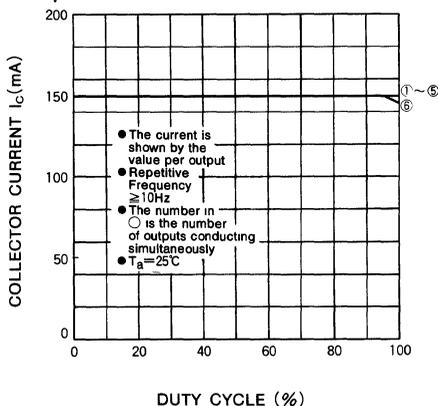
ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CEO} = 100\mu\text{A}$	40			V
$V_{CE(sat)}$	Output saturation voltage	$V_I = 7\text{V}, I_C = 150\text{mA}$		1.4	1.7	V
		$V_I = 7\text{V}, I_C = 100\text{mA}$		1.2	1.4	
I_i	Input current	$V_I = 18\text{V}$		0.9	1.8	mA
		$V_I = 35\text{V}$		1.9	5	
I_R	Input leakage current	$V_I = -35\text{V}$			-20	μA
$V_{F(D)}$	Clamp diode forward voltage	$I_{F(D)} = 150\text{mA}$		1.15	1.6	V
$I_{R(D)}$	Clamp diode leakage current	$V_{R(D)} = 40\text{V}$			100	μA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 150\text{mA}, T_a = 25^\circ\text{C}$	800	2500		—

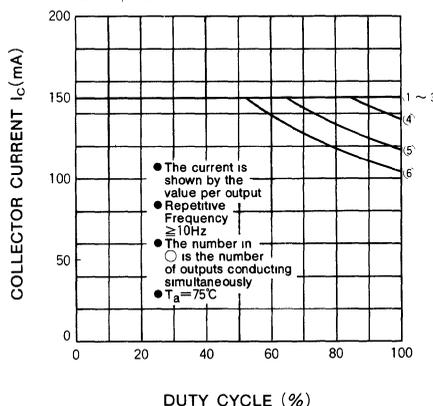
* : All typical values are at $T_a = 25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

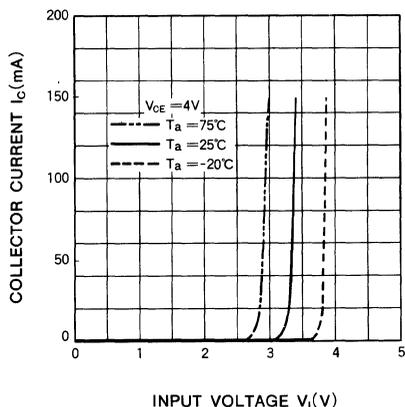
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



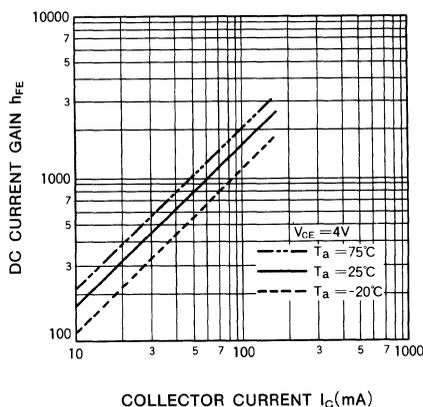
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



OUTPUT CURRENT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54571P

6-UNIT 350mA TRANSISTOR ARRAY AND MOTOR DRIVER

DESCRIPTION

The M54571P, 6-channel sink driver and voltage regulator, is designed for use with a small printer.

FEATURES

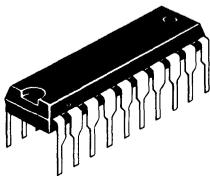
- High output sustaining voltage to 40V
- High output sink current to 350mA
- Voltage regulator with a control circuit
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Small calculator printer driver

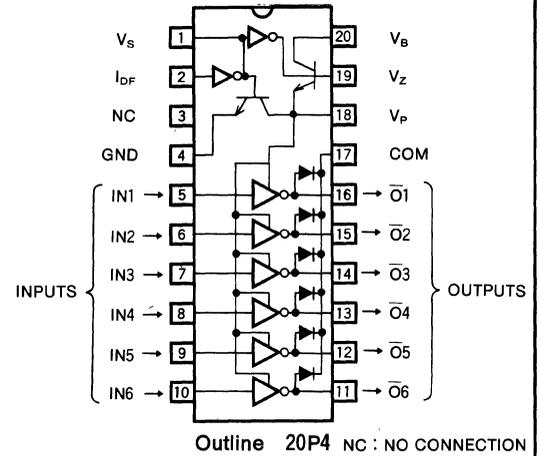
FUNCTION

The M54571P is designed for driving a small serial printer made by CITIZEN and EPSON, and consists of 6 relay drivers and 1.2A motor driver. Each driver has 4.3kΩ series input resistor and output transient suppression diode. The driver outputs are capable of sinking 350mA and will withstand 43V in the OFF state. The output of the motor driver at pin 18 can drive 1.2A.

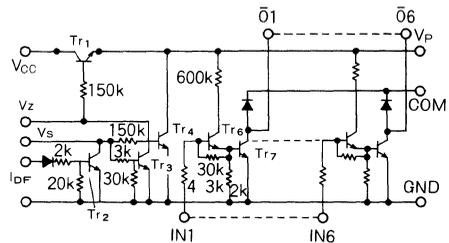


20-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		40	V
V_{CEO}	Output sustaining voltage		$-0.5 \sim +40$	V
I_{C1}	Collector current	T_{r1}	100	nA
I_{C2}		T_{r2}	100	mA
I_{C3}		T_{r3}	100	mA
I_{C4}		Spike current 2A max Pulse width ≤ 5 ms, Duty cycle $\leq 5\%$	1200	mA
I_{C7}		T_{r7} (Per channel)	350	nA
V_i	Input voltage	IN1~IN6	40	V
$V_{I(DF)}$			40	V
$V_{R(DF)}$	Input reverse voltage		-45	V
$V_{R(D)}$	Clamp diode reverse voltage		40	V
$I_{F(D)}$	Clamp diode forward current		350	mA
Pd	Power dissipation	$T_a = 25^\circ\text{C}$	1.79	W
T_{OPR}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{STG}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

6-UNIT 350mA TRANSISTOR ARRAY AND MOTOR DRIVER

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{CC}	Supply voltage	8		40	V
V_P	Supply voltage	4		18	V
V_S	Reference voltage		10		V
I_C	Collector current	$\bar{O}1 \sim \bar{O}6$	5	250	mA
		$\bar{O}1 \sim \bar{O}6$	0	100	
$V_{(IDF)}$	Input voltage			-35	V
			9	17	
V_I		$IN1 \sim IN6$	9	17	V
V_O	Output voltage	$\bar{O}1 \sim \bar{O}6$	0	40	V

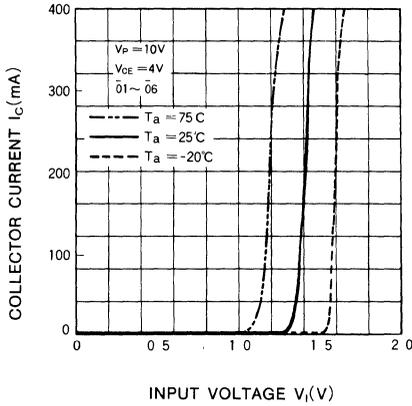
ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{BR(CEO)}$	Output sustaining voltage	$I_{CEO} = 100\mu\text{A}$, $V_P = 5\text{V}$, ($\bar{O}1 \sim \bar{O}7$)	40			V
$V_{CE(sat)}$	Collector emitter saturation voltage	$\bar{O}1$	$V_P = 6.5\text{V}$, $V_I = 3\text{V}$, $I_C = 250\text{mA}$		0.8	V
		$\bar{O}7$	$V_P = 3\text{V}$, $V_I = 2.4\text{V}$, $I_C = 120\text{mA}$		0.5	
		T_{r1}	$I_B = 1\text{mA}$, $I_C = 10\text{mA}$, $V_P = 0\text{V}$		0.5	
		T_{r2}	$V_{(IDF)} = 10\text{V}$, $I_{VS} = 100\text{mA}$		0.5	
		T_{r3}	$V_{(VS)} = 3\text{V}$, $I_{VZ} = 30\text{mA}$, $V_{(IDF)} = 0\text{V}$		0.4	
		T_{r4}	$I_{VS} = 50\text{mA}$, $I_{VP} = 0.3\text{V}$, $V_{(IDF)} = 0\text{V}$		0.45	
		$I_{VS} = 80\text{mA}$, $I_{VP} = 1\text{A}$, $V_{(IDF)} = 0\text{V}$		1.2		
I_I	Input current	$V_P = 6\text{V}$, $V_I = 10\text{V}$, ($IN1 \sim IN6$)			3.5	mA
$I_{(IDF)}$		$V_{(IDF)} = 10\text{V}$			6.5	mA
$I_{(VS)}$		$V_{(VS)} = 3\text{V}$, $V_{(IDF)} = 0\text{V}$			26	mA
$I_{R(IDF)}$	Input leakage current	$V_{R(IDF)} = -35\text{V}$			-20	μA
$V_{F(D)}$	Clamp diode forward voltage	$I_{F(D)} = 250\text{mA}$			2.4	V
I_{VP}	Supply current	$V_P = 17\text{V}$, $V_I = 10\text{V}$ (all input)			240	mA
		$V_P = 5\text{V}$, $V_I = 10\text{V}$ (all input)			60	
$h_{FE} 1$	DC forward current gain	T_{r4} , $I_C = 50\text{mA}$, $V_{CE} = 4\text{V}$, $T_a = 25^\circ\text{C}$	100			—
$h_{FE} 2$	DC forward current gain	T_{r4} , $I_C = 1\text{A}$, $V_{CE} = 4\text{V}$, $T_a = 25^\circ\text{C}$	80			—
$h_{FE} 3$	DC forward current gain	$\bar{O}1 \sim \bar{O}7$, $V_P = 6.5\text{V}$, $I_C = 350\text{mA}$, $V_{CE} = 4\text{V}$, $T_a = 25^\circ\text{C}$	1000			—

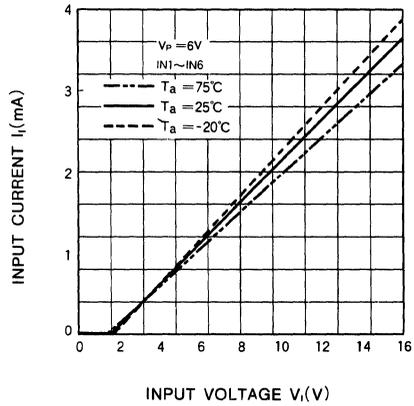
6-UNIT 350mA TRANSISTOR ARRAY AND MOTOR DRIVER

TYPICAL CHARACTERISTICS

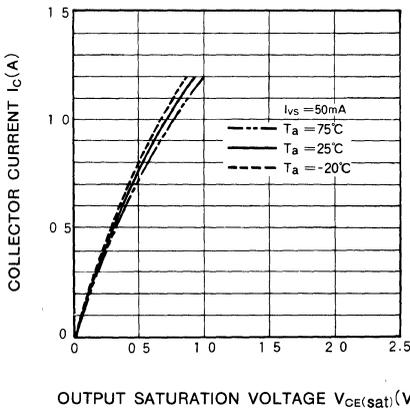
OUTPUT CURRENT CHARACTERISTICS



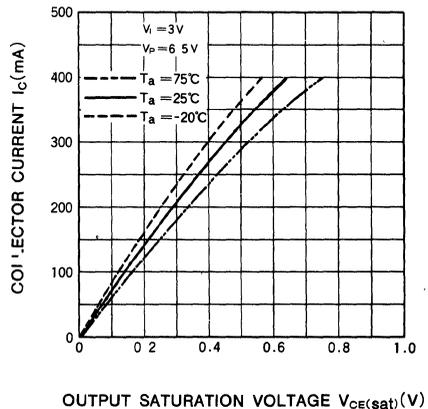
INPUT CHARACTERISTICS



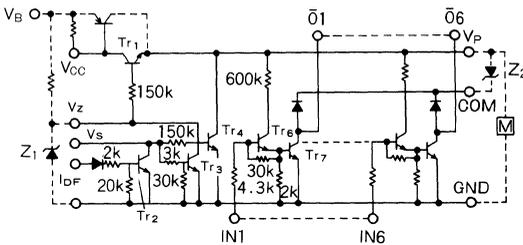
MOTOR DRIVER OUTPUT SATURATION CHARACTERISTICS



MAGNET RELAY DRIVER OUTPUT SATURATION CHARACTERISTICS



TYPICAL APPLICATION



NOTE

	V_B	V_{Z1}	Z_2	Magnet Relay Drive Current
EPSON Printer	15~40V	18V	—	90mA
CITIZEN Printer	3~9V	6V	connect between the V_P and the COM	250mA

M54514AP

7-UNIT 50mA TRANSISTOR ARRAY

DESCRIPTION

The M54514AP, 7-channel sink drivers, consists of 7 NPN transistors with 2.8kΩ series input resistors.

FEATURES

- Output breakdown voltage to 20V
- 50mA output sink current capability
- Low output saturation voltage
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

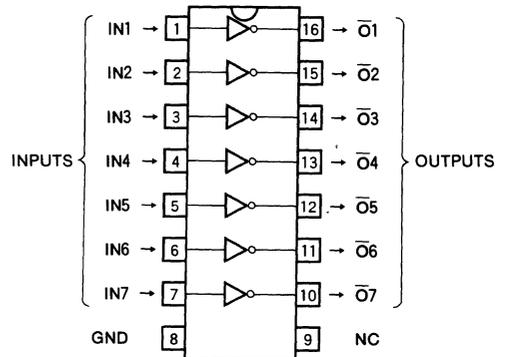
APPLICATION

LED or incandescent display digit driver

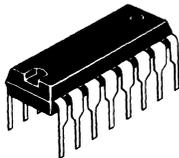
FUNCTION

The M54514AP is comprised of seven NPN drivers. Each input has a voltage divider by 2.8kΩ and 10kΩ resistors. All emitters and the substrate are connected together to pin 8. The open collector outputs are capable of sinking 50mA and will withstand 20V in the OFF state.

PIN CONFIGURATION (TOP VIEW)

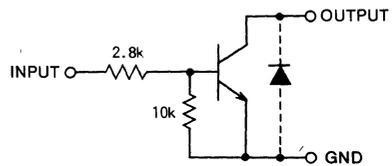


NC : NO CONNECTION



16-pin molded plastic DIP

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +20	V
I_C	Collector current	Transistor ON	50	mA
V_i	Input voltage		10	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_o	Output voltage	0		20	V
I_C	Collector current	0		20	mA
V_{IH}	"H" Input voltage	$I_C = 50\text{mA}$		8	V
V_{IL}	"L" Input voltage	0		0.2	V

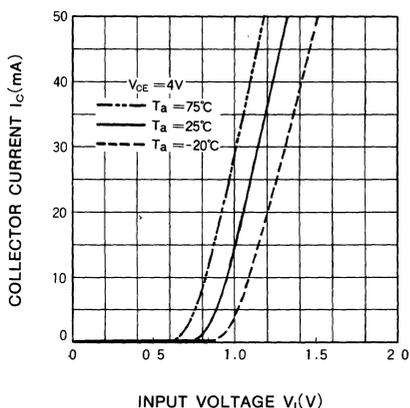
7-UNIT 50mA TRANSISTOR ARRAY

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

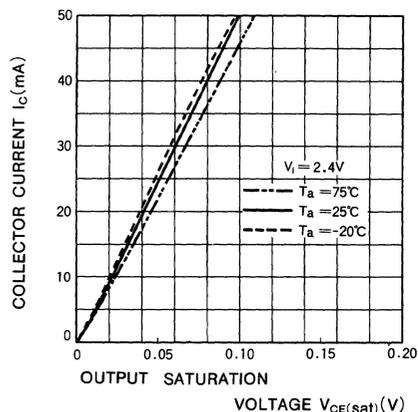
Symbol	Parameter	Test conditions	Limits			Unit	
			Min	Typ	Max		
$I_{O(leak)}$	Output leakage current	$V_{CE}=20\text{V}$			20	μA	
$V_{CE(sat)}$	Output saturation voltage	$V_I=2.4\text{V}$	$I_C=20\text{mA}$		0.04	0.17	V
			$I_C=40\text{mA}$		0.08	0.23	
I_I	Input current	$V_I=2.4\text{V}$		0.7	1.1	mA	
h_{FE}	DC forward current gain	$V_{CE}=4\text{V}, I_C=40\text{mA}, T_a=25^\circ\text{C}$	80	200		—	

TYPICAL CHARACTERISTICS

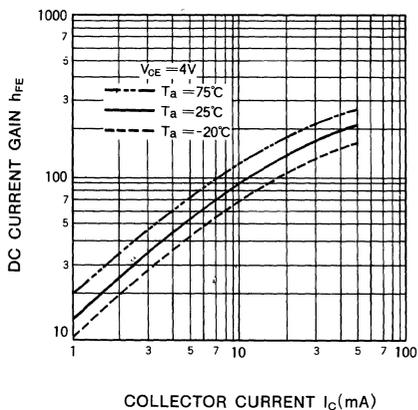
OUTPUT CURRENT CHARACTERISTICS



OUTPUT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54515P

7-UNIT 16mA TRANSISTOR ARRAY

DESCRIPTION

The M54515P, transistor array, consists of seven NPN transistors and is connected in a common-emitter configuration.

FEATURES

- Output breakdown voltage to 17V
- 16mA output sink current capability
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

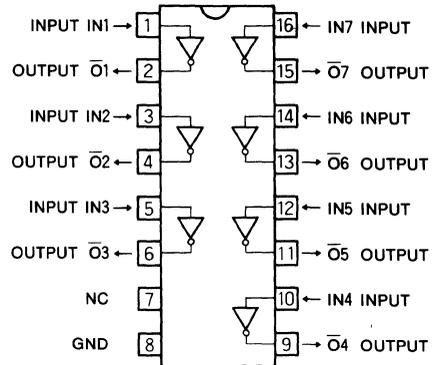
APPLICATION

LED or incandescent display driver

FUNCTION

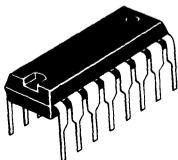
The M54515P is comprised of seven NPN transistors. All emitters and the substrate are connected together to pin 8. The outputs are capable of sinking 16mA and will withstand 17V in the OFF state.

PIN CONFIGURATION (TOP VIEW)



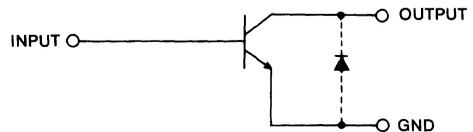
Outline 16P4

NC : NO CONNECTION



16-pin molded plastic DIP

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +17$	V
V_i	Input voltage		1.2	V
I_C	Collector current	Transistor ON	16	mA
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

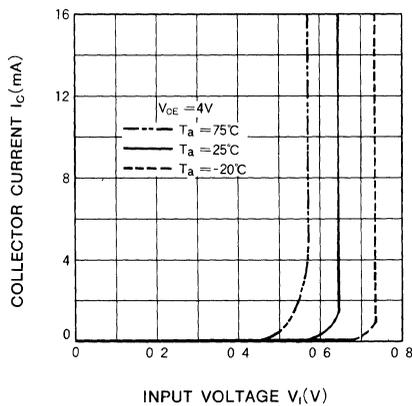
Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_O	Output voltage	0		17	V
I_C	Collector current per channel	0		16	mA

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

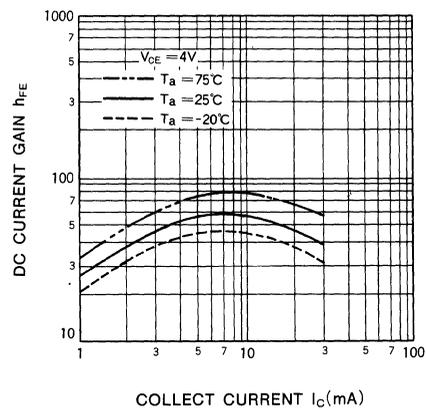
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{OH} = 100\mu\text{A}$	17			V
$V_{CE(sat)}$	Output saturation voltage	$I_{OL} = 16\text{mA}$, $I_B = 0.5\text{mA}$		0.14	0.5	V
$V_{BE(sat)}$	Base-emitter saturation voltage	$I_{OL} = 16\text{mA}$, $I_B = 0.5\text{mA}$		0.64	1.2	V
h_{FE}	DC forward current gain	$V_{CE} = 5\text{V}$, $I_C = 16\text{mA}$, $T_a = 25^\circ\text{C}$	32	50		—
$I_{O(leak)}$	Output leakage current	$V_O = 17\text{V}$, $V_i = 0.2\text{V}$			700	μA

TYPICAL CHARACTERISTICS

OUTPUT CURRENT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54577P, FP

7-UNIT 30mA TRANSISTOR ARRAY

DESCRIPTION

The M54577P, FP, 7-channel sink driver, consists of 14 NPN transistors connected to form high current gain driver pairs.

FEATURES

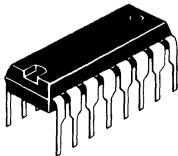
- Output breakdown voltage to 30V
- Output sink current to 30mA
- PMOS, CMOS Compatible input
- Low output saturation voltage
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

LED or incandescent display digit driver

FUNCTION

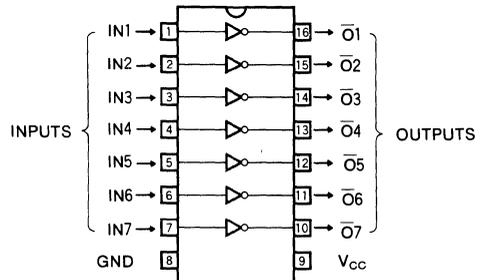
The M54577P, FP uses a predriver stage with a diode and $23\text{k}\Omega$ resistor in series to the input. The power supply of the predrivers is connected to pin 9. The outputs are capable of sinking 30mA and will withstand 30V in the OFF state. The M54577FP features a small flat mold package.



16-pin molded plastic DIP

16-pin molded plastic FLAT

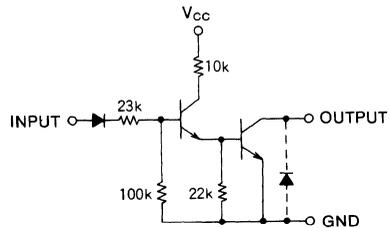
PIN CONFIGURATION (TOP VIEW)



Outline 16P2 (M54577FP)

Outline 16P4 (M54577P)

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		13	V
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +30$	V
I_C	Collector current	Transistor ON	30	mA
V_i	Input voltage		$-20 \sim +30$	V
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

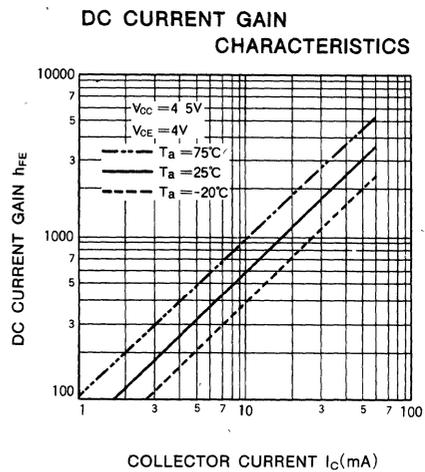
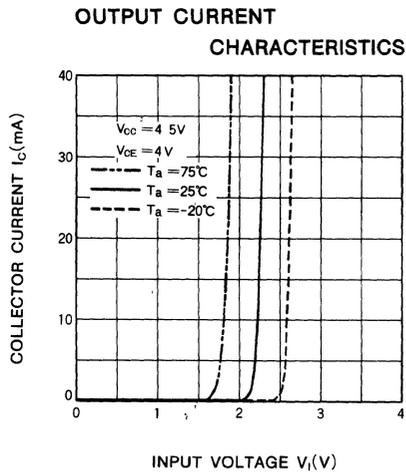
Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{CC}	Supply voltage	4.5	5	13	V
I_C	Collector current per channel	0	10	20	mA
V_{IH}	"H" Input voltage	3		V_{CC}	V
V_{IL}	"L" Input voltage	0		1	V

7-UNIT 30mA TRANSISTOR ARRAY

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{O(\text{leak})}$	Output leak current	$V_{CE} = 30\text{ V}$			100	$\mu\text{ A}$
$V_{CE(\text{sat})}$	Output saturation voltage	$V_{CC} = 4.5\text{ V}, V_i = 3\text{ V}, I_C = 10\text{ mA}$ $V_{CC} = 6\text{ V}, V_i = 3\text{ V}, I_C = 20\text{ mA}$			0.25 0.35	V
I_i	Input current	$V_{CC} = 4.5\text{ V}, V_i = 3\text{ V}$	30		90	$\mu\text{ A}$
I_{CC}	Supply current per channel (an only output conducting)	$V_{CC} = 4.5\text{ V}, V_i = 3\text{ V}$ $V_{CC} = 13\text{ V}, V_i = 3\text{ V}$		0.4 1.3	0.9 2.3	mA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{ V}, V_{CC} = 4.5\text{ V}, I_C = 20\text{ mA}, T_a = 25^\circ\text{C}$	500	1200		—

TYPICAL CHARACTERISTICS



M54537P

7-UNIT 350mA TRANSISTOR ARRAY

DESCRIPTION

The M54537P, 7-channel sink driver, consists of 14 NPN transistors connected to form high current gain driver pairs.

FEATURES

- Output breakdown voltage to 20V
- High output sink current to 250mA
- PMOS Compatible input
- Low output saturation voltage
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

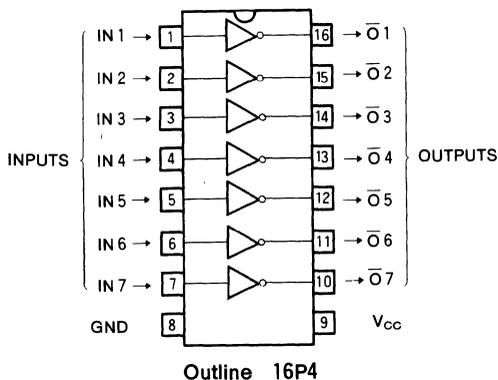
APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

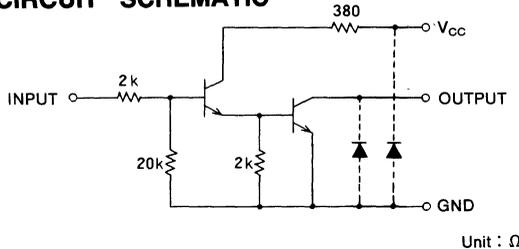
FUNCTION

The M54537P uses a predriver stage with $2k\Omega$ series input resistor. The power supply of the predrivers is connected to pin 9. All emitters and the substrate are connected together to pin 8. The outputs are capable of sinking 250mA and will withstand 20V in the OFF state.

PIN CONFIGURATION (TOP VIEW)

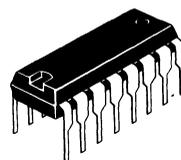


CIRCUIT SCHEMATIC



RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -25 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit	
		Min	Typ	Max		
V_{CC}	Supply voltage	3		8	V	
V_O	Output voltage	0		20	V	
I_C	Collector current per channel	$V_{CC}=6.5\text{V}$ Percent duty cycle less than 40%	0		250	mA
		$V_{CC}=6.5\text{V}$ Percent duty cycle less than 65%	0		150	
V_{IH}	"H" Input voltage	$I_C=250\text{mA}$	3		6	V
V_{IL}	"L" Input voltage	$I_O(\text{leak})=50\mu\text{A}$	0		0.3	V



16-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -25 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +20$	V
I_C	Collector current	Transistor ON	350	mA
V_I	Input voltage		10	V
P_d	Power dissipation	$T_a=25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

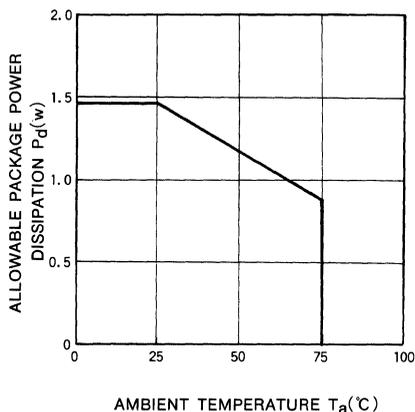
7-UNIT 350mA TRANSISTOR ARRAY

ELECTRICAL CHARACTERISTICS ($T_a = -25 \sim +75^\circ\text{C}$, unless otherwise noted)

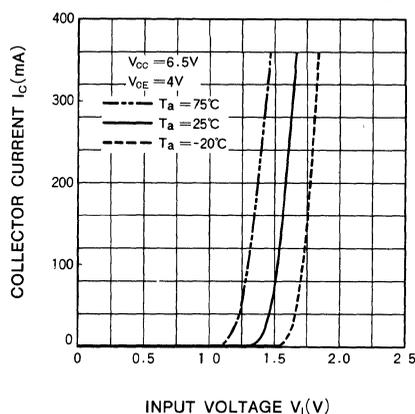
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$V_{CC}=8\text{V}, I_{CEO}=100\mu\text{A}$	20			V
$V_{CE(sat)}$	Output saturation voltage	$V_I=3\text{V}$ $V_{CC}=6.5\text{V}, I_C=250\text{mA}$ $V_{CC}=3\text{V}, I_C=150\text{mA}$		0.28 0.17	0.5 0.35	V
I_i	Input current	$V_{CC}=8\text{V}, V_I=3.2\text{V}$ $V_{CC}=8\text{V}, V_I=10\text{V}$		0.7 3.8	1.5 7.3	mA
I_{OC}	Supply current	$V_{CC}=8\text{V}, V_I=3.2\text{V}$		130	190	mA
h_{FE}	DC forward current gain	$V_{CE}=4\text{V}, V_{CC}=6.5\text{V}, I_C=250\text{mA}, T_a=25^\circ\text{C}$	1000	7000		

TYPICAL CHARACTERISTICS

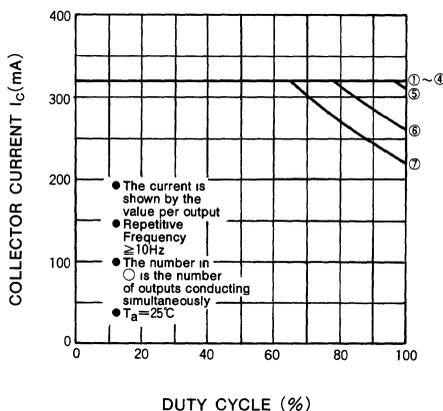
ALLOWABLE AVERAGE



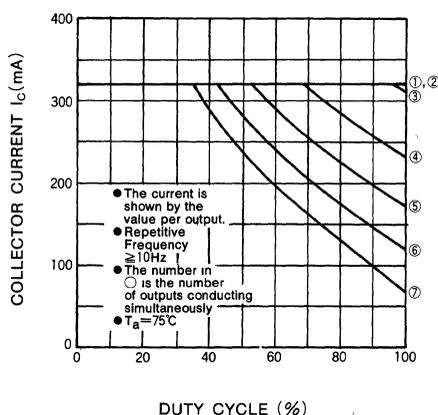
OUTPUT CURRENT CHARACTERISTICS

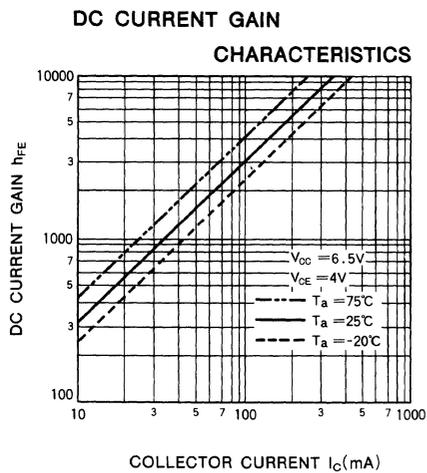


ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE





M54535P

7-UNIT 150mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

DESCRIPTION

The M54535P, 7-channel sink driver, consists of 14 NPN transistors connected to form high current gain driver pairs.

FEATURES

- Output sink current to 150mA
- Strobe input control
- PMOS Compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

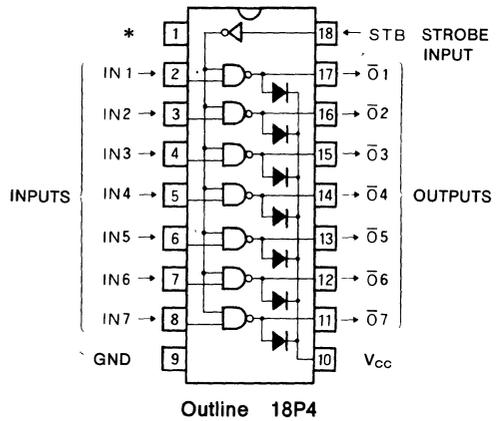
FUNCTION

The M54535P uses a predriver stage. Each input has a diode and $20\text{k}\Omega$ resistor in series to allow a negative voltage input. All inputs can be controlled simultaneously by a strobe input at pin 18. Each output has an integral diode for inductive load transient suppression.

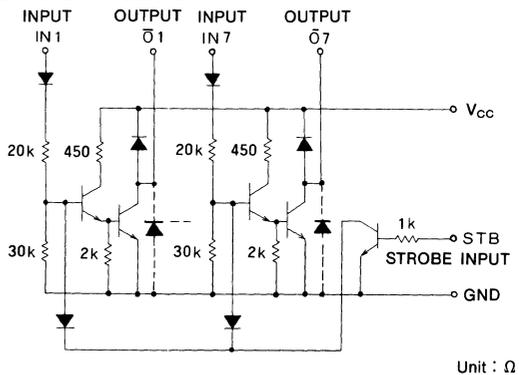
The cathodes of the diodes and the power supply of the predrivers are connected to pin 10. All emitters and the substrate are connected together to pin 9.

The outputs are capable of sinking 150mA and will withstand 10V in the OFF state.

PIN CONFIGURATION (TOP VIEW)

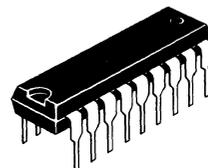


CIRCUIT SCHEMATIC



FUNCTIONAL TABLE

IN	STB	OUT
L	L	H
H	L	L
L	H	H
H	H	H



18-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_O	Output voltage	Transistor OFF	$-0.5 \sim V_{CC}$	V
I_C	Collector current	Transistor ON	150	mA
V_I	Input voltage		$-25 \sim +20$	V
$V_{I(STB)}$	Strobe input voltage		20	V
$V_{R(D)}$	Clamp diode reverse voltage		10	V
$I_{F(D)}$	Clamp diode forward current		150	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

7-UNIT 150mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

RECOMMENDED OPERATIONAL CONDITIONS (T_a = -20~+75°C, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage		3		8	V
V _O	Output voltage	V _{CC} =10V	0		10	V
I _C	Collector current per channel	Percent duty cycle less than 65%	0		150	mA
V _{IH}	"H" Input voltage	I _C =100mA	7		18	V
V _{IL}	"L" Input voltage		0		0.8	V
V _{IH(STB)}	"H" Input voltage (strobe input)	V _I =12V	1.3		6	V
		V _I =20V	2.4		6	
V _{IL(STB)}	"L" Input voltage (strobe input)		0		0.2	V

ELECTRICAL CHARACTERISTICS (T_a = -20~+75°C, unless otherwise noted)

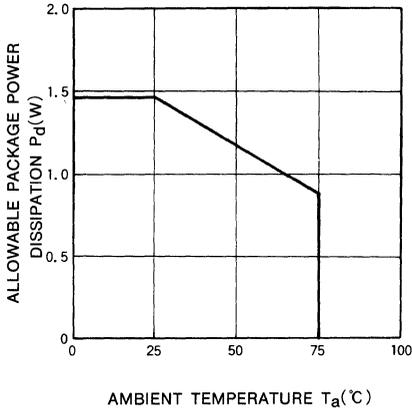
Symbol	Parameter	Test conditions	Limits			Unit	
			Min	Typ*	Max		
V _{CE(sat)}	Output saturation voltage	V _{I(STB)} =0.2V V _{CC} =6V, I _I =300μA, I _C =100mA		0.1	0.3	V	
		V _I =7V V _{I(STB)} =0.2V	V _{CC} =3V I _C =100mA		0.1		0.3
			V _{CC} =8V I _C =150mA		0.16		0.5
I _{O(leak)}	Output leak current	V _{CC} =8V, V _I =0.8V, V _O =8V V _{I(STB)} =0.2V			50	μA	
I _I	Input current	V _{CC} =8V, V _I =12V V _{I(STB)} =0.2V		0.5	1	mA	
I _R	Input leakage current	V _{CC} =8V, V _I =-25V			-20	μA	
I _{I(STB)}	Strobe input current	V _I =12V, V _{I(STB)} =2.4V		0.6	3	mA	
V _{F(D)}	Clamp diode forward voltage	I _{F(D)} =150mA		1.1	2.1	V	
V _{R(D)}	Clamp diode reverse voltage	I _{R(D)} =100μA	10			V	
I _{CC}	Supply current	V _{CC} =8V, V _I =12V all input V _{I(STB)} =0.2V		120	200	mA	
h _{FE}	DC forward current gain	V _{CE} =4V, V _{CC} =6V, I _C =150mA, T _a =25°C	700	2500		—	

* : The typical value is at T_a=25°C

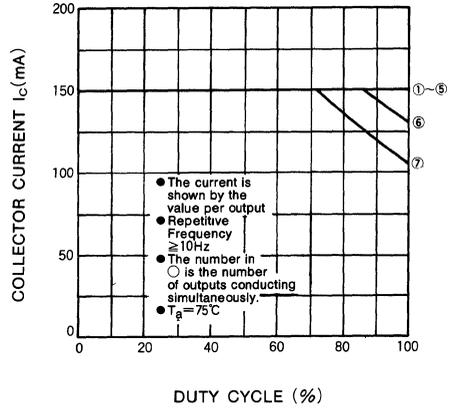
7-UNIT 150mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

TYPICAL CHARACTERISTICS

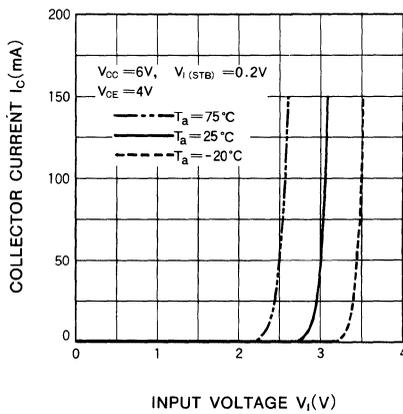
ALLOWABLE AVERAGE POWER DISSIPATION



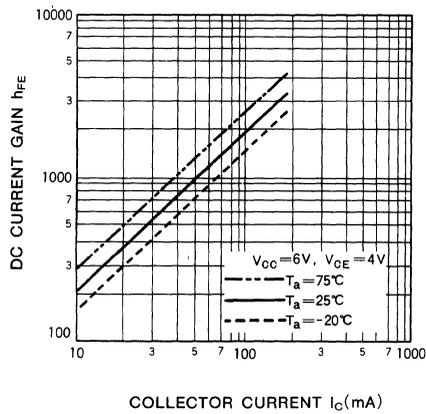
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



OUTPUT CURRENT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54536P

7-UNIT 150mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

DESCRIPTION

The M54536P, 7-channel sink driver, consists of 14 NPN transistors connected to form high current gain driver pairs.

FEATURES

- Output sink current to 150mA
- Strobe input control
- TTL Compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

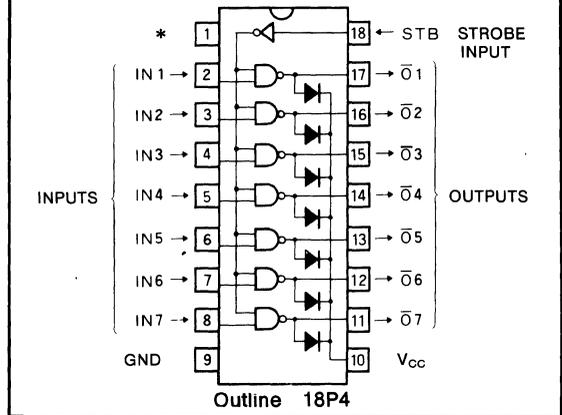
FUNCTION

The M54536P uses a predriver stage. Each input has a diode and $20\text{k}\Omega$ resistor in series to allow a negative voltage input. All inputs can be controlled simultaneously by a strobe input at pin 18. Each output has an integral diode for inductive load transient suppression.

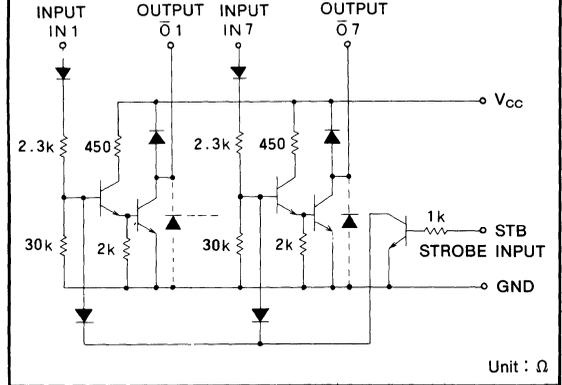
The cathodes of the diodes and the power supply of the predrivers are connected to pin 10. All emitters and the substrate are connected together to pin 9.

The outputs are capable of sinking 150mA and will withstand 10V in the OFF state.

PIN CONFIGURATION (TOP VIEW)

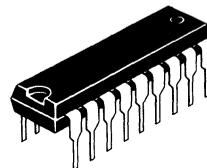


CIRCUIT SCHEMATIC



FUNCTION TABLE

IN	STB	OUT
L	L	H
H	L	L
L	H	H
H	H	H



18-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -25 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_O	Output voltage	Transistor OFF	$-0.5 \sim V_{CC}$	V
I_C	Collector current	Transistor ON	150	mA
V_I	Input voltage		$-25 \sim +10$	V
$V_{I(STB)}$	Strobe input voltage		20	V
$V_{R(D)}$	Clamp diode reverse voltage		10	V
$I_{F(D)}$	Clamp diode forward current		150	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

7-UNIT 150mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -25 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		3		8	V
V_O	Output voltage	$V_{CC}=10\text{V}$	0		10	V
I_C	Collector current per channel	Percent duty cycle less than 65%	0		150	mA
V_{IH}	"H" Input voltage	$I_C=100\text{mA}$	3.2		6	V
V_{IL}	"L" Input voltage		0		0.8	V
$V_{IH(STB)}$	"H" Input voltage (strobe input)	$V_I=3.5\text{V}$ $V_I=10\text{V}$	1.3		6	V
$V_{IL(STB)}$	"L" Input voltage (strobe input)		0		0.2	V

ELECTRICAL CHARACTERISTICS ($T_a = -25 \sim +75^\circ\text{C}$, unless otherwise noted)

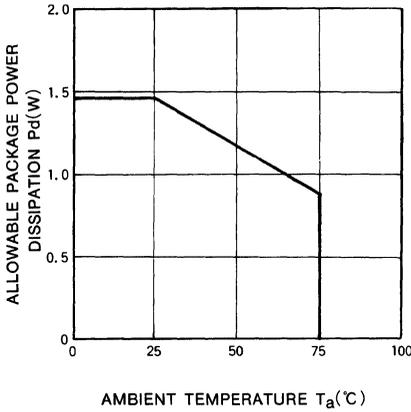
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$V_{CE(sat)}$	Output saturation voltage	$V_{I(STB)}=0.2\text{V}$ $V_{CC}=6\text{V}$, $I_I=300\mu\text{A}$, $I_C=100\text{mA}$		0.1	0.3	V
		$V_I=3.2\text{V}$ $V_{I(STB)}=0.2\text{V}$		0.1	0.3	
		$V_{CC}=3\text{V}$ $I_C=100\text{mA}$ $V_{CC}=8\text{V}$ $I_C=150\text{mA}$		0.16	0.5	
$I_{O(leak)}$	Output leak current	$V_{CC}=8\text{V}$, $V_I=0.8\text{V}$, $V_O=8\text{V}$ $V_{I(STB)}=0.2\text{V}$			50	μA
I_I	Input current	$V_{CC}=8\text{V}$, $V_I=3.5\text{V}$ $V_{I(STB)}=0.2\text{V}$		0.6	1.2	mA
I_R	Input leakage current	$V_{CC}=8\text{V}$, $V_I=-25\text{V}$			-20	μA
$I_{I(STB)}$	Strobe input current	$V_I=3.5\text{V}$, $V_{I(STB)}=2.4\text{V}$		0.9	3	mA
$V_{F(D)}$	Clamp diode forward voltage	$I_{R(D)}=150\text{mA}$		1.1	2.1	V
$V_{R(D)}$	Clamp diode reverse voltage	$I_{R(D)}=100\mu\text{A}$	10			V
I_{CC}	Supply current	$V_{CC}=8\text{V}$, $V_I=3.5\text{V}$ (all input) $V_{I(STB)}=0.2\text{V}$		120	200	mA
h_{FE}	DC forward current gain	$V_{CE}=4\text{V}$, $V_{CC}=6\text{V}$, $I_C=150\text{mA}$, $T_a=25^\circ\text{C}$	700	3000		—

* : A typical value is at $T_a=25^\circ\text{C}$.

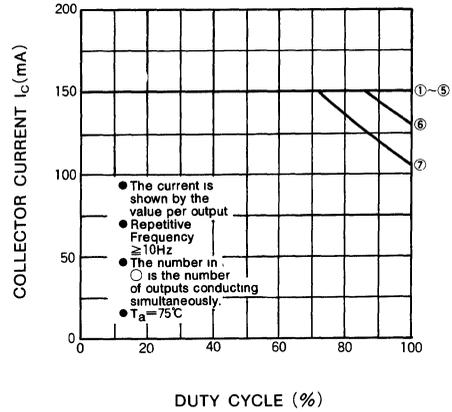
7-UNIT 150mA TRANSISTOR ARRAY WITH CLAMP DIODE AND STROBE

TYPICAL CHARACTERISTICS

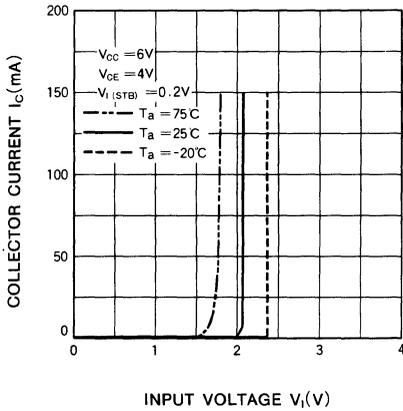
ALLOWABLE AVERAGE POWER DISSIPATION



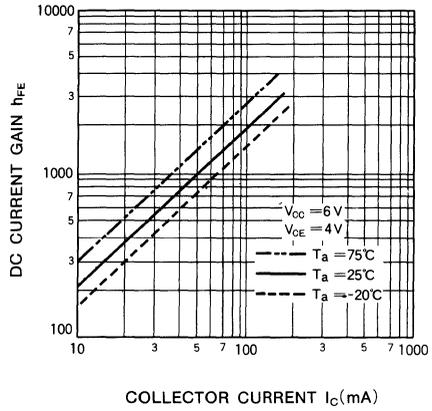
ALLOWABLE COLLECT CURRENT AS A FUNCTION OF DUTY CYCLE



OUTPUT CURRENT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54517P

7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54517P, 7-channel sink driver, consists of 14 NPN transistors connected to form seven high current gain driver pairs.

FEATURES

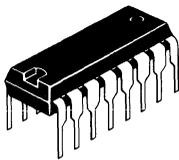
- Output sustaining voltage to 25V
- High output sink current to 400mA
- PMOS Compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

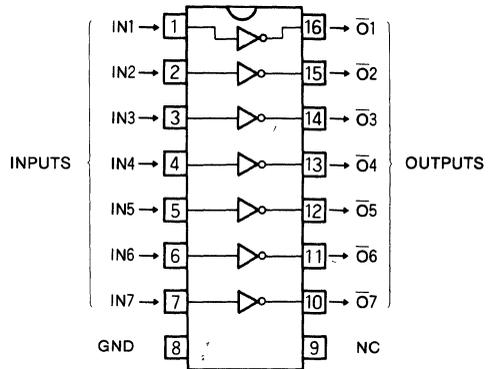
FUNCTION

The M54517P is comprised of seven NPN darlington driver pairs with $20\text{k}\Omega$ series input resistors. All emitters and the substrate are connected to pin 8. The output are capable of sinking 400mA and will withstand 25V in the OFF state.



16-pin molded plastic DIP

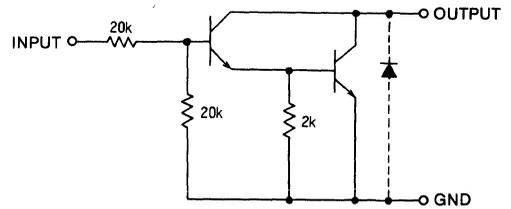
PIN CONFIGURATION (TOP VIEW)



Outline 16P4

NC : NO CONNECTION

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +25$	V
I_C	Collector current	Transistor ON	400	mA
V_I	Input voltage		25	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

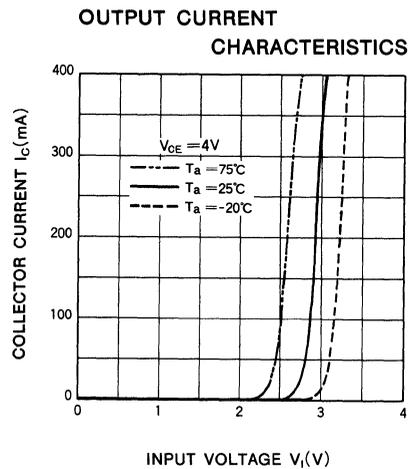
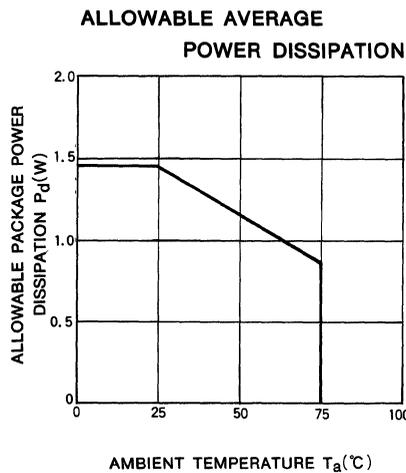
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		25	V
I_C	Collector current per channel	Percent duty cycle less than 8%	0		400	mA
		Percent duty cycle less than 40%	0		200	
V_{IH}	"H" Input voltage	$I_C = 400\text{mA}$	8		20	V
		$I_C = 100\text{mA}$	5		20	
V_{IL}	"L" Input voltage	$I_{oleak} = 50\mu\text{A}$	0		0.5	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CEO} = 100\mu\text{A}$	25			V
$V_{CE(sat)}$	Output saturation voltage	$V_i = 8\text{V}, I_C = 400\text{mA}$		1.15	2.2	V
		$V_i = 5\text{V}, I_C = 200\text{mA}$		0.95	1.4	
I_i	Input current	$V_i = 17\text{V}$		0.8	1.8	mA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 400\text{mA}, T_a = 25^\circ\text{C}$	1000	4500		—

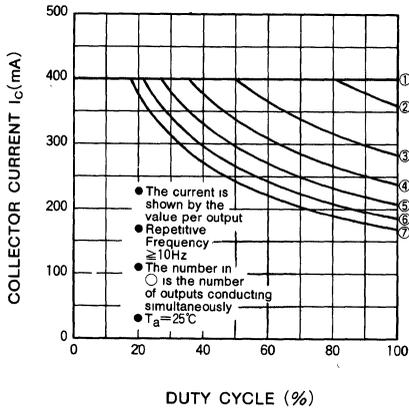
* : A typical value is at $T_a = 25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

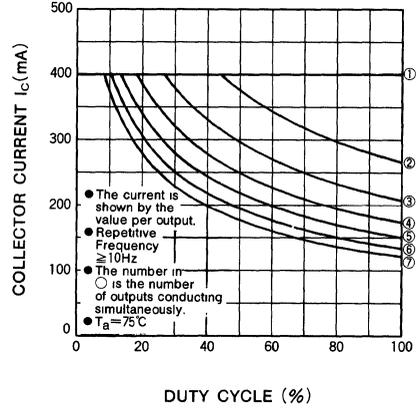


7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

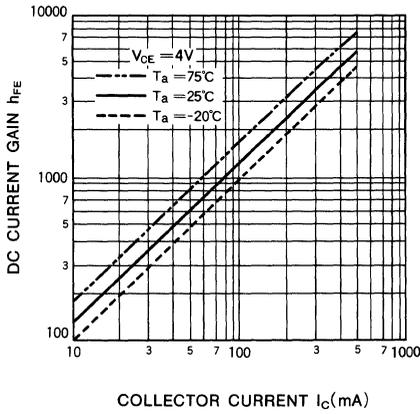
**ALLOWABLE COLLECTOR CURRENT
AS A FUNCTION OF DUTY CYCLE**



**ALLOWABLE COLLECTOR CURRENT
AS A FUNCTION OF DUTY CYCLE**



**DC CURRENT GAIN
CHARACTERISTICS**



M54519P

7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54519P, 7-channel sink driver, consists of 14 NPN transistors connected to form seven high current gain driver pairs.

FEATURES

- High output sustaining voltage to 40V
- High output sink current to 400mA
- PMOS Compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

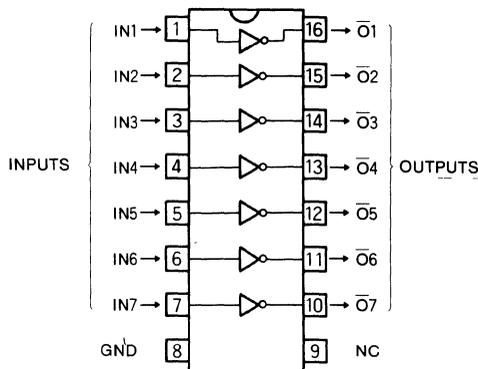
Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

FUNCTION

The M54519P is comprised of seven NPN darlington driver pairs with $20\text{k}\Omega$ series input resistors.

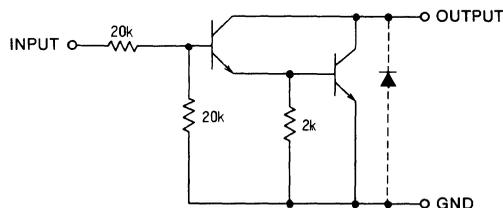
All emitters and the substrate are connected together to pin 8. The output are capable of sinking 400mA and will withstand 40V in the OFF state.

PIN CONFIGURATION (TOP VIEW)

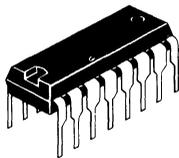


NC : NO CONNECTION

CIRCUIT SCHEMATIC



Unit : Ω



16-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CE0}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +40$	V
I_C	Collector current	Transistor ON	400	mA
V_i	Input voltage		40	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

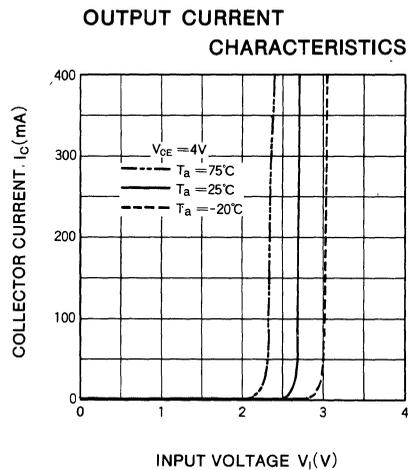
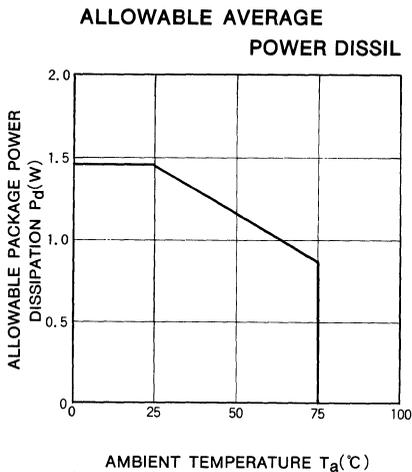
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		40	V
I_C	Collector current per channel	Percent duty cycle less than 8%	0		400	mA
		Percent duty cycle less than 30%	0		200	
V_{IH}	"H" Input voltage	$I_C = 400\text{mA}$	8		30	V
		$I_C = 100\text{mA}$	5		30	
V_{IL}	"L" Input voltage	$I_{O(leak)} = 50\mu\text{A}$	0		0.5	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CEO} = 100\mu\text{A}$	40			V
$V_{CE(sat)}$	Output saturation voltage	$V_I = 8\text{V}, I_C = 400\text{mA}$		1.2	2.4	V
		$V_I = 5\text{V}, I_C = 200\text{mA}$		0.9	1.6	
I_I	Input current	$V_I = 17\text{V}$	0.3	0.8	1.8	mA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 400\text{mA}, T_a = 25^\circ\text{C}$	1000	6000		—

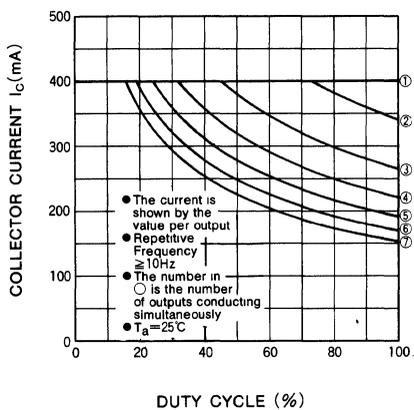
* : A typical value is at $T_a = 25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

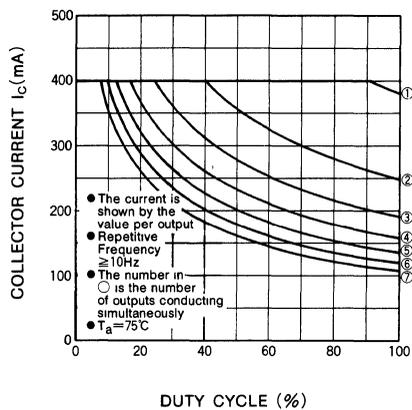


7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

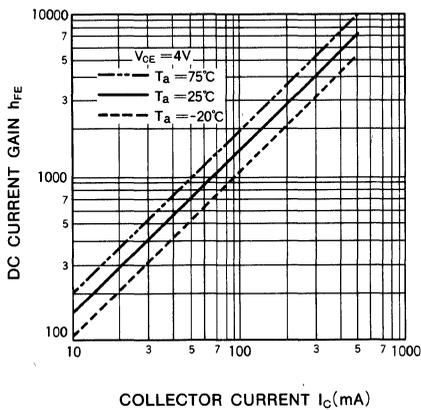
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54528P

7-UNIT 150mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54528P, 7-channel sink driver, consists of 14 NPN transistors connected to form high current gain driver pairs.

FEATURES

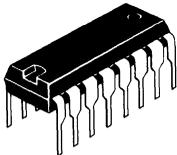
- High output sustaining voltage to 40V
- Output sink current to 150mA
- Efficient I/O pin layout
- PMOS compatible input
- Integral diodes for transient suppression
- Wide input voltage range from $-40V$ to $+40V$
- Wide operating temperature range ($T_a = -20 \sim +75^\circ C$)

APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

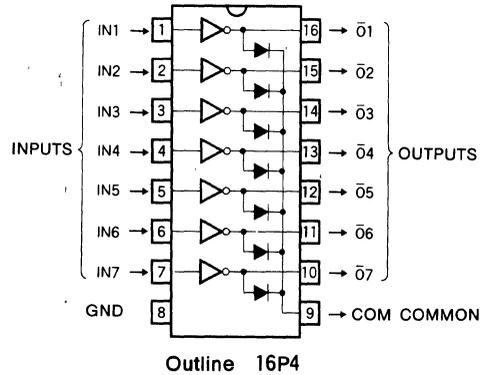
FUNCTION

The M54528P is comprised of seven darlington driver pairs. Each input has a diode and $20k\Omega$ resistor in series to allow a negative voltage input. Between pin 9 and each output, there are integral diodes for inductive load transient suppression. All emitters and the substrate are connected together to pin 8. The outputs are capable of sinking 150mA and will withstand 40V in the OFF state.

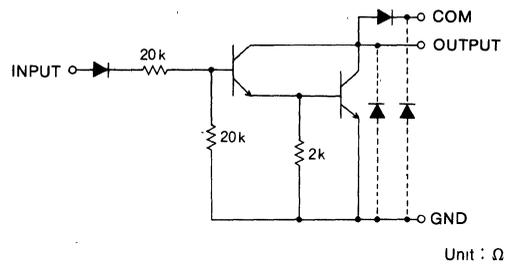


16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ C$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +40$	V
I_C	Collector current	Transistor ON	150	mA
V_i	Input voltage		-40	V
			40	V
I_F	Clamp diode forward current		150	mA
V_R	Clamp diode reverse voltage		40	V
P_d	Power dissipation	$T_a = 25^\circ C$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ C$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ C$

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ C$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		40	V
I_C	Collector current per channel	Percent duty cycle less than 40%	0		150	mA
V_{IH}	"H" Input voltage	$I_C = 150mA$	7		35	V
V_{iL}	"L" Input voltage	$I_{oleak} = 50\mu A$	0		1	V

7-UNIT 150mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

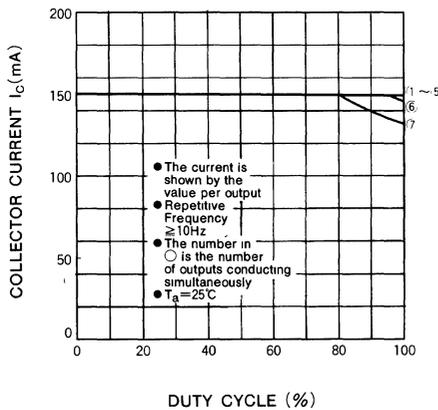
ELECTRICAL CHARACTERISTICS (T_a = -20~+75°C, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
V _{(BR)CEO}	Output sustaining voltage	I _{CEO} = 100μA	40			V
V _{CE(sat)}	Output saturation voltage	V _I = 7V, I _C = 150mA		1.4	1.7	V
		V _I = 7V, I _C = 100mA		1.2	1.4	
I _I	Input current	V _I = 18V		0.9	1.8	mA
		V _I = 35V		1.9	5	
I _R	Input leakage current	V _I = -35V			-20	μA
V _{F(D)}	Clamp diode forward voltage	I _{F(D)} = 150mA		1.15	1.6	V
I _{R(D)}	Clamp diode leakage current	V _{R(D)} = 40V			100	μA
h _{FE}	DC forward current gain	V _{CE} = 4V, I _C = 150mA, T _a = 25°C	800	2500		—

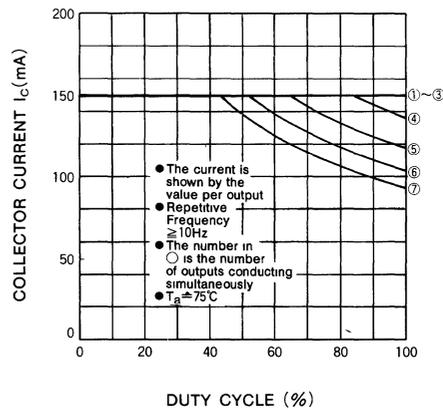
* : All typical values are at T_a = 25°C

TYPICAL CHARACTERISTICS

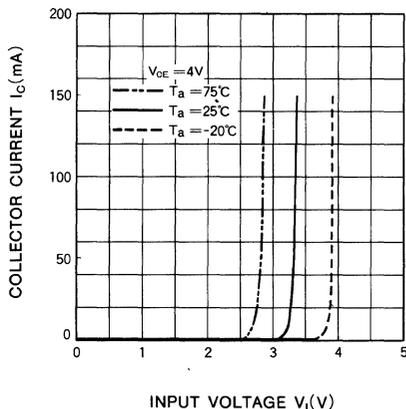
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



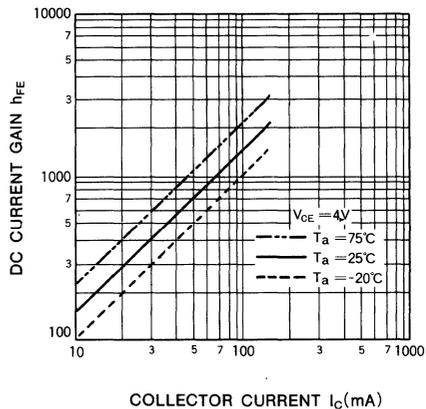
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



OUTPUT CURRENT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54523P

7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54523P, 7-channel sink driver, consists of 14 NPN transistors connected to form seven high current gain driver pairs.

FEATURES

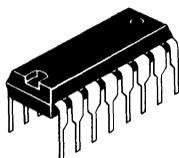
- High output sustaining voltage to 50V
- High output sink current to 500mA
- Integral diodes for transient suppression
- PMOS Compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics.

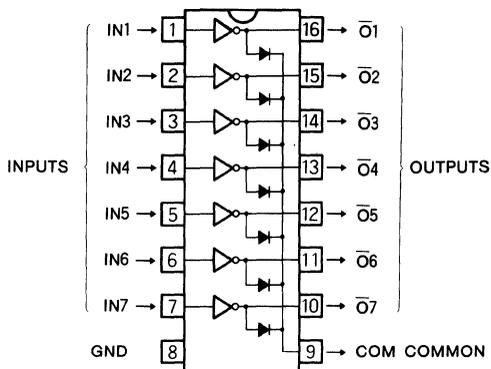
FUNCTION

The M54523P is comprised of seven NPN darlington driver pairs with $2.7\text{k}\Omega$ series input resistors. Between pin 9 and each output, there are integral diodes for inductive load transient suppression. All emitters and the substrate are connected together to pin 8. The outputs are capable of sinking 500mA and will withstand 50V in the OFF state.



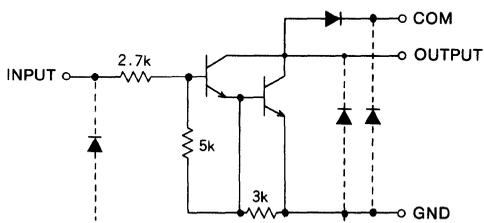
16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



Outline 16P4

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +50$	V
I_C	Collector current	Transistor ON	500	mA
V_I	Input voltage		30	V
I_F	Clamp diode forward current		500	mA
V_R	Clamp diode reverse voltage		50	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

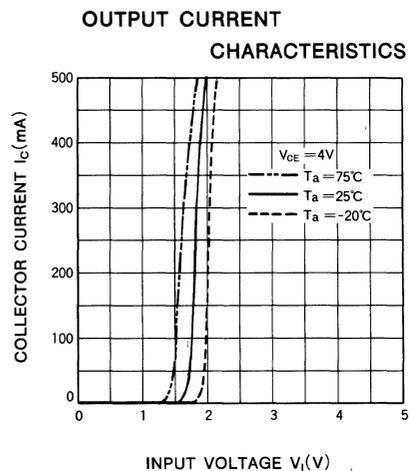
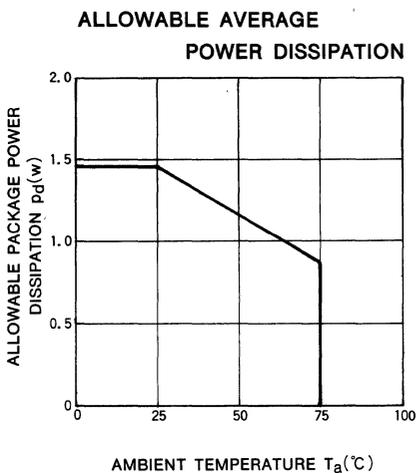
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		50	V
I_C	Collector current per channel	Percent duty cycle less than 8%	0		400	mA
		Percent duty cycle less than 30%	0		200	
V_{IH}	"H" Input voltage	$I_C = 400\text{mA}$	3.85		25	V
		$I_C = 100\text{mA}$	3.4		25	V
V_{IL}	"L" Input voltage		0		0.5	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$I_{O(leak)}$	Output leakage current	$V_{CE} = 50\text{V}$			100	μA
$V_{CE(sat)}$	Output saturation voltage	$V_I = 3.85\text{V}, I_C = 400\text{mA}$		1.3	2.4	V
		$V_I = 3.85\text{V}, I_C = 200\text{mA}$		0.95	1.6	
I_i	Input current	$V_I = 3.85\text{V}$		0.95	1.8	mA
		$V_I = 25\text{V}$		9	18	
V_F	Clamp diode forward voltage	$I_F = 400\text{mA}$		1.5	2.4	V
I_R	Clamp diode leakage voltage	$V_R = 50\text{V}$			100	μA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 350\text{mA}, T_a = 25^\circ\text{C}$	1000	2500		—

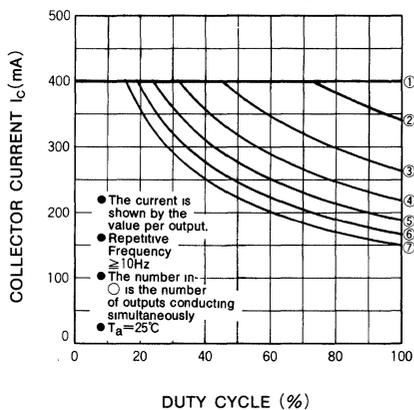
* : All typical values are at $T_a = 25^\circ\text{C}$

TYPICAL CHARACTERISTICS

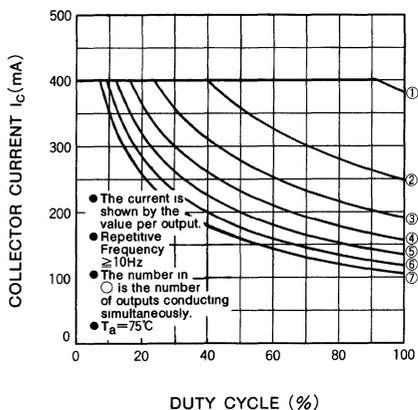


7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

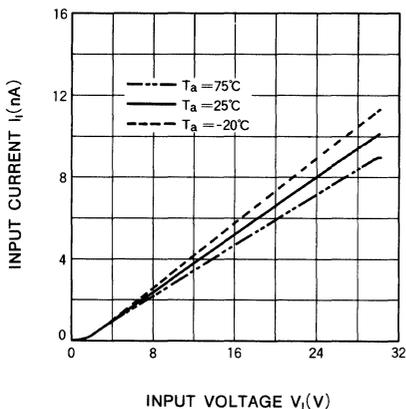
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



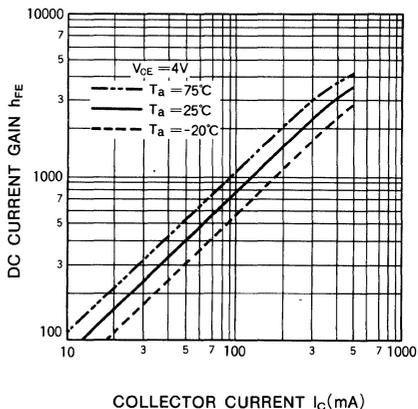
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



INPUT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54524P

7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54524P, 7-channel sink driver, consists of 14 NPN transistors connected to form high current gain driver pairs.

FEATURES

- High output sustaining voltage to 50V
- High output sink current to 500mA
- Integral diodes for transient suppression
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

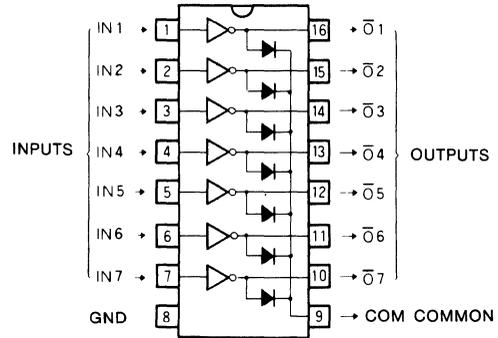
APPLICATION

Relay and printer drivers, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

FUNCTION

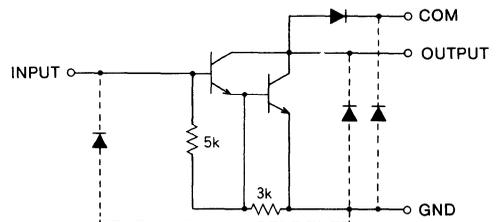
The M54524P is comprised of seven NPN darlington driver pairs. Between pin 9 and each output, there are integral diodes for inductive load transient suppression. All emitters and the substrate are connected together to pin 8. The outputs are capable of sinking 500mA and will withstand 50V in the OFF state.

PIN CONFIGURATION (TOP VIEW)

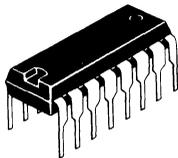


Outline 16P4

CIRCUIT SCHEMATIC



Unit : Ω



16-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +50	V
I_C	Collector current	Transistor ON	500	mA
I_F	Clamp diode forward current		500	mA
V_R	Clamp diode reverse voltage		50	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

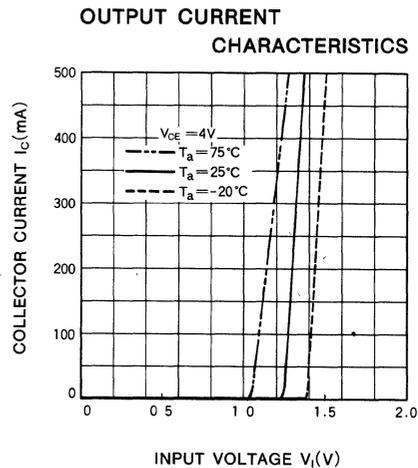
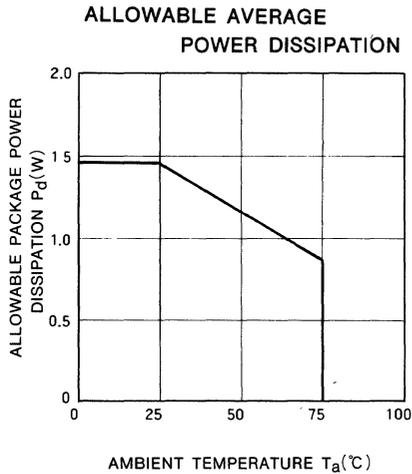
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		50	V
I_C	Collector current per channel	Percent duty cycle less than 8%	0		400	mA
		Percent duty cycle less than 30%	0		200	
I_{IH}	"H" Input current	$I_C = 400\text{mA}$	1		20	mA
I_{IL}	"L" Input current		0		20	μA

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$I_{O(Leak)}$	Output leakage current	$V_{CE} = 50\text{V}$			100	μA
$V_{CE(sat)}$	Output saturation voltage	$I_I = 1\text{mA}, I_C = 400\text{mA}$		1.3	2.4	V
		$I_I = 1\text{mA}, I_C = 200\text{mA}$		0.95	1.6	
V_I	Input voltage	$I_I = 1\text{mA}$		1.35	1.7	V
V_F	Clamp diode forward voltage	$I_F = 400\text{mA}$		1.5	2.4	V
I_R	Clamp diode leakage current	$V_R = 50\text{V}$			100	μA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 350\text{mA}, T_a = 25^\circ\text{C}$	1000	2500		—

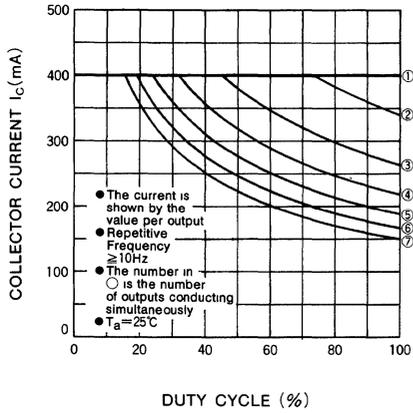
* : All typical values are at $T_a = 25^\circ\text{C}$

TYPICAL CHARACTERISTICS

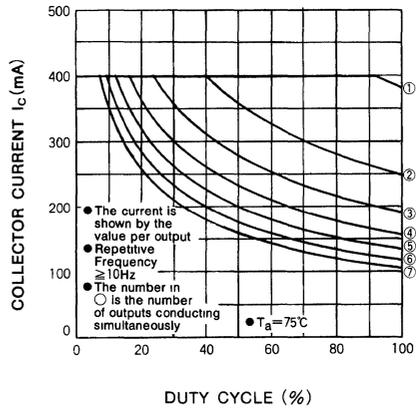


7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

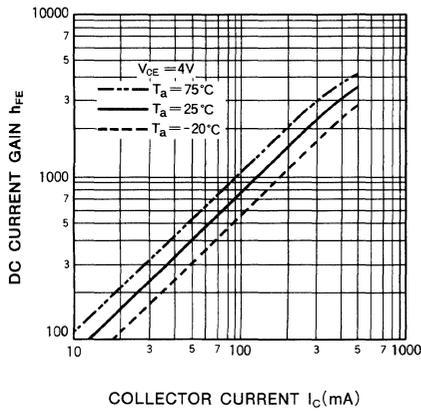
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54525P

7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54525P, 7-channel sink driver, consists of 14 NPN transistors connected to form high current gain driver pairs.

FEATURES

- High output sustaining voltage to 50V
- High output sink current to 500mA
- Integral diodes for transient suppression
- 24V PMOS compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

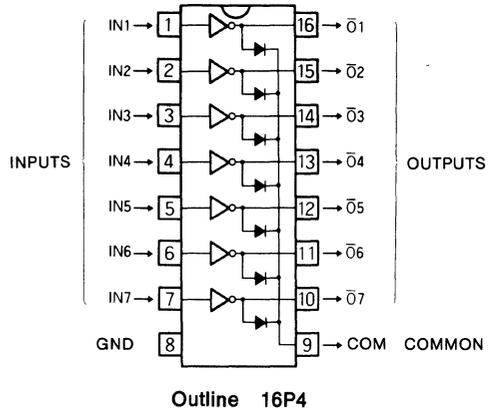
APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

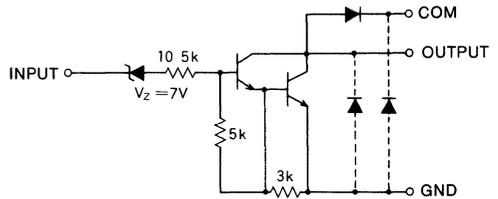
FUNCTION

The M54525P is comprised of seven NPN darlington driver pairs. Each input has a Zener diode and 10.5kΩ resistor in series to limit the input current. Between pin 9 and each output, there are integral diodes for inductive load transient suppression. All emitters and the substrate are connected together to pin 8. The outputs are capable of sinking 500mA and will withstand 50V in the OFF state.

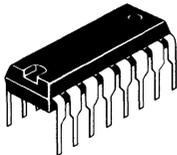
PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



Unit : Ω



16-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +50	V
I_C	Collector current	Transistor ON	500	mA
V_i	Input voltage		30	V
I_F	Clamp diode forward current		500	mA
V_R	Clamp diode reverse voltage		50	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

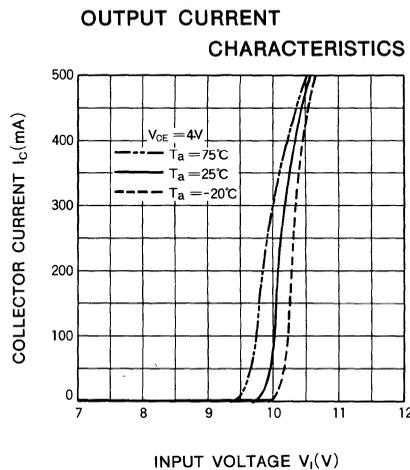
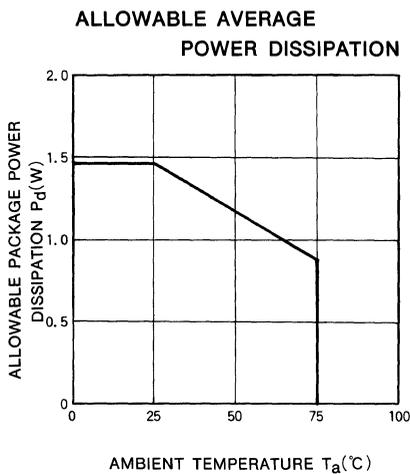
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		50	V
I_C	Collector current per channel	Percent duty cycle less than 8%	0		400	mA
		Percent duty cycle less than 30%	0		200	
V_{IH}	"H" Input voltage	$I_C = 400\text{mA}$	17		25	V
V_{IL}	"L" Input voltage		0		6	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$I_{O(leak)}$	Input leakage current	$V_{CE} = 50\text{V}$ $I_i = 0\text{mA}$ $V_i = 6\text{V}$			100 500	μA
$V_{CE(sat)}$	Output saturation voltage	$V_i = 17\text{V}, I_C = 400\text{mA}$		1.3	2.4	V
		$V_i = 17\text{V}, I_C = 200\text{mA}$		0.95	1.6	
I_i	Input current	$V_i = 17\text{V}$		0.85	1.8	mA
		$V_i = 25\text{V}$		1.6	3.2	
V_F	Clamp diode forward voltage	$I_F = 400\text{mA}$		1.5	2.4	V
I_R	Clamp diode leakage current	$V_R = 50\text{V}$			100	μA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 350\text{mA}, T_a = 25^\circ\text{C}$	1000	2500		—

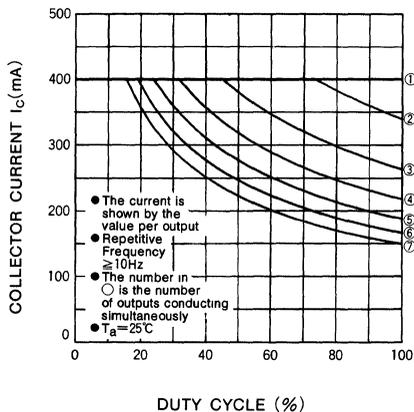
* : All typical values are at $T_a = 25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

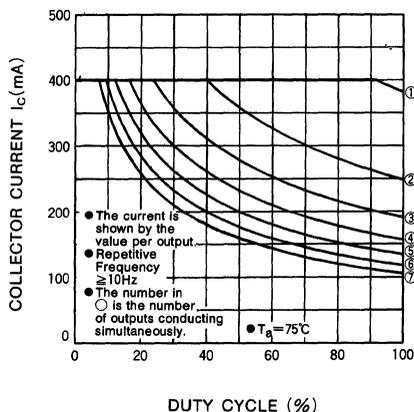


7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

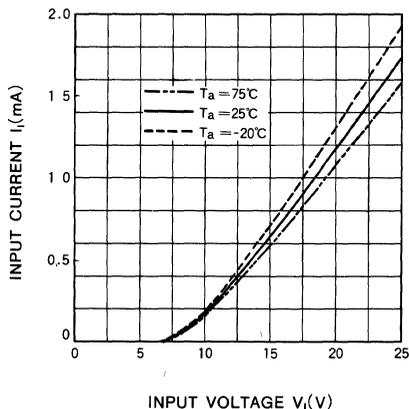
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



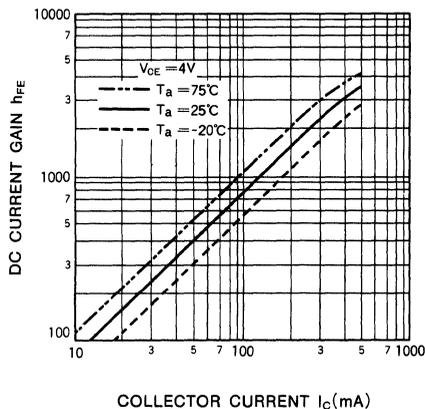
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



INPUT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54526P

7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54526P, 7-channel sink driver, consists of 14 NPN transistors connected to form high current gain driver pairs.

FEATURES

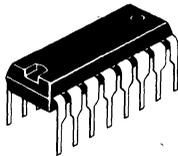
- High output sustaining voltage to 50V
- High output sink current to 500mA
- Integral diodes for transient suppression
- PMOS compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED or incandescent display digit driver, interfacing for standard MOS/BIPOLAR logics

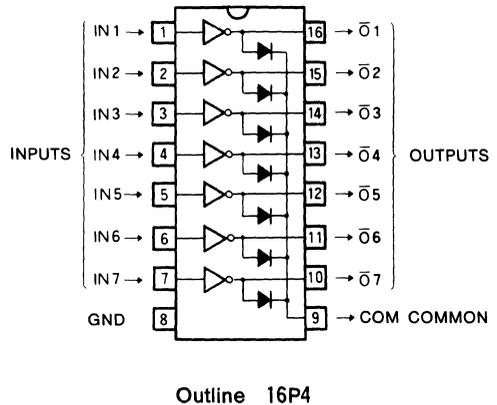
FUNCTION

The M54526P is comprised of seven darlington driver pairs with 10.5kΩ series input resistors. Between pin 9 and each output, there are integral diodes for inductive load transient suppression. All emitters are connected together to pin 8. The outputs are capable of sinking 500mA and will withstand 50V in the OFF state.

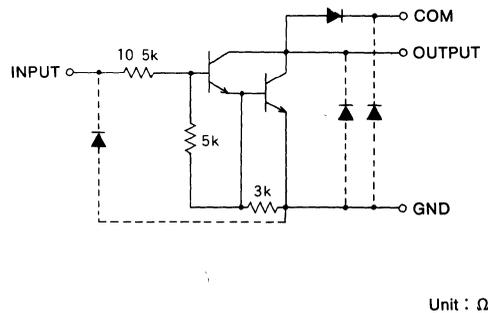


16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CE0}	Output sustaining voltage	Transistor OFF	-0.5~+50	V
I_C	Collector current	Transistor ON	500	mA
V_i	Input voltage		30	V
I_F	Clamp diode forward current		500	mA
V_R	Clamp diode reverse voltage		50	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

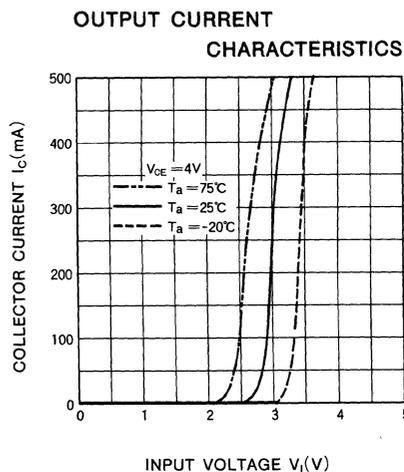
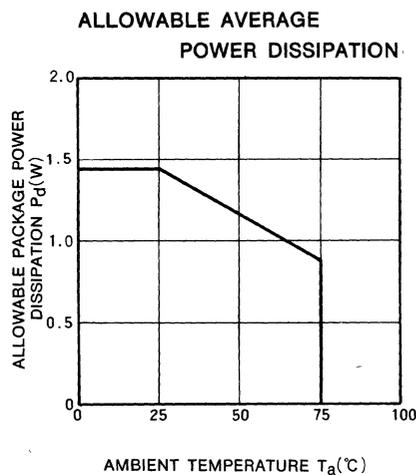
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		50	V
I_C	Collector current per channel	Percent duty cycle less than 8%	0		400	mA
		Percent duty cycle less than 30%	0		200	
V_{IH}	"H" Input voltage	$I_C = 400\text{mA}$	8	10	25	V
V_{IL}	"L" Input voltage		0		0.5	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$I_{O(leak)}$	Output leakage current	$V_{CE} = 50\text{V}$			100	μA
$V_{CE(sat)}$	Output saturation voltage	$V_I = 8\text{V}, I_C = 400\text{mA}$		1.3	2.4	V
		$V_I = 8\text{V}, I_C = 200\text{mA}$		0.95	1.6	
I_I	Input current	$V_I = 10\text{V}$		0.9	1.5	mA
		$V_I = 25\text{V}$		2.8	4.1	
V_F	Clamp diode forward voltage	$I_F = 400\text{mA}$		1.5	2.4	V
I_R	Clamp diode leakage current	$V_R = 50\text{V}$			100	μA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 350\text{mA}, T_a = 25^\circ\text{C}$	1000	2500		—

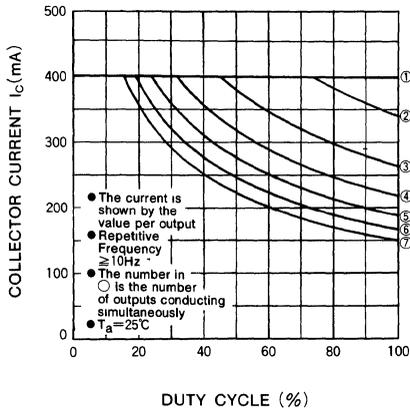
* : All typical values are at $T_a = 25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

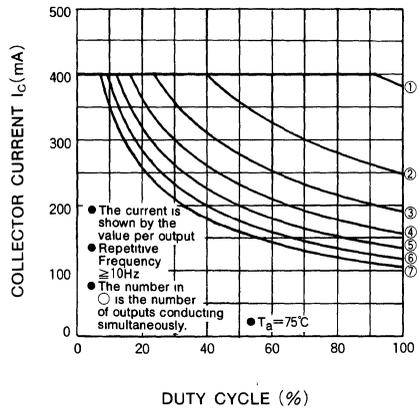


7-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

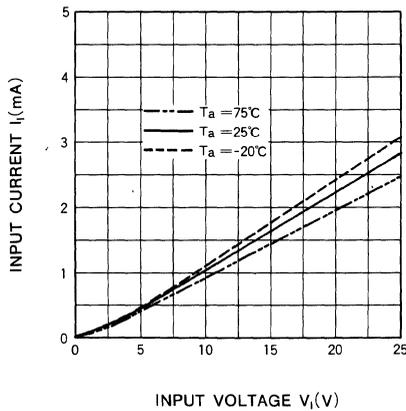
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



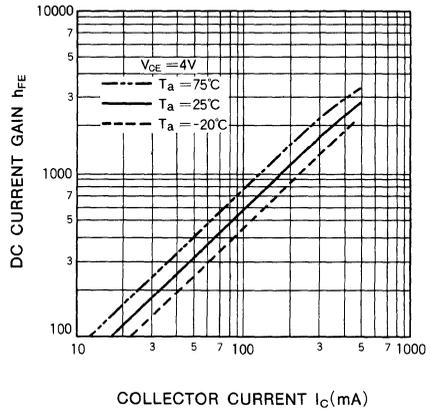
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



INPUT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54530P

7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54530P, 7-channel sink driver, consists of 14 NPN transistors connected to form seven high current gain driver pairs.

FEATURES

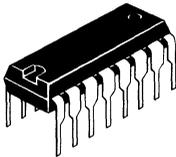
- High output sustaining voltage to 40V
- High output sink current to 400mA
- Integral diodes for transient suppression
- PMOS compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

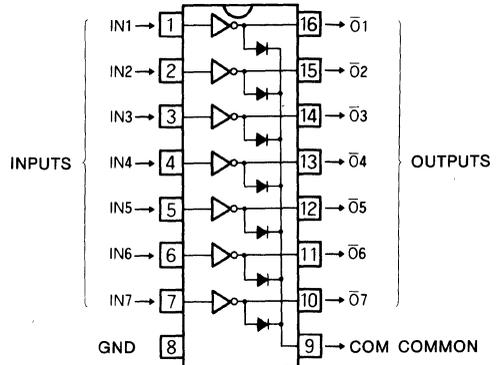
FUNCTION

The M54530P is comprised of seven NPN darlington driver pairs with $20\text{k}\Omega$ series input resistors. Between pin 9 and each output, there are integral diodes for inductive load transient suppression. All emitters and the substrate are connected together to pin 8. The outputs are capable of sinking 400mA and will withstand 40V in the OFF state.



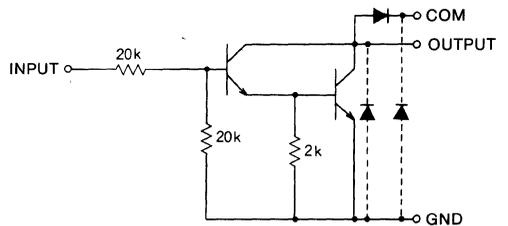
16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



Outline 16P4

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	$-50 \sim +40$	V
I_C	Collector current	Transistor ON	400	mA
V_i	Input voltage		40	V
I_F	Clamp diode forward current		400	mA
V_R	Clamp diode reverse voltage		40	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

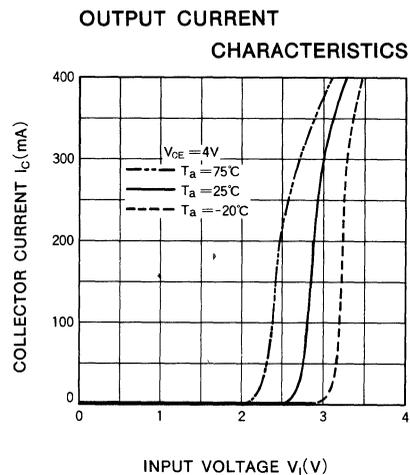
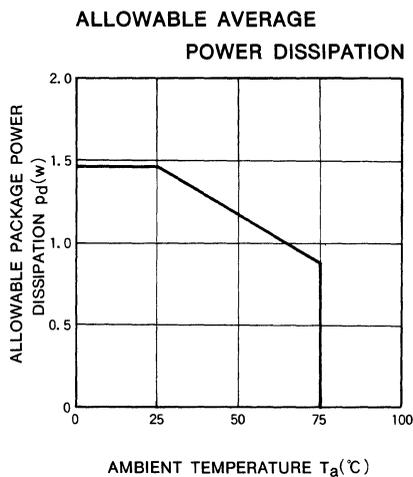
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		40	V
I_C	Collector current per channel	Percent duty cycle less than 8%	0		400	mA
		Percent duty cycle less than 30%	0		200	
V_{IH}	"H" Input voltage	$I_C = 400\text{mA}$	8		35	V
		$I_C = 200\text{mA}$	5		35	V
V_{IL}	"L" Input voltage	$I_{O(LEAK)} = 50\mu\text{A}$	0		0.5	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CER} = 100\mu\text{A}$	40			V
$V_{CE(sat)}$	Output saturation voltage	$V_I = 8\text{V}, I_C = 400\text{mA}$		1.3	2.4	V
		$V_I = 5\text{V}, I_C = 200\text{mA}$		1	1.6	
I_I	Input current	$V_I = 17\text{V}$		0.85	1.8	mA
		$V_I = 35\text{V}$		2.0	3.8	
V_F	Clamp diode forward voltage	$I_{RD} = 400\text{mA}$		1.5	2.4	V
V_R	Clamp diode reverse voltage	$V_{RD} = 100\mu\text{A}$	40			V
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 300\text{mA}, T_a = 25^\circ\text{C}$	1000	3500		—

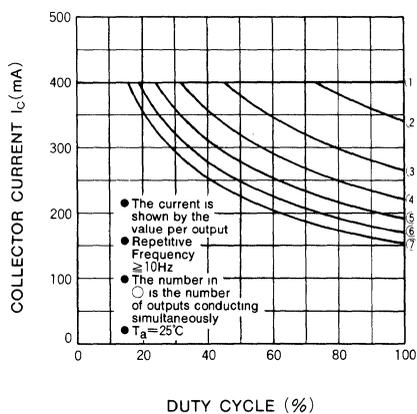
* : All typical values are at $T_a = 25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

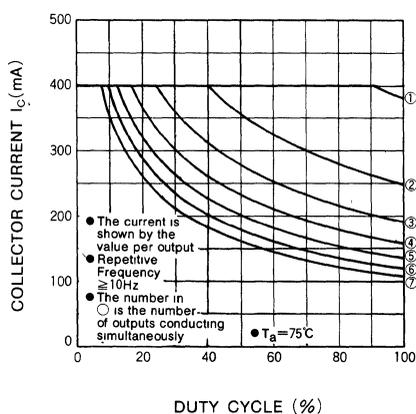


7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

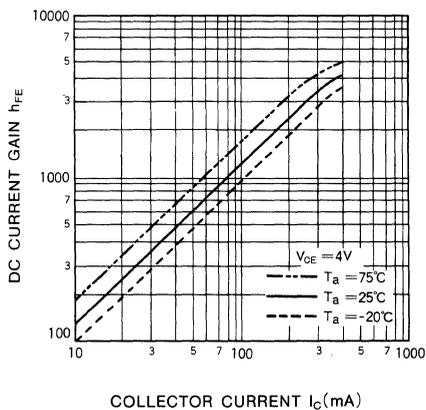
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54531P

7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54531P, 7-channel sink driver, consists of 14 NPN transistors connected to form high current gain driver pairs.

FEATURES

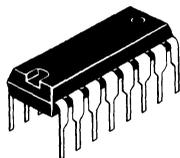
- High output sustaining voltage to 40V
- High output sink current to 400mA
- Integral diodes for transient suppression
- PMOS compatible input
- Wide input voltage range from $-40V$ to $+40V$
- Wide operating temperature range ($T_a = -20 \sim +75^\circ C$)

APPLICATION

Relay and printer driver, LED and incandescent display digit driver, Interfacing for standard MOS/BIPOLAR logics

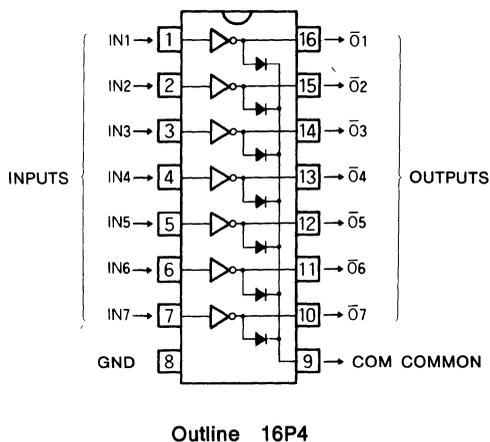
FUNCTION

The M54531P is comprised of seven NPN darlington driver pairs. Each input has a diode and $20k\Omega$ resistor in series to allow a negative voltage input. Between pin 9 and each out, there are integral diodes for inductive load transient suppression. All emitters and the substrate are connected together to pin 8. The outputs are capable of sinking 400mA and will withstand 40V in the OFF state.

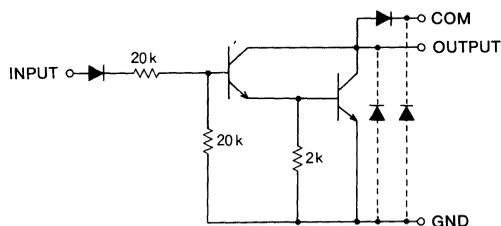


16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ C$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +40$	V
I_C	Collector current	Transistor ON	400	mA
V_I	Input voltage		$-40 \sim +40$	V
$I_{F(D)}$	Clamp diode forward current		400	mA
$V_{R(D)}$	Clamp diode reverse voltage		40	V
P_D	Power dissipation	$T_a = 25^\circ C$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ C$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ C$

7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		40	V
I_C	Collector current per channel	Percent duty cycle less than 8%	0		400	mA
		Percent duty cycle less than 30%	0		200	
V_{IH}	"H" Input voltage	$I_C = 400\text{mA}$	9		35	V
		$I_C = 200\text{mA}$	6		35	
V_{iL}	"L" Input voltage	$I_{C(\text{leak})} = 50\mu\text{A}$	0		1	V

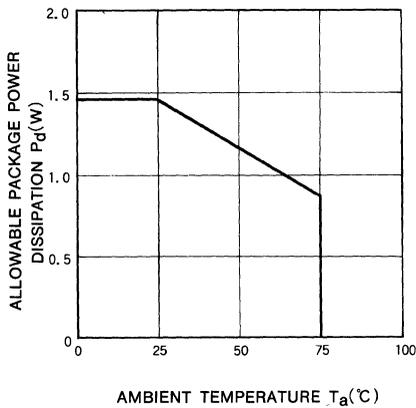
ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ*	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CE0} = 100\mu\text{A}$	40			V
$V_{CE(\text{sat})}$	Output saturation voltage	$V_I = 9\text{V}, I_C = 400\text{mA}$		1.3	2.4	V
		$V_I = 6\text{V}, I_C = 200\text{mA}$		1	1.6	
I_i	Input current	$V_I = 18\text{V}$		0.85	1.8	mA
		$V_I = 35\text{V}$		2.0	3.8	
I_R	Input leakage current	$V_I = -35\text{V}$			-20	μA
$V_{F(D)}$	Clamp diode forward voltage	$I_{F(D)} = 400\text{mA}$		1.5	2.4	V
$V_{R(D)}$	Clamp diode reverse voltage	$I_{R(D)} = 100\mu\text{A}$	40			V
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 300\text{mA}, T_a = 25^\circ\text{C}$	1000	3500		—

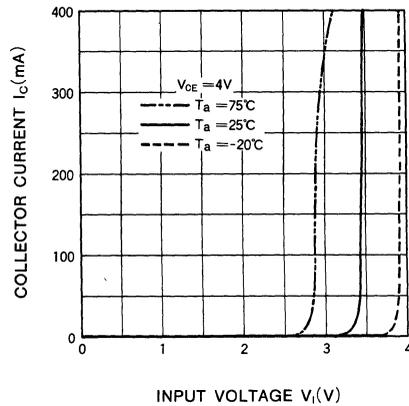
* : All typical values are at $T_a = 25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION

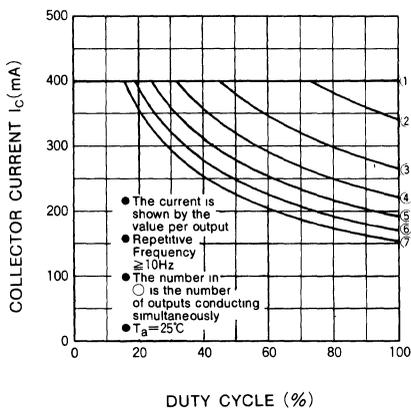


OUTPUT CURRENT CHARACTERISTICS

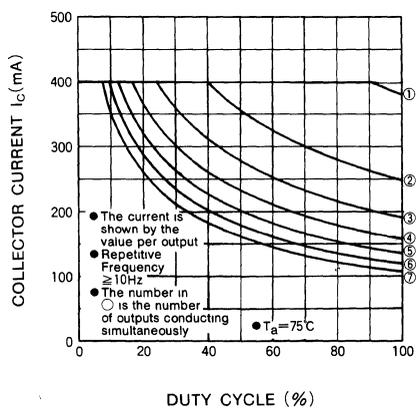


7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

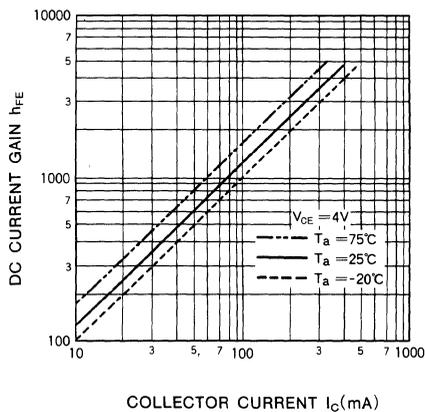
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54513P

8-UNIT 50mA TRANSISTOR ARRAY

DESCRIPTION

The M54513P, 8-channel sink drivers, consists of 8 NPN transistors with 2 kΩ series input resistors.

FEATURES

- High output sustaining voltage of 40V
- 50mA output sink current capability
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

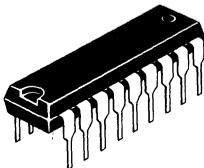
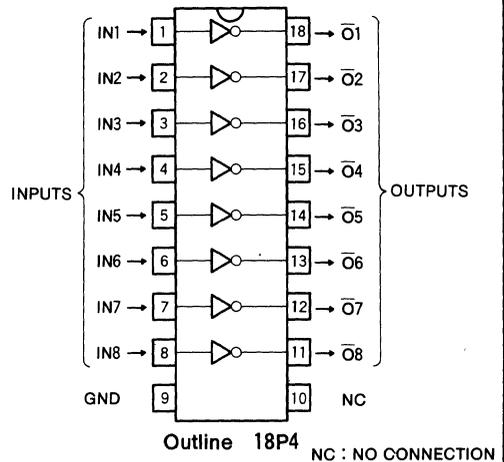
APPLICATION

LED or incandescent display digit driver

FUNCTION

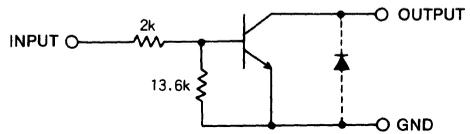
The M54513P is comprised of eight NPN drivers. Each input has a voltage divider by 2kΩ and 13.6kΩ resistors. All emitters and the substrate are connected together to pin 9. The open collector outputs are capable of sinking 50mA and will withstand 40V in the OFF state.

PIN CONFIGURATION (TOP VIEW)



18-pin molded plastic DIP

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +40	V
I_C	Collector current	Transistor ON	50	mA
V_I	Input voltage		10	V
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

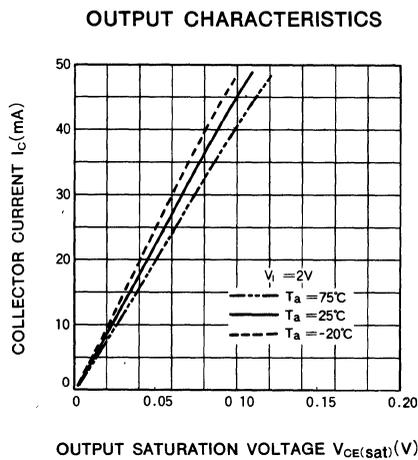
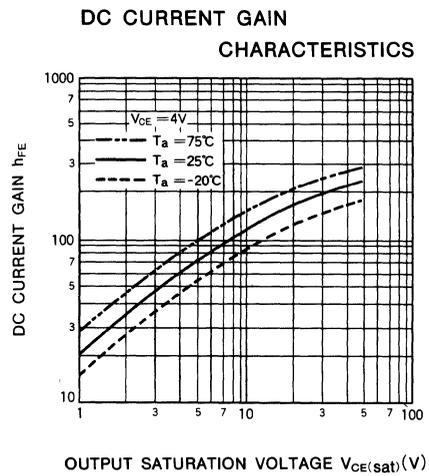
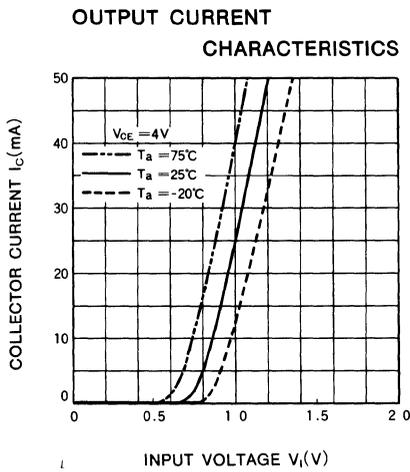
Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_O	Output voltage	0		40	V
I_C	Collector current per channel	0		20	mA
V_{IH}	"H" Input voltage	2		8	V
V_{IL}	"L" Input voltage	0		0.2	V

8-UNIT 50mA TRANSISTOR ARRAY

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{O(Leak)}$	Output leakage current	$V_o = 40\text{V}$			50	μA
$V_{CE(sat)}$	Output saturation voltage	$V_i = 2\text{V}, I_c = 12\text{mA}$		30	100	mV
		$V_i = 2.5\text{V}, I_c = 30\text{mA}$		70	170	
I_i	Input current	$V_i = 2.5\text{V}$		0.85	1.7	mA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_c = 30\text{mA}, T_a = 25^\circ\text{C}$	80	200		—

TYPICAL CHARACTERISTICS



M54538P

7-UNIT 350mA TRANSISTOR ARRAY AND MOTOR DRIVER

DESCRIPTION

The M54538P, 7-channel sink driver and a motor driver, is designed for use in a thermal printer.

FEATURES

- Output breakdown voltage to 20V
- High output sink current to 350mA
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

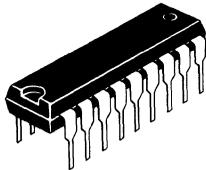
Thermal printer driver, LED or incandescent display driver, Interfacing for standard MOS/BIPOLAR logic

FUNCTION

The M54538P is designed for use in a thermal printer, consisting 7-channel thermal head driver and a D-C or stepper motor driver.

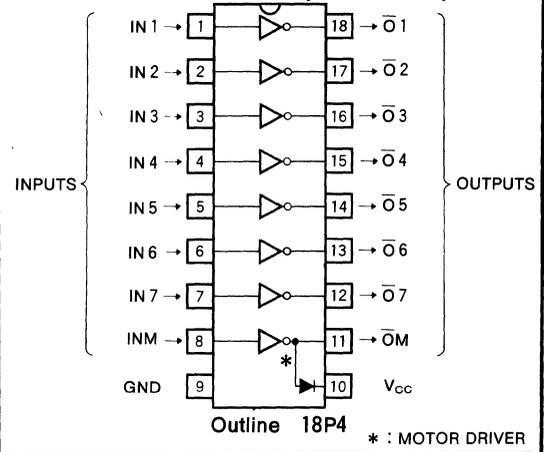
The output of the motor driver has a diode for inductive load transient suppression.

The outputs of the sink drivers are capable of sinking 350mA and will withstand 20V in the OFF state.

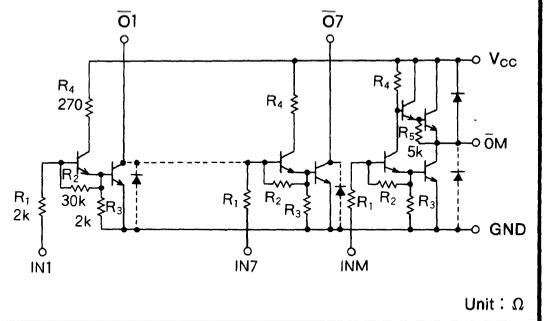


18-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit	
V_{CC}	Supply voltage		10	V	
V_{CEO}	Output sustaining voltage	Transistor OFF	$\bar{O}1$ to $\bar{O}7$ Outputs $\bar{O}M$ Output	$-0.5 \sim +20$ $-0.5 \sim V_{CC}$	V
I_C	Collector current	Transistor ON	350	mA	
V_I	Input voltage		10	V	
I_F	Clamp diode forward current	Pulse width $\leq 35\text{ms}$, Percent duty cycle $\leq 5\%$	700	mA	
			350		
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W	
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$	
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$	

7-UNIT 350mA TRANSISTOR ARRAY AND MOTOR DRIVER

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

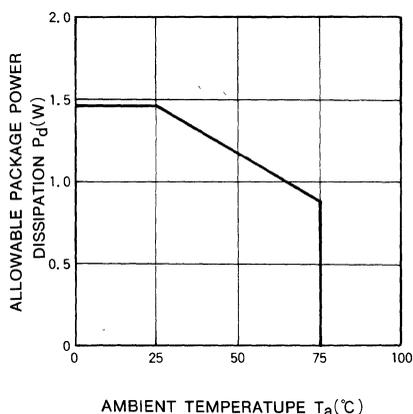
Symbol	Parameter	Limits			Unit	
		Min	Typ	Max		
V_{CC}	Supply voltage	3		6	V	
V_O	Output voltage	0		20	V	
I_C	Collector current per channel	Percent duty cycle less than 30%, $V_{CC}=6\text{V}$	0		250	mA
		Percent duty cycle less than 35%, $V_{CC}=6\text{V}$	0		170	
V_{IH}	"H" Input voltage	$I_C=250\text{mA}$	3.2		6	V
		$I_C=150\text{mA}$	2.4		6	
V_{IL}	"L" Input voltage	0		0.3	V	

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

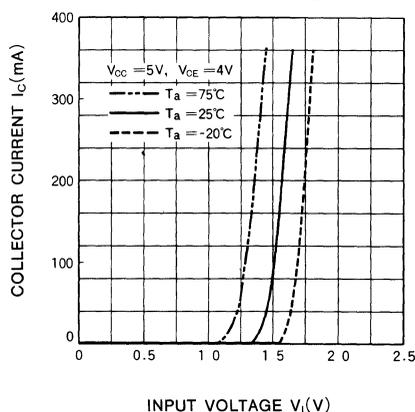
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_O (leak)	Input leakage current	$V_{CC}=6\text{V}$, $V_I=0$, $V_{CE}=20\text{V}$			50	μA
$V_{CE(sat)}$	Output saturation voltage	$V_{CC}=3.5\text{V}$, $V_I=3\text{V}$, $I_C=250\text{mA}$		0.23	0.6	V
		$V_{CC}=3\text{V}$, $V_I=2.4\text{V}$, $I_C=150\text{mA}$		0.14	0.4	
$V_{OH(M)}$	"H" Output voltage (motor driver)	$V_{CC}=6\text{V}$, $I_{OH(M)}=-250\text{mA}$	2.4			V
I_i	Input current	$V_{CC}=6\text{V}$, $V_I=3.2\text{V}$		0.8	1.5	mA
		$V_{CC}=6\text{V}$, $V_I=10\text{V}$		4.6	7.3	
$V_{F(M)}$	Clamp diode forward voltage	$I_{F(M)}=350\text{mA}$		1.6	3	V
I_{CC}	Supply current	$V_{CC}=6\text{V}$, $V_I=3.2\text{V}$ (all input)			235	mA
h_{FE}	DC forward current gain	$V_{CC}=5\text{V}$, $V_{CE}=4\text{V}$, $I_C=250\text{mA}$, $T_a=25^\circ\text{C}$	1000	6000		—

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION

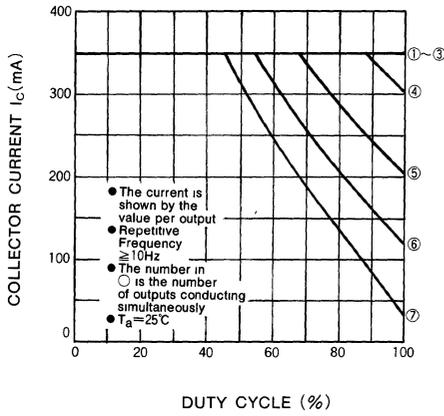


OUTPUT CURRENT CHARACTERISTICS

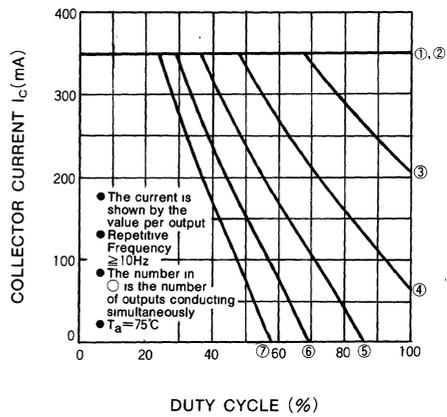


7-UNIT 350mA TRANSISTOR ARRAY AND MOTOR DRIVER

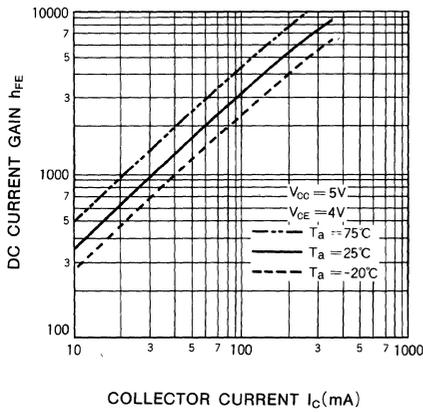
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54522P

8-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54522P, 8-channel sink driver, consists of 16 NPN transistors connected to form eight high current gain driver pairs.

FEATURES

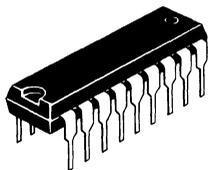
- High output sustaining voltage to 40V
- High output sink current to 400mA
- Integral diodes for transient suppression
- PMOS Compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED or incandescent display digit driver, Interfacing between MOS/BIPOLAR logics and high power loads

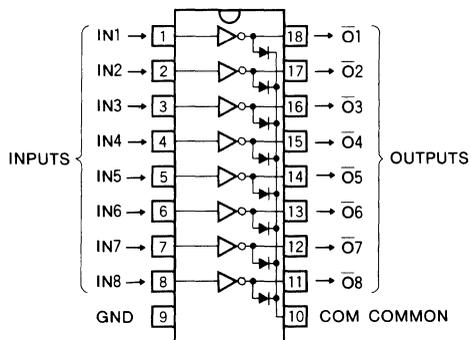
FUNCTION

The M54522P is comprised of eight NPN darlington driver pairs with $20\text{k}\Omega$ series input resistors. Each output has an integral diode for inductive load transient suppression. The cathodes of the diodes are connected together to pin 10. All emitters and the substrate are connected to pin 9. The outputs are capable of sinking 400mA and will withstand 40V in the OFF state.



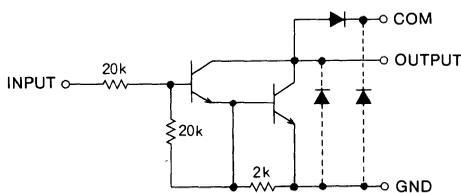
18-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



Outline 18P4

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +40$	V
I_C	Collector current	Transistor ON	400	mA
V_I	Input voltage		40	V
I_F	Clamp diode forward current		400	mA
V_R	Clamp diode reverse voltage		40	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.79	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

8-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

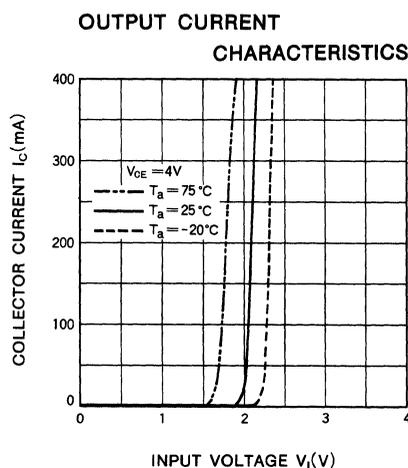
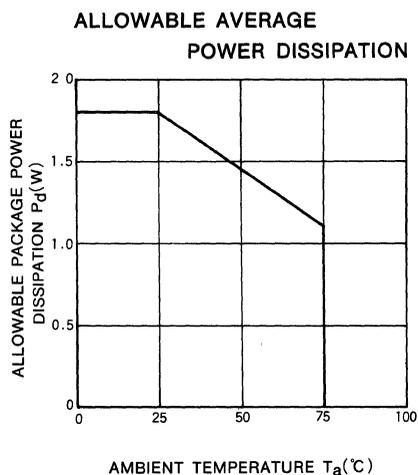
RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		40	V
I_C	Collector current per channel	Percent duty cycle less than 7%	0		400	mA
		Percent duty cycle less than 30%	0		200	
V_{IH}	"H" Input voltage	$I_C = 400\text{mA}$	8		30	V
		$I_C = 200\text{mA}$	4		30	
V_{IL}	"L" Input voltage	$I_{OL(leak)} = 50\mu\text{A}$	0		0.5	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

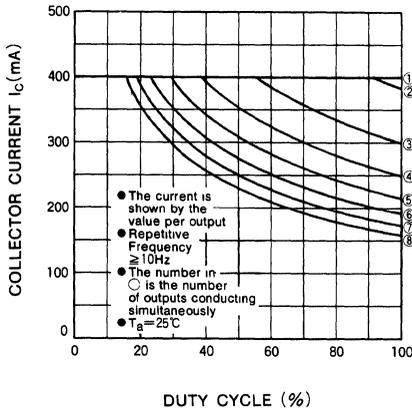
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CEO} = 100\mu\text{A}$	40			V
$V_{CE(sat)}$	Output saturation voltage	$V_I = 8\text{V}, I_C = 400\text{mA}$		1.15	2.4	V
		$V_I = 4\text{V}, I_C = 200\text{mA}$		0.95	1.6	
I_I	Input current	$V_I = 17\text{V}$	0.3	0.9	1.8	mA
V_F	Clamp diode forward voltage	$I_F = 400\text{mA}$		1.5	2.4	V
V_R	Clamp diode reverse voltage	$I_R = 100\mu\text{A}$	40			V
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, I_C = 300\text{mA}, T_a = 25^\circ\text{C}$	1000	8000		—

TYPICAL CHARACTERISTICS

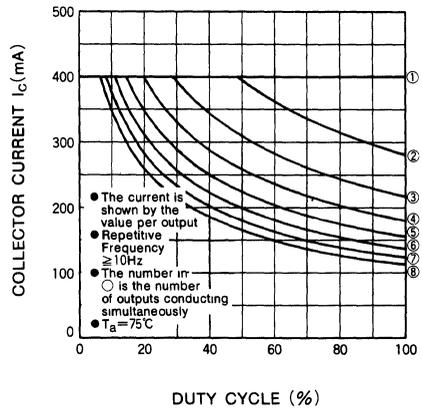


8-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

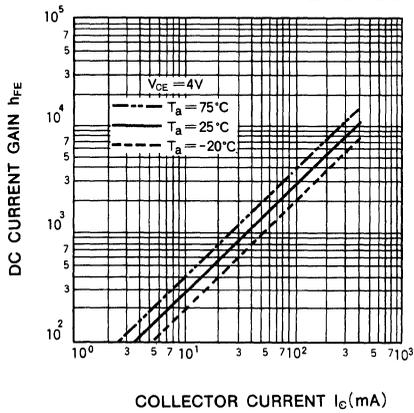
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54585P

8-UNIT 500mA TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54585P, 8-channel sink driver, consists of 16 NPN darlington transistors with internal clamp diodes connected to form high current gain driver pairs with low input current.

FEATURES

- High output sustaining voltage to 50V
- High output sink current to 500mA
- Built-in clamp diode
- TTL, PMOS IC output for drive
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

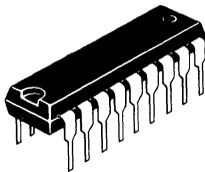
Relay and small printer driver, LED or incandescent display digit driver, Output for microcomputer and interface with high voltage system

FUNCTION

The M54585P is composed of eight NPN darlington transistor pairs. A resistor of $2.7\text{k}\Omega$ is connected between the base of input transistor and the input pin.

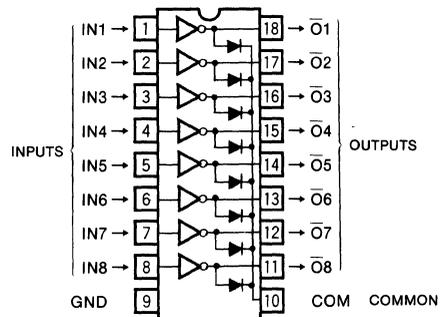
A clamp diode for inductive load transient suppression is connected for the output pin (collector) and COM pin (pin 10). All emitters of the output transistors are connected to GND (pin 9).

The outputs are capable of sinking 500mA and will withstand 40V between collector and emitter.



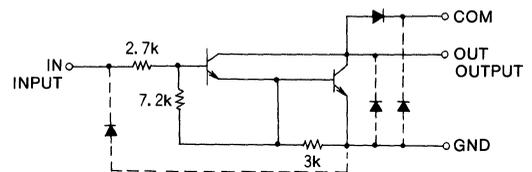
18-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



Outline 18P4

CIRCUIT SCHEMATIC (EACH CIRCUIT)



COM and GND are all common to 8 circuits.
The diodes shown by broken line are parasite diodes and must not be used.

Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	When the output is "H"	-0.5 ~ +50	V
I_C	Collector current	Per channel current, when the output is "L"	500	mA
V_I	Input voltage		-0.5 ~ +30	V
I_F	Clamp diode forward current		500	mA
V_R	Clamp diode reverse current		-0.5 ~ +50	V
P	Power dissipation	$T_a = 25^\circ\text{C}$	1.79	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

8-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

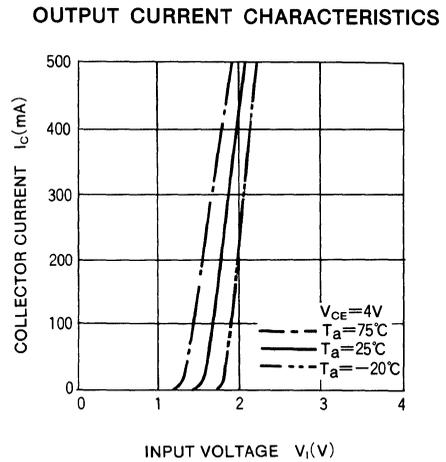
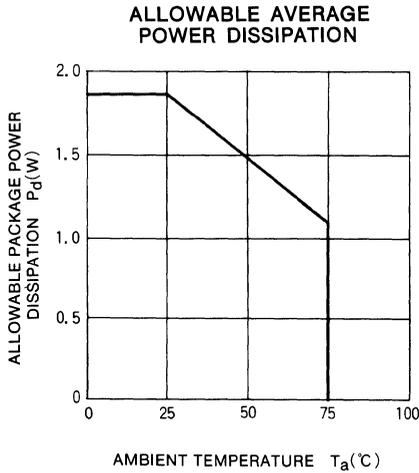
RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_O	Output voltage		0		50	V
I_i	Collector current per channel	Percent duty cycle less than 6%	0		400	mA
		Percent duty cycle less than 34%	0		200	
V_{IH}	"H" Input voltage		$I_c = 400\text{mA}$	3.85	30	V
			$I_c = 200\text{mA}$	3.4	30	
V_{IL}	"L" Input voltage		$I_{O(\text{leak})} = 50\mu\text{A}$	0	0.6	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

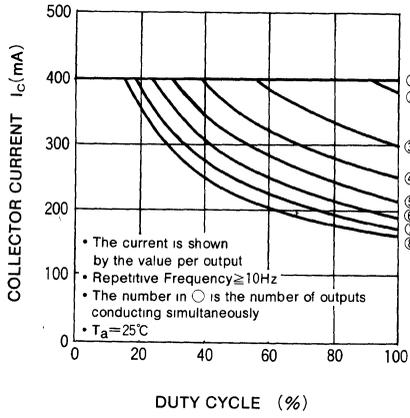
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{O(\text{leak})}$	Output leakage current	$V_{CE} = 50\text{V}$			100	μA
$V_{CE(\text{sat})}$	Output saturation voltage		$V_i = 3.85\text{V}, I_c = 400\text{mA}$	1.3	2.4	V
			$V_i = 3.85\text{V}, I_c = 200\text{mA}$	1.0	1.6	
I_i	Input current		$V_i = 3.85\text{V}$	0.95	1.8	mA
			$V_i = 25\text{V}$	11	18	
V_F	Clamp diode forward voltage	$I_F = 400\text{mA}$		1.5	2.4	V
I_R	Clamp diode reverse voltage	$V_R = 50\text{V}$			100	μA
h_{FE}	DC forward current transfer ratio	$V_{CC} = 4\text{V}, I_c = 350\text{mA}, T_a = 25^\circ\text{C}$	1000	2500		—

TYPICAL CHARACTERISTICS

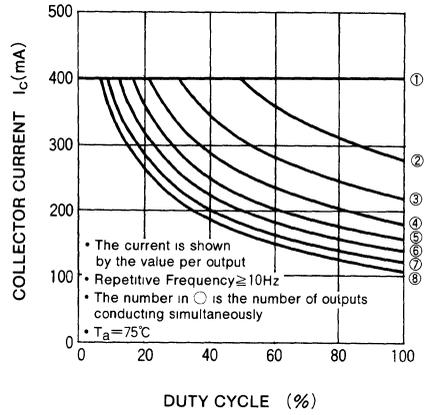


8-UNIT 500mA DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

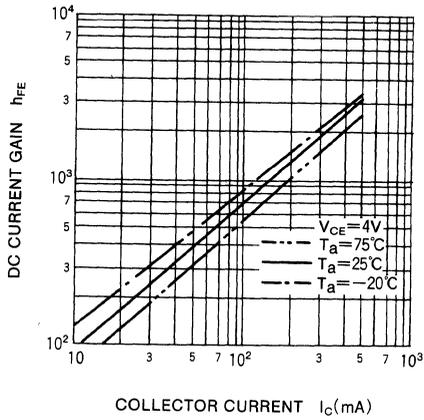
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



CURRENT GAIN CHARACTERISTICS



M54584P

8-UNIT 350mA TRANSISTOR ARRAY

DESCRIPTION

The M54584P, 8-channel sink driver, consists of 16 NPN transistors connected to form high current gain driver pairs with low input current.

FEATURES

- High output sustaining voltage to 20V
- High output sink current to 350mA
- PMOS IC output for drive
- Low output saturation voltage
($V_{CE(sat)}=0.5V$ at $I_C=250mA$)
- Wide operating temperature range ($T_a=-20\sim+75^\circ C$)

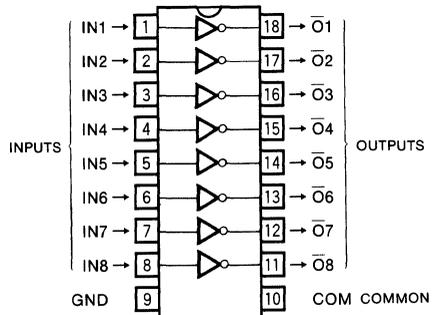
APPLICATION

Relay and thermal printer dot driver, LED or incandescent display digit driver, Interface for MOS-bipolar logic ICs

FUNCTION

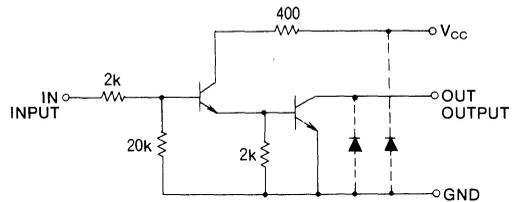
The M54584P is composed of eight NPN transistors with the emitters of output transistors connected to GND pin (pin 9). The collectors of NPN predriver transistors are connected to the V_{CC} (pin 10) via a resistor of 400Ω . The outputs are capable of sinking 350mA and will withstand 20V between collector and emitter.

PIN CONFIGURATION (TOP VIEW)



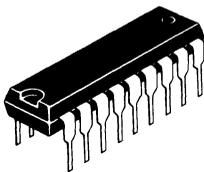
Outline 18P4

CIRCUIT SCHEMATIC (EACH CIRCUIT)



V_{CC} and GND are all common to 8 circuits
The diodes shown by broken line are parasite diodes and must not be used

Unit : Ω



18-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a=-20\sim+75^\circ C$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		-0.5~10	V
V_{CEO}	Output sustaining voltage	When the output is "H"	-0.5~+20	V
I_C	Collector current	Per channel current, when the output is "L"	350	mA
V_i	Input voltage		-0.5~+10	V
P_d	Power dissipation	$T_a=25^\circ C$	1.79	W
T_{opr}	Operating ambient temperature range		-20~+75	$^\circ C$
T_{sqg}	Storage temperature range		-55~+125	$^\circ C$

8-UNIT 350mA TRANSISTOR ARRAY

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

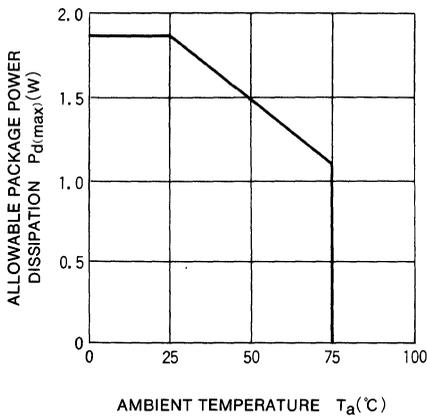
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		3	5	8	V
V_O	Output voltage		0		20	V
I_C	Collector current per channel	Percent duty cycle less than 45%, $V_{CC}=6.5\text{V}$			250	mA
		Percent duty cycle less than 70%, $V_{CC}=6.5\text{V}$			150	
V_{IH}	"H" Input voltage	$I_C \geq 250\text{mA}$	3		V_{CC}	V
V_{IL}	"L" Input voltage	$I_{O(\text{leak})} \geq 50\mu\text{A}$	0		0.4	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

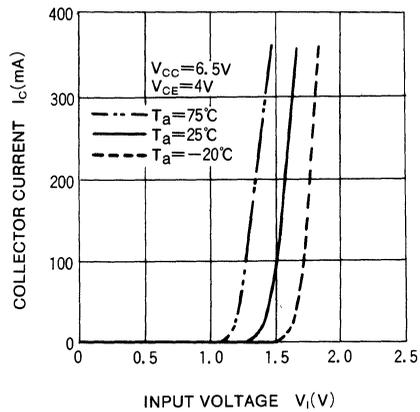
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{O(\text{leak})}$	Output leakage voltage	$V_{CC}=8\text{V}, V_{CE}=20\text{V}$			100	μA
$V_{CE(\text{sat})}$	Output saturation voltage	$V_{CC}=6.5\text{V}, V_I=3\text{V}, I_C=250\text{mA}$		0.3	0.5	V
		$V_{CC}=3\text{V}, V_I=3\text{V}, I_C=150\text{mA}$		0.7	0.35	
I_I	Input current	$V_{CC}=8\text{V}, V_I=3\text{V}$		0.7	1.5	mA
		$V_{CC}=8\text{V}, V_I=10\text{V}$		4.3	7.3	
I_{CC}	Supply current (all output ON)	$V_{CC}=8\text{V}, V_I=3\text{V}$			220	mA
h_{FE}	DC forward current transfer ratio	$V_{CC}=6.5\text{V}, V_{CE}=4\text{V}, I_C=250\text{mA}, T_a=25^\circ\text{C}$	1000	7000		—

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION

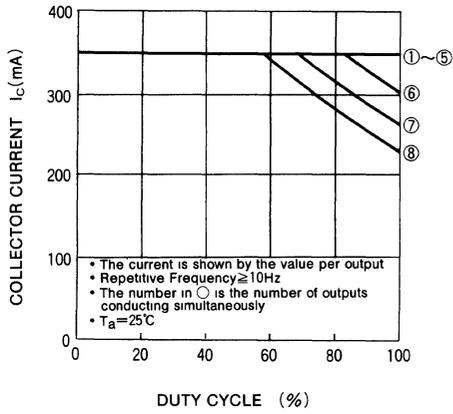


OUTPUT CURRENT CHARACTERISTICS

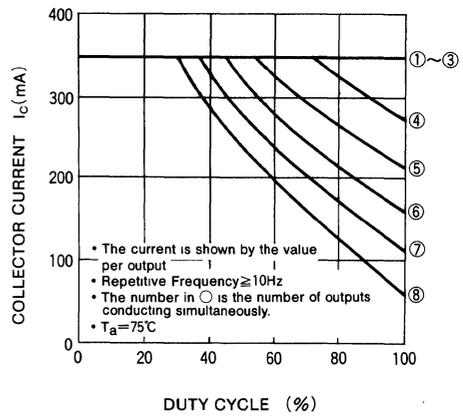


8-UNIT 350mA TRANSISTOR ARRAY

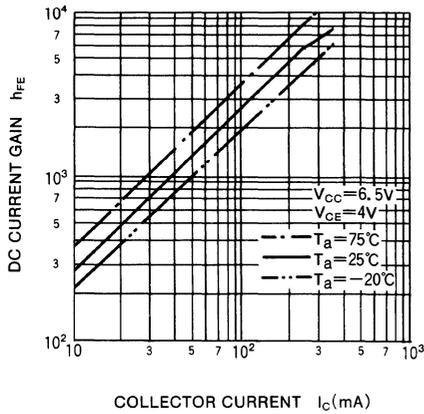
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



CURRENT GAIN CHARACTERISTICS



M54562P

8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54562P, 8-channel source driver, is designed for use with +6 to +16V MOS logic systems.

FEATURES

- High output sustaining voltage to 50V
- High output source current to 500mA
- Integral diode for transient suppression
- 6~16V CMOS compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

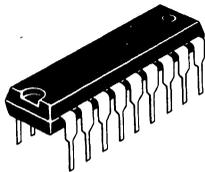
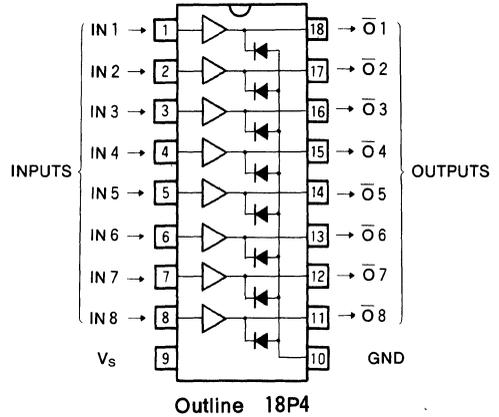
APPLICATION

Relay and printer driver, LED, incandescent or fluorescent display driver, Interfacing for standard MOS/BIPO-LAR logics

FUNCTION

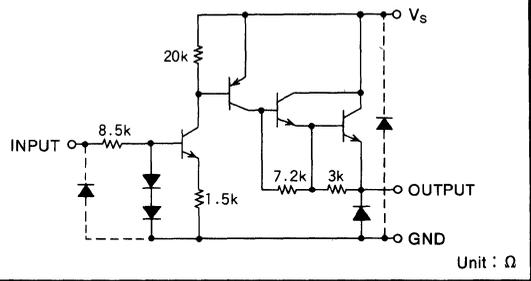
The driver of the M54562P is comprised of a NPN inverter and compound PNP/NPN/NPN output source driver, and the output is turned ON by an active high input level. Each output has an integral diode for inductive load transient suppression. The outputs are capable of driving 500mA and are rated for operation with output voltage up to 50V.

PIN CONFIGURATION (TOP VIEW)



18-pin molded plastic DIP

CIRCUIT SCHEMATIC



ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Output is in "L"	-0.5~+50	V
V_S	Supply voltage		50	V
V_I	Input voltage		0~+30	V
I_O	Output current	Per channel current at "H" output	-500	mA
I_F	Clamp diode forward current		-500	mA
V_R	Clamp diode reverse voltage		50	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.79	W
T_{opr}	Operating ambient temperature range		-20~+75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

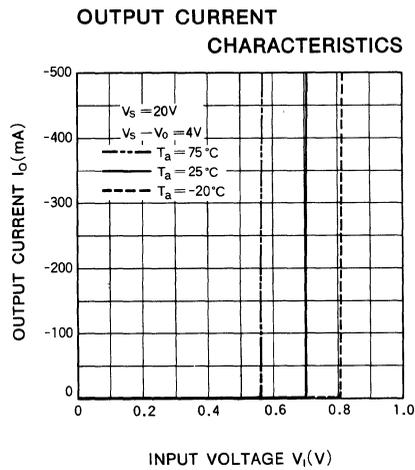
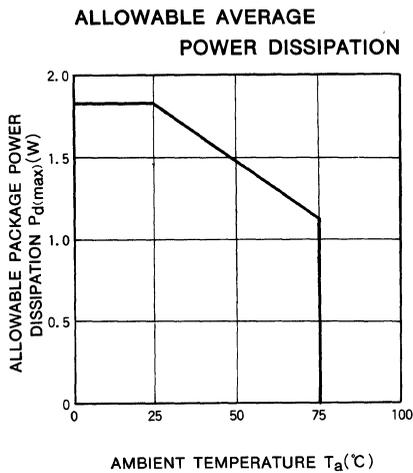
RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_S	Supply voltage		0		50	V
I_O	Output current per channel	Percent duty cycle less than 8%	0		-350	mA
		Percent duty cycle less than 55%	0		-100	
V_{IH}	"H" Input voltage	$I_O = -350\text{mA}$	2.4	5	30	V
V_{IL}	"L" Input voltage		0		0.2	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

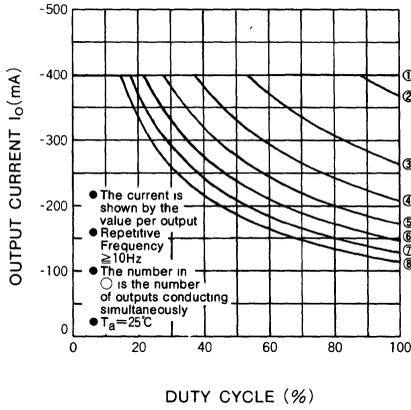
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_S (leak)	Supply leak current	$V_S = 50\text{V}, V_I = 0.2\text{V}$			100	μA
V_{CE} (sat)	Output saturation voltage	$V_S = 10\text{V}, V_I = 4\text{V}, I_O = -350\text{mA}$		1.6	2.4	V
		$V_S = 10\text{V}, V_I = 4\text{V}, I_O = -100\text{mA}$		1.45	2.0	
I_I	Input current	$V_I = 5\text{V}$		0.48	0.75	mA
		$V_I = 25\text{V}$		2.8	4.7	
I_S	Supply current	$V_S = 50\text{V}, V_I = 5\text{V}$		5.6	6.5	mA
V_F	Clamp diode forward voltage	$I_F = -350\text{mA}$		-1.2	-2.4	V
V_R	Clamp diode reverse voltage	$I_R = 100\mu\text{A}$	50			V

TYPICAL CHARACTERISTICS

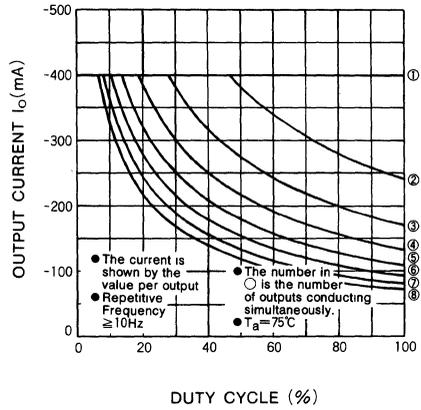


8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

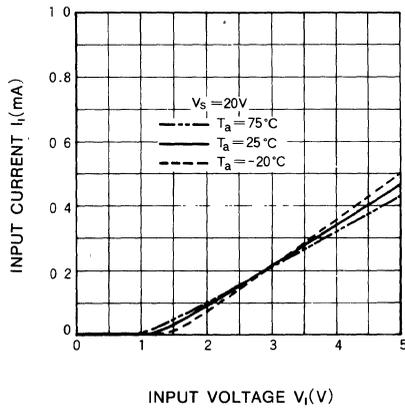
ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



INPUT CHARACTERISTICS



M54563P

8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54563P, 8-channel source driver, is designed for use with +6 to +16V MOS logic systems.

FEATURES

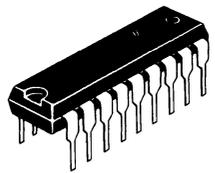
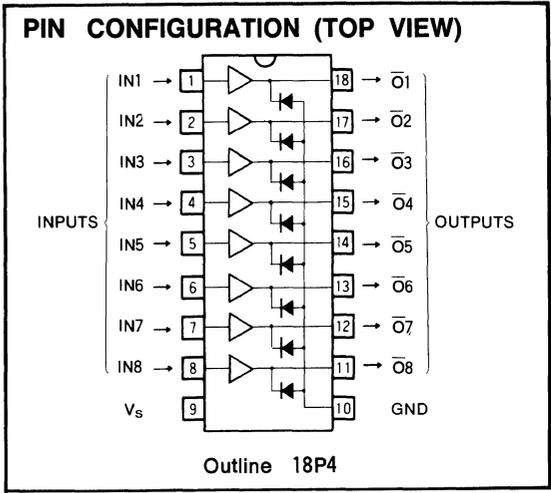
- High output sustaining voltage to 50V
- High output source current to 500mA
- Integral diode for transient suppression
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

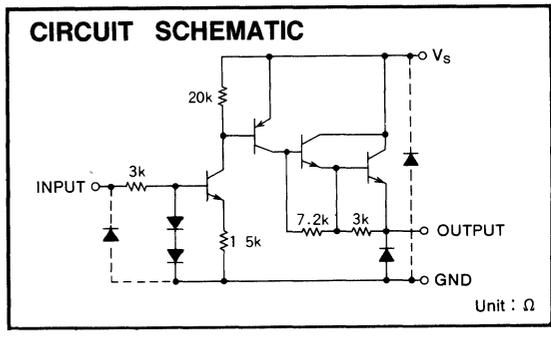
Relay and printer driver, LED, incandescent or fluorescent display driver, Interfacing for standard MOS/BIPO-LAR logics

FUNCTION

The driver of the M54563P is comprised of a NPN inverter and compound PNP/NPN/NPN output source driver and the output is turned ON by an active high input level. Each output has an integral diode for inductive load transient suppression. The outputs are capable of driving 500mA and are rated for operating with output voltage up to 50V.



18-pin molded plastic DIP



ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +50	V
V_S	Supply voltage		50	V
V_I	Input voltage		0 ~ +10	V
I_O	Output current	Transistor OFF	-500	mA
I_F	Clamp diode forward current		-500	mA
V_R	Clamp diode reverse voltage		50	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.79	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

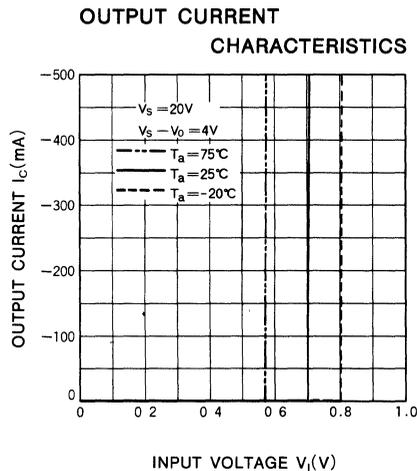
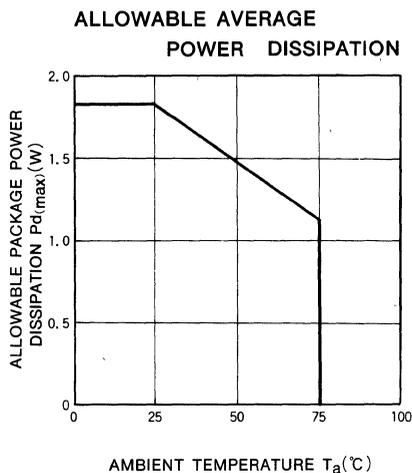
RECOMMENDED OPERATIONAL CONDITIONS (T_a = -20~+75°C, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V _S	Supply voltage		0		50	V
I _O	Output current per channel	Percent duty cycle less than 8%	0		-350	mA
		Percent duty cycle less than 55%	0		-100	
V _{IH}	"H" Input voltage	I _O = -350mA	2.4		6	V
V _{IL}	"L" Input voltage		0		0.2	V

ELECTRICAL CHARACTERISTICS (T_a = -20~+75°C, unless otherwise noted)

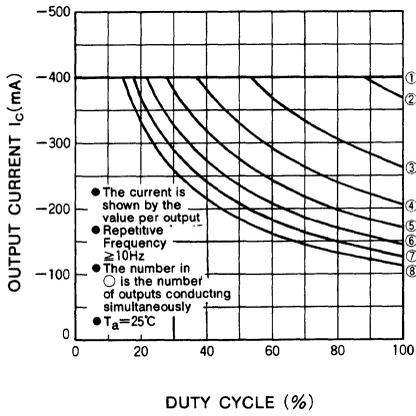
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I _S (leak)	Supply leakage current	V _S = 50 V, V _I = 0.2 V			100	μA
V _{CE} (sat)	Output saturation voltage	V _S = 10 V, V _I = 2.4 V, I _O = -350 mA		1.6	2.4	V
		V _S = 10 V, V _I = 2.4 V, I _O = -100 mA		1.45	2	
I _I	Input current	V _I = 3 V		0.6	1	mA
		V _I = 10 V		2.9	5	
I _S	Supply current	V _S = 50 V, V _I = 3 V		5.6	6.5	mA
V _F	Clamp diode forward voltage	I _F = -350 mA		-1.2	-2.4	V
V _R	Clamp diode reverse voltage	I _R = 100 μA	50			V

TYPICAL CHARACTERISTICS

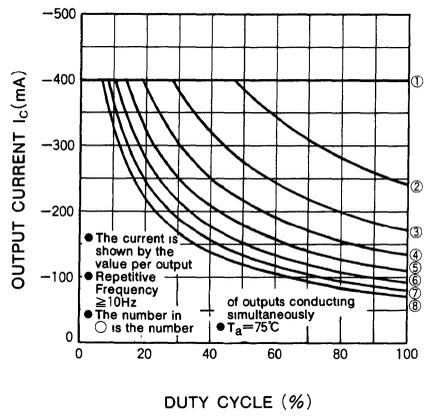


8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

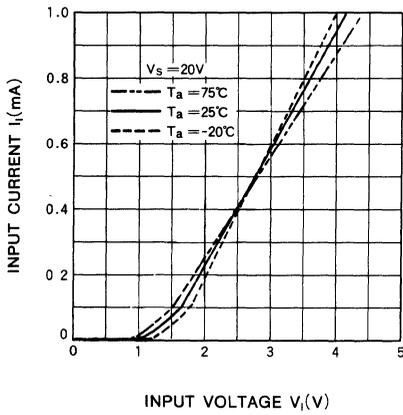
ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



INPUT CHARACTERISTICS



M54564P

8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54564P, 8-channel source driver, is designed for interfacing between low power digital logic and a fluorescent display.

FEATURES

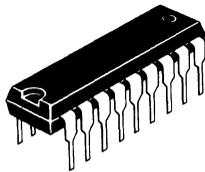
- High output sustaining voltage to 50V
- High output source current to 500mA
- CMOS, TTL Compatible input
- Internal pull-down resistors
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED, incandescent or fluorescent display driver, Interfacing for standard MOS/BIPOLAR logic

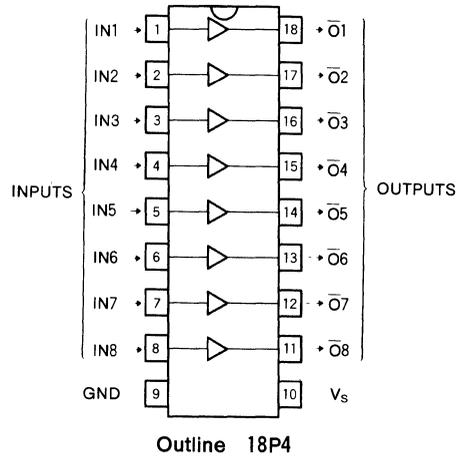
FUNCTION

The driver of the M54564P is comprised of a NPN inverter and compound PNP/NPN/NPN output source driver and the output is turned ON by an active high input level. Each output has 50kΩ pull-down resistor suitable for driving fluorescent displays. The outputs are capable of driving 500mA and are rated for operation with output voltage up to 50V.

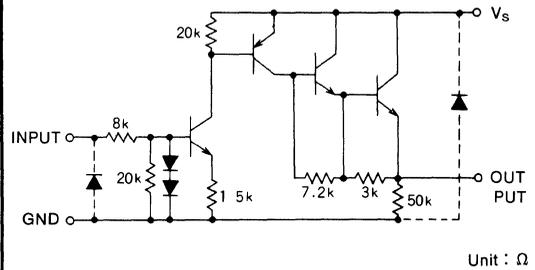


18-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage		$-0.5 \sim +50$	V
V_S	Supply voltage		50	V
V_i	Input voltage		$0 \sim +30$	V
I_o	Output current		-500	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.79	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

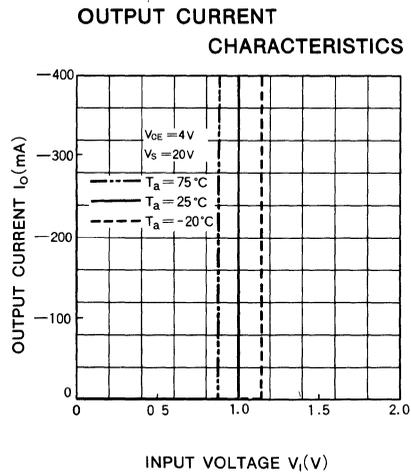
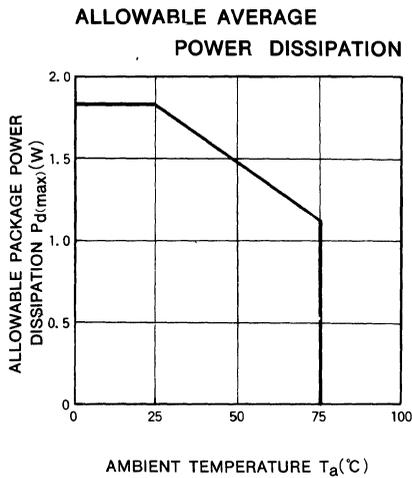
RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_S	Supply voltage		0		50	V
I_O	Output current per channel	Percent duty cycle less than 8%	0		-350	mA
		Percent duty cycle less than 60%	0		-100	
V_{IH}	"H" Input voltage	$I_O = -350\text{mA}$	4		25	V
V_{IL}	"L" Input voltage		0		0.2	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

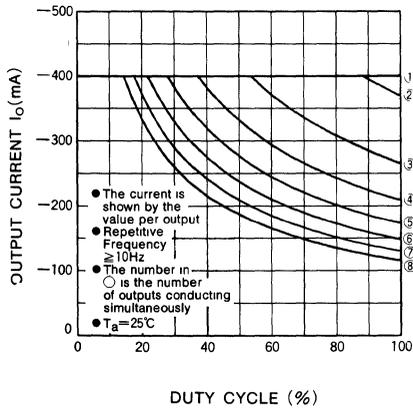
Symbol	Parameter	Test conditions	Limits			Unit	
			Min	Typ	Max		
$I_S(\text{leak})$	Supply leak current	$V_S = 50\text{V}$, $V_I = 0.2\text{V}$			100	μA	
$V_{CE(\text{sat})}$	Output saturation voltage	$V_S = 10\text{V}$			1.6	V	
		$V_I = 4\text{V}$	$I_O = -350\text{mA}$		2.4		
I_I	Input current	$V_I = 4\text{V}$			0.4	mA	
		$V_I = 25\text{V}$	$I_O = -100\text{mA}$		4.7		
I_S	Supply current	$V_S = 50\text{V}$, $V_I = 4\text{V}$			5.6	6.5	mA

TYPICAL CHARACTERISTICS

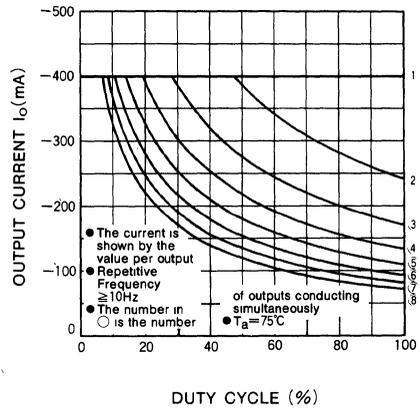


8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

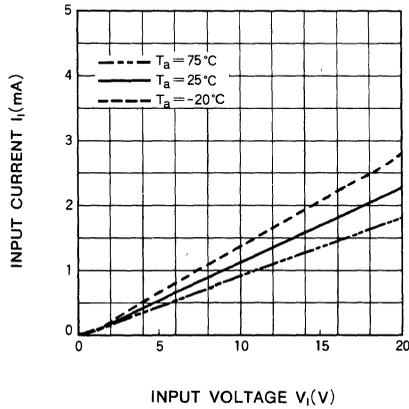
ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



INPUT CHARACTERISTICS



M54567P**4-UNIT 1.5A DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE****DESCRIPTION**

The M54567P, 4-channel sink driver, consists of 4 PNP and 14 NPN transistors to form high current gain driver pairs.

FEATURES

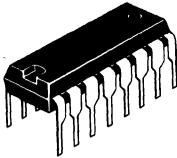
- High output sustaining voltage to 50V
- High output current to 1.5A
- Integral diodes for transient suppression
- NMOS Compatible input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

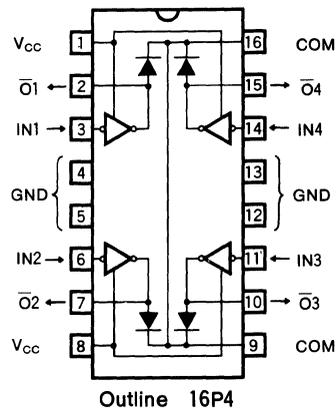
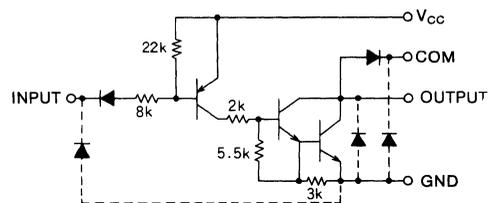
Relay and printer driver, LED or incandescent display digit driver

FUNCTION

The M54567P is comprised of four PNP invertors with $8k\Omega$ series input resistors and NPN darlington sink drivers. Each output has an integral diode for inductive load transient suppression and the anodes of the diode connected to pins 9 and 16. The outputs are capable of sinking 1.5A and will withstand 50V in the OFF state.



16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)**CIRCUIT SCHEMATIC**

Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +50$	V
V_i	Input voltage		30	V
I_C	Collector current	Transistor ON	1.5	A
V_R	Clamp diode reverse voltage		50	V
I_F	Clamp diode forward current	Pulse width $\leq 10\text{ms}$, Repetitive cycle $\leq 10\text{Hz}$	1.5	A
			1	A
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.92	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

4-UNIT 1.5A DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

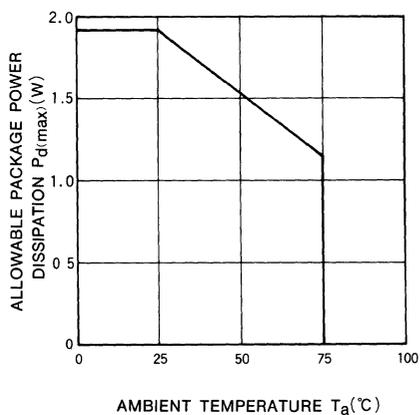
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	5	6	V
V_O	Output voltage		0		40	V
I_C	Collector current per channel	All units ON Percent duty cycle $\leq 4\%$	0		1.25	A
		All units ON Percent duty cycle $\leq 18\%$	0		0.7	
V_{IH}	"H" Input voltage	$I_O(\text{leak}) = 50\mu\text{A}$	$V_{CC} - 0.5$		V_{CC}	V
V_{IL}	"L" Input voltage	$I_C = 1.25\text{A}$	0		$V_{CC} - 3.5$	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

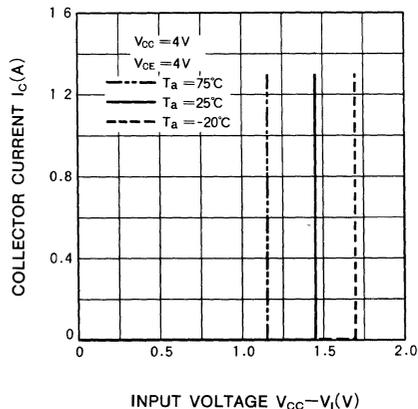
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CEO} = 100\mu\text{A}$	50			V
I_{CC}	Supply current	$V_{CC} = 6\text{V}, V_I = 0.5\text{V}$		3.0	4.5	mA
$V_{CE(\text{sat})}$	Output saturation voltage	$V_{CC} = 4\text{V}$		1.6	2.2	V
		$V_I = 0.5\text{V}$	$I_C = 1.25\text{A}$ $I_C = 0.7\text{A}$	1.1	1.7	
I_I	Input current	$V_I = V_{CC} - 3.5\text{V}$		-0.3	-0.6	mA
		$V_I = V_{CC} - 6\text{V}$		-0.58	-0.95	
V_R	Clamp diode reverse voltage	$I_R = 100\mu\text{A}$	50			V
V_F	Clamp diode forward voltage	$I_F = 1.25\text{A}$		1.6	2.3	V
η_{FE}	DC forward current gain	$V_{CC} = 4\text{V}, V_{CE} = 4\text{V}, I_C = 1\text{A}, T_a = 25^\circ\text{C}$	4000	30000		—

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION

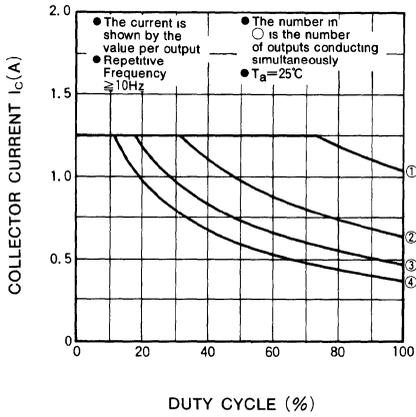


OUTPUT CURRENT CHARACTERISTICS

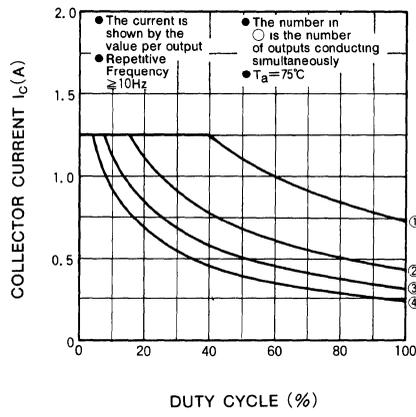


4-UNIT 1.5A DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

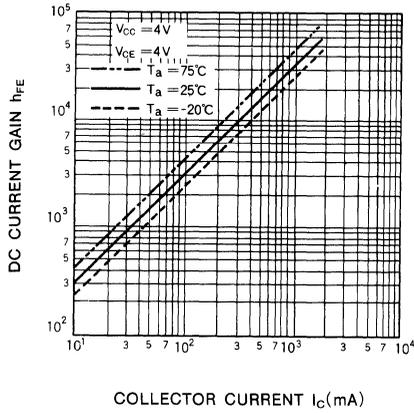
ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE COLLECTOR CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54576P,FP

7-UNIT 30mA TRANSISTOR ARRAY (INPUT "L" ACTIVE)

DESCRIPTION

The M54576P,FP, 7-channel sink driver, consists of 28 NPN transistors connected to form high current gain driver pairs.

FEATURES

- 30V output breakdown
- 30mA output sink current capability
- CMOS compatible input
- Low output saturation voltage
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

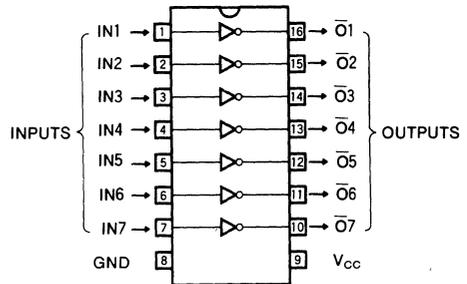
APPLICATION

LED or incandescent display digit driver

FUNCTION

The M54576P,FP is comprised of seven NPN inverters with diodes and 23k Ω resistors in series to the input and non darlington NPN sink drivers. The output is turned ON by switching the input low. The outputs are capable of sinking 30mA and will withstand 30V in the OFF state. The M54576FP features a small flat mold package.

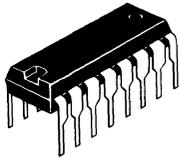
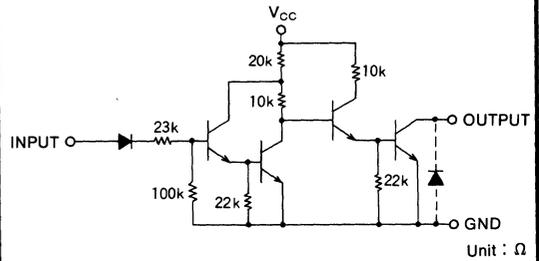
PIN CONFIGURATION (TOP VIEW)



Outline 16P2 (M54576FP)

Outline 16P4 (M54576P)

CIRCUIT SCHEMATIC



16-pin molded plastic DIP 16-pin molded plastic FLAT

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		13	V
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +30	V
I_C	Collector current	Transistor ON	30	mA
V_I	Input voltage		-20, 13	V
P_d	Power dissipation	$T_a=25^\circ\text{C}$	1.47/0.56	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

RECOMMENDED OPERATIONAL CONDITIONS ($T_a=-20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{CC}	Supply voltage	4	5	13	V
I_C	Collector current per channel	0	10	20	mA
V_{IH}	"H" Input voltage	3		V_{CC}	V
V_{IL}	"L" Input voltage	0		1	V

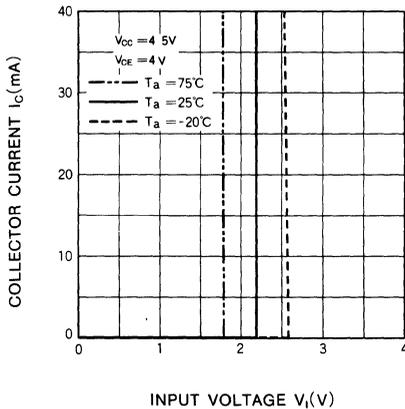
7-UNIT 30mA TRANSISTOR ARRAY (INPUT "L" ACTIVE)

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

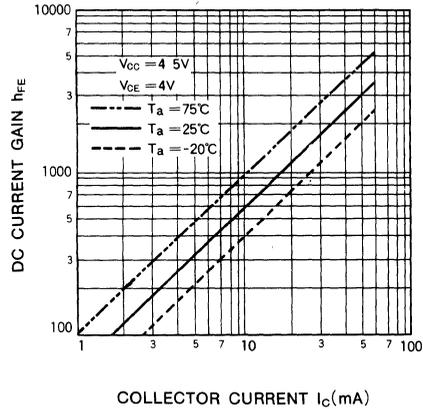
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{o(Leak)}$	Output leakage current	$V_{CE}=30V, V_I=3V, V_{CC}=6V$			100	μA
$V_{CE(sat)}$	Output saturation voltage	$V_{CC}=4.5V, V_I=1V, I_C=10\text{mA}$		0.02	0.25	V
		$V_{CC}=6V, V_I=1V, I_C=20\text{mA}$		0.04	0.35	
I_i	Input current	$V_{CC}=4.5V, V_I=3V$	30	60	90	μA
I_{CC}	Supply current	$V_{CC}=4.5V, V_I=1V$			6.3	mA
		$V_{CC}=13V, V_I=1V$			18	
h_{FE}	DC forward current gain	$V_{CE}=4V, V_{CC}=4.5V, I_C=20\text{mA}, T_a=25^\circ\text{C}$	500	1200		—

TYPICAL CHARACTERISTICS

OUTPUT CURRENT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54566P

7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY (INPUT "L" ACTIVE)

DESCRIPTION

The M54566P, 7-channel sink driver, consists of 7 PNP and 14 NPN transistors connected to form high current gain driver pairs.

FEATURES

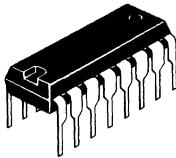
- High output sustaining voltage to 50V
- High output sink current to 400mA
- "L" Active input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, Interfacing between standard MOS/BIPOLAR logics

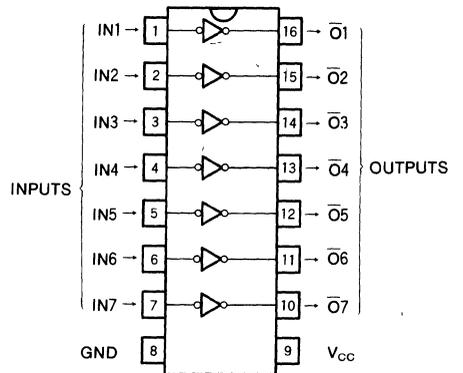
FUNCTION

The M54566P is comprised of seven PNP invertors with 8 k Ω series input resistors and NPN darlington sink drivers. The output is turned ON by switching the input low. The outputs are capable of sinking 400mA and will withstand 50V in the OFF state.



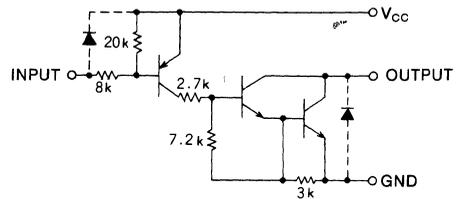
16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



Outline 16P4

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +50$	V
V_i	Input voltage		$0 \sim V_{CC}$	V
I_c	Collector current	Transistor ON	400	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

**7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY
 (INPUT "L" ACTIVE)**

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

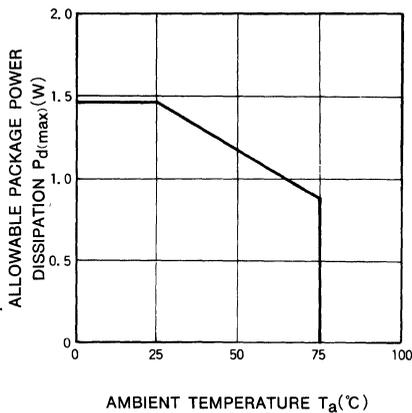
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	5	8	V
I_C	Collector current per channel	Percent duty cycle less than 10%	0		350	mA
		Percent duty cycle less than 30%	0		200	
V_{IH}	"H" Input voltage	$I_{O(LEAK)} = 50 \mu\text{A}$	$V_{CC} - 0.2$		V_{CC}	V
V_{IL}	"L" Input voltage	$I_C = 350\text{mA}$	0		$V_{CC} - 3$	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

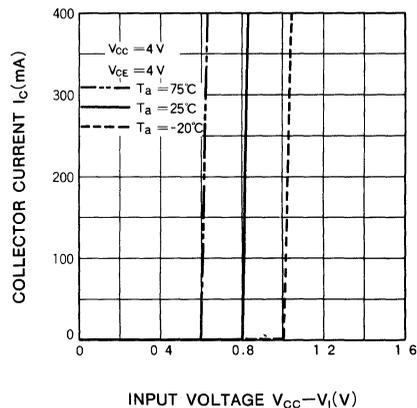
Symbol	Parameter	Test conditions	Limits			Unit	
			Min	Typ	Max		
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CEO} = 100 \mu\text{A}$	50			V	
$V_{CE(sat)}$	Output saturation voltage	$V_I = V_{CC} - 3\text{V}$		$I_C = 350\text{mA}$	1.1	2.2	V
				$I_C = 200\text{mA}$	0.9	1.6	
I_I	Input current	$V_I = V_{CC} - 3.5\text{V}$		-0.38	-0.58	mA	
I_{CC}	Supply current	$V_{CC} = 5\text{V}$, $V_I = V_{CC} - 3.5\text{V}$		1.4	3	mA	
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}$, $V_{CC} = 5\text{V}$, $I_C = 350\text{mA}$, $T_a = 25^\circ\text{C}$	2000	10000		—	

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION

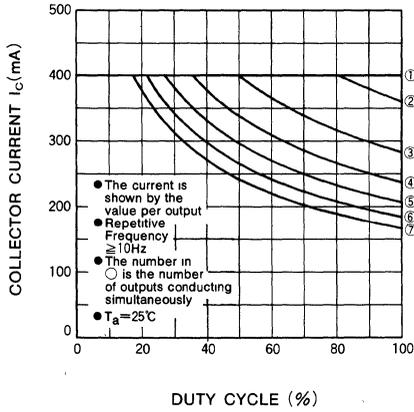


OUTPUT CURRENT CHARACTERISTICS

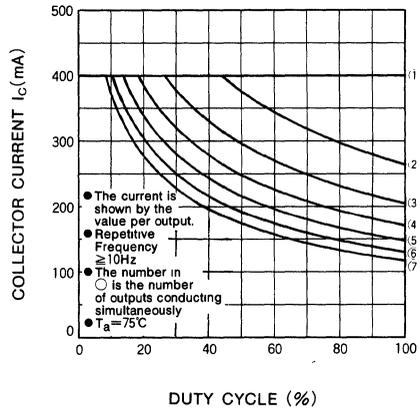


**7-UNIT 400mA DARLINGTON TRANSISTOR ARRAY
 (INPUT "L" ACTIVE)**

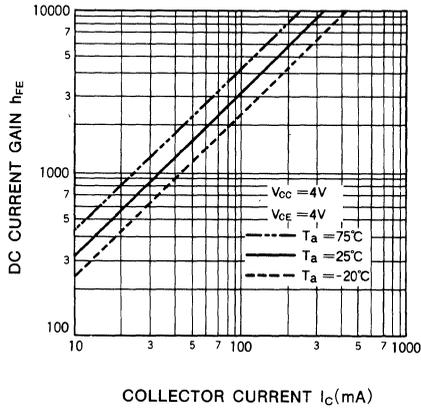
**ALLOWABLE COLLECTOR CURRENT
 AS A FUNCTION OF DUTY CYCLE**



**ALLOWABLE COLLECTOR CURRENT
 AS A FUNCTION OF DUTY CYCLE**



**DC CURRENT GAIN
 CHARACTERISTICS**



M54565P

8-UNIT 50mA TRANSISTOR ARRAY
(INPUT "L" ACTIVE)

DESCRIPTION

The M54565P, 8-channel sink driver, consists of 7 PNP and 7 NPN transistors connected to form high current gain driver pairs.

FEATURES

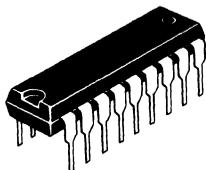
- Output breakdown voltage to 20V
- Output sink current to 50mA
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)
- "L" Active Input

APPLICATION

LED or incandescent display driver, Interfacing for standard MOS/BIPOLAR logics

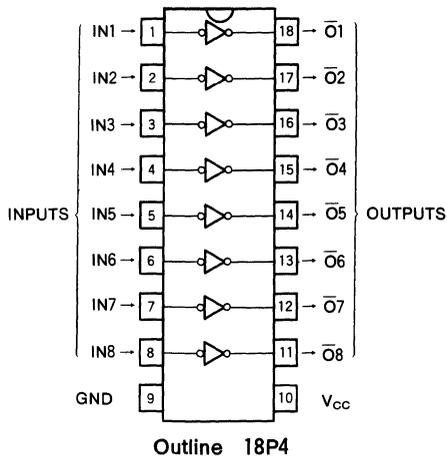
FUNCTION

The M54565P is comprised of eight PNP-NPN non darlington sink drivers. It functions from 2 V of supply voltage and features low output saturation voltage. The output is turned ON by switching the input low.

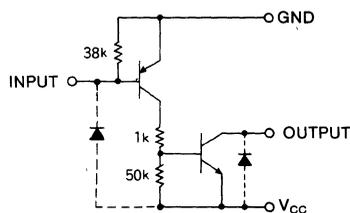


18-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		10	V
V_{CEO}	Output sustaining voltage	Transistor OFF	$-0.5 \sim +20$	V
I_C	Collector current	Transistor ON	50	mA
V_I	Input voltage		$0 \sim V_{CC}$	V
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

**8-UNIT 50mA TRANSISTOR ARRAY
 (INPUT "L" ACTIVE)**

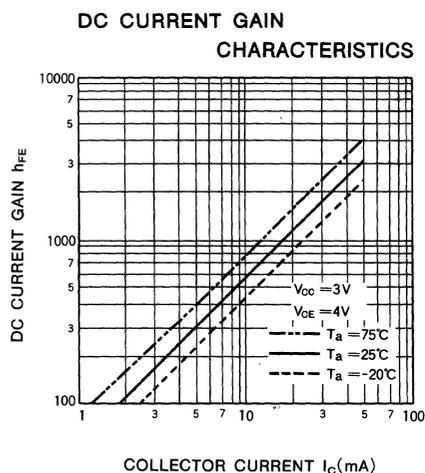
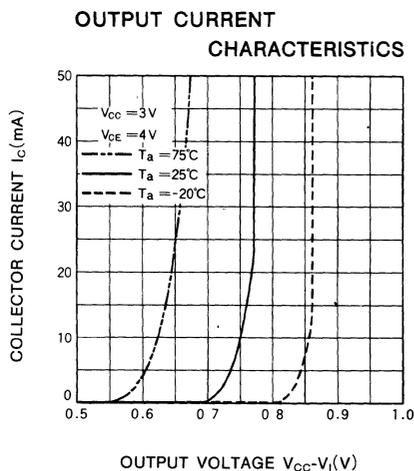
RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{CC}	Supply voltage	2		6	V
V_O	Output voltage	0		20	V
I_C	Collector current	0		20	mA
I_{IH}	"H" Input current	-8		8	μA
I_{IL}	"L" Input current $I_o = 40\text{mA}$	-200		-5000	μA

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{O(Leak)}$	Output leakage current	$V_{CC} = 6\text{V}$, $V_O = 20\text{V}$			50	μA
$V_{CE(sat)}$	Output saturation voltage	$V_{CC} = 3\text{V}$ $I_i = -200\mu\text{A}$		$I_C = 20\text{mA}$ $I_C = 40\text{mA}$	0.03 0.17	V
V_i	Input voltage	$V_{CC} = 2\text{V}$, $I_i = -200\mu\text{A}$	1	1.25		V
I_{CC}	Supply current	$V_{CC} = 3\text{V}$, $I_i = -200\mu\text{A}$		2.3	4	mA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}$, $V_{CC} = 3\text{V}$, $I_C = 40\text{mA}$, $T_a = 25^\circ\text{C}$	800	2500		—

TYPICAL CHARACTERISTICS



M54583P

8-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54583P, 8-channel source driver, is composed of 16 NPN and PNP current sink darlington transistors which form high current gain driver pairs at low input current.

FEATURES

- High output sustaining voltage to 50V
- High output source current to 400mA
- "L" active level input
- Internal input diodes
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Interfacing for microcomputer and high voltage and high current driver system, Interfacing for standard MOS/ BIPOLAR logics, Relay

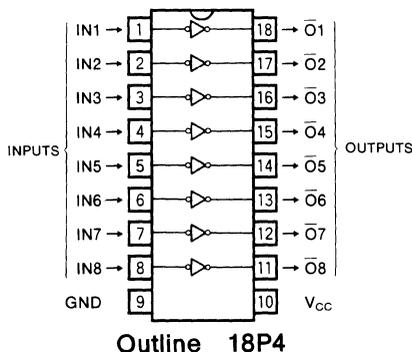
FUNCTION

The driver of the M54583P is composed of an input circuit of the M54523P with additional PNP transistors and "L" active input. A resistor of $7k\Omega$ is connected between the input and the base of PNP transistors. The input diode is intended to prevent the flow of current from the input to the V_{CC} . Without this diode, the current flows from "H" input to the V_{CC} and the "L" input circuit is activated, in such a case where one of the inputs of the 8 circuits is "H" and the others are "L" to save power consumption. The diode is inserted to prevent such misoperation.

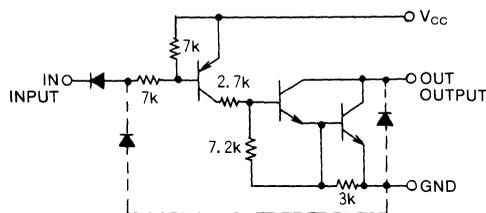
The outputs are capable of driving 400mA and are rated for operation with output voltage up to 50V.

This device is most suitable for a driver using NMOS IC output, especially for the driver of current sink.

PIN CONFIGURATION (TOP VIEW)

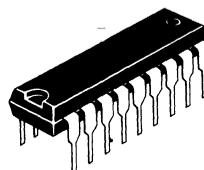


CIRCUIT SCHEMATIC (EACH CIRCUIT)



V_{CC} and GND are common to the 8 pairs
The diodes shown by broken line are parasite diodes and must not be used

Unit : Ω



18-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		$-0.5 \sim +10$	V
V_{CEO}	Output sustaining voltage	Output is in "H"	$-0.5 \sim +50$	V
V_i	Input voltage		$-0.5 \sim V_{CC}$	V
I_c	Collector current	Per channel current at "L" output	400	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.79	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

8-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

RECOMMENDED OPERATING CONDITONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	5	8	V
I_C	Collector current per channel	Percent duty cycle less than 10%, $V_{CC}=5\text{V}$	0		350	mA
		Percent duty cycle less than 34%, $V_{CC}=5\text{V}$	0		200	
V_{IH}	"H" Input voltage	$I_{O(\text{leak})} \leq 50\mu\text{A}$	$V_{CC}-0.7$		V_{CC}	V
V_{IL}	"L" Input voltage	$I_C \leq 350\text{mA}$	0		$V_{CC}-3.6$	V

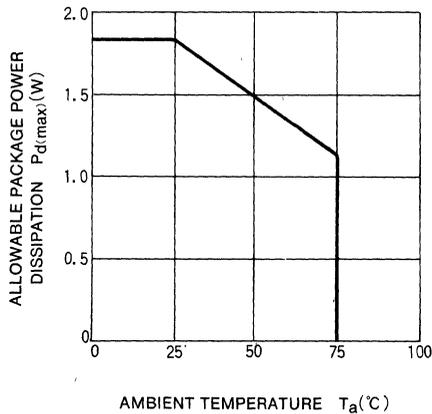
ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits		Unit		
			Min	Typ* , Max			
$I_{O(\text{leak})}$	Output leakage current	$V_{CE0}=50\text{V}$			100	μA	
$V_{CE(\text{sat})}$	Output saturation voltage	$V_I = V_{CC} - 3.6\text{V}$				V	
							$I_C = 350\text{mA}$
					0.98	1.6	
I_I	Input current	$V_I = V_{CC} - 3.6\text{V}$			-320	-600	μA
I_{CC}	Supply current (an only input)	$V_{CC}=5\text{V}, V_I = V_{CC} - 3.6\text{V}$				3	mA
h_{FE}	DC forward current gain	$V_{CE}=4\text{V}, V_{CC}=5\text{V}, I_C=350\text{mA}, T_a=25^\circ\text{C}$	2000				—

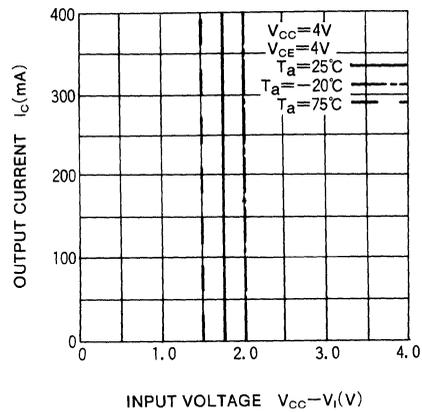
* : A typical value is at $T_a = 25^\circ\text{C}$.

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION

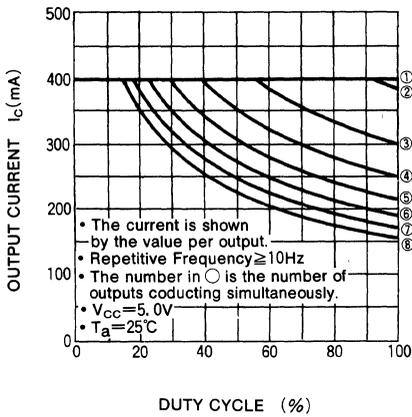


OUTPUT CURRENT CHARACTERISTICS

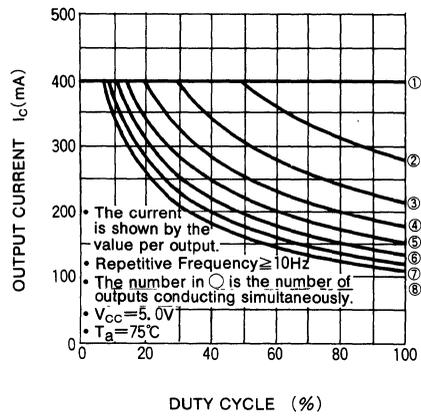


8-UNIT 400mA DARLINGTON TRANSISTOR ARRAY

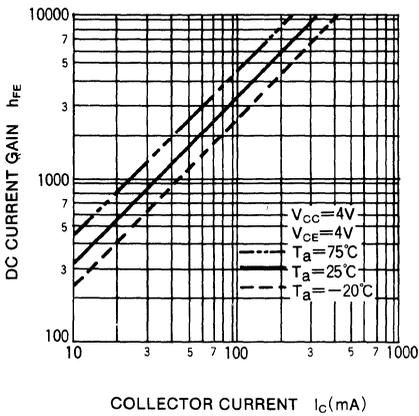
ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54568L

4-UNIT 30mA PNP TRANSISTOR ARRAY

DESCRIPTION

The M54568L, general purpose transistor array, consists of 4 PNP transistors connected in a common-emitter configuration.

FEATURES

- 20V breakdown
- 30mA output source current capability
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

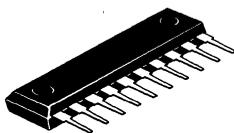
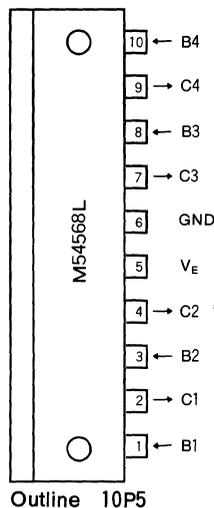
APPLICATION

LED or incandescent display driver

FUNCTION

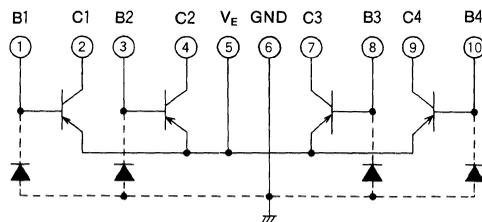
The M54568L is comprised of 4 PNP transistors. ALL emitters are connected to pin 5. Each transistor is capable of switching 30mA and will withstand 20V in the OFF state.

PIN CONFIGURATION (TOP VIEW)



10-pin molded plastic SIP

CIRCUIT SCHEMATIC



ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CBO}	Collector-base sustaining voltage	Base voltage : 0V	-40	V
V_{EBO}	Emitter-base sustaining voltage	Base voltage : 0V	-40	V
V_{CEO}	Collector-emitter sustaining voltage	Emitter voltage : 0V	-20	V
I_C	Collector current per transistor		-30	mA
I_B	Base current per transistor		-20	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1000	mW
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

4-UNIT 30mA PNP TRANSISTOR ARRAY

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

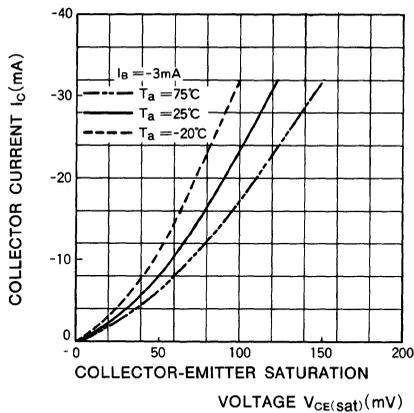
Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
I_C	Collector current $I_B = -3\text{mA}$	0		-20	mA
I_B	Base current	0		-10	mA
V_E	Emitter current	-0.3		20	V
V_B	Base voltage	-0.3		V_E	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

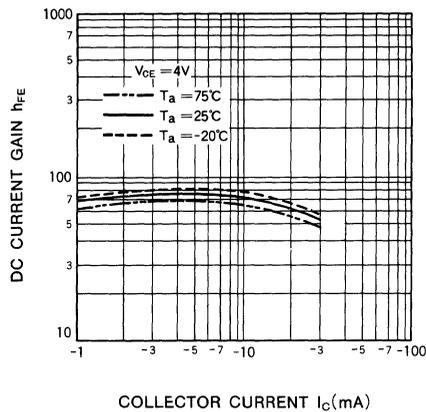
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CBO}$	Collector-emitter sustaining voltage	$I_C = -10\mu\text{A}$, $V_B = 0\text{V}$, V_E : OPEN	-40			V
$V_{(BR)EBO}$	Emitter-base sustaining voltage	$I_E = -10\mu\text{A}$, $V_B = 0\text{V}$, V_C : OPEN	-40			V
$V_{(BR)CEO}$	Collector-emitter sustaining voltage	$I_C = -100\mu\text{A}$, $V_E = 0\text{V}$, V_B : OPEN	-20			V
$V_{CE(sat)}$	Collector-emitter saturation voltage	$I_C = -20\text{mA}$, $I_B = -3\text{mA}$, $V_E = 5\text{V}$		-0.09	-0.3	V
		$I_C = -2\text{mA}$, $I_B = -0.2\text{mA}$, $V_E = 5\text{V}$		-0.02	-0.28	
h_{FE}	DC forward current gain	$V_{CE} = -4\text{V}$ $T_a = 25^\circ\text{C}$	$I_C = -2\text{mA}$	20	80	—
			$I_C = -20\text{mA}$	15	60	

TYPICAL CHARACTERISTICS

OUTPUT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS



M54560P

7-UNIT 150mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54560P, 7-channel source driver, consists of 7 PNP and 7 NPN transistors, connected to form high current gain driver with PNP action.

FEATURES

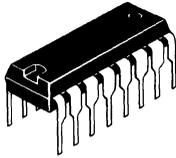
- High output sustaining voltage to 40V
- Output source current to 150mA
- Integral diode for transient suppression
- Active "L" input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED, incandescent or fluorescent display driver, Interfacing for standard MOS/BIPO-LAR logics

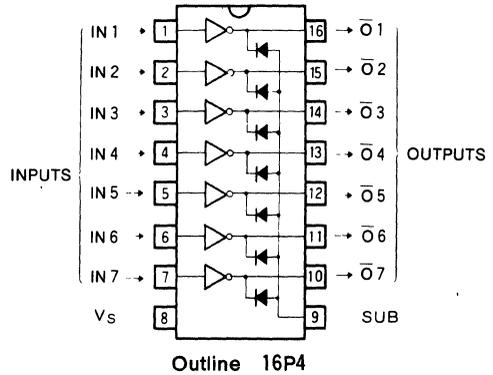
FUNCTION

The M54560P is comprised of seven PNP-NPN darlington source driver pairs with $20\text{k}\Omega$ series input resistors. Each output has an integral diode for inductive load transient suppression. The anodes of the diodes and the substrate connected together to pin 9. The outputs are capable of driving 150mA and are rated for operation with output voltages of up to 40V. The output is turned ON by switching the input low.

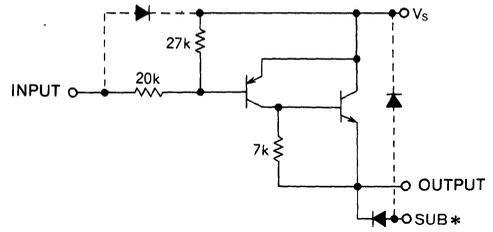


16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Output is in "L"	$-0.5 \sim +40$	V
V_S	Supply voltage		40	V
V_I	Input voltage		$0 \sim +40$	V
I_O	Output current	Per channel current at "H" output	-150	mA
I_F	Clamp diode forward current		-150	mA
V_R	Clamp diode reverse voltage		40	V
P_D	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

7-UNIT 150mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

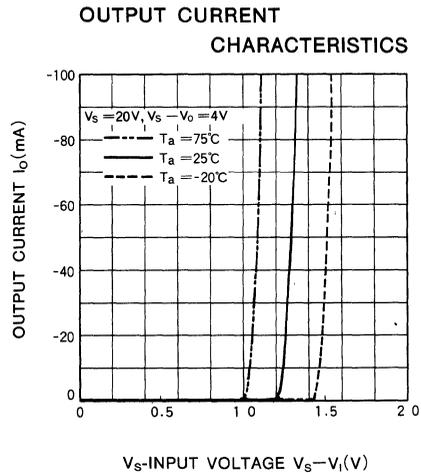
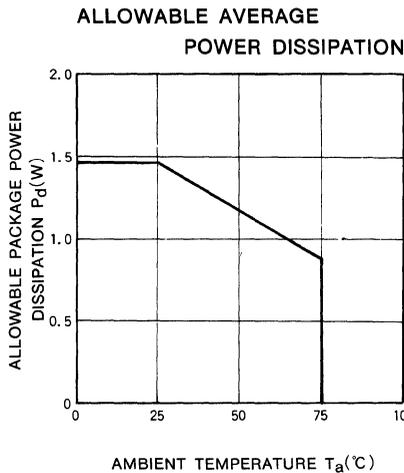
RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_S	Supply voltage		0		40	V
I_O	Output current per channel	Percent duty cycle less than 90%	0		-100	mA
		Percent duty cycle less than 100%	0		-50	
V_{IH}	"H" input voltage		$V_S - 0.2$		$V_S + 0.3$	V
V_{IL}	"L" input voltage	$I_O = -100\text{mA}$	0		$V_S - 5$	V
		$I_O = -50\text{mA}$	0		$V_S - 3.5$	

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

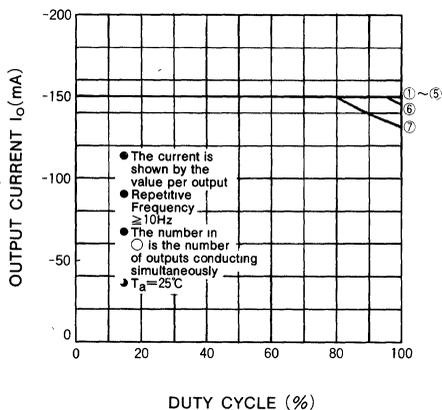
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{S(\text{leak})}$	Supply leakage current	$V_S = 40\text{V}$			100	μA
$V_{CE(\text{sat})}$	Output saturation voltage	$V_I = V_S - 5\text{V}, I_O = -100\text{mA}$		0.82	1.5	V
		$V_I = V_S - 3.5\text{V}, I_O = -50\text{mA}$		0.75	1.2	
I_I	Input voltage	$V_I = V_S - 8.5\text{V}$		-380	-670	μA
V_F	Clamp diode forward voltage	$I_F = -100\text{mA}$		-1.1	-2.4	V
V_R	Clamp diode reverse voltage	$I_R = 100\mu\text{A}$	40			V
h_{FE}	DC forward current gain	$V_S - V_O = 4\text{V}, I_O = -100\text{mA}, T_a = 25^\circ\text{C}$	500	2800		—

TYPICAL CHARACTERISTICS

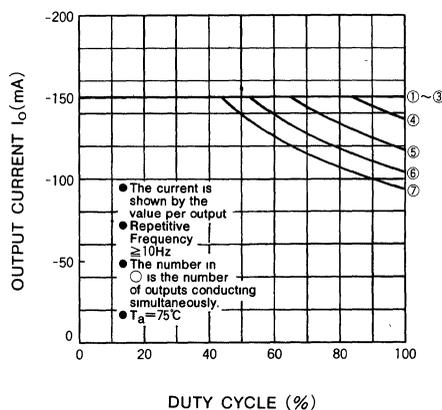


7-UNIT 150mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

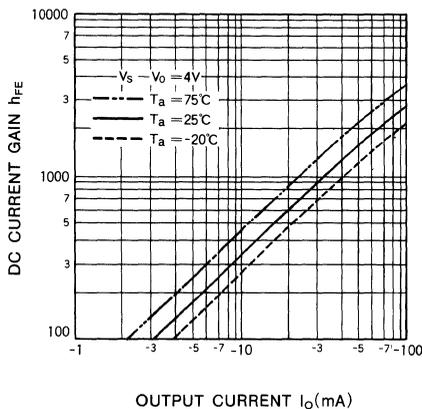
ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54561P

7-UNIT 300mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54561P, 7-channel source driver, consists of 7 PNP and 14 NPN transistors connected to form high current gain driver with PNP action.

FEATURES

- High output sustaining voltage to 40V
- High output source current to 300mA
- Integral diode for transient suppression
- Active "L" input
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

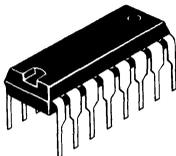
Relay and printer driver, LED, incandescent or fluorescent display driver, Active "L" input, Interfacing for standard MOS/BIPOLAR logics

FUNCTION

The M54561P functions like a PNP transistor and the compound PNP/NPN/NPN output provides high current gain. Each output has an integral diode for inductive load transient suppression and the anodes of the diodes and the substrate are connected together to pin 9.

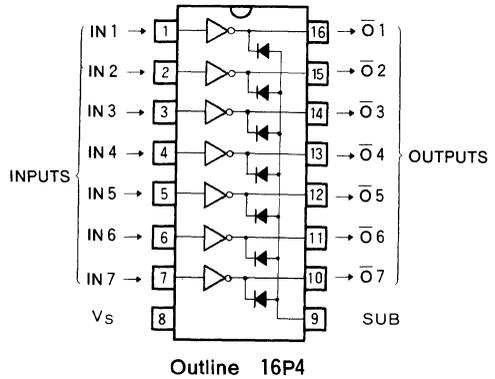
The output are capable of driving 300mA and are rated for operation with output voltage up to 40V.

The output is turned ON by switching the input low.

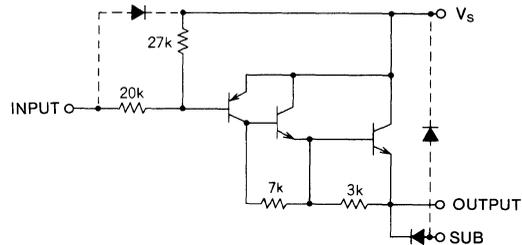


16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Output is in "L"	$-0.5 \sim +40$	V
V_S	Supply voltage		40	V
V_I	Input voltage		$0 \sim +40$	V
I_O	Output current	Per channel current at "H" output	-300	mA
I_F	Clamp diode forward current		-300	mA
V_R	Clamp diode reverse voltage		40	V
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{Opr}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

7-UNIT 300mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

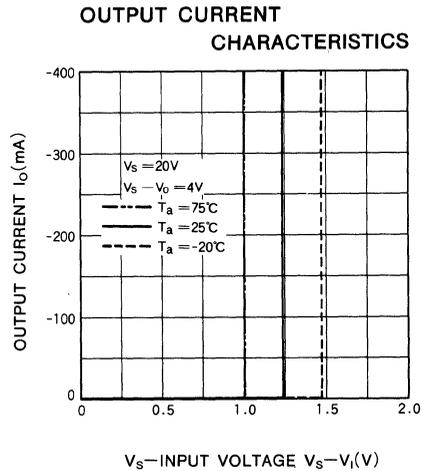
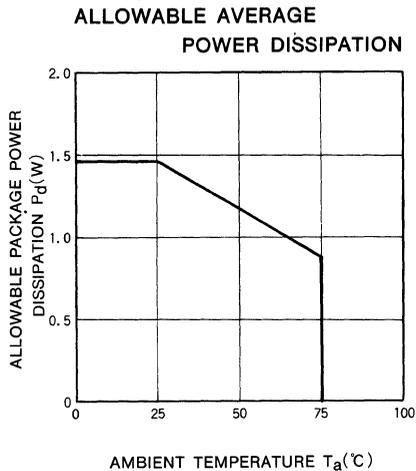
RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_S	Supply voltage		0		40	V
I_O	Output current per channel	Percent duty cycle less than 15%	0		-250	mA
		Percent duty cycle less than 50%	0		-100	
V_{IH}	"H" Input voltage		$V_S - 0.2$		$V_S + 0.3$	V
V_{IL}	"L" Input voltage		0		$V_S - 3$	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

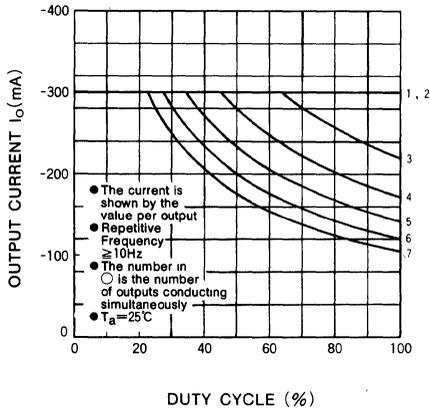
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{S(\text{leak})}$	Supply leakage current	$V_S = 40\text{V}$			100	μA
$V_{CE(\text{sat})}$	Output saturation voltage	$V_I = V_S - 3\text{V}, I_O = -250\text{mA}$		1.6	2.3	V
		$V_I = V_S - 3\text{V}, I_O = -100\text{mA}$		1.45	2.0	
I_I	Input current	$V_I = V_S - 3.5\text{V}$		-150	-250	μA
V_F	Clamp diode forward voltage	$I_F = -300\text{mA}$		-1.6	-2.4	V
V_R	Clamp diode reverse voltage	$I_R = 100\mu\text{A}$	40			V
h_{FE}	DC forward current gain	$V_S - V_O = 4\text{V}, I_O = -300\text{mA}, T_a = 25^\circ\text{C}$	1000	8000		—

TYPICAL CHARACTERISTICS

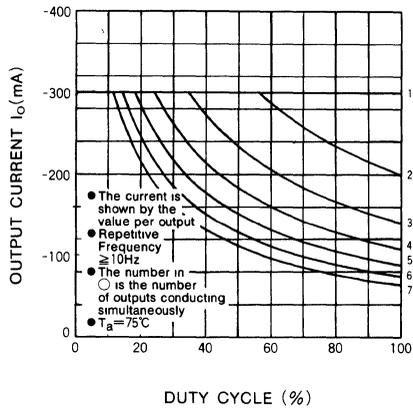


7-UNIT 300mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

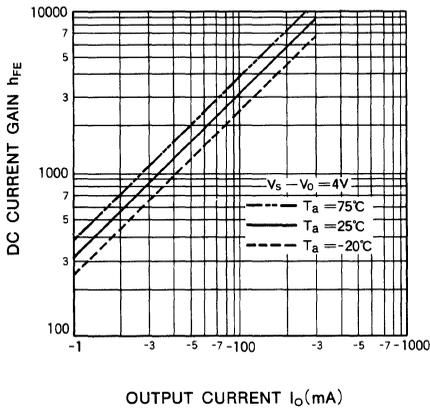
ALLOWABLE OUTPUT CURRENT AS A FUNCTIONAL OF DUTY CYCLE



ALLOWABLE OUTPUT CURRENT AS A FUNCTIONAL OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54580P

7-UNIT 150mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The M54580P, 7-channel source driver, consists of 7 PNP and 7 NPN transistors connected to form high current gain driver with PNP action.

FEATURES

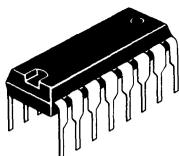
- High output sustaining voltage to 50V
- High output source current to 150mA
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

APPLICATION

Relay and printer driver, LED, incandescent or fluorescent display driver, Interfacing for standard MOS/BIPO-LAR logics

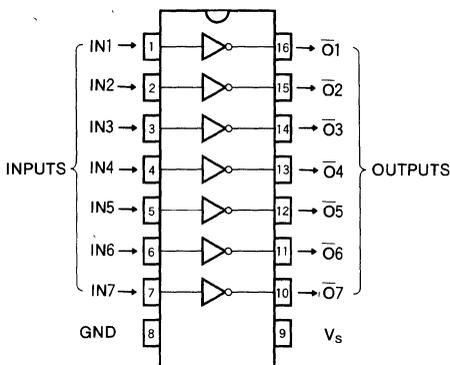
FUNCTION

The M54580P is comprised of seven PNP-NPN darlington source driver pairs with a diode and 7 k Ω resistor in series to the input. The output is turned ON by switching the input low. Each output has 50k Ω pull-down resistor suitable for driving fluorescent displays. The outputs are capable of driving 100mA and are rated for operation with output voltage up to 50V.



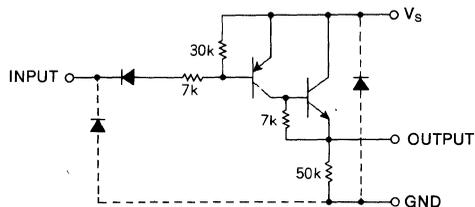
16-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



Outline 16P4

CIRCUIT SCHEMATIC



Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_s	Supply voltage		50	V
V_{CEO}	Output sustaining voltage	Transistor OFF	-0.5 ~ +50	V
V_i	Input voltage		0 ~ V_s	V
I_o	Output current	Transistor OFF	-150	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ\text{C}$

7-UNIT 150mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

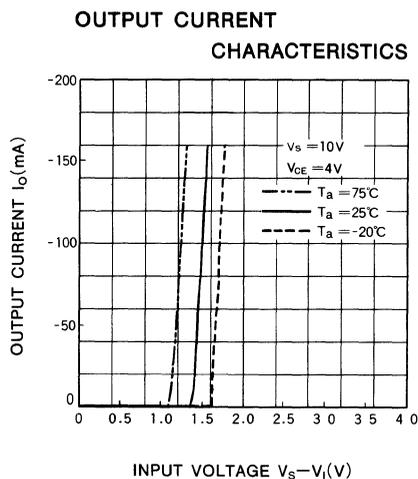
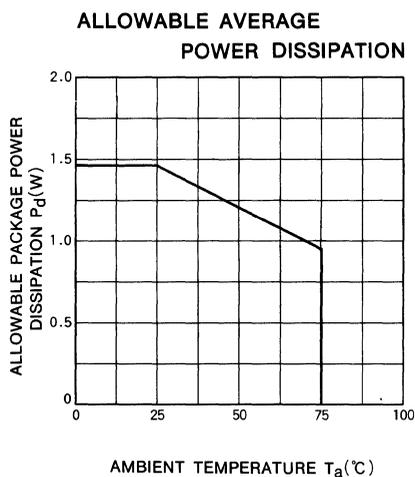
RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_S	Supply voltage		4		50	V
I_O	Output current per channel	All outputs conducting simultaneously Percent duty cycle less than 85%	0		-100	mA
		All outputs conducting simultaneously Percent duty cycle less than 100%	0		-50	
V_{IH}	"H" Input voltage	$I_{O(\text{leak})} = 50\mu\text{A}$	$V_S - 0.4$		V_S	V
V_{IL}	"L" Input voltage	$I_O = -100\text{mA}$	0		$V_S - 3.2$	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

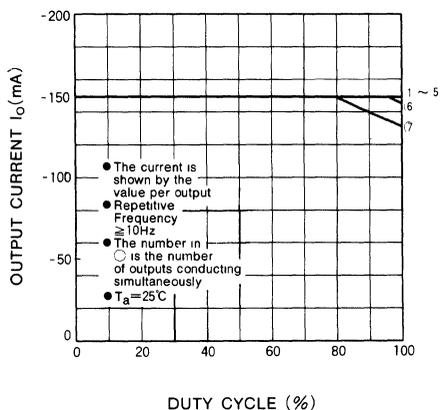
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CEO}$	Output sustaining voltage	$I_{CEO} = 100\mu\text{A}$	50			V
$V_{CE(\text{sat})}$	Output saturation voltage	$V_I = V_S - 3.2\text{V}$	$I_O = -100\text{mA}$	0.9	1.5	V
			$I_O = -50\text{mA}$	0.8	1.2	
I_I	Input current	$V_I = V_S - 3.5\text{V}$		-0.3	-0.6	mA
		$V_I = V_S - 6\text{V}$		-0.65	-0.95	
I_R	Input leakage current	$V_I = 40\text{V}$			100	μA
h_{FE}	DC forward current gain	$V_{CE} = 4\text{V}, V_S = 10\text{V}, I_C = -100\text{mA}, T_a = 25^\circ\text{C}$	800	3000		—

TYPICAL CHARACTERISTICS

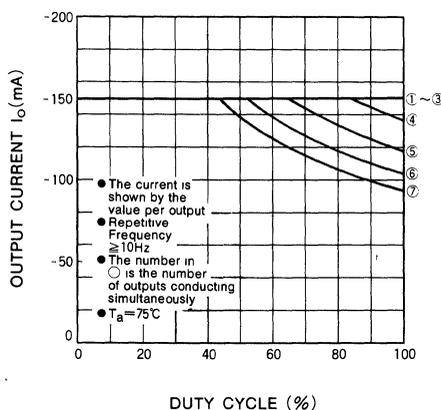


7-UNIT 150mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

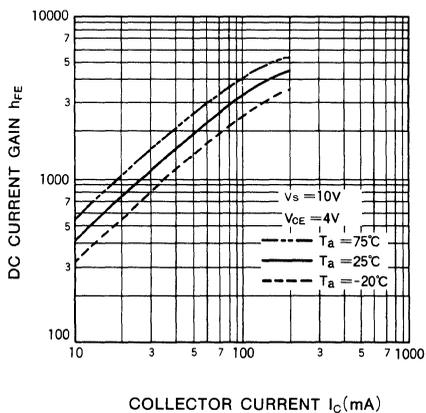
ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



DC CURRENT GAIN CHARACTERISTICS



M54581P

8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

DESCRIPTION

The M54581P, 8-channel source driver, consists of 8 NPN and 8 PNP source type darlington transistors connected to form high current gain driver.

FEATURES

- High output sustaining voltage to 50V ($BV_{CEO} > 50V$)
- High output source current to 500mA
($I_{O(max)} = -500mA$)
- "L" active input level
- Internal input diode
- Integral clamp diode for transient suppression
- Wide operating temperature range ($T_a = -20 \sim +75^\circ C$)

APPLICATION

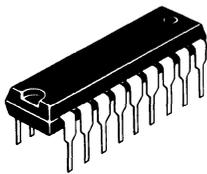
Relay and printer driver, LED or incandescent or fluorescent display driver, Interfacing for standard MOS/BIPOLAR logics and interfacing for relay, solenoid or small printer

FUNCTION

The M54581P is composed of 8 PNP and 8 NPN source type darlington transistors. A diode and a resistor of 7k Ω is connected between the input pin and the base of PNP transistor. The emitter and the collector of NPN transistor are connected to V_S (pin 9), and the output NPN transistors are in darlington configuration. An integral clamp diode is inserted between each output and GND, and V_S (pin 9) and GND (pin 10) are common to the 8 circuits.

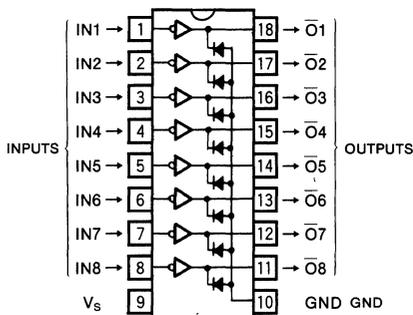
The outputs are capable of driving 500mA and are rated for operation with output voltage up to 50V.

The device is activated with "L" level input.



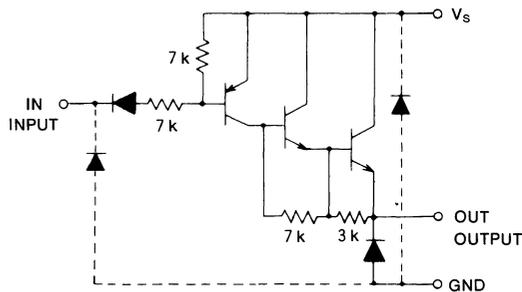
18-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



Outline 18P4

CIRCUIT SCHEMATIC (EACH CIRCUIT)



V_S and GND are common to the 8 circuits

The diodes shown by broken line are parasite diodes and must not be used

Unit : Ω

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ C$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CEO}	Output sustaining voltage	Output is in "L"	-0.5 ~ +50	V
V_S	Supply voltage		-0.5 ~ +50	V
V_I	Input voltage		0, $V_S - 30$	V
I_O	Output current	Per channel current at "H" output	-500	mA
I_F	Clamp diode forward current	Per channel current	-500	mA
V_R	Clamp diode reverse voltage		-0.5 ~ +50	V
P_d	Power dissipation	$T_a = 25^\circ C$	1.79	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ C$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ C$

M54581P

8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 + 75^\circ\text{C}$, unless otherwise noted)

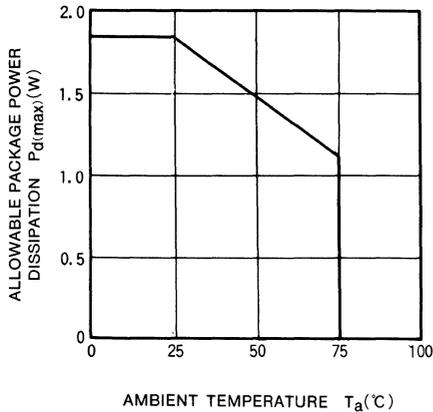
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_S	Supply voltage		0		50	V
I_O	Output current per channel	Percent duty cycle less than 8%	0		-350	mA
		Percent duty cycle less than 55%	0		-100	
V_{IH}	"H" Input voltage	$I_{O(\text{leak})} = -50\mu\text{A}$	$V_S - 0.7$		V_S	V
V_{IL}	"L" Input voltage	$I_C = -350\text{mA}$	0		$V_S - 3.6$	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 + 75^\circ\text{C}$, unless otherwise noted)

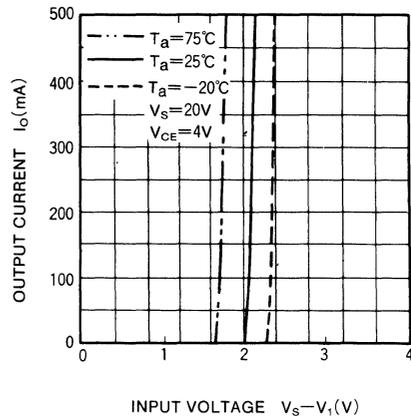
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CEO}	Output leakage current	$V_{CE0} = 50\text{V}$			-100	μA
$V_{CE(\text{sat})}$	Output saturation voltage	$V_I = V_S - 3.2\text{V}$, $I_O = -100\text{mA}$			2.0	V
		$V_I = V_S - 3.6\text{V}$, $I_O = -350\text{mA}$			2.4	
I_I	Input current	$V_I = V_S - 3.6\text{V}$			-600	μA
		$V_I = V_S - 15\text{V}$			-3.2	mA
V_R	Clamp diode reverse voltage	$I_R = 100\mu\text{A}$			50	V
V_F	Clamp diode forward voltage	$I_F = -350\text{mA}$			-2.4	V

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE
POWER DISSIPATION

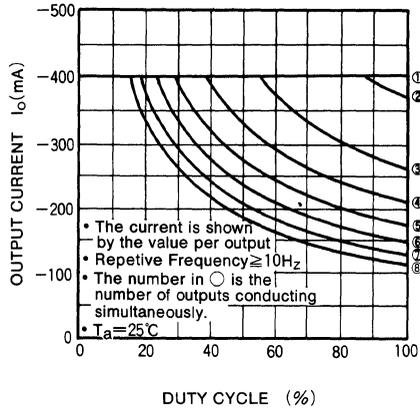


OUTPUT CURRENT CHARACTERISTICS

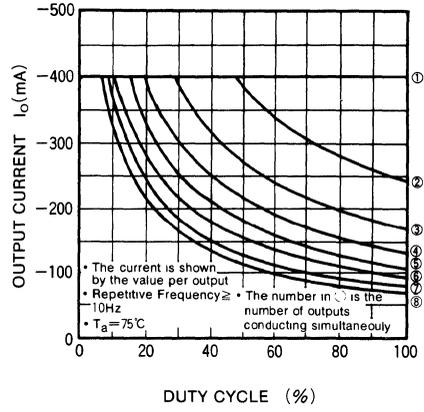


**8-UNIT 500mA SOURCE TYPE
 DARLINGTON TRANSISTOR ARRAY WITH CLAMP DIODE**

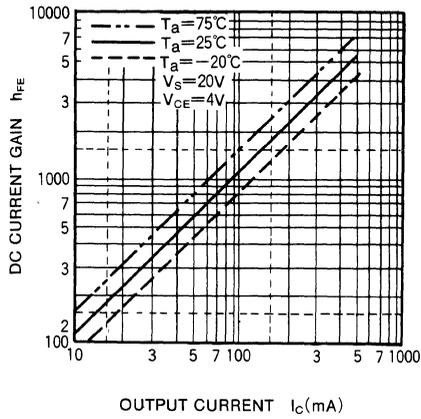
**ALLOWABLE OUTPUT CURRENT
 AS A FUNCTION OF DUTY CYCLE**



**ALLOWABLE OUTPUT CURRENT
 AS A FUNCTION OF DUTY CYCLE**



**CURRENT GAIN VS
 OUTPUT CURRENT**



M54586P

8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

DESCRIPTION

The 54586P, 8-channel source driver, consists of 8 NPN and 8 PNP source type darlington transistors connected to form high current gain driver.

FEATURES

- High output sustaining voltage to 50V ($BV_{CEO} > 550V$)
- High output source current to 500mA
($I_{O(max)} = -500mA$)
- "L" active input level
- Internal input diode
- Wide operating temperature range ($T_a = -20 \sim +75^\circ C$)

APPLICATION

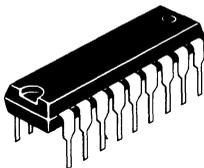
Relay and printer driver, LED or incandescent or fluorescent display driver, Interfacing for standard MOS/BIPOLAR logics and interfacing for relay, solenoid or small printer

FUNCTION

The M54586P is composed of 8 PNP and 8 NPN source type darlington transistors. A diode and a resistor of $7k\Omega$ is connected between the input pin and the base of PNP transistors. The emitter of the transistor and the collector of NPN transistor are connected to V_s (pin 9), and a resistor of $50k\Omega$ is connected between each output pin and GND pin (pin 10).

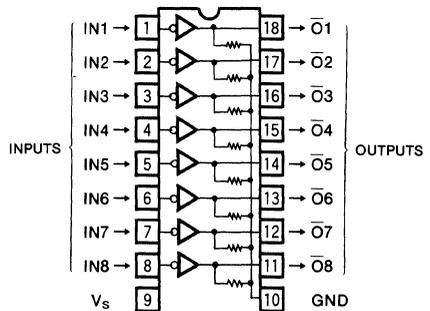
The outputs are capable of driving 500mA and are rated for operation with output voltage up to 50V.

The device is activated with "L" level input.



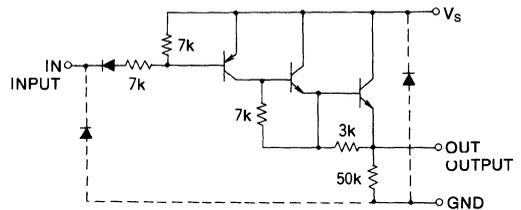
18-pin molded plastic DIP

PIN CONFIGURATION (TOP VIEW)



Outline 18P4

CIRCUIT SCHEMATIC (EACH CIRCUIT)



V_s and GND are common to the 8 circuits.
The diodes shown by broken line are parasite diodes and must not be used.

Unit : Ω

18-pin molded plastic DIP

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ C$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_s	Supply voltage		-0.5 ~ +50	V
V_{CEO}	Output sustaining voltage	Output is in "L"	-0.5 ~ +50	V
V_i	Input voltage		-0.5 ~ V_s	V
I_o	Output current		-500	mA
P_d	Power dissipation	$T_a = 25^\circ C$	1.79	W
T_{opr}	Operating ambient temperature range		-20 ~ +75	$^\circ C$
T_{stg}	Storage temperature range		-55 ~ +125	$^\circ C$

8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

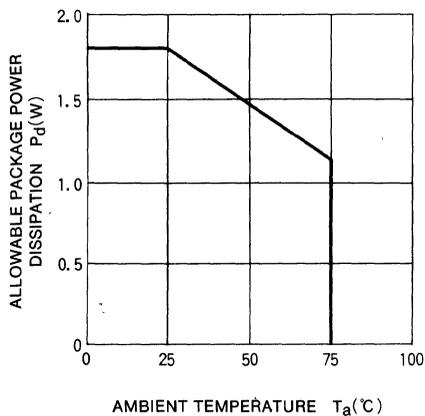
Symbol	Parameter		Limits			Unit
			Min	Typ	Max	
V_S	Supply voltage		4		50	V
I_O	Output current per channel when 8 outputs are conducting simultaneously	Percent duty cycle less than 8%	0		-350	mA
		Percent duty cycle less than 60%	0		-100	
V_{IH}	"H" Input voltage	$I_{O(\text{leak})} \geq 50\mu\text{A}$	$V_S - 0.7$		V_S	V
V_{IL}	"L" Input voltage	$I_O \geq -350\text{mA}$	0		$V_S - 3.6$	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

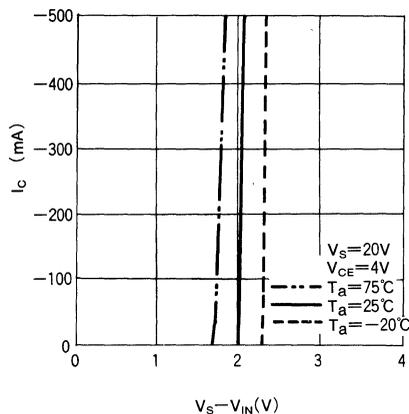
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{O(\text{leak})}$	Output leakage voltage	$V_{CEO} = 50\text{V}$			-100	μA
$V_{CE(\text{s}\hat{\text{a}}\text{t})}$	Output saturation voltage	$V_I = V_S - 3.2\text{V}, I_O = -100\text{mA}$		1.6	2.0	V
		$V_I = V_S - 3.6\text{V}, I_O = -350\text{mA}$		1.8	2.4	
I_i	Input current	$V_I = V_S - 3.6\text{V}$		-320	-600	μA
		$V_I = V_S - 15\text{V}$		-1.6	-3.2	mA
h_{FE}	Collector-emitter saturation voltage	$V_{CE} = 4\text{V}, V_S = 20\text{V}, I_O = -350\text{mA}, T_a = 25^\circ\text{C}$	800	3500		—

TYPICAL CHARACTERISTICS

ALLOWABLE AVERAGE POWER DISSIPATION

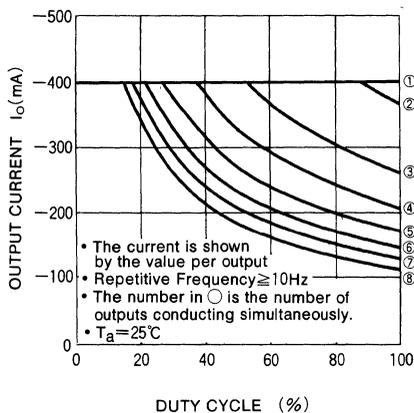


OUTPUT CURRENT CHARACTERISTICS

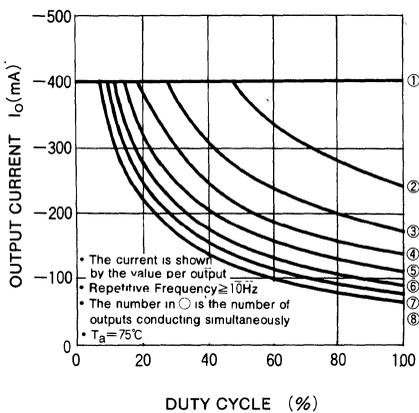


8-UNIT 500mA SOURCE TYPE DARLINGTON TRANSISTOR ARRAY

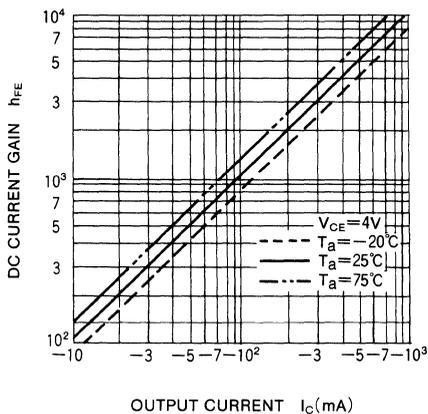
ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



ALLOWABLE OUTPUT CURRENT AS A FUNCTION OF DUTY CYCLE



CURRENT GAIN VS OUTPUT CURRENT



M54569P

8-UNIT 30mA PNP TRANSISTOR ARRAY

DESCRIPTION

The M54569P, general purpose transistor array, consists of 8 PNP transistors connected in a common-emitter configuration.

FEATURES

- 20V breakdown
- 30mA output source current capability
- Wide operating temperature range ($T_a = -20 \sim +75^\circ\text{C}$)

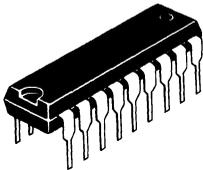
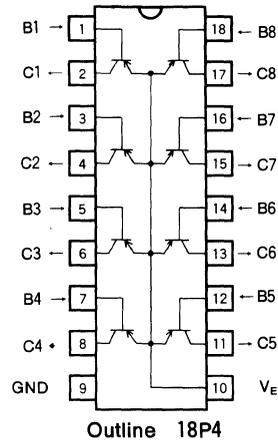
APPLICATION

LED or incandescent display driver

FUNCTION

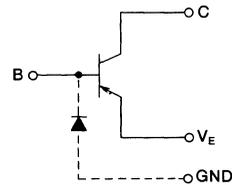
The M54569P is comprised of 8 PNP transistors. All emitters are connected to pin 10. Each transistor is capable of switching 30mA and will withstand 20V in the OFF state.

PIN CONFIGURATION (TOP VIEW)



18-pin molded plastic DIP

CIRCUIT SCHEMATIC



ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CBO}	Collector-base sustaining voltage	Base voltage : 0V	-40	V
V_{EBO}	Emitter-base sustaining voltage	Base voltage : 0V	-40	V
V_{CEO}	Collector-emitter sustaining voltage	Emitter voltage : 0V	-20	V
I_C	Collector current per transistor		-30	mA
I_B	Base current per transistor		-20	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.47	W
T_{OPR}	Operating ambient temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

8-UNIT 30mA PNP TRANSISTOR ARRAY

RECOMMENDED OPERATIONAL CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

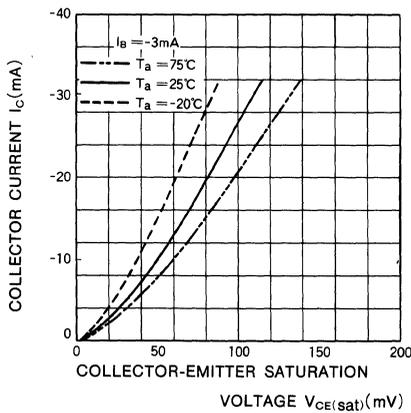
Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
I_C	Collector current $I_B = -3\text{mA}$	0		-20	mA
I_B	Base current	0		-10	mA
V_E	Emitter voltage	-0.3		20	V
V_B	Base voltage	-0.3		V_E	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

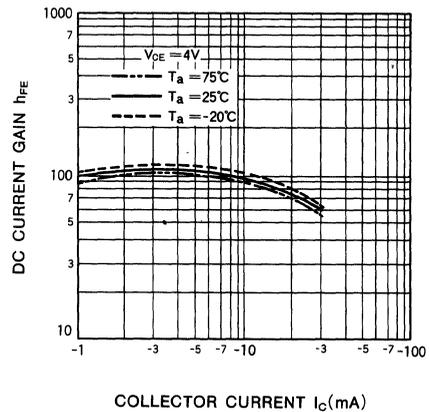
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{(BR)CBO}$	Collector-emitter sustaining voltage	$I_C = -10\mu\text{A}$, $V_B = 0\text{V}$ V_E : OPEN	-40			V
$V_{(BR)EBO}$	Emitter-base sustaining voltage	$I_E = -10\mu\text{A}$, $V_B = 0\text{V}$ V_C : OPEN	-40			V
$V_{(BR)CEO}$	Collector-emitter sustaining voltage	$I_C = -100\mu\text{A}$, $V_E = 0\text{V}$ V_B : OPEN	-20			V
$V_{CE(sat)}$	Collector-emitter saturation voltage	$I_C = -20\text{mA}$, $I_B = -3\text{mA}$, $V_E = 5\text{V}$ $I_C = -2\text{mA}$, $I_B = -0.2\text{mA}$, $V_E = 5\text{V}$		-0.09	-0.3	V
h_{FE}	DC forward current gain	$V_{CE} = -4\text{V}$		20	100	—
		$T_a = 25^\circ\text{C}$	$I_C = -2\text{mA}$	15	70	

TYPICAL CHARACTERISTICS

OUTPUT CHARACTERISTICS



DC CURRENT GAIN CHARACTERISTICS

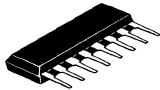


MISCELLANEOU

QUICK REFERENCE TABLE OF ELECTRONIC VOLUME CONTROL, BATTERY CHECKER

Classification	Type	Function	Features	Maximum ratings		Electrical characteristics		Package outlines	Interchangeable products
				Supply voltage range V_{CC} (V)	Power dissipation P_d (mW)	Circuit current I_{CC} (mA)	Remarks		
Electronic volume	M5222	Two-channel voltage controlled amplifier (VCA) for use in a low-voltage electronic volume control	<ul style="list-style-type: none"> ● A-curve characteristics (logarithmic response) ● Low-voltage operation ● Large attenuation. ATT=90dB 	1.8~20	800/625/440	3.6		L: 8-pin SIP P: 8-pin DIP	—
	★ M5241L	Two-channel VCA with a high withstand voltage for use in a hi-fi electronic volume control	<ul style="list-style-type: none"> ● Wide supply voltage range ● Low distortion: THD=0.02% ● Large attenuation. ATT=90dB ● Independent control capability for R and L channels 	$\pm 3 \sim \pm 18$	800	9.0		10-pin SIP	—
Battery checker	M5232L	Voltage detector/on-off alarm circuit	<ul style="list-style-type: none"> ● Built-in comparator, oscillator circuit, reference voltage generator and voltage regulating circuits ● Variable LED flash rate 	5~20	800	2.0	$V_{ref}=1.31V$ $f=1.8Hz$	8-pin SIP	—

★. New product



8-pin SIP



8-pin DIP



10-pin SIP

MITSUBISHI LINEAR ICs

M5222L, P, FP

**VOLTAGE CONTROLLED AMPLIFIER(VCA)FOR USE
IN A LOW-VOLTAGE ELECTRONIC VOLUME CONTROL**

DESCRIPTION

The M5222 is a semiconductor integrated circuit consisting of a dual voltage controlled amplifier (VCA) designed for use in an electronic volume control. Operable over a wide supply voltage range from 1.8V to 20V, the M5222 is available in a compact 8-pin SIP, DIP or FP. The two built-in channels and especially the attenuation characteristics that change logarithmically with respect to the external DC control voltage (a response equivalent to the A-curve volume) permit the M5222 to be applied widely in portable stereo radio/cassette recorders, car stereo system, and electronic musical instruments.

FEATURES

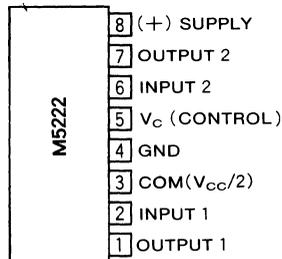
- Operable at low voltages $V_{CC}=1.8\sim 20V$
- Two built-in channels
 - Simultaneous control of both channels is possible with V_C (control) at Pin ⑤
- Logarithmic response VCA
 - Logarithmic response equivalent to A-curve volume
- Wide attenuation range
 - $0dB(V_C \approx 0) \sim -90dB(V_C \approx -270mV)$ (typ.)
- High maximum input voltage
 - $V_i=1.0V_{rms}$ (typ.) (@ $V_{CC}=3V$)
- Low distortion THD=0.05%
- Similar characteristics between the two channels

APPLICATION

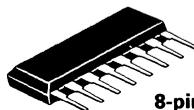
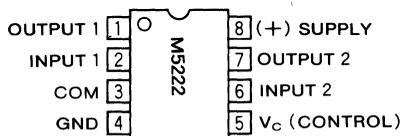
As an electronic volume control in portable stereo radio/cassette recorders, car stereos, electronic musical instruments, VCAs.

PIN CONFIGURATION (TOP VIEW)

SIP



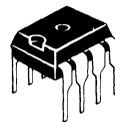
DIP, MINI FLAT



8-pin molded plastic SIP

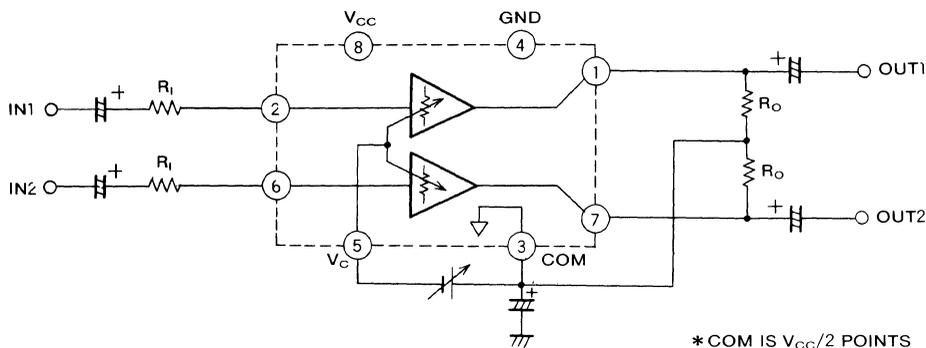


8-pin molded plastic FP
(MINI FLAT)



8-pin molded plastic DIP

BLOCK DIAGRAM



Note: 1. R_1 is used to convert input voltage to current.

2. R_O is an output resistor used to convert the current output signal to voltage. Connect this output with COM pin (3) to fix the DC output potential

3. The COM pin is used for making a 1/2 pint supply voltage within the IC. It is used in connecting R_O and in V_C control

MITSUBISHI LINEAR ICs M5222L, P, FP

VOLTAGE CONTROLLED AMPLIFIER(VCA)FOR USE IN A LOW-VOLTAGE ELECTRONIC VOLUME CONTROL

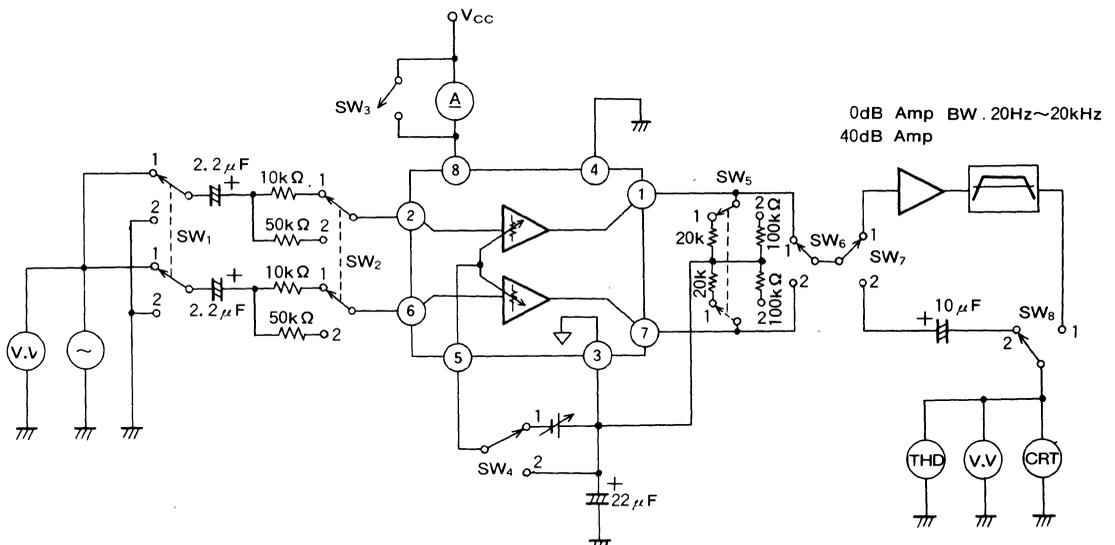
ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$ unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		20	V
P_d	Power dissipation		800(SIP)/625(DIP)/440(FP)	mW
K_{θ}	Thermal derating	$T_a \geq 25^{\circ}\text{C}$	8(SIP)/6.25(DIP)/4.4(FP)	mW/ $^{\circ}\text{C}$
T_{opr}	Operating temperature range		$-20 \sim +75$	$^{\circ}\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^{\circ}\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit	
			V_{CC}	Min	Typ		Max
I_{CC}	Circuit current	$V_i=0, V_c=0$	3V	2.5	3.6	5.5	mA
V_{iM1}	Maximum input voltage	$f=1\text{kHz}$ $V_c=0$	3V	0.7	1.0		Vrms
V_{iM2}	Maximum input voltage	THD=1%					
ATT_M	Maximum attenuation	$R_i=10\text{k}\Omega, R_o=20\text{k}\Omega$ $V_c=-270\text{mV}$	3V	80	90		dB
ATT_{01}	Attenuation error	$f=1\text{kHz}$ $V_c=0$	3V	-4.4	-1.4	+1.6	dB
ATT_{02}	Attenuation error	$V_i=0\text{dBm}$					
ΔATT	Attenuation deviation between channels	$f=1\text{kHz}, V_c=0, V_i=0\text{dBm}$ $R_i=10\text{k}\Omega, R_o=20\text{k}\Omega$	3V		0.1	3.0	dB
V_{NO1}	Noise output voltage	$V_c=0(ATT=-1.4\text{dB}), R_i=10\text{k}\Omega$ $R_o=20\text{k}\Omega, BW=20\text{Hz} \sim 20\text{kHz}$	3V		30	60	μVrms
V_{NO2}	Noise output voltage	$ATT=-40\text{dB}, R_i=10\text{k}\Omega$ $R_o=20\text{k}\Omega, BW=20\text{Hz} \sim 20\text{kHz}$	3V		5		μVrms

TEST CIRCUIT



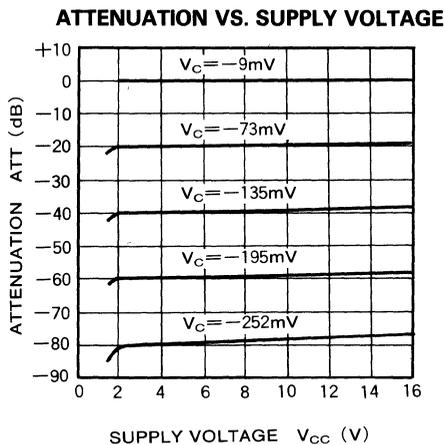
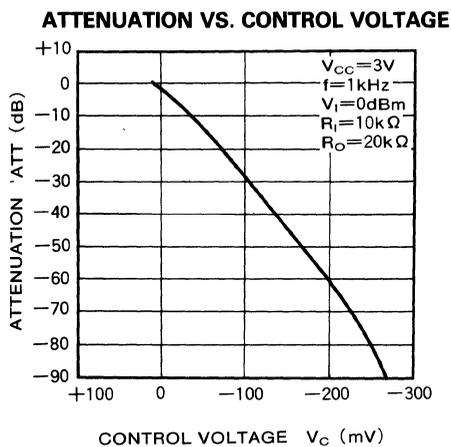
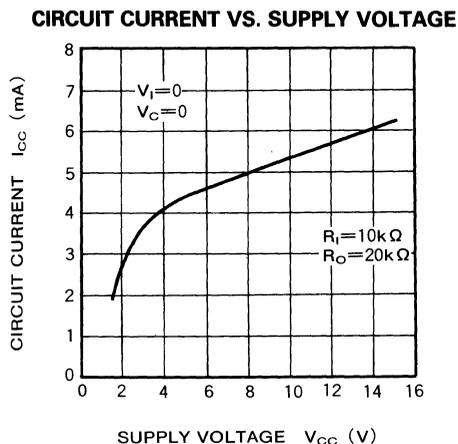
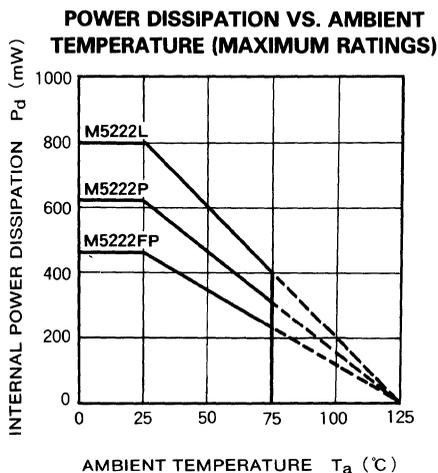
**VOLTAGE CONTROLLED AMPLIFIER(VCA)FOR USE
 IN A LOW-VOLTAGE ELECTRONIC VOLUME CONTROL**

SWITCH MATRIX

Parameter	SW ₁	SW ₂	SW ₃	SW ₄	SW ₅	SW ₆	SW ₇	SW ₈
I _{CC}	2	1	OFF	2	1	1/2	2	2
V _{IM}	1	1	1	ON	2	1	1/2	1
	2	1	2	ON	2	2	1/2	1
ATTM	1	1	ON	1	1	1/2	2	2
ATT	01	1	1	ON	2	1	1/2	2
	02	1	2	ON	2	2	1/2	2
V _{NO}	1	2	1	ON	2	1	1/2	1
	2	2	1	ON	1	1	1/2	1

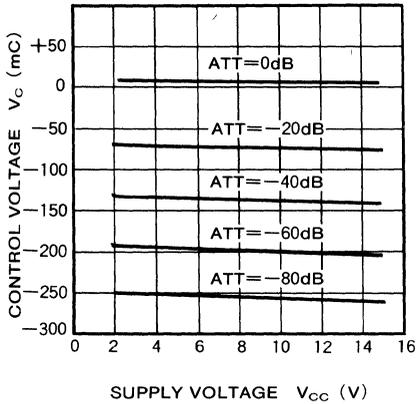
- (Note)1. Use 0dB amplification when measuring V_{IM}
 2. Use 40dB amplification when measuring V_{NO}
 3. V_{NO}=measurement value/100(40dB) (μVrms)

TYPICAL CHARACTERISTICS

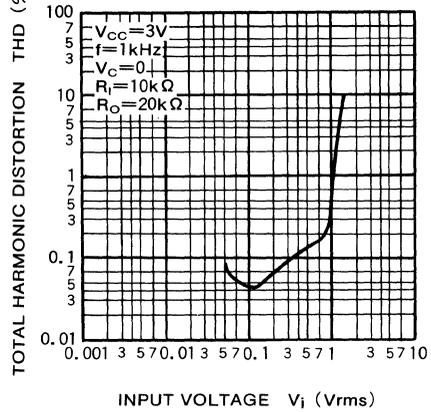


**VOLTAGE CONTROLLED AMPLIFIER(VCA)FOR USE
 IN A LOW-VOLTAGE ELECTRONIC VOLUME CONTROL**

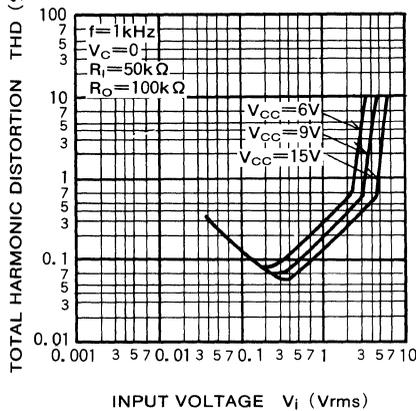
CONTROL VOLTAGE VS. SUPPLY VOLTAGE



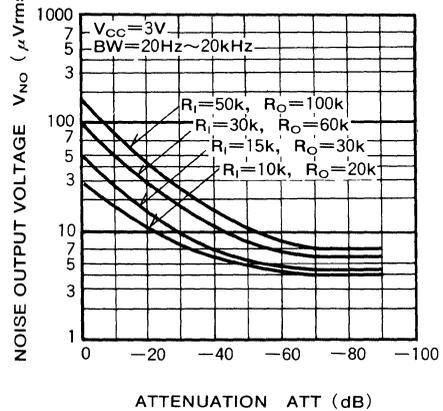
**TOTAL HARMONIC DISTORTION
 VS. INPUT VOLTAGE**



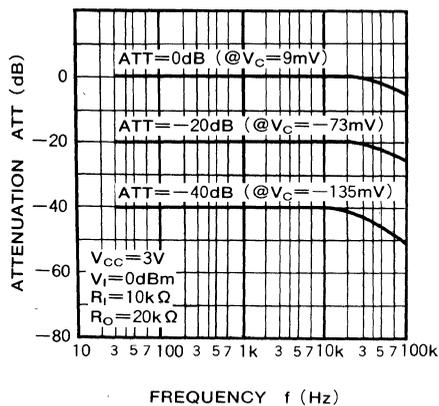
**TOTAL HARMONIC DISTORTION
 VS. INPUT VOLTAGE**



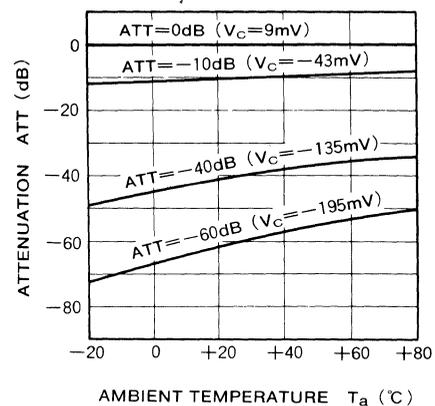
**NOISE OUTPUT VOLTAGE
 VS. ATTENUATION**



ATTENUATION VS. FREQUENCY



ATTENUATION VS. AMBIENT TEMPERATURE

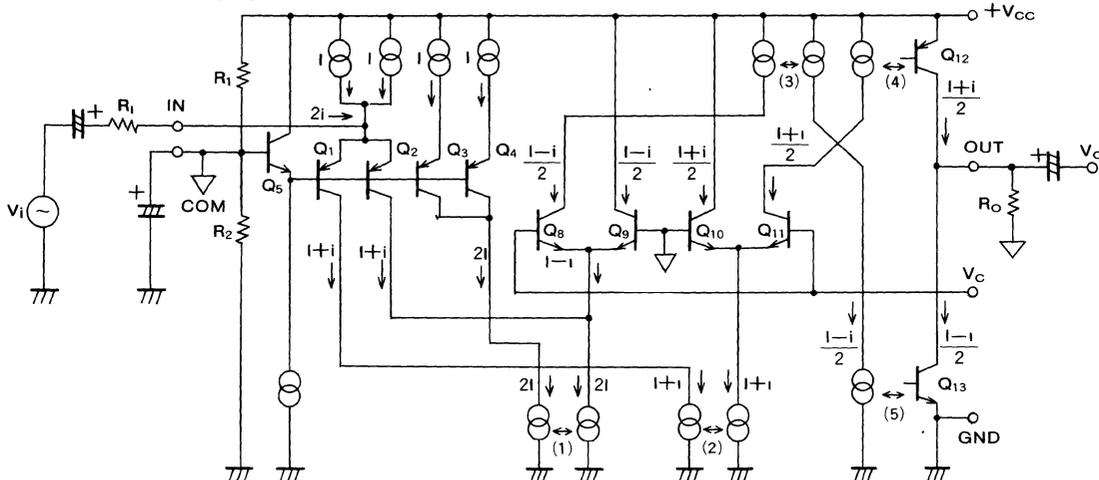


VOLTAGE CONTROLLED AMPLIFIER(VCA)FOR USE
IN A LOW-VOLTAGE ELECTRONIC VOLUME CONTROL

BASIC PRINCIPLE OF OPERATION

The M5222 is a current input, current output type of VCA IC. This amplifier uses the principle by which changing the balance of the differential circuit with external control voltage V_C will change gm. The circuit is also called

a variable transconductance (variable gm) OP amp. The basic principle of operation will be simply explained below.



※(↓) is the $V_{CC}/2$ pin (COM pin) derived by voltage division by B_1, B_2 and so on. ALL coupling is internal.

Fig. 1 M5222 EQUIVALENT CIRCUIT

Basic voltage-current conversion mechanism for input and output

Applying the input signal V_i which flows through external input resistor R_1 results in a change to a current signal at input terminal IN. The V_{BE} level shift of $R_1, R_2, Q_5, Q_1,$ and Q_2 will cause input pin IN to become ground level by means of $V_{CC}/2$ in terms of direct current and to become ground level by means of the externally-connected capacitor in terms of alternating current. The signal input in this way will be sent to the output pin as a current signal by the current mirror and differential circuit. By taking this current signal through the externally-connected output resistor (load resistor), the signal can go through a current-to-voltage conversion and be obtained as output signal V_o .

The output transistors combine the currents by means of the joined PNP and NPN collector circuits. Basically, the DC potential floats and is not determined in this joining of currents. This is why one end of externally-connected resistor R_o is connected to the $V_{CC}/2$ pin and the DC level ($V_{CC}/2$) at the time of no signal is set.

Basic mechanism of attenuation

The output is controlled by means of changing the control voltage applied to the V_C pin with respect to the COM pin ($V_{CC}/2$ pin). By applying voltage from the COM pin to the base of one side of a differential circuit and applying voltage from the V_C pin to the other base, the current distribution of the differential circuit is changed and the gain of this circuit is changed.

Let us first consider when V_C equals zero (V_C -COM is shorted). Input signal V_i is converted to current by in-

put resistor R_1 and the i currents ($2i = V_i/R_1$) flow through the collectors of Q_1 and Q_2 . When the current flowing in Q_i becomes $i+i$, the overall emitter current of the differential circuit consisting of Q_{10} and Q_{11} will also be determined as $i+i$ by means of current mirror (2). Since the base potential of Q_{10} and Q_{11} is the same, the current will be divided equally and current $(i+i)/2$ will flow in each of Q_{10} and Q_{11} . The current of current mirror (4) will also be determined as $(i+i)/2$ because of this.

Since the current of current mirror (1) is determined as $2i$ by the current flowing in Q_3 and Q_4 , the total of the current flowing in Q_2 and the current flowing in differential circuit Q_8, Q_9 will also be $2i$. The current from Q_2 which will become $i+i$ flows here and as a result, the overall emitter current of the differential circuit will be $2i - (i+i) = i-i$. This current is divided the same way as in the differential circuit consisting of Q_{10} and Q_{11} with current $(i-i)/2$ flowing in each of Q_8 and Q_9 . From this, the current of current mirror (3) is determined as $(i-i)/2$ and the current of current mirror (5) becomes $(i+i)/2$.

Now, current $(i-i)/2$ from current mirror (4) flows in transistor Q_{12} of the output stage. Since the current flowing in transistor Q_{13} from current mirror (5) is held at $(i-i)/2$, connecting output resistor R_o between the output pin and the COM pin will result in current i flowing through R_o and providing a voltage signal $V_o = i \cdot R_o$.

Here, by selecting $R_o = 2R_1, V_o = i \cdot R_o = 2i \cdot R_1 = V_i$ and the amplifier will have a gain of 1.

Next, we will consider case of when control voltage V_C is applied with regard to the selection of this resistance.

VOLTAGE CONTROLLED AMPLIFIER(VCA)FOR USE
IN A LOW-VOLTAGE ELECTRONIC VOLUME CONTROL

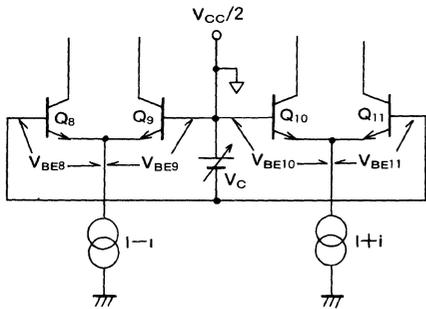


Fig. 2 DIFFERENTIAL CIRCUIT

The values of V_{BE} of the differential stage will be as follows:

$$V_{BE8} \doteq \frac{kT}{q} \ln \left(\frac{I_{C8}}{I_s} \right)$$

$$V_{BE9} \doteq \frac{kT}{q} \ln \left(\frac{I_{C9}}{I_s} \right)$$

$$V_{BE10} \doteq \frac{kT}{q} \ln \left(\frac{I_{C10}}{I_s} \right)$$

$$V_{BE11} \doteq \frac{kT}{q} \ln \left(\frac{I_{C11}}{I_s} \right)$$

where, I_s : the saturation current
 k : the Boltzmann constant
 q : the amount of electric charge on the electrons
 T : the absolute temperature

From this,

$$-V_c = V_{BE8} - V_{BE9} = \frac{kT}{q} \ln \frac{I_{C8}}{I_{C9}}$$

$$-V_c = V_{BE11} - V_{BE10} = \frac{kT}{q} \ln \frac{I_{C11}}{I_{C10}}$$

Here,

$$I_{C8} + I_{C9} \doteq I - i$$

$$I_{C10} + I_{C11} \doteq I - i$$

$$-V_c = \frac{kT}{q} \ln \frac{I_{C8}}{I - i - I_{C8}}$$

$$-V_c = \frac{kT}{q} \ln \frac{I_{C11}}{I - i - I_{C11}}$$

The current flowing through Q_8 and Q_{11} will be

$$I_{C8} = \frac{(I-i) \exp\left(-\frac{q}{kT} V_c\right)}{1 + \exp\left(-\frac{q}{kT} V_c\right)} = \frac{I-i}{1 + \exp\left(-\frac{q}{kT} V_c\right)}$$

$$I_{C11} = \frac{(I+i) \exp\left(-\frac{q}{kT} V_c\right)}{1 + \exp\left(-\frac{q}{kT} V_c\right)} = \frac{I-i}{1 + \exp\left(-\frac{q}{kT} V_c\right)}$$

Current I_{C11} is the current of current mirror (4), and I_{C8} will be the same as the current of current mirror (5).

At this time, the current that will flow through the output pin will be the same as that in the explanation when V_c was equal to zero, and is expressed as

$$i_o = \frac{2i}{1 + \exp\left(\frac{q}{kT} \cdot V_c\right)}$$

The gain will be

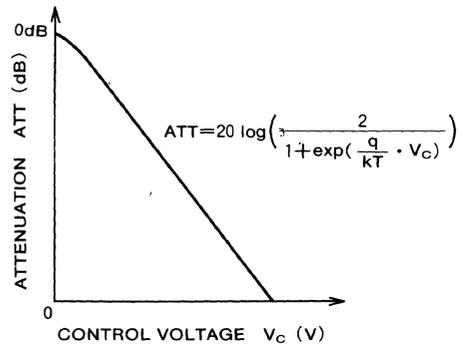
$$\frac{V_o}{V_i} = \frac{i_o \cdot R_o}{2i \cdot R_i} = \frac{2}{1 + \exp\left(\frac{q}{kT} \cdot V_c\right)}$$

and when calculated in dB,

$$ATT = 20 \log \left(\frac{2}{1 + \exp\left(\frac{q}{kT} \cdot V_c\right)} \right)$$

As in the graph below, the attenuation will change logarithmically with respect to the change of V_c .

ATTENUATION VS. CONTROL VOLTAGE



Setting and connection of input/output resistance

As explained above, the input signal is converted to current, but since the transistor of the input stage is biased at a fixed current of $I = 76 \mu A$, the maximum value of the input current is determined at the least upper bound of I (FIG.3). Accordingly, when a large signal is input it is necessary to select a large input/output resistance and decrease the input current. Note that increasing the resistance will also increase the noise distortion factor, so the value of the setting should be made to suit the particular application.

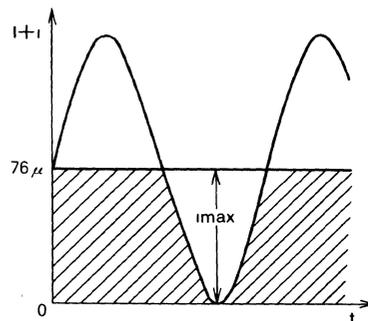


Fig. 3 MAXIMUM CURRENT SIGNAL

**VOLTAGE CONTROLLED AMPLIFIER(VCA)FOR USE
 IN A LOW-VOLTAGE ELECTRONIC VOLUME CONTROL**

The M5222 has a floating-type output stage with the collectors of Q₁₂ and Q₁₃ joined as shown in FIG. 4. Here, the difference of the combined currents will become the output current that will flow through the load. Note that it is necessary to set the DC potential of this output pin by externally-connected resistor R_O and that it is generally DC-connected to the V_{CC}/2 pin (or to pin (3)).

In terms of AC, it is necessary to set the output pin to ground level so that capacitor C is required. Since the voltage gain (amount of attenuation) is determined by R_O, the value of the input impedance connected to the next stage is sometimes affected. (Placing Z_I in parallel with R_O will lower the impedance.) Generally, a buffer amplifier composed of a transistor or OP amp connected.

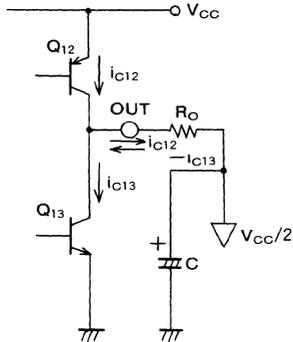


Fig. 4 EQUIVALENT CIRCUIT OF OUTPUT STAGE

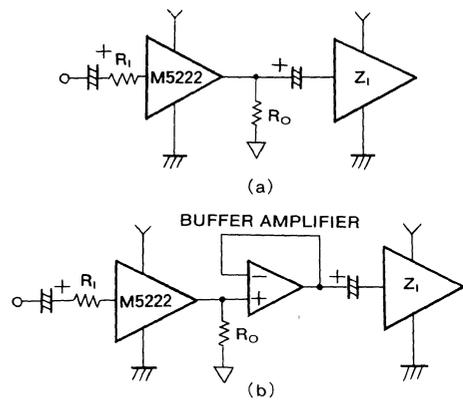
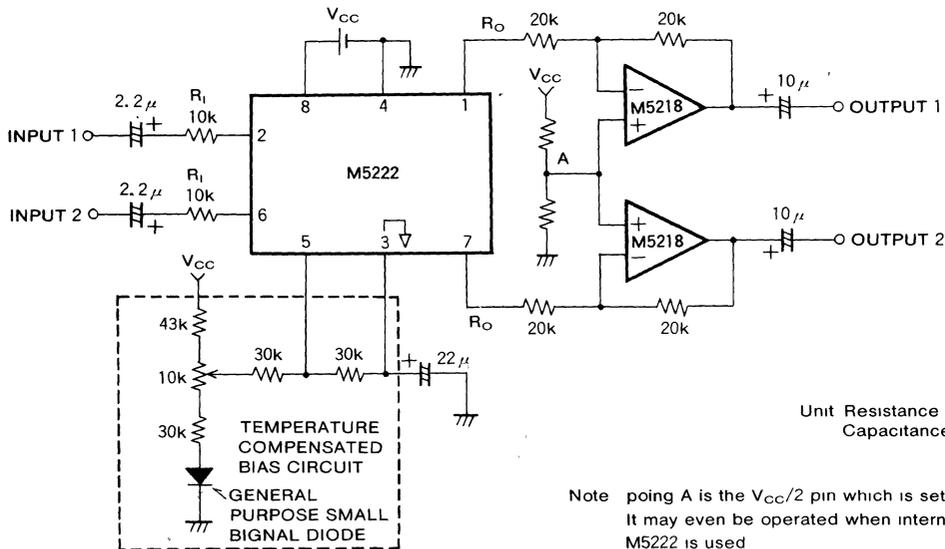


Fig. 5

APPLICATION EXAMPLES

(1) TEMPERATURE COMPENSATED BIAS AND OUTPUT BUFFER CIRCUITS

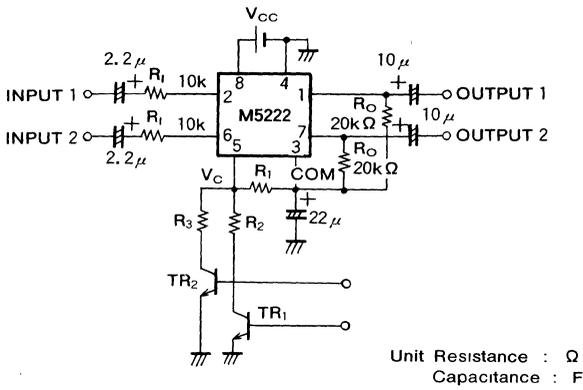


Unit Resistance Ω
 Capacitance F

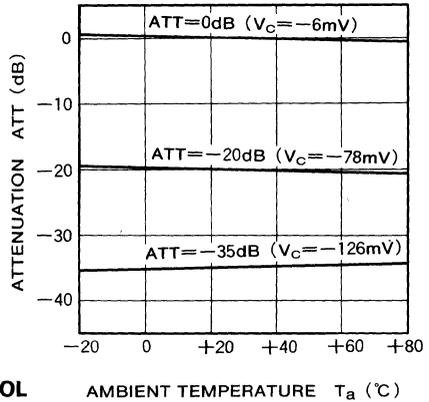
Note point A is the V_{CC}/2 pin which is set externally
 It may even be operated when internal pin (3) of
 M5222 is used

**VOLTAGE CONTROLLED AMPLIFIER(VCA)FOR USE
 IN A LOW-VOLTAGE ELECTRONIC VOLUME CONTROL**

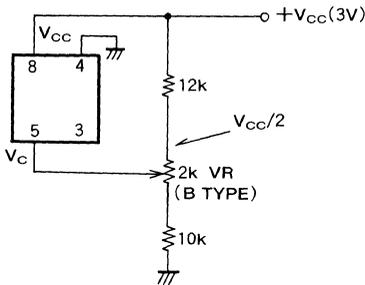
(2) PROGRAMMABLE ATTENUATION CIRCUIT



**ATTENUATION VS. AMBIENT TEMPERATURE
 (TEMPERATURE COMPENSATION)**

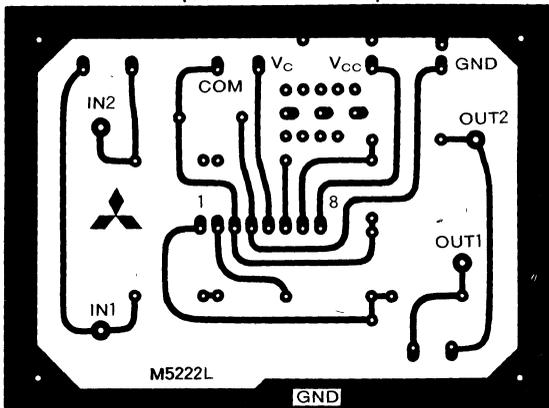


(3) CONTROL APPLICATION WITH EXISTING VOLTAGE CONTROL

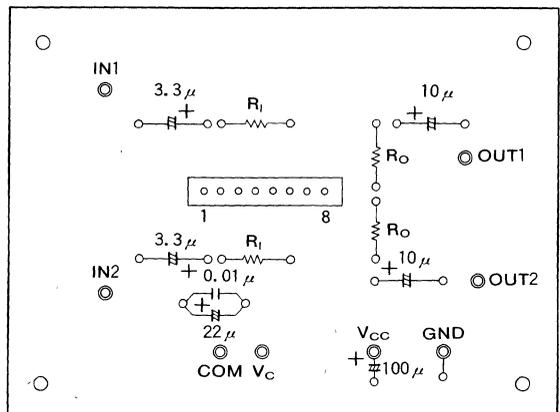


PRINTED CIRCUIT BOARD FOR CIRCUIT TESTING

**PRINTED CIRCUIT BOARD WIRING DIAGRAM
 (COPPER FOIL SIDE)**



(PARTS SIDE)



M5241L

VOLTAGE CONTROLLED AMPLIFIER(VCA)FOR USE IN AN ELECTRONIC VOLUME CONTROL

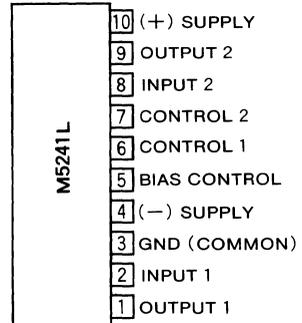
DESCRIPTION

The M5241L is a semiconductor integrated circuit consisting of a dual voltage controlled amplifier (VCA) with a 2-power supply system designed for use in an electronic volume control for analog signals. Housed in a 10-pin SIP, the M5241L as two built-in channels and is designed to hi-fi specifications with a high dynamic range, low distortion factor, and high signal-to-noise ratio. There are control pins for each channel permitting independent control of each channel with an external DC control voltage. The attenuation characteristics change logarithmically with respect to the external DC control voltage (a response equivalent to the A-curve volume) and make the M5241L and ideal electronic volume control for use in stereos, radio/cassette recorders, car audio systems, hi-fi VTRs, and electronic musical instruments.

FEATURES

- Two built-in channels with independent control pins
 -Independent control of each channel is possible with V_C (control) at pins ⑥ and ⑦ for channels 1 and 2, respectively.
- High maximum input voltage
 - $V_i=3V_{rms}$ (when THD=0.5%)
- Low distortion THD=0.02%(when $V_o=1V_{rms}$)
- Wide attenuation range 0~100dB
- High signal-to-noise ratio (dynamic range) 94dB (when ATT=0dB, $R_i=15k\Omega$, $R_o=30k\Omega$, and $R_c=1.8k\Omega$)
- Logarithmic response VCA
 - ... Logarithmic response equivalent to A-curve volume
- Bias voltage can be adjusted with external resistor
- Can be operated with a single power supply
 - Built-in COM Pin ③ $V_{CC}/2$ bias pin

PIN CONFIGURATION (TOP VIEW)

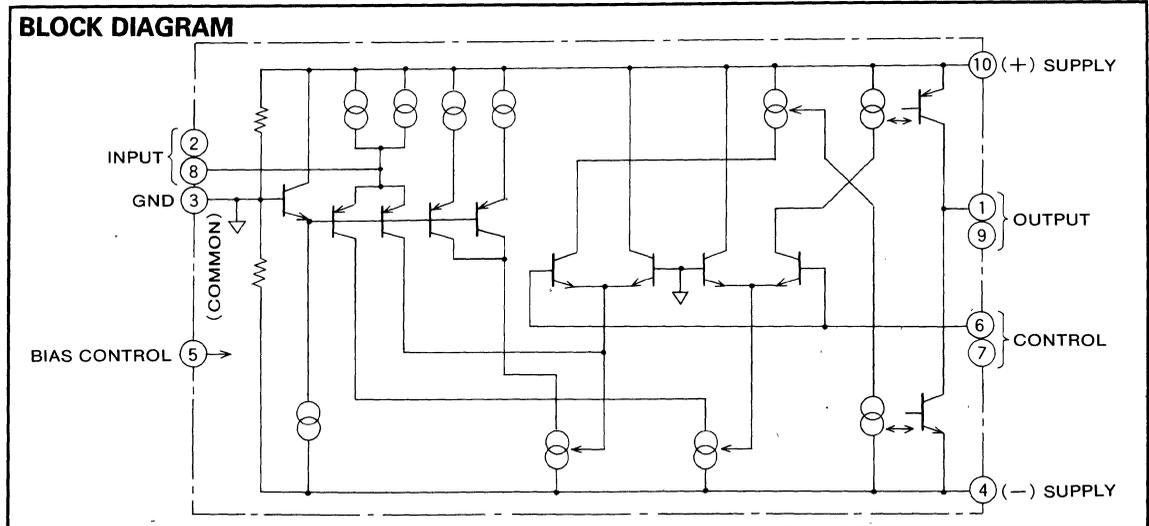


10-pin moded plastic SIP

APPLICATION

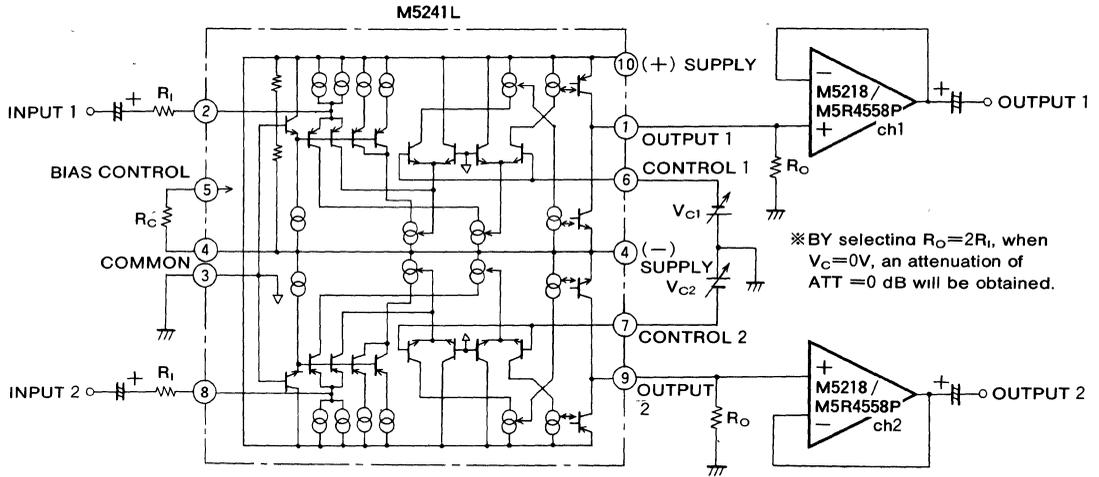
As an electronic volume control in stereo equipment, radio/cassette recorders, car audio systems, hi-fi VTRs, and as a program attenuator in electronic musical instruments.

BLOCK DIAGRAM



**VOLTAGE CONTROLLED AMPLIFIER(VCA)FOR USE
IN AN ELECTRONIC VOLUME CONTROL**

APPLICATION EXAMPLE



ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$, unless otherwise noted)

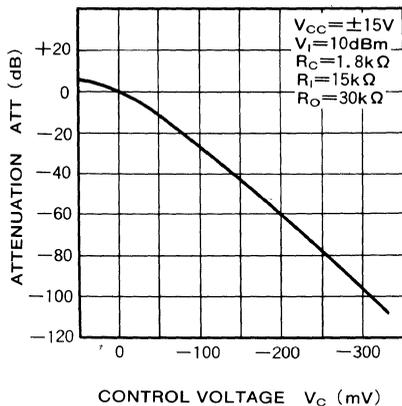
Symbol	Parameter	Ratings	Unit
V_{CC}	Supply voltage	$\pm 18(36)$	V
P_d	Power dissipation	800	mW

ELECTRICAL CHARACTERISTICS ($T_a=25^{\circ}\text{C}$, $V_{CC}=\pm 15\text{V}$, $R_i=15\text{k}\Omega$, $R_o=30\text{k}\Omega$, $V_c=0\text{V}$, $V_i=10\text{dBm}$, $f=1\text{kHz}$, unless otherwise noted)

Symbol	Parameter	Conditions	Typ	Unit
I_{CC}	Circuit current	$V_i=0\text{V}$	3.8	mA
V_{IM1}	Maximum input voltage	THD=0.5%	3.0	Vrms
ATT_M	Maximum attenuation	$V_c=-300\text{mV}$	100	dB
ATT	Attenuation error		-0.5	dB
THD	Total harmonic distortion	$V_o=1\text{Vrms}$	0.02	%
V_{NO}	Noise output voltage	$V_i=0$ input shorted	57	μVrms

TYPICAL CHARACTERISTICS

ATTENUATION VS. CONTROL VOLTAGE



MITSUBISHI LINEAR ICs
M5232L

VOLTAGE DETECTOR/ON-OFF ALARM CIRCUIT

DESCRIPTION

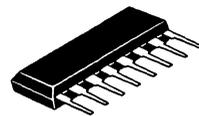
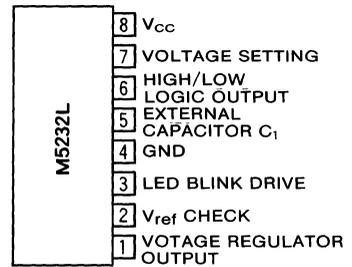
The M5232L is a unique semiconductor integrated circuit designed for use as a voltage detector/on-off alarm circuit.

Housed in a compact 8-pin SIP, the M5232L contains a comparator, reference voltage source, a vibrator circuit for turning the LED on and off, and a voltage regulation circuit. When the input voltage of the comparator at Pin ⑦ is higher than the internal reference voltage, the LED lights up, and when it is lower, the LED turns on and off. Also provided is an output pin (Pin ⑥) which does not operate intermittently but permits a relay or micro buzzer to be driven while the LED is being turned on and off by Pin ③. Signals from a low voltage checker for batteries, or from optical or thermal sensors are detected at the input pin of the comparator (Pin ⑦) allowing the M5232L to be applied widely in the alarm and protection circuits of electronic equipment.

FEATURES

- Starting supply voltage at which the LED will blink can be set optionally by using external resistors R_1 and R_2 (in the case of a low voltage checker for batteries)
- LED on/off frequency can be set optionally with external capacitor C_1
- Built-in logic output pin (Pin ⑥) causes a high-to-low level transition as soon as the blinking begins
- Hysteresis operation is possible at the blink starting voltage using Pin ⑥
- LED lights when the input voltage of the comparator at Pin ⑦ is higher than the internal reference voltage, permitting the M5232L to be used as a pilot lamp for power ON indication

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic SIP

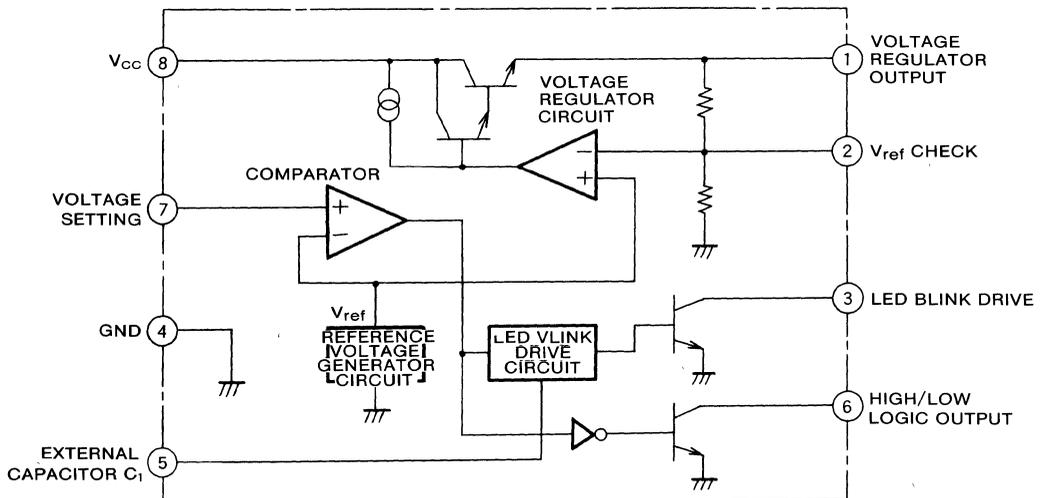
APPLICATION

Low voltage checker for batteries in equipment such as radio/cassette recorders, portable VTRs, cameras. Alarm and protection circuits of electronic equipment.

RECOMMENDED OPERATING CONDITIONS

Supply voltage range $V_{CC}=5\sim 18V$

BLOCK DIAGRAM



VOLTAGE DETECTOR/ON-OFF ALARM CIRCUIT

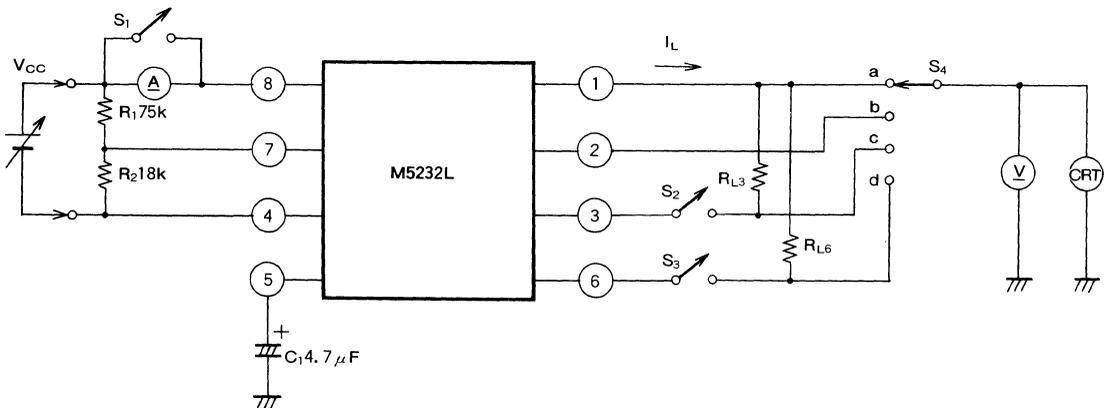
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		20	V
P_d	Power dissipation		800	mW
I_{LP}	Load current		50	mA
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	8	mW/°C
T_{opr}	Operating temperature range		-20~+75	°C
T_{stg}	Storage temperature range		-55~+125	°C

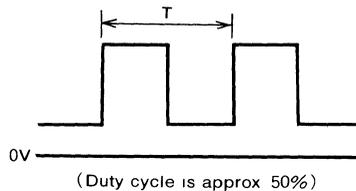
ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	$V_{CC}=9\text{V}, I_L=0$		2.0	3.0	mA
V_2	Reference voltage	$V_{CC}=9\text{V}, R_{L3}=400\ \Omega$	1.22	1.31	1.40	V
V_1	Output voltage	$V_{CC}=9\text{V}, R_{L3}=400\ \Omega$	3.6	4.0	4.4	V
V_3	Saturation voltage	$V_{CC}=9\text{V}, R_{L3}=400\ \Omega$		0.2	0.5	V
V_6	Saturation voltage	$V_{CC}=6\text{V}, R_{L6}=400\ \Omega$		0.2	0.5	V
f	Oscillation frequency	$V_{CC}=6\text{V}, C_1=4.7\ \mu\text{F}, R_{L3}=400\ \Omega$	1.8			Hz

TEST CIRCUIT



☆ MEASUREMENT OF f ON CRT
PIN ③ WAVEFORM $f=1/T$ (Hz)



Parameter	V_{CC}	S_1	S_2	S_3	S_4
I_{CC}	9V	OFF	OFF	OFF	—
V_2	9V	ON	ON	OFF	b
V_1	9V	ON	ON	OFF	a
V_3	9V	ON	ON	OFF	c
V_6	6V	ON	OFF	ON	d
f	6V	ON	ON	OFF	c

VOLTAGE DETECTOR/ON-OFF ALARM CIRCUIT

1. Basic principle of M5232L operation

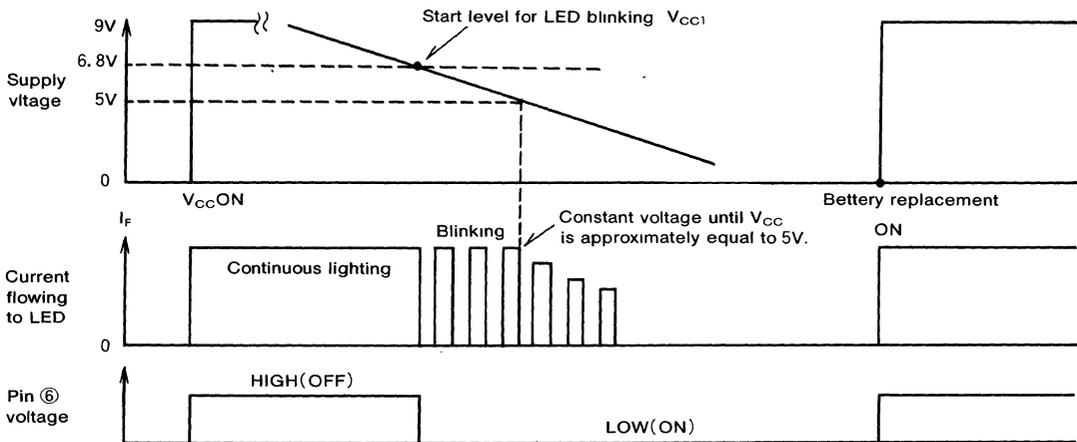
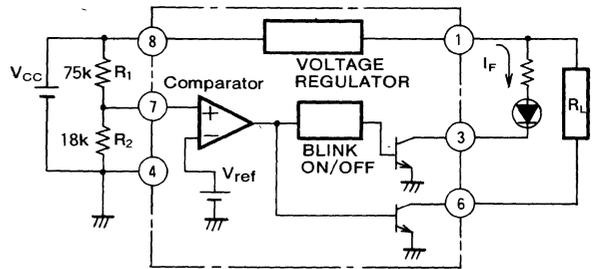
- When supply voltage V_{CC} is normal, the LED lights and functions as a pilot lamp. In this case, Pin ③ drives the LED with open-collector output.
- V_{CC} drops, becoming V_{CC1} and when the Pin ⑦ potential becomes

$$V_7 = \frac{R}{R_1 + R_2} \cdot V_{CC1} < V_{ref}$$

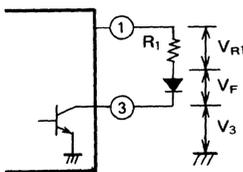
the comparator inverts, the blink circuit is switched on, and the LED blinks on and off. (V_{ref} , produced by the internal reference voltage source, is 1.31V typ.)

- The on/off alarm circuit shown on the right will activate when the voltage is 6.8V, which is 25% less than $V_{CC} = 9V$ (six 1.5V cells).
- Pin ⑥ is an open-collector output that causes a high-to-low level transition simultaneously with the Pin ③ on/off operation. A micro buzzer, relay or other load can be connected across this pin and Pin ① of V_{CC} (Pin ⑧) for a wide range of applications.

LOW VOLTAGE CHECKER FOR BATTERIES (SIMPLIFIED DIAGRAM)



2. LED Drive current I_F



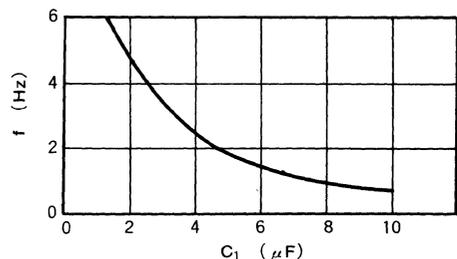
- Assuming that Pin ① output voltage is 4V, LED forward voltage is V_F , and V_3 is 0.2V, then

$$I_F = \frac{4V - 0.2V - V_F}{R_1}$$

I_F is approximately equal to 4.6mA with $V_F = 2V$, and $R_1 = 390\Omega$ (in a typical application circuit).

3. On/off oscillation frequency

The on/off oscillation frequency can be varied by changing the value of external capacitor C_1 .

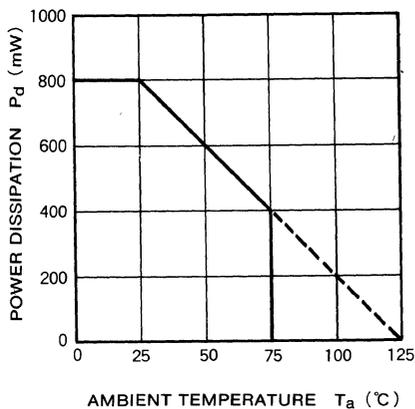


* If capacitance C_1 is even further reduced, oscillation will be possible up to a frequency of about 10 kHz

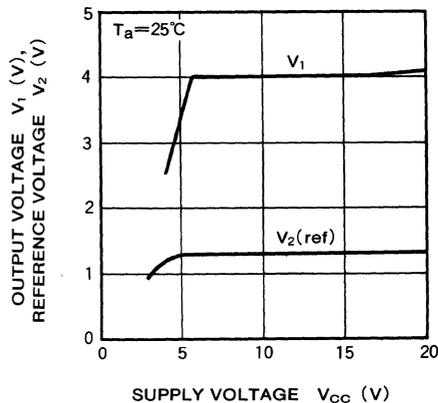
VOLTAGE DETECTOR/ON-OFF ALARM CIRCUIT

TYPICAL CHARACTERISTICS

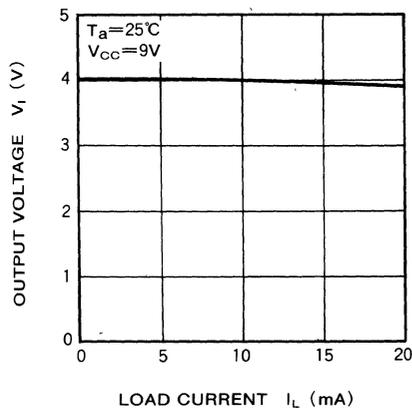
POWER DISSIPATION VS. AMBIENT TEMPERATURE



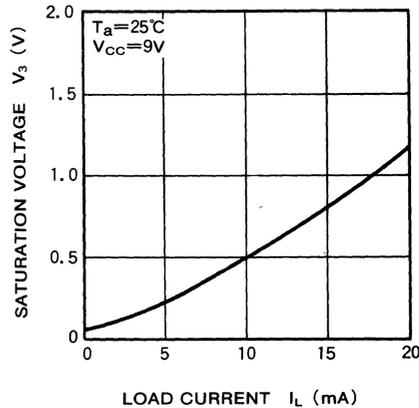
OUTPUT VOLTAGE, REFERENCE VOLTAGE VS. SUPPLY VOLTAGE



OUTPUT VOLTAGE VS. LOAD CURRENT

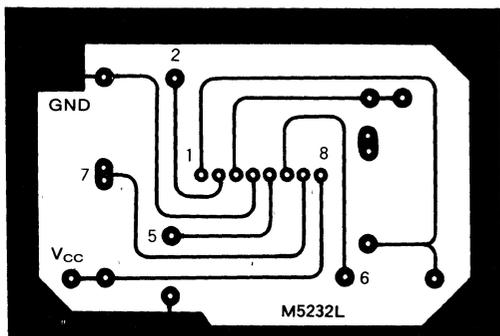


SATURATION VOLTAGE VS. LOAD CURRENT

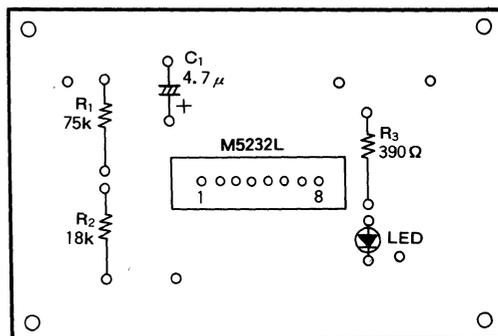


PRINTED CIRCUIT BOARD FOR CIRCUIT TESTING (TYPICAL APPLICATION EXAMPLE)

PRINTED CIRCUIT BOARD WIRING DIAGRAM (COPPER FOIL SIDE)



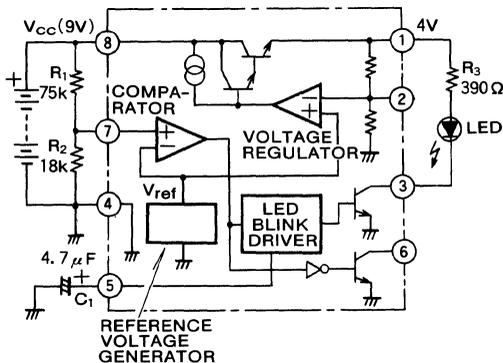
(PARTS SIDE)



VOLTAGE DETECTOR/ON-OFF ALARM CIRCUIT

APPLICATION EXAMPLES

1. Low voltage checker for batteries

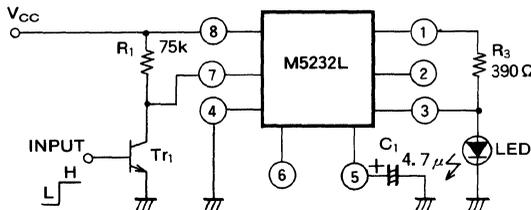


This low voltage checker for batteries is set to start the LED blinking when supply voltage $V_{CC} = 9V$ (six 1.5V cells) is reduced by 25% (to 6.8V).

$$C_1 = 4.7 \mu F \rightarrow f \approx 1.8 \text{ Hz}$$

C_2 , which has a value of 100pF, prevents oscillation. It should be inserted when the input/output leads are long or when parasitic oscillation is generated by the load.

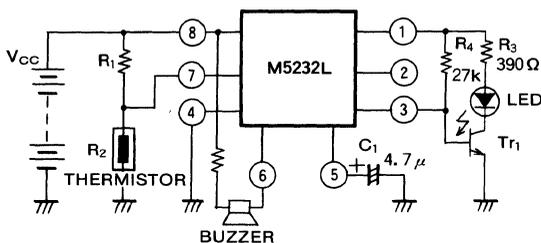
2. Trouble indicator



When the base of transistor T_{r1} is set low (the normal condition), Pin 7 comparator voltage is set high, Pin 3 is set low, and the LED is switched off.

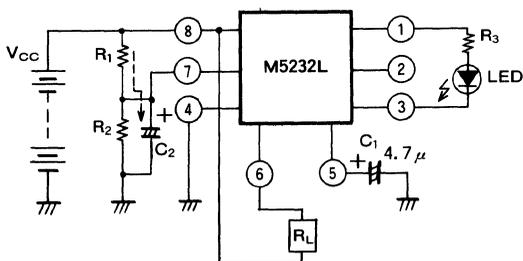
When the base of transistor T_{r1} is set high (signifying trouble), Pin 7 comparator input voltage is set low, the internal vibrator circuit is switched on, Pin 3 is repeatedly set high and low, and the LED blinks on and off. At the same time, an electronic buzzer can be sounded using Pin 6 or a relay can be driven. (An ordinary switch may be used in place of transistor T_{r1} .)

3. Abnormal temperature indicator



In normal circumstances, the LED is off and power dissipation is kept low. In abnormal circumstances, the LED blinks on and off. It is also possible to sound a buzzer or drive a relay using Pin 6.

4. Timer and muting indicator



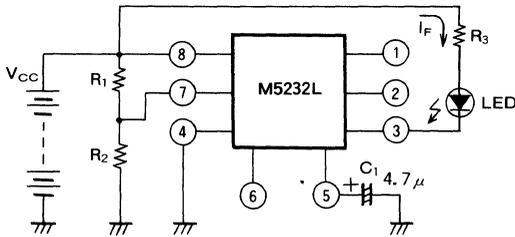
By connecting C_2 in parallel with R_2 , V_{CC} is switched on, the charging current indicated by the dotted line in the figure flows, and the LED blinks on and off until the Pin 7 voltage reaches:

$$V_{CC} \cdot \frac{R_2}{R_1 + R_2}$$

When C_2 is charged up, the LED will light. These operations can be applied to timer and muting circuits.

VOLTAGE DETECTOR/ON-OFF ALARM CIRCUIT

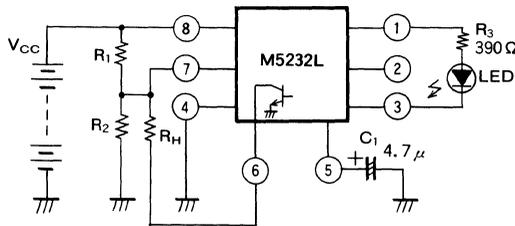
5. Low voltage checker for batteries (of 5V or less)



Since output Pin ① of the M5232L is regulated at 4V, the output is not stabilized at V_{CC} less than 5V.

When an LED is connected directly from V_{CC} as shown in the figure at left, it is possible to construct a battery checker for batteries of less than V_{CC} 5V, (for instance, $V_{CC}=3\sim 5V$). Note that in this case, the I_F of the LED will fluctuate in accordance with the changes in V_{CC} .

6. Hysteresis operation of the on/off starting voltage



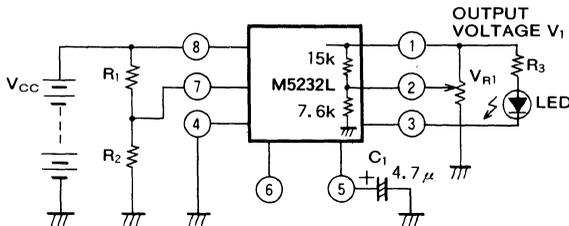
By connecting R_H across pins ⑥ and ⑦, as shown in the figure on the left, the on/off starting voltage is set at:

$$V_{2(\text{ref})} \cdot \frac{R_1 + R_2}{R_2}$$

After the start of the on/off blinking, Pin ⑥ (the open collector) goes on, and so it is possible to apply hysteresis and the voltage expressed below:

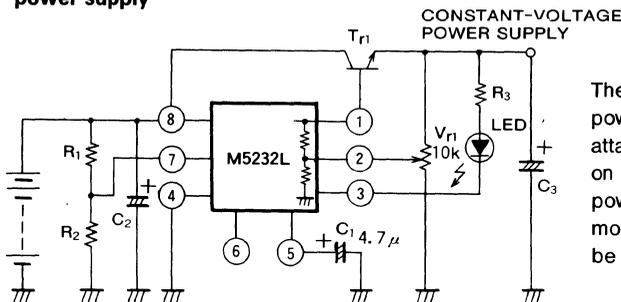
$$V_{2(\text{ref})} \cdot \frac{R_1 + R_2 // R_H}{R_2 // R_H}$$

7. Modification of output voltage V



The output voltage V_1 of the M5232L is set by the internal resistor as shown in the figure at left, but this can be changed by connecting a semi-fixed resistor across GND and pins ① and ②.

8. Increased current capacity of constant-voltage power supply



The current capacity of the built-in constant-voltage power supply is approximately 20mA. However, by attaching external transistor T_{r1} as shown in the figure on the left, it is possible to obtain a constant-voltage power supply with a large current capacity of 1A or more. The output voltage of the power supply can also be varied with variable resistor V_{r1} .

Note · Oscillation may be generated when the input or output leads are long. In cases like this, input and output capacitors C_1 and C_2 (1~10µF) should be inserted near the IC.

ELECTRONIC VOLUME CONTROL CMOS D-A CONVERTER

DESCRIPTION

The M50601P is a semiconductor integrated circuit designed to digitally control the control voltage of an electronic volume control. Fabricated using aluminum gate CMOS technology, this device contains a PWM-type 6-bit D-A converter that functions to control the amplitude of the analog signal by dividing it into 64 steps.

FEATURES

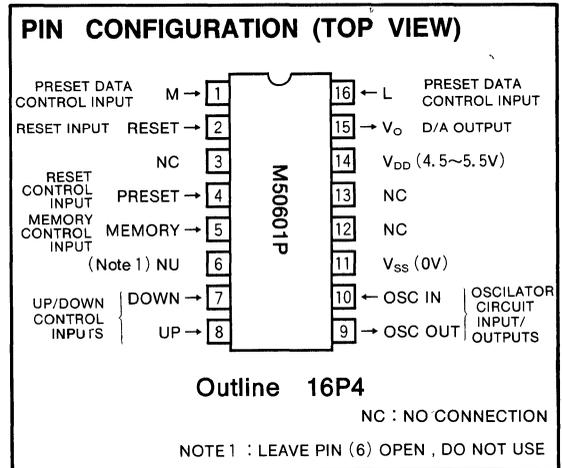
- Low power dissipation
- Built-in ceramic oscillator for high frequency stability
- Analog signal amplitude may be controlled over 64 steps
- Includes a memory function
- Includes a preset function

APPLICATION

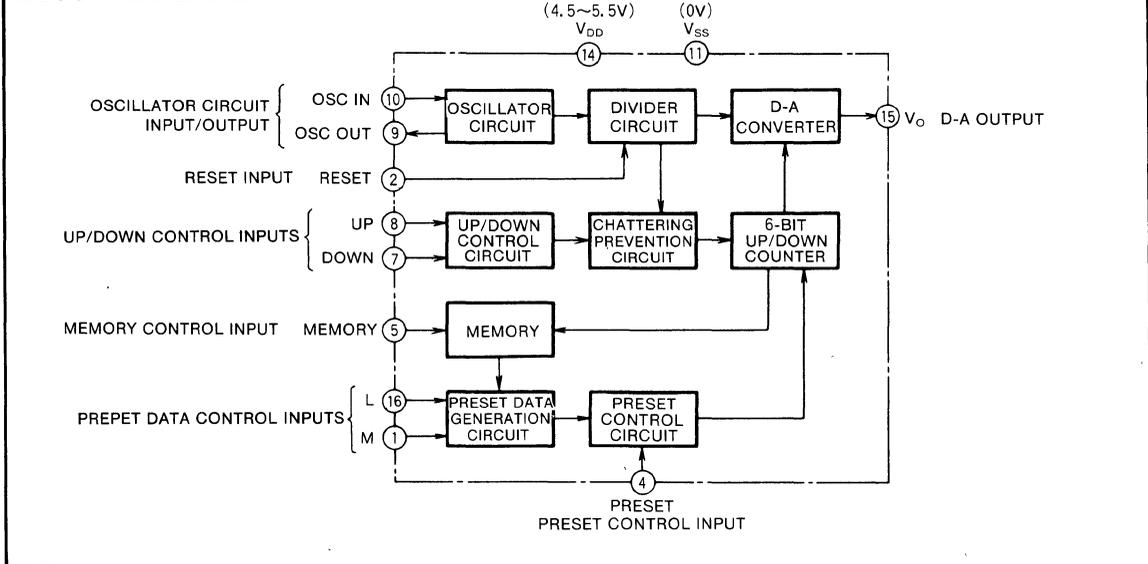
All types of electronic volume controls.

FUNCTIONAL DESCRIPTION

The M50601P is a CMOS integrated circuit containing a 6-bit up/down counter, oscillator circuit, divider circuit, D-A converter, and the other circuits required to digitally process the control voltage of an electronic volume control. The major functions include both manual and automatic operation of the 6-bit up/down counter, and a memory function designed to hold the contents of the up/down counter. It also has a preset function that allows the contents of the memory, or "0" or "32" to be set into the up/down counter.



BLOCK DIAGRAM



ELECTRONIC VOLUME CONTROL CMOS D-A CONVERTER

FUNCTIONAL DESCRIPTION

Oscillator Circuit

The internal oscillator circuit generates a reference signal by connecting one ceramic resonator and two capacitors to the OSC IN and OSC OUT terminals. An example circuit is shown in Fig. 1.

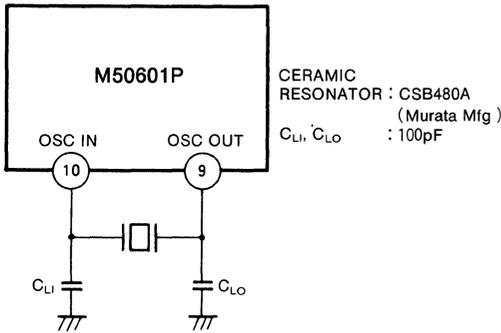


Fig. 1. Example Oscillator Circuit (using a ceramic resonator)

Up/Down Counter Function

Each time the UP pin receives a high-level input, the contents of the 6-bit up/down counter are incremented by one step. The count will continue to "63", after which it will not increase.

If the high level is held at the UP pin for 0.41 to 0.52 seconds, the contents of the up/down counter will be incremented at a rate of approximately 9.77 steps/sec. The counter will stop when it reaches "63".

Each time the DOWN pin receives a high-level input, the contents of the 6-bit up/down counter are decremented by one step. The count will continue to "0", after which it will not decrease.

If the high level is held at the DOWN pin for 0.41 to 0.52 seconds, the contents of the up/down counter will be decremented at a rate of approximately 9.77 steps/sec. The counter will stop when it reaches "0".

When both the UP and DOWN pins receive a simultaneous high-level input, the contents of the up/down counter will not change.

The chattering prevention circuit also prevents data from being entered into the counter (for both of the UP and DOWN pins) if the duration of the high-level signal is less than 25 to 50ms.

Memory Function

When a pulse similar to the one shown in Fig. 2 is applied to the MEMORY pin, the contents of the 6-bit up/down counter are stored in memory.

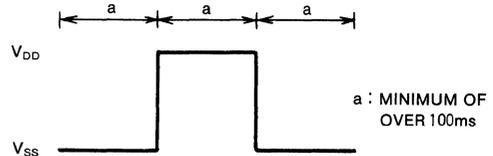


Fig. 2. Input Signal Waveform at the MEMORY Terminal

Preset Function

When the PRESET pin and the the M pin are at low level and a pulse similar to the one shown in Fig. 3 is applied to the L preset to "0".

When the PRESET pin and the L pin are at low level and a pulse signal similar to the one shown in Fig. 3 is applied to the M pin, the contents of the 6-bit up/down counter are preset to "32".

When the L pin and the M pin are at low level and a pulse similar to the one shown in Fig. 3 is applied to the PRESET pin, the contents of the 6-bit up/down counter will not be preset.

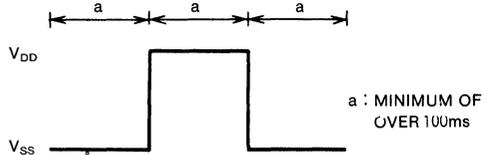


Fig. 3. Input Signal Waveform for the L, M, and PRESET Terminals

Reset Function

Reset is a test function. When a high-level signal is applied to the RESET pin, the divider (480 kHz-4.88Hz) is reset.

D-A Output

The internal 6-bit D/A converter functions to control the analog signal amplitude in 64 steps.

Output is repetitious at a frequency of 1.25 kHz ; it is a pulse-width modulated output with a minimum pulse width of 12.5us.

Operation of the UP or DOWN terminal, if set in the automatic mode, will increment or decrement the analog signal amplitude at a rate of approximately 9.77 steps/sec. Consequently, the time required to cover the full range from minimum to maximum analog value, or vice versa is approximately 6.45 seconds.

ELECTRONIC VOLUME CONTROL CMOS D-A CONVERTER

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Conditions	Ratings	Unit
V_{DD}	Supply voltage	With respect to V_{SS} pin	-0.3~7	V
V_I	Input voltage		$V_{SS} \leq V_I \leq V_{DD}$	—
V_O	Output voltage		$V_{SS} \leq V_O \leq V_{DD}$	—
P_d	Maximum power dissipation	$T_a=25^\circ\text{C}$	300	mW
T_{opr}	Operating temperature range		-30~70	$^\circ\text{C}$
T_{stg}	Storage temperature range		-40~125	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS

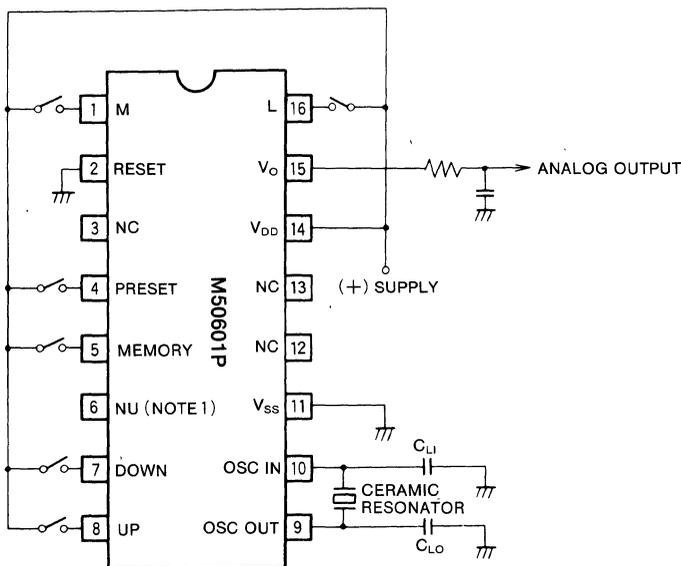
Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{DD}	Supply voltage	4.5	5	5.5	V
f_{OSC}	Oscillation frequency	480			kHz
V_{IH}	"H" input voltage (Note 2)	$0.7 \times V_{DD}$	V_{DD}	V_{DD}	—
V_{IL}	"L" input voltage (Note 2)	0	0	$0.3 \times V_{DD}$	—

Note 2 : Applicable to the seven terminals, M, PRESET, RESET, MEMORY, DOWN, UP, and L.

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{DD}=5\text{V}$, $V_{SS}=0\text{V}$, $C_{L1}=C_{L0}=100\text{pF}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{DD}	Operating supply voltage	$T_a=-30\sim70^\circ\text{C}$, $f_{osc}=480\text{kHz}$	4.5	5	5.5	V
V_{DD}	Supply current	$f_{osc}=480\text{kHz}$		0.3	2	mA
R_I	Pull-down resistor (Note 2)	$V_I=5\text{V}$		100		k Ω
I_{OH}	"H" output current	$V_O=0\text{V}$	2			mA
I_{OL}	"L" output current	$V_O=5\text{V}$	2			mA

APPLICATION EXAMPLE



FLEXIBLE DISK DRIVE (FDD) READ/WRITE AND LOGIC IC

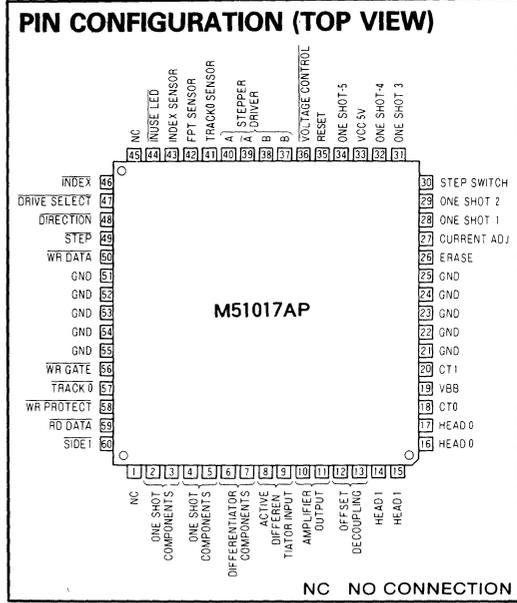
DESCRIPTION

The M51017AP is a semiconductor integrated circuit designed for use in flexible disk drives. It consists of read, write, and stepping motor drive pulse generator circuit sections, as well as a wide variety of control logic circuits.

The M51017AP can be used for double-sided recording. The magnetic head is selected by side select signal.

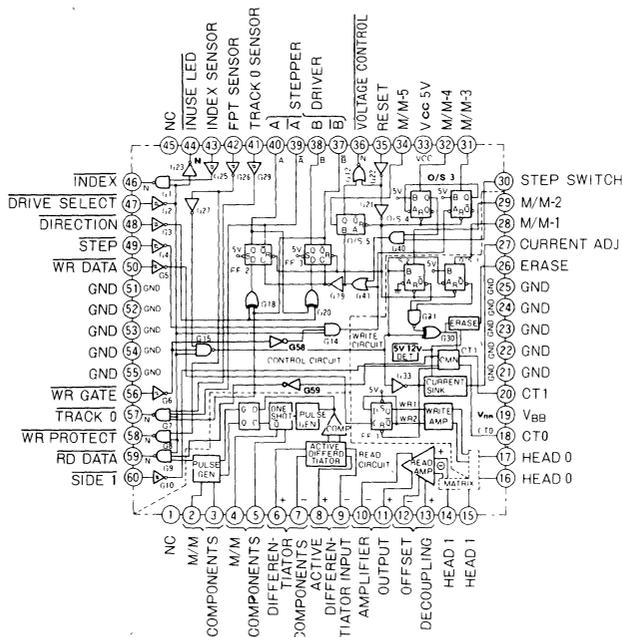
FEATURES

- Read circuit section
 Amplifies the signals from the magnetic head and outputs the required read data.
- Write circuit section
 Drives the write switching circuit according to the write data and the recording current is made to flow in the magnetic head. The recording current is set externally. Also drives the erasing circuit (tunnel erase system) according to the on/off switching of the write gate. The erase timing can be set by an external constant.
- Stepping motor drive pulse generator section
 Generates a drive pulse corresponding to the two-phase excitation system according to the step input signal. A switching terminal for selecting one step/one pulse or one step/two pulses has been provided. The output is TTL level.
- Control logic circuit sections
 (functions of each section are described later)
 The input circuit section provides hysteresis operation and LSTTL level, while the output circuit section has open collector output.
- Housed in a 60-pin molded plastic quad flat package (with a lead pitch of 0.8mm)



60-pin molded plastic quad flat package

BLOCK DIAGRAM



FLEXIBLE DISK DRIVE (FDD) READ/WRITE AND LOGIC IC

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		7.0	V
V _{BB}	Supply voltage		15.0	V
T _{opr}	Operating temperature range		0~+60	°C
T _{stg}	Storage temperature range		-40~+125	°C
V _{hd}	Voltage applied to head input pins (pins 14, 15, 16, 17)	Pulse applied for 2 μs	25	V
V ₂₆	Voltage applied to erase-current pin (pin 26)	Pulse applied for 2 μs	20	V
I ₂₆	Erase-current pin output current (pin 26)	Erase mode	100	mA
I ₁₀	Center-tap pin output current (pin 10)	Write/erase mode	110	mA
I ₂₀	(pin 18, 20)			

ELECTRICAL CHARACTERISTICS (T_a=25°C, V_{CC}=5V, V_{BB}=12V, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
G _V	Voltage gain	f=125kHz	47	49	51	dB
B _w	Preamplifier high range gain attenuation	f=5MHz		-3		dB
V _O	Differential output voltage width	Pins 10-11		6		V _{P-P}
I _{OS}	Differential output current width	Pins 10-11		8		mA _{P-P}
r _i	Differential input resistance	Pins 14-15/16-17	20	25		kΩ
r _O	Differential output resistance	Pins 10-11		30		Ω
N _i	Input conversion noise voltage	B _w =10kHz~1MHz		8		μVrms
V _{OFF}	Output offset voltage	Pins 10-11	-0.6		+0.6	V
CMRR	In-phase signal suppression ratio	V _i =5mVrms, f=10kHz		50		dB
SVRR	Supply fluctuation suppression ratio	ΔV _{BB} =100mVrms, f=10kHz		60		dB
r _{id}	Differential input resistance	Pins 8-9	22	30		kΩ
I _{OS}	Output sink current	Pins 6, 7		1.5		mA
P _s	Peak shift		-2		+2	%
—	Monostable one-pulse width setting range		500	2000	2400	ns
—	Monostable two-pulse width setting range		150	1000	1200	ns
—	Monostable one-pulse width precision	R ₁ =5.6kΩ, C ₁ =560pF, t _w =2000ns	-15		+15	%
—	Monostable two-pulse width precision	R ₂ =5.6kΩ, C ₂ =300pF, t _w =1200ns	-20		+20	%
I _{WR}	Write current	R _{WR} =5.6kΩ		6		mA
—	Write current precision		-10		+10	%
—	Write current supply dependency	V _{CC} =5V, V _{BB} =10.8~13.2V		±0.1		%/V
—	Write current temperature dependency	T _a =0~60°C		±0.05		%/°C
ΔI _{WR}	Write current pair quality	I _{WR1} -I _{WR2}	-1		+1	%
—	Write current setting range				10	mA
V _{sat}	Output saturation voltage	I _{OL} =7.5mA			4	V
I _{OH1}	Off-state leakage current	Head on non-selected side			250	μA
I _{OH2}	Off-state leakage current	Head on selected side			100	μA
V _{OL}	Output saturation voltage	I _{OL} =70mA			0.6	V
I _{OH}	Output leakage current	V _{OH} =20V			250	μA
I _{er}	Erase current range				100	mA
V _{OH}	Output voltage at time of write selection		10.2	11.3		V
V _{OL}	Output voltage at time of read selection			2		V
—	Output current range				110	mA
t _{w1}	M/M-1 output pulse width	R ₁ =100kΩ, C ₁ =0.01μF, k≐0.28		285		μs
t _{w2}	M/M-2 output pulse width	R ₂ =82kΩ, C ₂ =0.033μF, k≐0.27		730		μs
t _{w3}	M/M-3 output pulse width	R ₃ =100kΩ, C ₃ =0.12μF, k≐0.29		3.7		ms
t _{w5}	M/M-5 output pulse width	R ₅ =100kΩ, C ₅ =2.2μF, k≐0.31		68		ms

FLEXIBLE DISK DRIVE (FDD) READ/WRITE AND LOGIC IC

ELECTRICAL CHARACTERISTICS (CONTINUED) ($T_a=25^\circ\text{C}$, $V_{CC}=5\text{V}$, $V_{BB}=12\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{OL}	"L" output voltage	$I_{OL}=48\text{mA}$			0.4	V
I_{OH}	"H" output current	$V_{OH}=12\text{V}$			250	μA
—	Applicable pins: ④④, ④⑥, ④⑦, ④⑧, ④⑨					
V_{OL}	"L" output voltage	$I_{OL}=12\text{mA}$			0.4	V
I_{OH}	"H" output current	$V_{OH}=12\text{V}$			250	μA
—	Applicable pins: ④⑩					
V_{OH}	"H" output voltage	$I_{OH}=1\text{mA}$	2.7	3.4		V
V_{OL}	"L" output voltage	$I_{OL}=0.1\text{mA}$			0.4	V
—	Applicable pins: ④⑦, ④⑧, ④⑨, ④⑩ Totem-pole output					
V_{th+}	Positive-going threshold voltage		1.2	1.6	1.9	V
V_{th-}	Negative-going threshold voltage		0.5	0.8	1.1	V
H_{ys}	Hysteresis		0.4	0.8		V
I_{IH}	"H" input current	$V_i=2.7\text{V}$			40	μA
I_{IL}	"L" input current	$V_i=0.4\text{V}$			-0.4	mA
—	Applicable pins: ④①, ④②, ④③, ④⑦, ④⑧, ④⑨, ④⑩, ④⑪, ④⑫					
V_{IH}	"H" input voltage		2.0			V
V_{IL}	"L" input voltage				0.8	V
I_{IH}	"H" input current	$V_i=2.4\text{V}$			40	μA
I_{IL}	"L" input current	$V_i=0.4\text{V}$			-0.4	mA
—	Applicable pins: ④⑩, ④⑪					
V_{th5}	5V-system detection voltage		3.4	3.9	4.4	V
V_{th12}	12V-system detection voltage		7.6	8.6	9.6	V
I_{CC}	5V-system circuit current	Read mode		68		mA
I_{BB}	12V-system circuit current	Read mode		18		mA

M51660L

SERVO MOTOR CONTROL FOR RADIO CONTROL

DESCRIPTION

The M51660L is a semiconductor integrated circuit for use in servo motor control in radio control applications.

Housed in a 14-pin molded plastic zig-zag inline package (ZIP), the M51660L contributes to the miniaturization of the set.

The built-in voltage regulating circuit, and the differential comparator used in the comparator circuit provide the M51660L with extremely stable power supply voltage fluctuation characteristics and temperature change characteristics.

FEATURES

- Small circuit current3.5mA typ. (When output is off)
- Excellent power supply and temperature stability
- Simple setting of dead band
- Includes protection circuit for continuous "H" level input

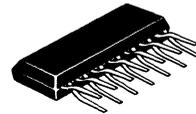
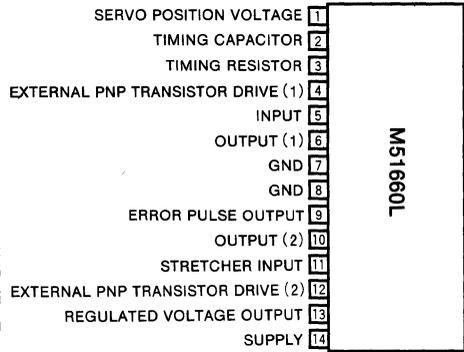
APPLICATION

Digital proportional systems for radio control, servo motor control circuits.

RECOMMENDED OPERATING CONDITIONS

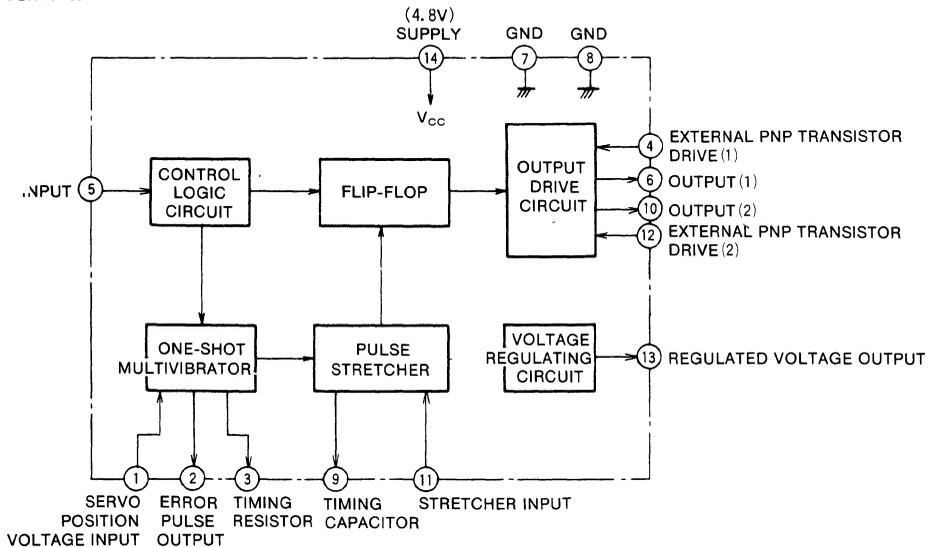
Supply voltage range 3.5~7V
 Rated supply voltage 4.8V

PIN CONFIGURATION (TOP VIEW)



14-pin molded plastic ZIP

BLOCK DIAGRAM



SERVO MOTOR CONTROL FOR RADIO CONTROL

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

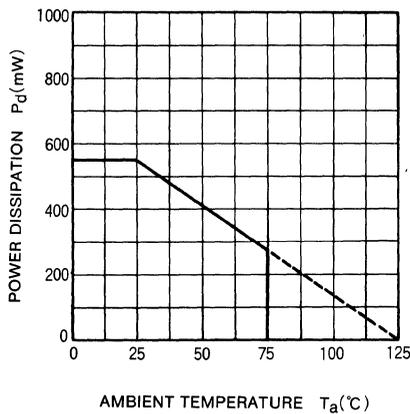
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		7.5	V
$I_{O\ SINK}$	Output sink current		500	mA
$I_{O\ SOURCE}$	Output source current		200	mA
P_d	Power dissipation		550	mW
K_θ	Thermal derating range	$T_a \geq 25^\circ\text{C}$	5.5	mW/°C
T_{opr}	Operating temperature		-20~+75	°C
T_{stg}	Storage temperature range		-40~-+125	°C

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=4.8\text{V}$ unless otherwise noted)

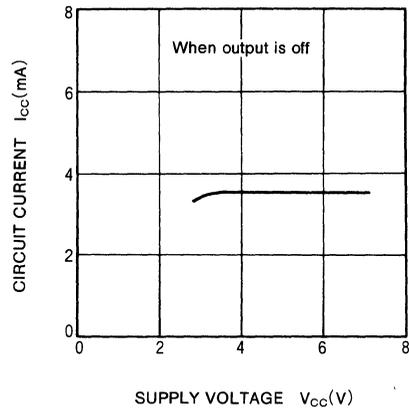
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current	When output is off		3.5	5	mA
		When output is on		20		
V_{OL}	Output voltage "L"	$I_{O\ SINK}=100\text{mA}$		0.1	0.2	V
		$I_{O\ SINK}=400\text{mA}$		0.4	0.7	
V_{OH}	Output voltage "H"	$I_{O\ SOURCE}=100\text{mA}$	3.4	3.8		V
I_{PNP}	External PNP transistor drive current		30			mA
V_{REG}	Internal regulated supply voltage		2.3	2.45	2.6	V
I_{REG}	Internal regulated supply output current				3.0	mA
T_{DB}	Minimum dead band width	$R_{DB}=510\Omega$, $C_S=0.1\mu\text{F}$			1.5	μs

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

THERMAL DERATING (MAXIMUM RATING)

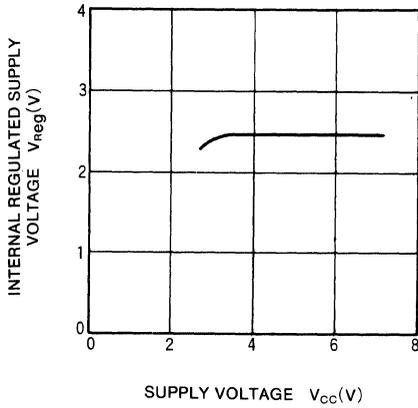


CIRCUIT CURRENT VS SUPPLY VOLTAGE

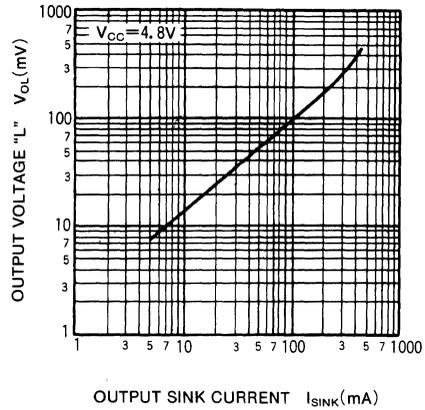


SERVO MOTOR CONTROL FOR RADIO CONTROL

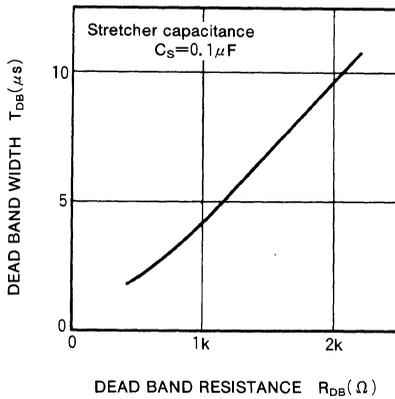
INTERNAL REGULATED SUPPLY VOLTAGE VS SUPPLY VOLTAGE



OUTPUT VOLTAGE "L" VS OUTPUT SINK CURRENT

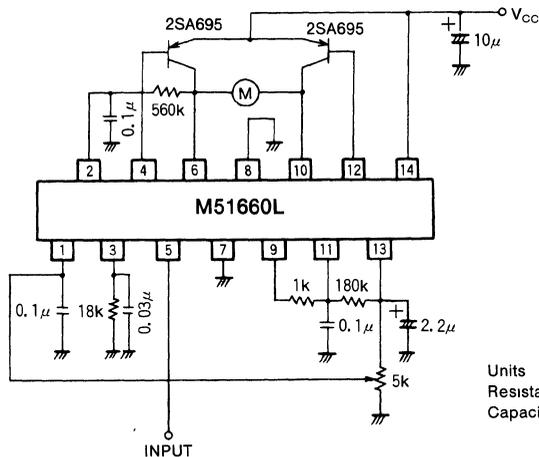


DEAD BAND WIDTH VS DEAD BAND RESISTANCE



APPLICATION EXAMPLE

Servo motor control circuit for radio control



Units
Resistance : Ω
Capacitance : F

SERVO MOTOR CONTROL FOR RADIO CONTROL

TECHNICAL APPLICATION NOTES**PIN DESCRIPTION****1. Servo Position Voltage Input Pin (Pin ①)**

Connect the potentiometer terminal for position detection that follows the output axis. Compare this voltage with the voltage of the triangular wave of pin ② and drive the motor. A capacitor of approximately $0.1\mu\text{F}$ should be connected for noise prevention.

2. Timing Capacitor Pin (Pin ②)

Connect a capacitor that will generate a triangular wave by constant current charging. A typical value is $0.1\mu\text{F}$. Also connect a feedback resistor from the output here.

3. Timing Resistor (pin ③)

Connect a resistor that will determine the value of the constant current of pin ②. A resistor of $18\text{k}\Omega$ will yield a current of 1.0mA . A capacitor of approximately $0.03\mu\text{F}$ should be connected in parallel with the resistor to increase stability.

4. External PNP Transistor Drive ① (Pin ④)

Connect to the base of the external PNP transistor.

5. Input Pin (Pin ⑤)

Operate with a positive pulse of peak value 3V or greater.

6. Output ① Pin (Pin ⑥)

Connect a feedback resistor between this pin and pin ②.

7. Ground (pins ⑦ and ⑧)**8. Error Pulse Output pin (Pin ⑨)**

Connect a resistor between this pin and pin ⑩. The dead band will change according to the value of this resistor.

9. Output ② pin (Pin ⑩)

This is the output ② pin.

10. Stretcher Input Pin (Pin ⑪)

Connect the capacitor and resistor of the pulse stretcher section.

11. External PNP Transistor Drive ② (Pin ⑫)

Connect to the base of the external PNP transistor.

12. Regulated Voltage Output Pin (Pin ⑬)

This is the output of the internal regulated supply voltage. Make connections from this pin to a potentiometer or pulse stretcher resistor. Connect a capacitor of approximately $2.2\mu\text{F}$ for stability.

13. Supply Voltage (Pin ⑭)

The supply voltage exhibits uniform characteristics from 3.5V to 7V . Connect a capacitor of approximately $10\mu\text{F}$.

MITSUBISHI LINEAR ICs

M51523AL

DUAL ELECTRONIC VOLUME CONTROL

DESCRIPTION

The M51523AL is a semiconductor integrated circuit containing a dual channel electronic volume control housed in a miniature 14-pin molded plastic zig-zag inline (ZIP) package.

Application of a DC voltage to the control pins permits the attenuation of the left and right channels to be changed as well as changes in the balance. Also included is a temperature-compensated reference voltage supply, enabling the M51523AL to be used as a control voltage source.

FEATURES

- High attenuation capacity92dB(typ.)
($f=1\text{kHz}$, $V_i=150\text{mVrms}$ JIS-A network)
- Low distortion 0.015%(typ.)
($f=1\text{kHz}$, $V_i=150\text{mVrms}$ at maximum volume)
- Low noise $3.6\mu\text{Vrms}$ (typ.)
(JIS-A network at minimum volume)
- Good matching between ch1 and ch2
- Includes a balance circuit
- Built-in stabilized power supply circuit makes device immune to fluctuations in supply voltage
- Good temperature characteristics

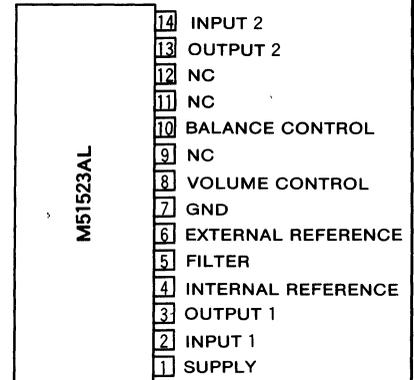
APPLICATION

Volume controls for car stereos, radio cassette recorders, TVs, VTRs, etc.

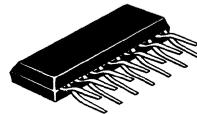
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range8~16V
- Rated supply voltage12V

PIN CONFIGURATION (TOP VIEW)

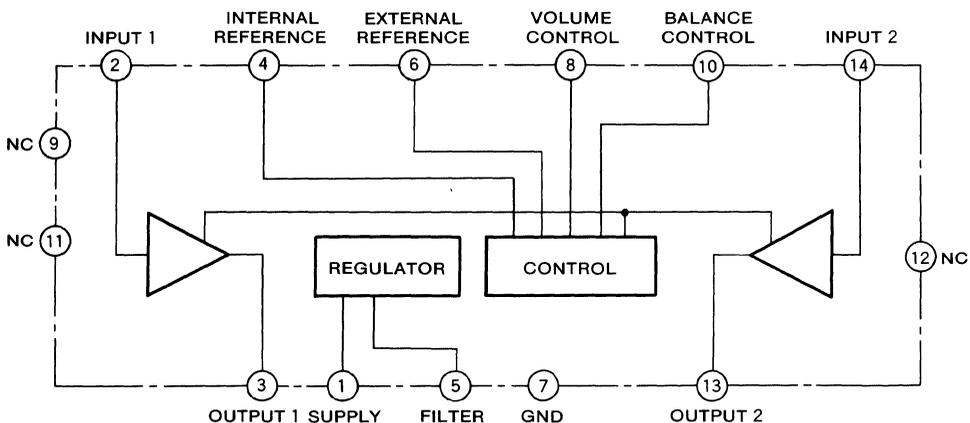


NC : NO CONNECTION



14-pin molded plastic ZIP

BLOCK DIAGRAM



DUAL ELECTRONIC VOLUME CONTROL

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

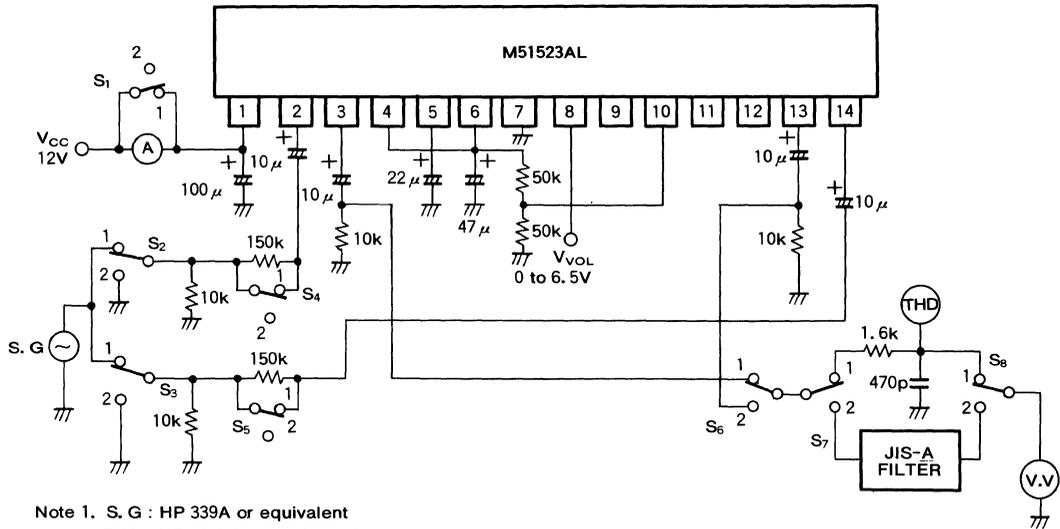
Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage	Quiescent	18	V
I_{CC}	Circuit current		30	mA
P_d	Power dissipation		550	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	5.5	mW/°C
T_{opr}	Operating temperature range		-20~+75	°C
T_{stg}	Storage temperature range		-40~+125	°C

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, $f=1\text{kHz}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CCO}	Quiescent circuit current	$V_{VOL}=0\text{V}$, $V_I=0$	7	12	20	mA
ATT	Attenuation level	$V_{VOL}=0\text{V}$, $V_I=150\text{mVrms}$	83	92		dB
C. B	Channel balance	$V_{VOL}=2.8\text{V}$, $V_I=1\text{Vrms}$	-3	0	3	dB
THD	Total harmonic distortion	$V_{VOL}=6.5\text{V}$, $V_I=150\text{mVrms}$		0.015	0.1	%
R_i	Input resistance	$V_{VOL}=6.5\text{V}$, $V_I=1\text{Vrms}$	50	150		k Ω
$V_{i(max)}$	Maximum input voltage	THD=1%	1.0	1.5		Vrms
N_o	Output noise voltage	$V_i=0$, JIS-A network		3.6	10	μVrms

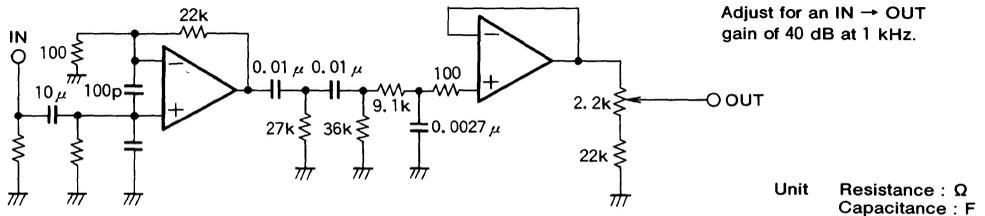
DUAL ELECTRONIC VOLUME CONTROL

TEST CIRCUIT



- Note 1. S. G : HP 339A or equivalent
 THD : HP : 339A or equivalent
 V. V . Kikusui Model 1635 or equivalent
- Note 2. A low-noise power supply should be used. (Less than $2\mu\text{V}$)

● Where the specified JIS-A filter is not used, the following circuit can be substituted. Note that output is increased by a factor of 100.



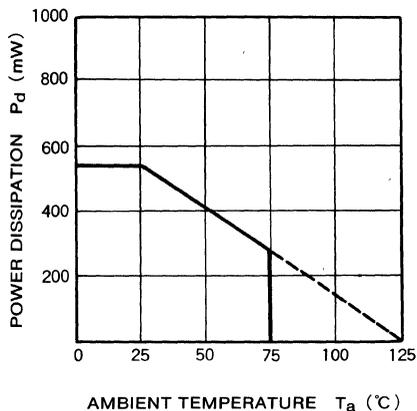
TEST METHODS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, $f=1\text{kHz}$, unless otherwise noted)

Parameter	Switch condition								Method
	S1	S2	S3	S4	S5	S6	S7	S8	
I_{CCO}	2	2	2	1	1		1	1	Measure with ammeter
ATT	1	1	1	1	1	1	1	1	Vary V_{VOL} from 0 to 6.5V and calculate using $ATT=20 \log(V_O/V_I)$ dB
C. B	1	1	1	1	1	1	1	1	Channel balance at $V_{VOL}=2.8\text{V}$
THD	1	1	1	1	1	1	1	1	At $f=1\text{kHz}$, $V_i=150\text{mVrms}$, and maximum volume, measure with distortion meter
R_i	1	1	1	1→2	1	1	1	1	Taking output V_{O1} at $S_4=1$, and output V_{O2} at $S_4=2$, calculate using $R_i=150/(V_{O1}/V_{O2}-1)$ kΩ
$V_{i(max)}$	1	1	1	1	1	1	1	1	At $f=1\text{kHz}$, the input voltage required to produce a THD of 1% at maximum volume
N_O	1	2	2	1	1	1	2	2	AT minimum volume level, $R_g=10\text{k}\Omega$, JIS-A filter

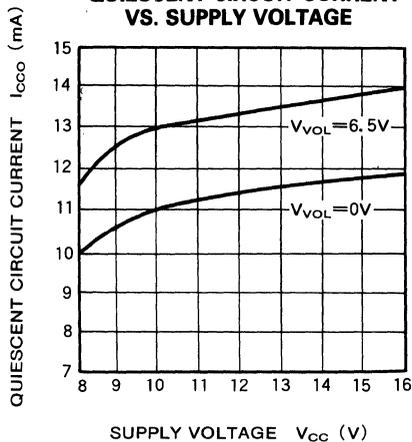
DUAL ELECTRONIC VOLUME CONTROL

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, $f=1\text{kHz}$, unless otherwise noted)

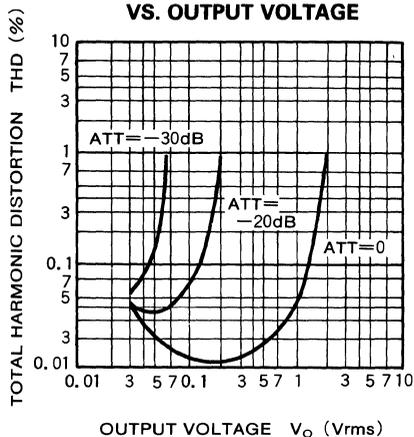
THERMAL DERATING (MAXIMUM RATING)



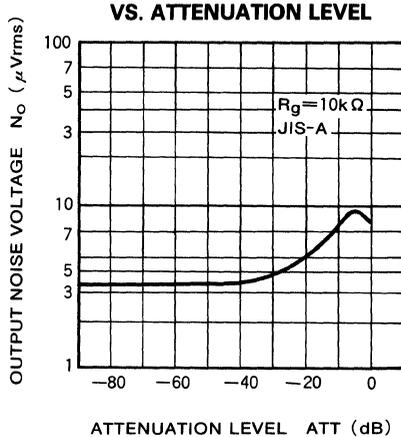
QUIESCENT CIRCUIT CURRENT VS. SUPPLY VOLTAGE



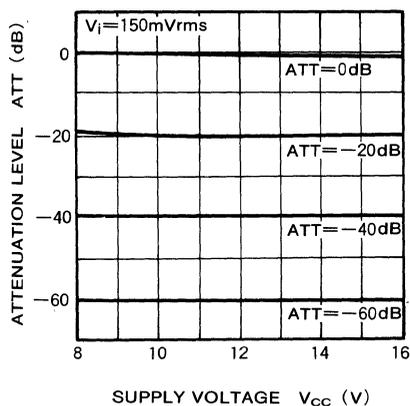
TOTAL HARMONIC DISTORTION VS. OUTPUT VOLTAGE



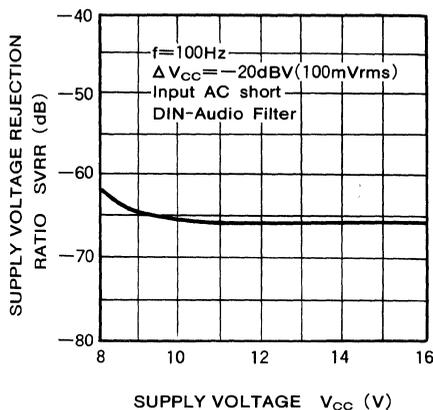
OUTPUT NOISE VOLTAGE VS. ATTENUATION LEVEL



ATTENUATION LEVEL VS. SUPPLY VOLTAGE

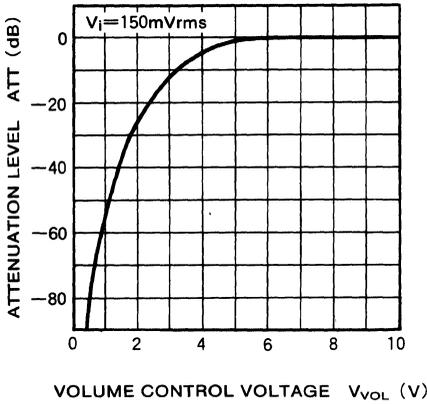


SUPPLY VOLTAGE REJECTION RATIO VS. SUPPLY VOLTAGE

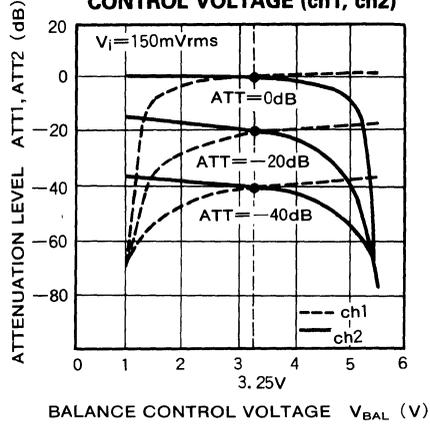


DUAL ELECTRONIC VOLUME CONTROL

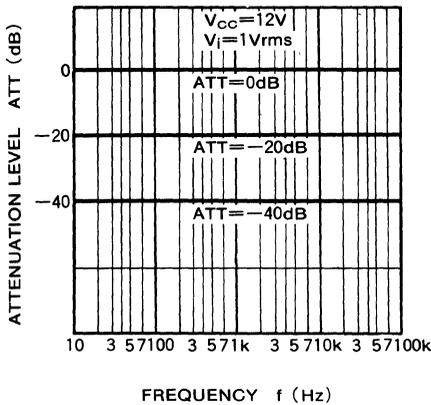
ATTENUATION LEVEL VS. VOLUME CONTROL VOLTAGE



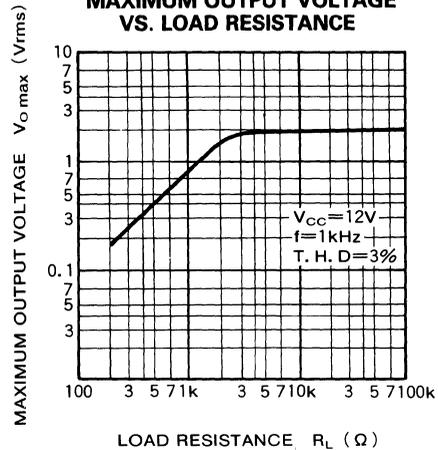
ATTENUATION LEVEL VS. BALANCE CONTROL VOLTAGE (ch1, ch2)



ATTENUATION LEVEL VS. FREQUENCY



MAXIMUM OUTPUT VOLTAGE VS. LOAD RESISTANCE

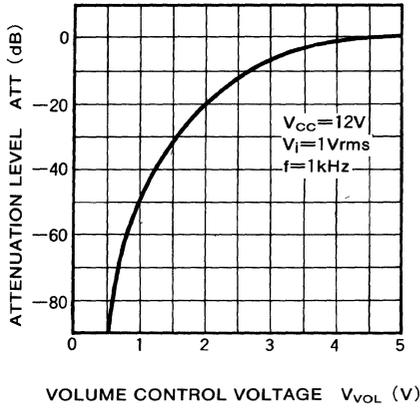


The attenuation characteristics and balance characteristics will change with the application of voltage to Pin ⑥. When the volume control voltage is equal to the Pin ⑥ voltage, there will be maximum volume. When the balance control voltage is half of the Pin ⑥ voltage, the balance will be centered.

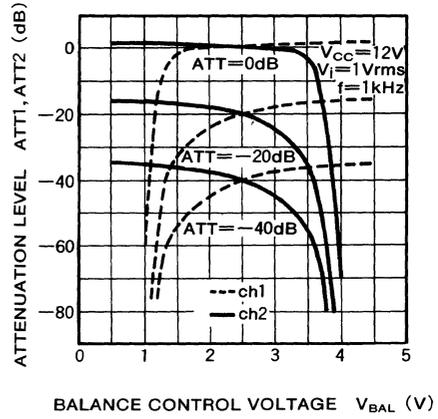
The attenuation level characteristics and balance characteristics for an applied voltage of 5V to Pin ⑥ are shown in the following graphs.

DUAL ELECTRONIC VOLUME CONTROL

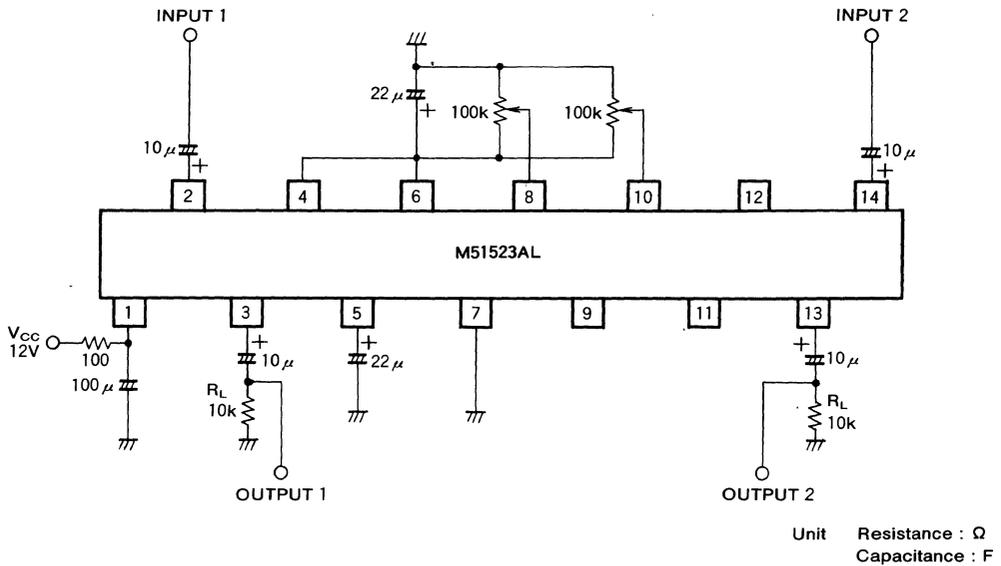
ATTENUATION LEVEL VS. VOLUME CONTROL VOLTAGE



ATTENUATION LEVEL VS. BALANCE CONTROL VOLTAGE (ch1, ch2)



APPLICATION EXAMPLE



M5172L

ZERO-CROSS POINT SWITCH

DESCRIPTION

The M5172L is a semiconductor integrated circuit designed for use in zero-point ignition temperature control circuits. It consists of a rectifier circuit, zero-point synchronous pulse generator circuit, temperature adjustment circuit using a differential amplifier, and a pulse generator circuit that is used in safety circuit.

The built-in zero-point ignition circuit and differential amplifier can operate directly from commercial power supply voltage through a resistor of 10k (at 100Vrms AC), permitting the M5172L to be widely applied in temperature control circuits using thyristors.

FEATURES

- Can be driven directly from commercial power supply voltage (100Vrms AC)
- Built-in zero-point ignition control circuit
- Can compensate for line voltage and line frequency fluctuations
- Includes a pulse generator circuit for a safety circuit

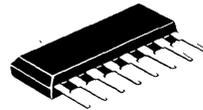
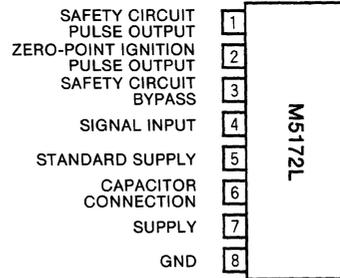
APPLICATION

Temperature control circuit for electric blankets, zero-point ignition circuit for thyristors, and all kinds of temperature control circuits.

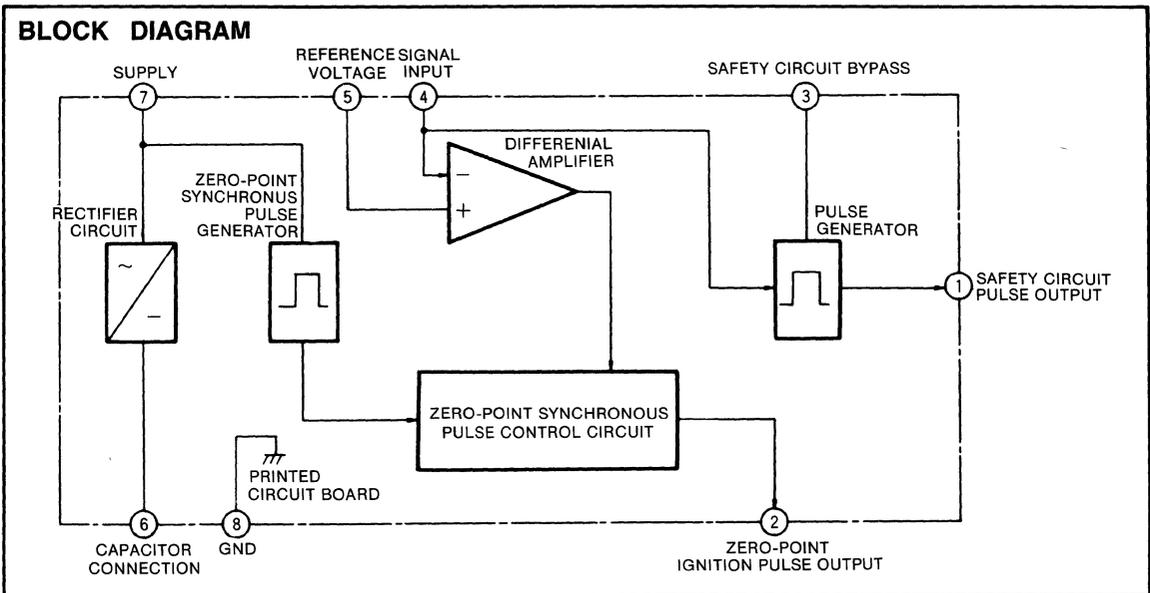
RECOMMENDED OPERATING CONDITIONS

AC supply voltage range 90~110Vrms(50~60Hz)
 Rated AC supply voltage 100Vrms(50~60Hz)
 (Note that a resistor of 10k or greater ($\geq 2W$) should be connected between pin ⑦ and the AC supply voltage.)

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic SIP



ZERO-POINT IGNITION TEMPERATURE CONTROL CIRCUIT

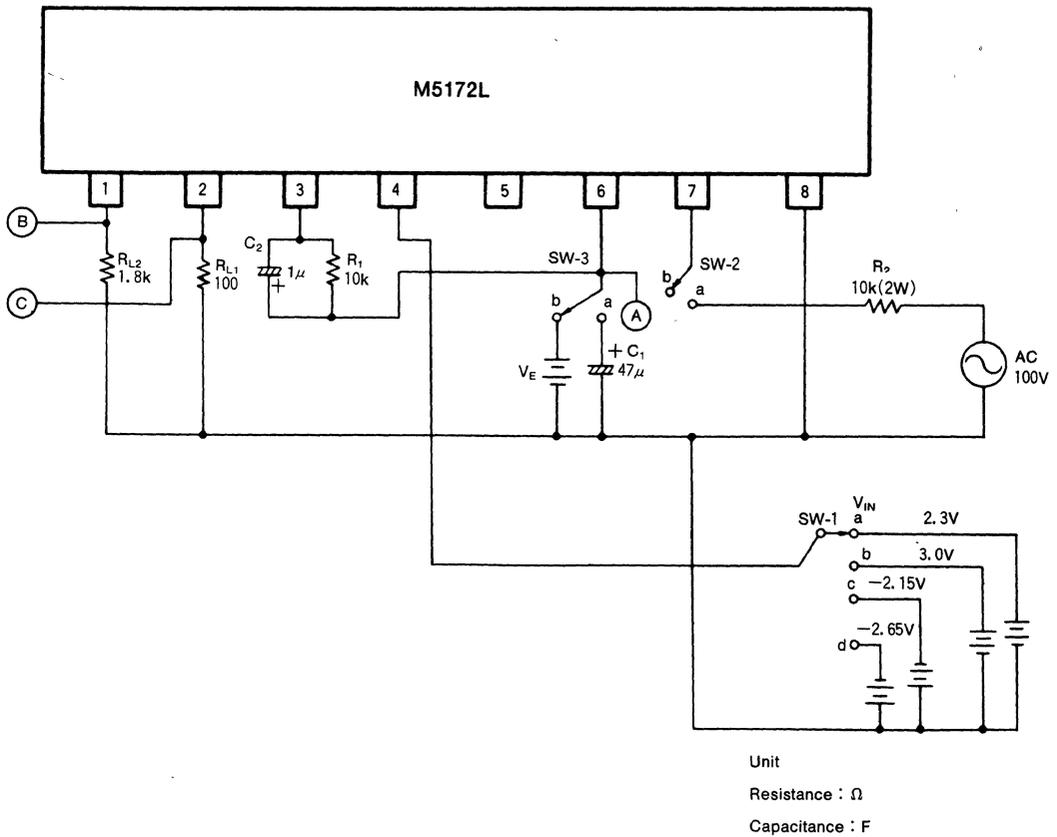
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage (between pins ⑦ and ⑧)		10	V
$I_{O⑦}$	Pin ⑦ sink current		10	mA
P_d	Power dissipation		360	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	3.5	mW/ $^\circ\text{C}$
T_{opg}	Operating temperature range		-20~+60	$^\circ\text{C}$
T_{stg}	Storage temperature range		-20~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{DC}	Rectification current (between pins ⑥ and ⑧)	$C_1=47\mu\text{F}$, $R_2=10\text{k}\Omega$	5.85		6.9	V
V_{TH-T}	Differential amplifier ON level	$V_E=5.9\text{V}$	2.3	2.7	3.0	V
V_{TH-S}	Safety circuit ON level	$V_E=5.9\text{V}$	-2.65	-2.4	-2.15	V
$V_{OH(T)}$	Zero-point synchronous pulse peak value	$R_{L1}=100\Omega$, $V_E=5.9\text{V}$	0.65			V
$V_{OH(S)}$	Safety circuit output pin "H" level	$R_{L2}=1.8\text{k}\Omega$, $V_E=5.9\text{V}$	0.59			V

TEST CIRCUIT



ZERO-POINT IGNITION TEMPERATURE CONTROL CIRCUIT

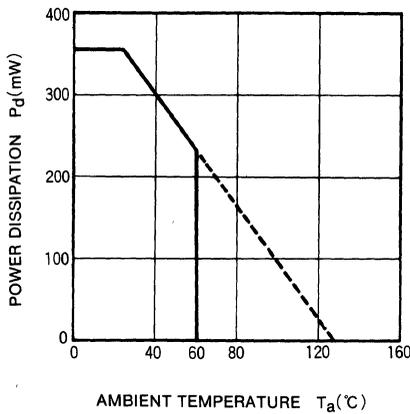
TEST METHODS

Symbol	SW-1	SW-2	SW-3	Measurement point
V_{DC}	a	a	a	A
V_{TH-T}	a b	b	b	C
V_{TH-T}	c d	b	b	B
$V_{OH(T)}$	a	b	b	C
$V_{OH(S)}$	d	b	b	B

TYPICAL CHARACTERISTICS

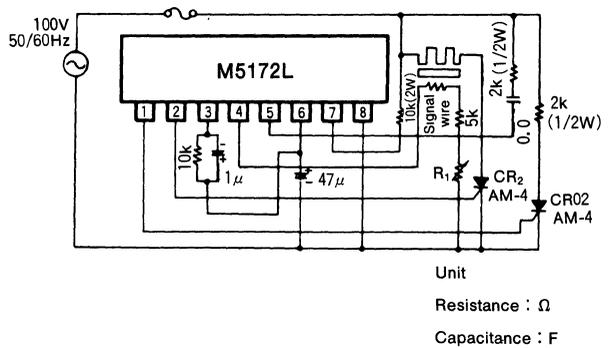
($T_a=25^\circ\text{C}$, unless otherwise noted)

**THERMAL DERATING
(MAXIMUM RATING)**

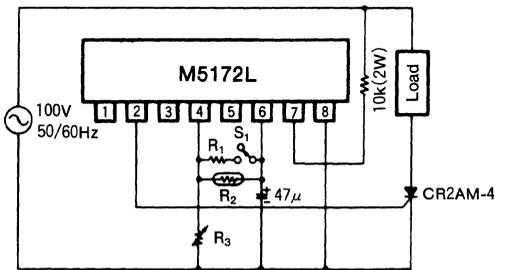


APPLICATION EXAMPLES

(1) Electric Blanket Temperature Control Circuit



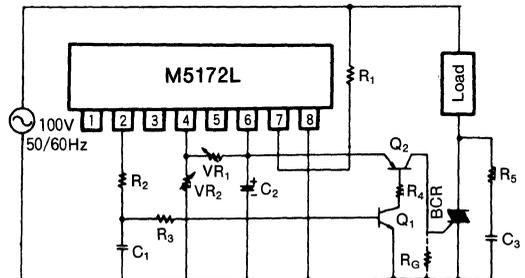
(2) Thyristor Zero-Point Ignition Circuit



R_2 : NTC thermistor 10k
 S_1 : OFF Only when thermistor used
 S_1 : ON Linear compensation of thermistor
 R_1 : 10k Ω

Unit
Resistance : Ω
Capacitance : F

(3) BCR Zero-Point Ignition Circuit



R_1 : 10k Ω (2W) R_2 : 1k Ω (1/4W) R_3 : 10k Ω (1/4W)
 R_4 : 1k Ω (1/4W) R_5 : 100 Ω (1/2W) C_1 : 0.068 μF (50WV)
 C_2 : 220 μF (25WV) C_3 : 0.1 μF (400WV) Q_1 : 2SC712-D
 Q_2 : 2SA696-D BCR : BCR3AM~BCR25A

M51743P

SEQUENCE CONTROLLER

DESCRIPTION

The M51743P is a semiconductor integrated circuit designed for use as sequence controller for oil combustion systems. It consists of five comparators and two voltage regulation circuits.

Three of the comparators can be used for sensor signal monitoring and two can be used for driving relays, etc.

Since one of the comparators for drive applications has a redundant "2" structure, the M51743P is especially suitable for use in oil combustion system, for which safety operation is required.

FEATURES

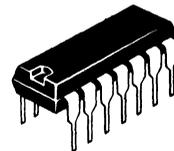
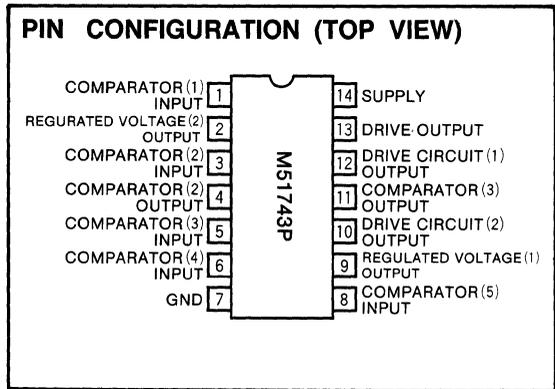
- Includes five comparators (three for signal monitoring and two for drive purposes)
- Includes two voltage regulation circuits
.....5.5V (typ), 6.2V (typ)
- Can directly drive relays, lamps, etc.
- Built-in memory function (comparator (2))
- Built-in hysteresis function (comparators (1), (4), (5))
- Redundancy "2" safety circuit design (comparator (5))
- Built-in reference power supply reduces the number of external parts
- Wide supply voltage range 8.5~20V

APPLICATION

Sequence controller (in oil and gas combustion systems), flame detection circuits.

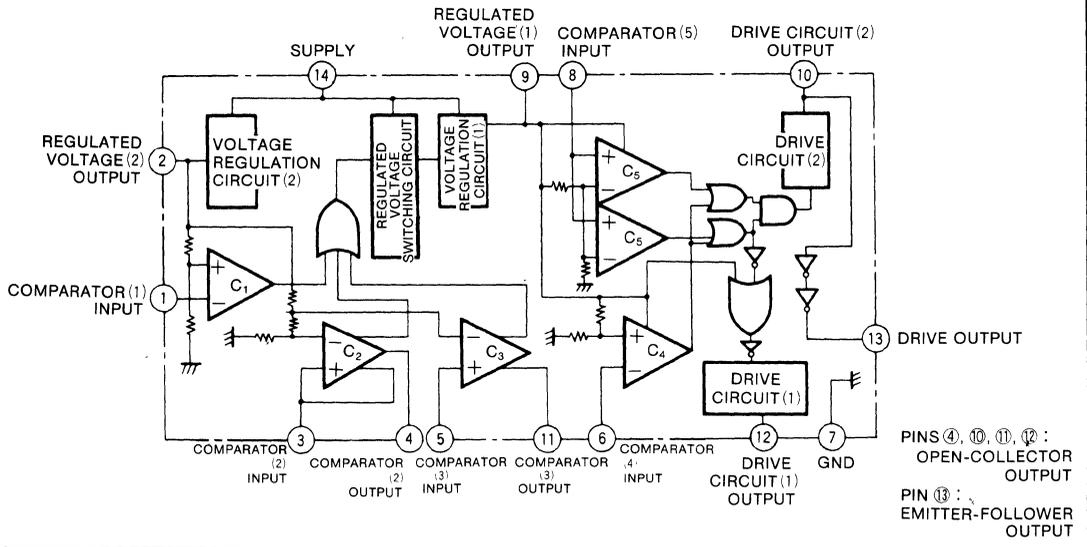
RECOMMENDED OPERATING CONDITIONS

- Supply voltage range 8.5~20V
- Rated supply voltage 12V



14-pin molded plastic DIP

BLOCK DIAGRAM



SEQUENCE CONTROLLER

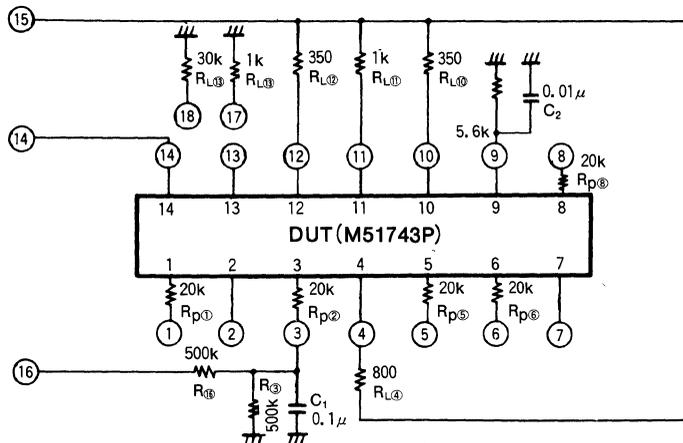
ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		20	V
I _{CC}	Circuit current		22	mA
I _{L④}	Pin ④ maximum current	*Instantaneous peak value	70	mA
I _{L⑩}	Pin ⑩ maximum current		60	mA
I _{L⑬}	Pin ⑬ maximum current		60	mA
P _d	Power dissipation		900	mW
K _θ	Thermal derating	T _a ≥25°C	9	mW/°C
T _{opr}	Operating temperature range		-20~+65	°C
T _{stg}	Storage temperature range		-40~+125	°C

ELECTRICAL CHARACTERISTICS (T_a=25°C, V_{CC}=15V)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{O (1)}	Voltage regulation circuit (1) output	Pin ⑩, pin ⑬ ON state	5.0	5.5	6.3	V
V _{O (2)}	Voltage regulation circuit (2) output		5.0	6.1	6.9	V
I _{CC (I)}	Circuit current (I)	V _⑩ =1.2V, V _⑬ =3V	6	14	20	mA
I _{CC (II)}	Circuit current (II)	V _⑩ =4.5V	2	10	17	mA
V _{ON (1)}	Comparator (1) ON voltage		3.8	4.4	5.3	V
V _{OFF (1)}	Comparator (1) OFF voltage		3.0	3.5	4.6	V
I _{IN (1)}	Comparator (1) input current	V _⑩ =5.1V		0.15	0.6	μA
V _{ON (2)}	Comparator (2) ON voltage		1.5	2.0	2.8	V
V _③	Pin ③ clamp voltage	R _③ =500kΩ	4.9	5.75	6.3	V
V _{sat ④}	Pin ④ output saturation voltage	R _{L④} =800Ω	0.9	1.12	1.5	V
V _{ON (3)}	Comparator (3) ON voltage		2.5	3.3	4.1	V
I _{IN (3)}	Comparator (3) input current	V _⑩ =4V		0.15	0.5	μA
V _{sat ⑩}	Pin ⑩ output saturation voltage	R _{L⑩} =1kΩ, V _⑬ =4.1V	1.3	1.83	3.0	V
V _{REF I (4)}	Comparator (4) reference voltage (I)		3.3	3.84	4.5	V
V _{REF II (4)}	Comparator (4) reference voltage (II)		2.9	3.55	3.9	V
I _{IA (4)}	Comparator (4) input current	V _⑩ =2.5V		50	250	nA
V _{ON (5)}	Comparator (5) ON voltage		2.0	2.3	2.7	V
V _{OFF (5)}	Comparator (5) OFF voltage		1.6	2.0	2.3	V
I _{IN (5)}	Comparator (5) input current	V _⑩ =2.5V		0.1	0.35	μA
V _{sat ⑩ (1)}	Pin ⑩ output saturation voltage (1)	R _{L⑩} =350Ω, V _{CC} =10V		0.3	0.9	V
V _{sat ⑬ (1)}	Pin ⑬ output saturation voltage (1)	R _{L⑬} =350Ω, V _{CC} =10V		0.3	0.9	V
V _{OH ⑬}	Pin ⑬ output voltage	R _{L⑬} =1kΩ	7.5			V

TEST CIRCUIT



Unit
Resistance : Ω
Capacitance : F

SEQUENCE CONTROLLER

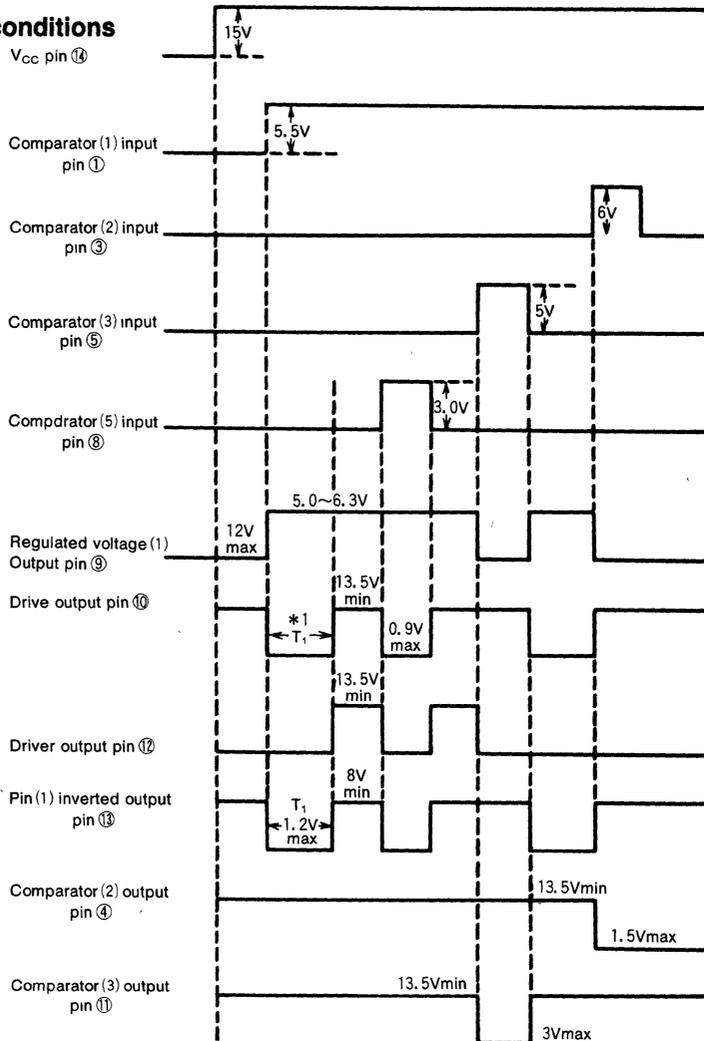
TEST METHODS

Symbol	Pin No.																		Measurement point
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	
V _o (1)	②*1	①	GND		GND	1.2V	GND	GND	Ⓜ*2				⑰	15V	15V		⑬		⑨
V _o (2)	5.2	Ⓜ				1.2V							⑰	15V			⑬		②
I _{cc} (1)						1.2V		3V					⑱	Ⓜ 15V				⑬	⑭
I _{cc} (II)						4.5V							⑱	Ⓜ 15V				⑬	⑭
V _{on} (1)	1.3V 3.8V					1.2V			Ⓜ				⑰	15V			⑬		⑨
V _{off} (1)	4.6V 3.0V								Ⓜ										⑨
I _{in} (1)	Ⓜ 5.1V		GND																①
V _{on} (2)	②	①				1.2V			Ⓜ										⑨
V _Ⓝ			Ⓜ					GND											③
V _{sat} ④				Ⓜ	GND	1.2V		GND											④
V _{on} (3)			GND		4.1V 2.5V				Ⓜ										⑨
I _{in} (3)					Ⓜ 4V														③
V _{sat} ⑩					4V	1.2V					Ⓜ								⑪
V _{REF I} (4)					GND	4.1V 3.3V				Ⓜ									⑩
V _{REF II} (4)						3.9V 2.9V				Ⓜ				15V					⑩
V _{sat} ⑫(1)						4.5V		2.6V				Ⓜ	⑰	10V					⑫
V _{OH} ⑬								2.6V					⑰ Ⓜ	15V					⑬
I _{in} (4)						Ⓜ 2.5V							⑰						④
V _{on} (5)						4.5V		2.7V 2.0V					Ⓜ						⑫
V _{off} (5)								2.3V 1.6V					Ⓜ						⑫
I _{in} (5)								2.5V						15V					⑤
V _{sat} ⑯(1)	②	①	GND		GND	4.5V	GND	2.6V		Ⓜ			⑰	10V	15V		⑬		⑩

* 1 : ○ indicates pin number
 * 2 : Ⓜ indicates measurement pin

M51743P OPERATION TIME CHART

Input application conditions



Output waveform

* 1 : T_1 is the timer output determined by the external connection of pin 6

M51743P INPUT/OUTPUT TRUTH TABLE
Operation comparators (1), (2), (3)

	Pin No.	State			
Input	Pin 1	0*1	1	1	1
	Pin 3	0	0	1	0
	Pin 5	0	0	0	1
	Pin 9	0	1	0	0
Output	Pin 10	1*2	0*4	1	1
	Pin 12	0*2	0*4	0	0
	Pin 13	1*3	0*4	1	1
	Pin 4	1*2	1	0	1
	Pin 11	1*2	1	1	0

Operation comparators (4), (5)

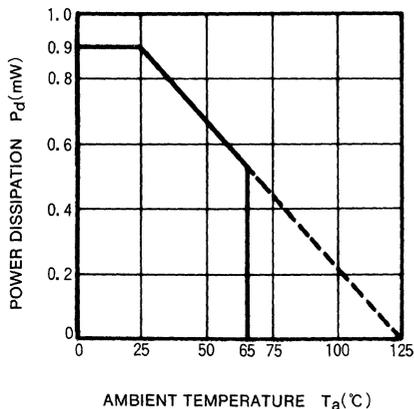
	Pin No.	State							
Input	Pin 9	1	1	1	1	0	0	0	0
	Pin 6	0	1	0	1	0	1	—	—
	Pin 8	0	0	1	1	—	—	0	1
Output	Pin 10	0	1	0	0	1	1	1	1
	Pin 12	0	1	0	0	0	0	0	0
	Pin 13	0	1	0	0	1	1	1	1

- * 1 : "1" and "0" indicate the "H" and "L" voltage levels of the pins
- * 2 : Pins 10, 12, 4, 11 are ON at "0" and OFF at "1" (open-collector output)
- * 3 : Pin 13 is ON at "1" and OFF at "0" (emitter-follower output)
- * 4 : Will be 0 or 1 depending on the condition of comparators (4) and (5)

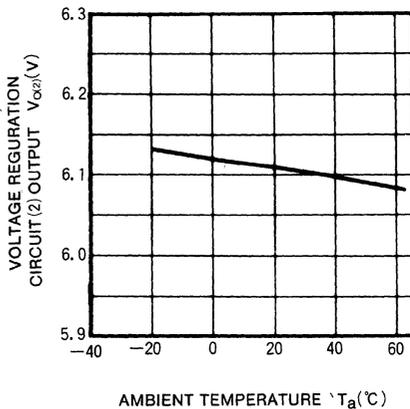
SEQUENCE CONTROLLER

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=9\text{V}$, unless otherwise noted)

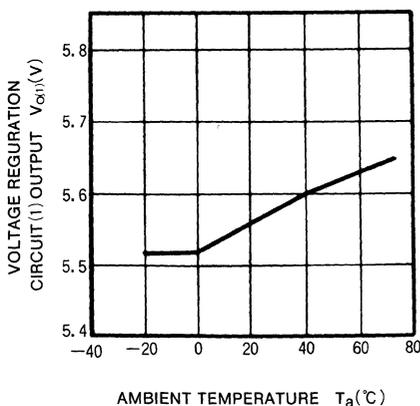
**THERMAL DERATING
(MAXIMUM RATING)**



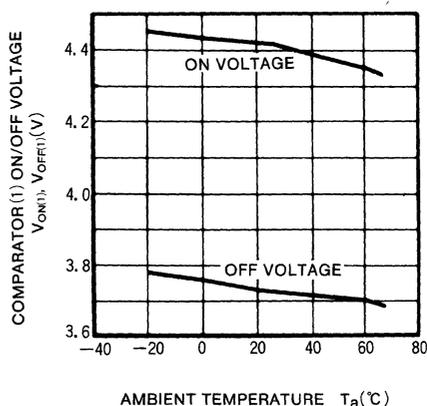
**VOLTAGE REGULATION CIRCUIT (2)
VS. AMBIENT TEMPERATURE**



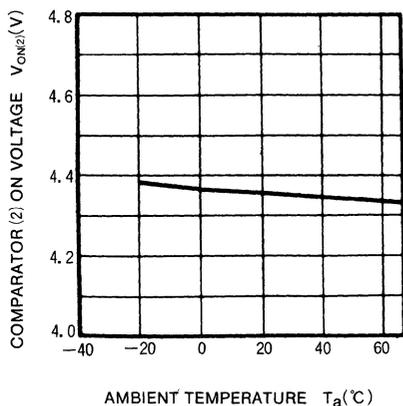
**VOLTAGE REGULATION CIRCUIT (1)
AMBIENT TEMPERATURE**



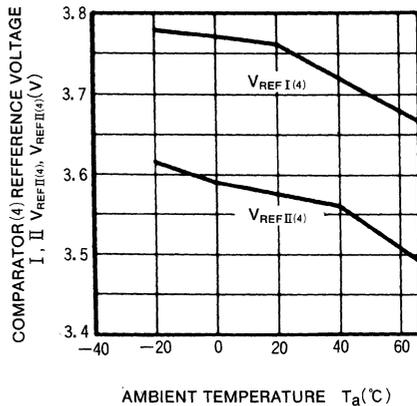
**COMPARATOR ON/OFF VOLTAGE
VS. AMBIENT TEMPERATURE**



**COMPARATOR (2) ON VOLTAGE
VS. AMBIENT TEMPERATURE**

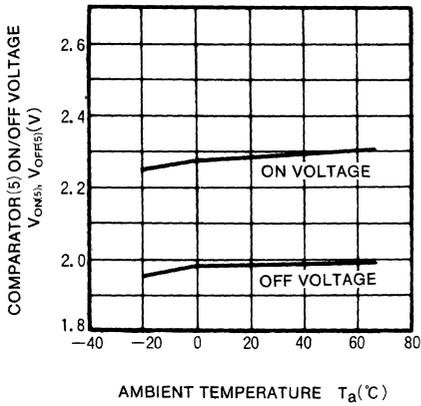


**COMPARATOR (4) REFERENCE VOLTAGE
I, II VS. AMBIENT TEMPERATURE**

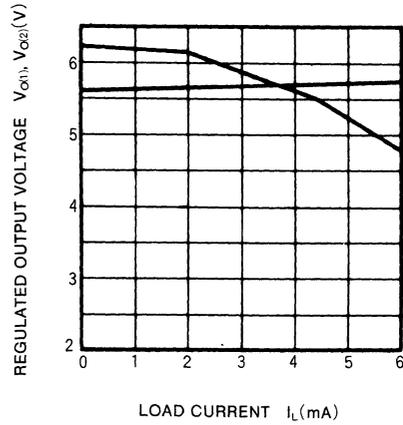


SEQUENCE CONTROLLER

COMPARATOR (5) ON/OFF VOLTAGE VS. AMBIENT TEMPERATURE

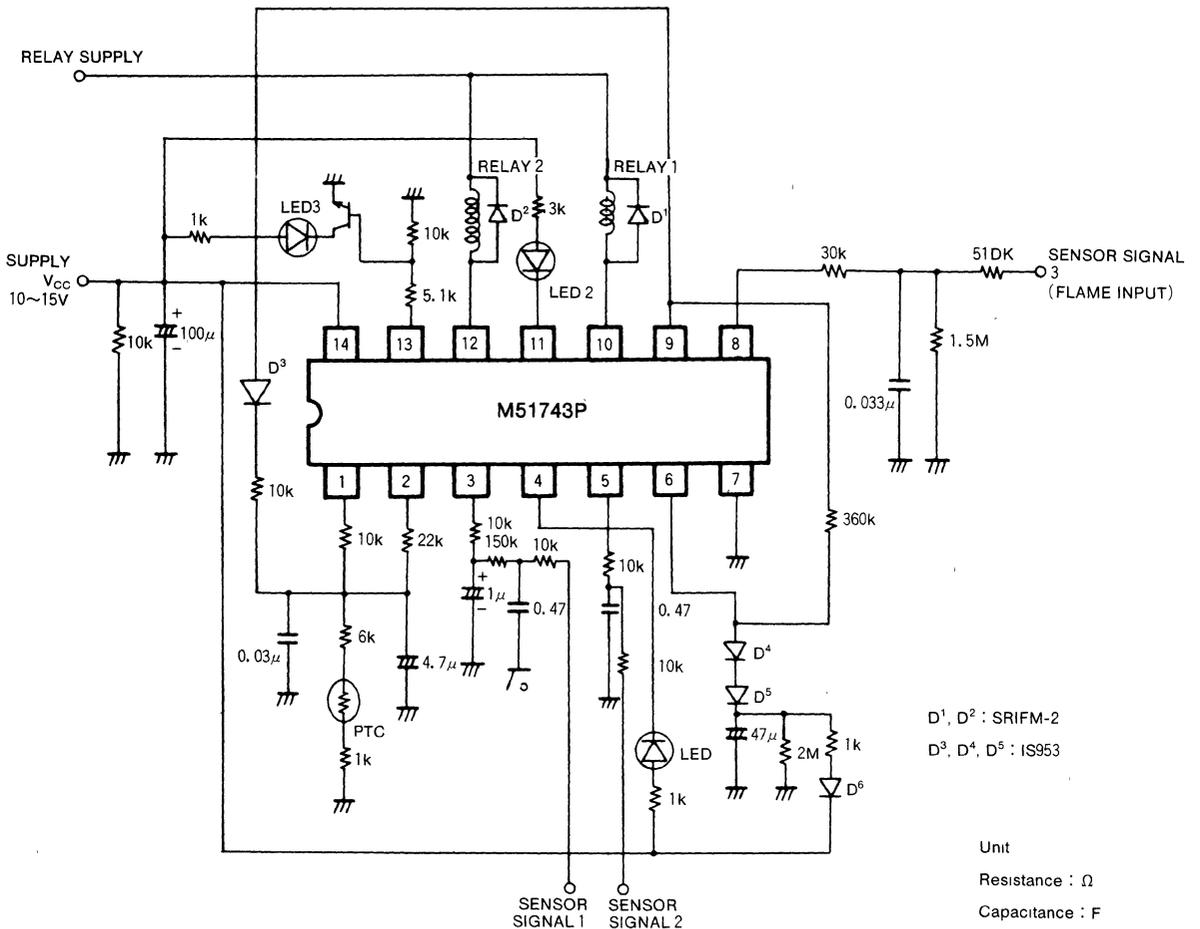


REGULATED OUTPUT VOLTAGE VS. LOAD CURRENT



APPLICATION EXAMPLE

Example of using comparator (4) as timer operation



M5174P

FLAME DETECTOR

DESCRIPTION

The M5174P is a semiconductor integrated circuit designed for use in a flame detection circuit.

It consists of a voltage regulation circuit, an input amplifier circuit that amplifies extremely small input currents and drives a thyristor, and a bias circuit that supplies current to a shunt circuit connected in parallel with the load (relay) and the externally-connected capacitor used as a gate trigger for the thyristor circuit. The thyristor circuit drives the external load (relay) and there are two shunt circuits that protect the load (relay) under abnormalities or at quiescence, in addition to an overcurrent protection circuit.

FEATURES

- Includes a fail-safe system (redundancy 2) that permits simple construction of a system featuring a high level of safety
- Includes a voltage regulation circuit and an over-current protection circuit
- Small threshold value of temperature fluctuation of the operating input current
 Characteristic value of the normal temperature 20%
 ($T_a = -20 \sim +60^\circ\text{C}$)
- High maximum input current $50\mu\text{A}$ (max)

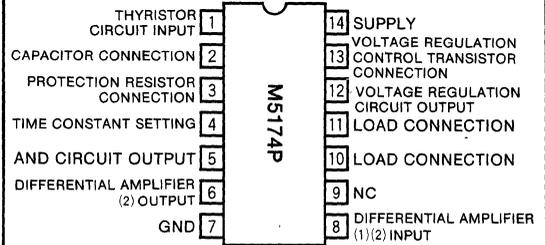
APPLICATION

Flame detection circuits (in oil and gas combustion systems), relay drivers.

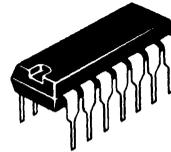
DESCRIPTION OF FALL-SAFE SYSTEM

1. This system uses two differential amplifiers and drives AND and OR circuits by two input signals. Noise on the input can lead to abnormal operation and surges, which can cause circuit destruction. This system can prevent

PIN CONFIGURATION (TOP VIEW)



NC : NO CONNECTION



14-Pin molded plastic DIP

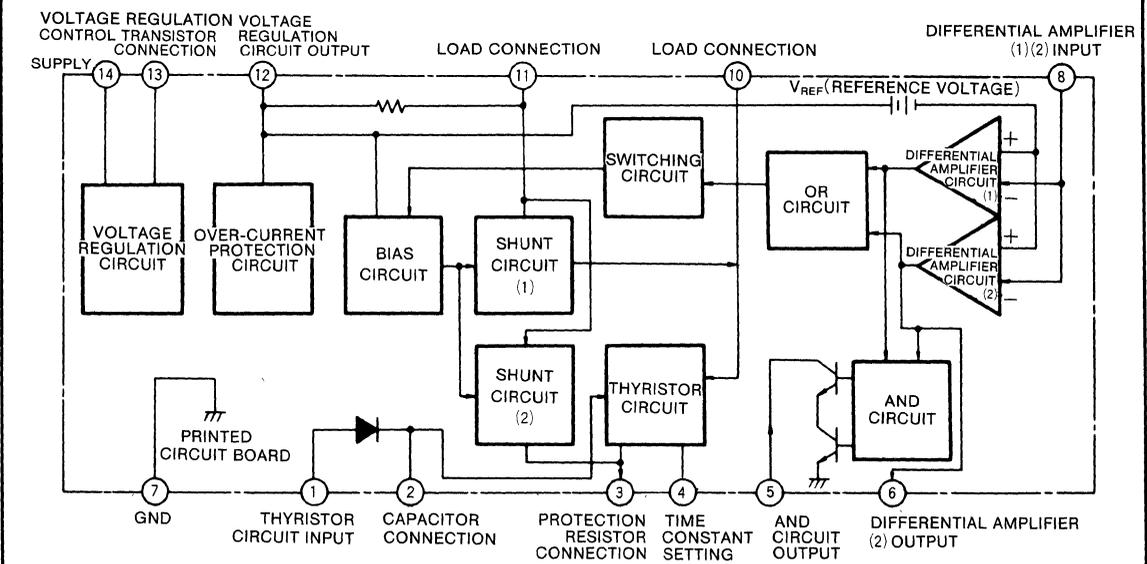
abnormal operation due to such events.

2. Even if the thyristor and AND circuits should be destroyed by a surge or some other cause, the shunt circuit will operate so that load (relay) dose not conduct.
3. The over-current protection circuit prevents the load (relay) form conducting even at times of abnormally high current flow.

RECOMMENDED OPERATING CONDITIONS

Supply voltage range $12 \sim 20\text{V}$
 Rated supply voltage 15V

BLOCK DIAGRAM



FLAME DETECTION CIRCUIT

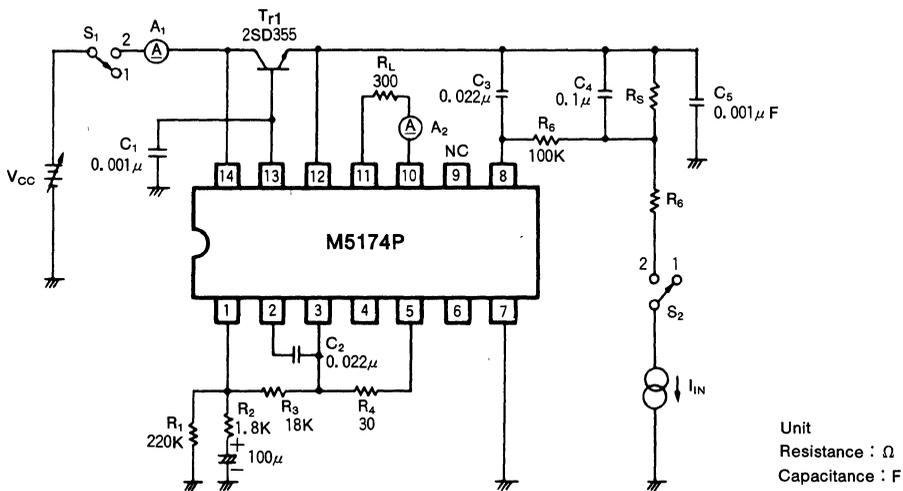
ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage	Stand-by	20	V
I_{CC}	Circuit current		40	mA
I_{IN}	Input current		50	μA
P_d	Power dissipation		650	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	6.5	mW/ $^\circ\text{C}$
T_{opg}	Operating temperature		-20~+60	$^\circ\text{C}$
T_{stg}	Storage temperature range		-40~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{CC1}	Supply voltage	Stand-by	12	15	20	V
V_{CC2}		Operating ($I_{CC} \leq 36\text{mA}$)	12	15	17	V
I_{CC1}	Quiescent circuit current	$V_{CC}=15\text{V}$	7	10	14	mA
I_{CC2}	Operating circuit current	$I_L=20\text{mA}$, $V_{CC}=17\text{V}$		30	36	mA
I_L	Load current	$R_L=300\Omega$, $V_{CC}=17\text{V}$	15			mA
V_{O1}	Voltage regulation circuit output voltage	Quiescent, $V_{CC}=12\text{V}$	9.5	10.3	11.5	V
V_{O2}		Operating, $V_{CC}=12\text{V}$	9.2	10	11.5	V
I_{IN}	Input ON current	$R_L=300\Omega$, $V_{CC}=17\text{V}$, $R_S=4.7\text{M}\Omega$	0.48	0.60	0.72	μA
$I_{IN(\text{max})}$	Maximum input current	$V_{CC}=17\text{V}$	50			μA
I_{LS1}	AND circuit short-circuit load current	Quiescent, $V_{CC}=15\text{V}$			5	mA
I_{LS2}	AND Circuit, thyristor circuit short-circuit load current	Quiescent, $V_{CC}=15\text{V}$			8	mA

TEST CIRCUIT



FLAME DETECTION CIRCUIT

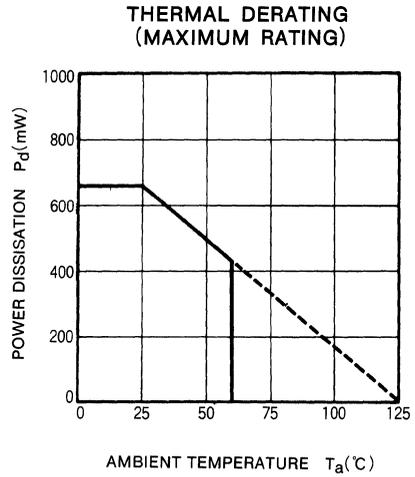
TEST CONDITION

Symbol	S ₁	S ₁	Measurement point
I _{CC1}	2	1	Note 3 : A ₁
I _{CC2}	2	2	A ₁
I _L	2	2	Note 2 : A ₂
V _{O1}	2	1	Pin (12)
V _{O2}	2	2	Pin (12)
I _{IN}	2	2	A ₂
I _{IN(max)}	2	2	A ₂
Note 3 : I _{LS1}	2	1	A ₂
Note 4 : I _{LS2}	2	1	A ₂

Note 1. A₁ : Current inflowing from supply V_{CC}
 Note 2. A₂ : Current inflowing from pin (11) to pin (10)
 Note 3. Short circuit condition between pins (5) and (7)
 Note 4. Short circuit condition between pins (5) and (7), (10) and (5)

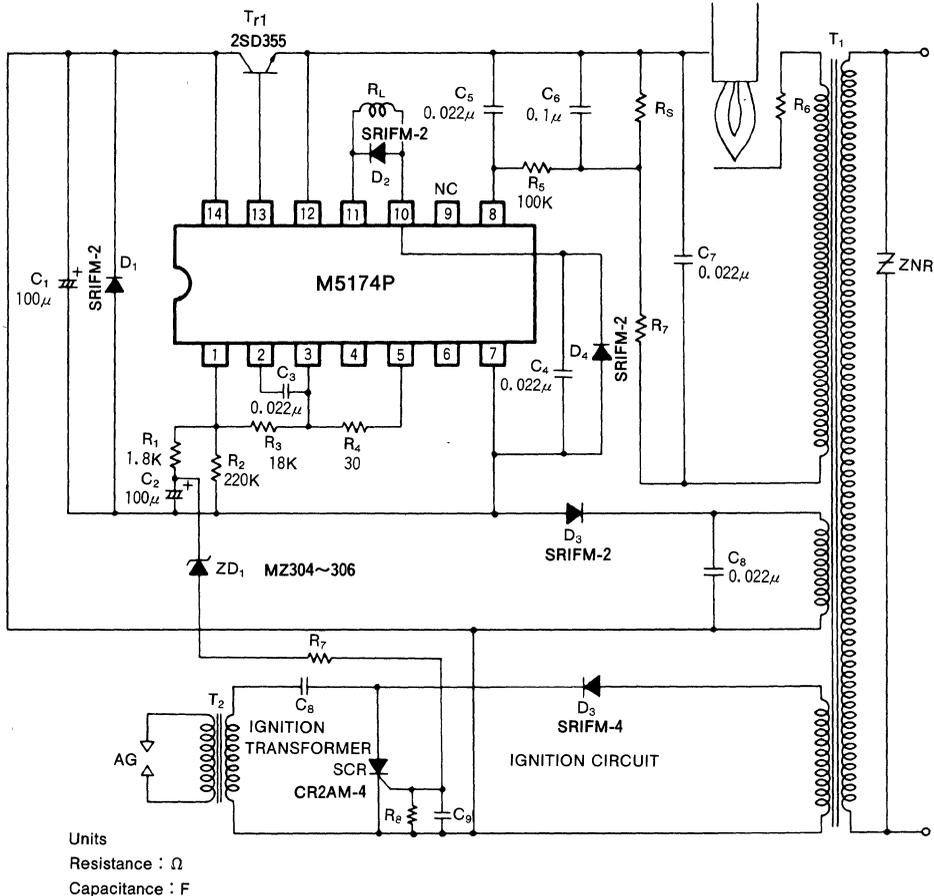
TYPICAL CHARACTERISTICS

(T_a=25°C, unless otherwise noted)



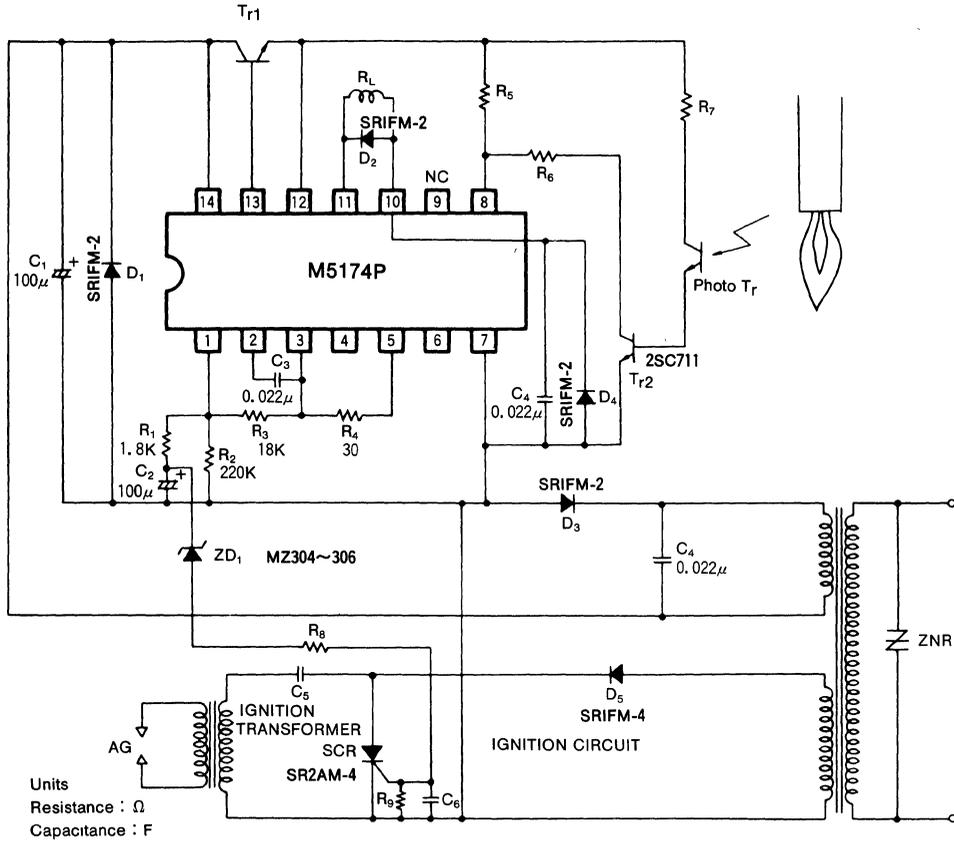
APPLICATION EXAMPLES

(1) Flame Detection Circuit Using Flame Current



FLAME DETECTION CIRCUIT

(2) Flame Detection Circuit Using a Photo Transistor

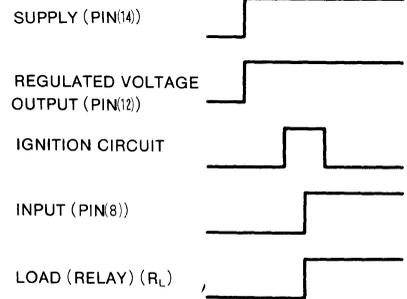


MAXIMUM VOLTAGE RATING FOR EACH PIN

Voltage is in reference to ground (pin (7))

Pin	Voltage rating		Pin condition														
	Negative	Positive	①	②	③	④	⑤	⑥	⑦	⑧	⑨	⑩	⑪	⑫	⑬	⑭	
①	-80V	50V	1. 8K Ω between GND, 100 μ F connected 0. 022 μ F connected between pin (3) 18K Ω connected between pin (1) Open 30 Ω connected between pin (3) Open GND 0. 022 μ F connected between pin (12) NC 0. 022 μ F connected between GND 300 Ω connected between pin (10) 0. 022 μ F connected between pin (8) 0. 001 μ F connected between GND Open														
②	-20V	60V															
③	-20V	30V															
④	-40V	70V															
⑤	-15V	40V															
⑥	-15V	10V															
⑦	Ground																
⑧	-80V	70V															
⑨	NC																
⑩	-20V	20V															
⑪	-20V	20V															
⑫	-10V	20V															
⑬	-20V	30V															
⑭	-20V	50V															

TIME SEQUENCE OUTLINE DIAGRAM



LOW POWER SUPPLY VOLTAGE INDICATOR, WARNING DRIVER

ABSOLUTE MAXIMUM RATINGS (T_a=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC1}	Supply voltage		-0.2~2.0	V
V _{CC2}	Supply voltage		-0.2~2.0	V
V _{CC3}	Supply voltage		-0.2~2.0	V
I _{AO}	Pin ② drive current	Output is saturated	300	mA
I _{L3}	Pin ④ drive current	Output is saturated	70	mA
P _d	Power dissipation		600	mW
T _{opr}	Operating temperature range		-15~+65	°C
T _{stg}	Storage temperature range		-40~+125	°C

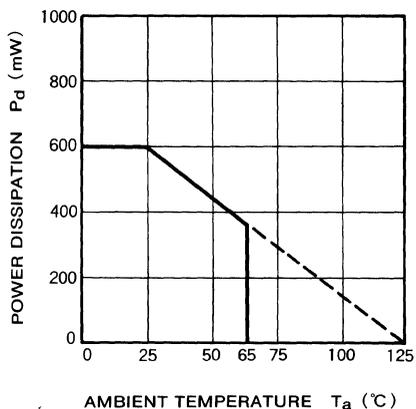
ELECTRICAL CHARACTERISTICS (T_a=25°C, V_{CC}=1.5V, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage		1.1		1.8	V
I _{ST1}	Standby current drain	V _{CC1} =1.5V, other pins open			5	μA
I _{ST2}		V _{CC1} , V _{CC2} =1.5V, other pins open			10	μA
I _{ST3}		V _{CC3} =1.5V, other pins open, L ₃ is "H"	420		630	μA
I _{opr1}	Operating current drain	V _{CC1} =1.5V, I _Q =350μA, other pins open		10	20	mA
I _{opr2}		V _{CC1} =1.5V, A12=GND, other pins open		10	20	mA
I _{opr3}		V _{CC2} =1.5V, M1, M2=GND, other pins open		2	4	mA
I _{opr4}		V _{CC3} =1.5V, M1, M2=GND, other pins open		5	10	mA
I _{IH1}	A11 input pin current	V _{CC1} =1.2V, A11=1.2V		450	700	μA
I _{IL1}	A11 input leakage current	V _{CC1} =1.2V, A11=GND	-1			μA
I _{IH2}	A12 input leakage current	V _{CC1} =1.2V, A11=1.2V			1	μA
I _{IL2}	A12 input pin current	V _{CC1} =1.2V, A12=GND	-4	-2		mA
I _{IHM1}	M1 input pin current	V _{CC1} , V _{CC2} =1.2V, M1=1.2V		50	75	μA
I _{ILM1}	M1 input leakage current	V _{CC1} , V _{CC2} =1.2V, M1=GND	-1			μA
I _{IHM2}	M2 input pin current	V _{CC1} , V _{CC2} =1.2V, M2=1.2V		50	75	μA
I _{ILM2}	M2 input leakage current	V _{CC1} , V _{CC2} =1.2V, M2=GND	-1			μA
I _{IHM1}	M1 input leakage current	V _{CC1} , V _{CC2} =1.2V, M1=1.2V			1	μA
I _{ILM1}	M1 input pin current	V _{CC1} , V _{CC2} =1.2V, M1=GND	-75	-50		μA
I _{IHM2}	M2 input leakage current	V _{CC1} , V _{CC2} =1.2V, M1=1.2V			1	μA
I _{ILM2}	M2 input pin current	V _{CC1} , V _{CC2} =1.2V, M2=GND	-75	-50		μA
V _{SAT1}	AO output saturation voltage	V _{CC1} =1.2V, R _L =6Ω		0.2	0.3	V
V _{SAT1'}	AO output saturation voltage	V _{CC1} =1.2V, R _L =13Ω		0.1	0.2	V
V _{SAT2}	L3 output saturation voltage	V _{CC3} =1.2V, R _L =25Ω		0.3	0.5	V
I _{OH1}	L1 output pin current	V _{CC2} =1.2V, L1=GND	-7.8	-6.0	-4.2	mA
I _{OH2}	C output pin current	V _{CC2} =1.2V, C=GND	-350	-270	-190	μA
I _{OL3}	L2 output pin current	L2=1.2V	190	270	350	μA
I _{LAO}	AO output leakage current	V _{CC1} =1.5V, A11=GND			1	μA

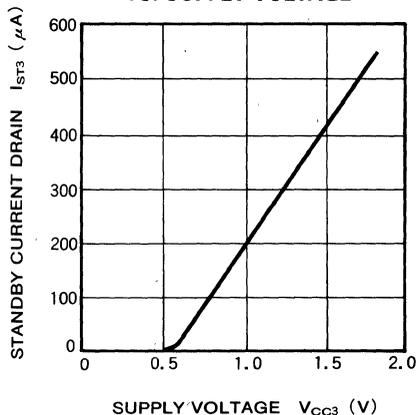
LOW POWER SUPPLY VOLTAGE INDICATOR, WARNING DRIVER

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

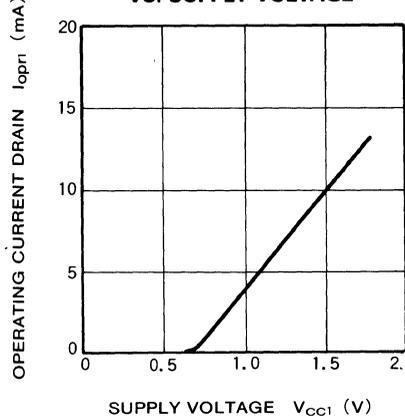
THERMAL DERATING (MAXIMUM RATING)



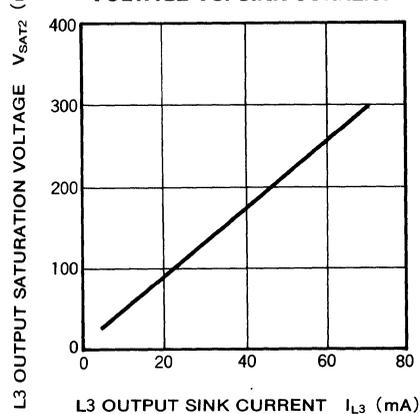
STANDBY CURRENT DRAIN VS. SUPPLY VOLTAGE



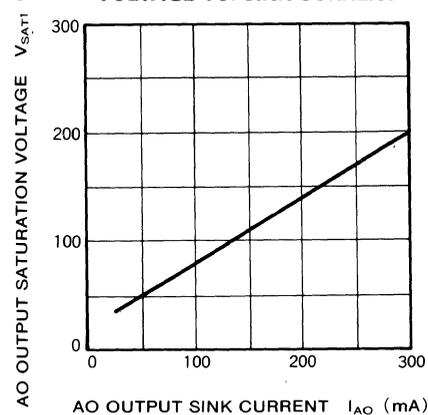
OPERATING CURRENT DRAIN VS. SUPPLY VOLTAGE



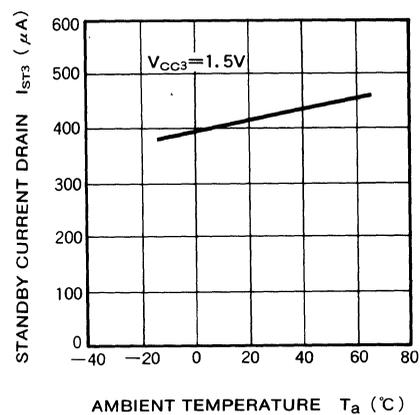
L3 OUTPUT SATURATION VOLTAGE VS. SINK CURRENT



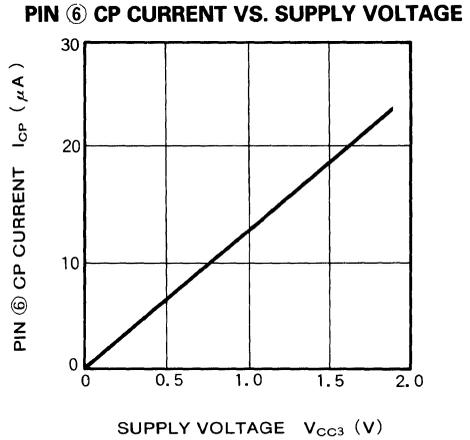
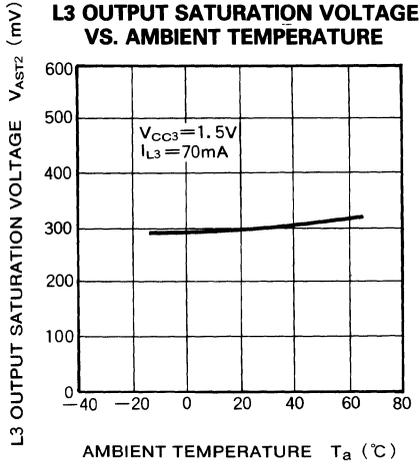
AO OUTPUT SATURATION VOLTAGE VS. SINK CURRENT



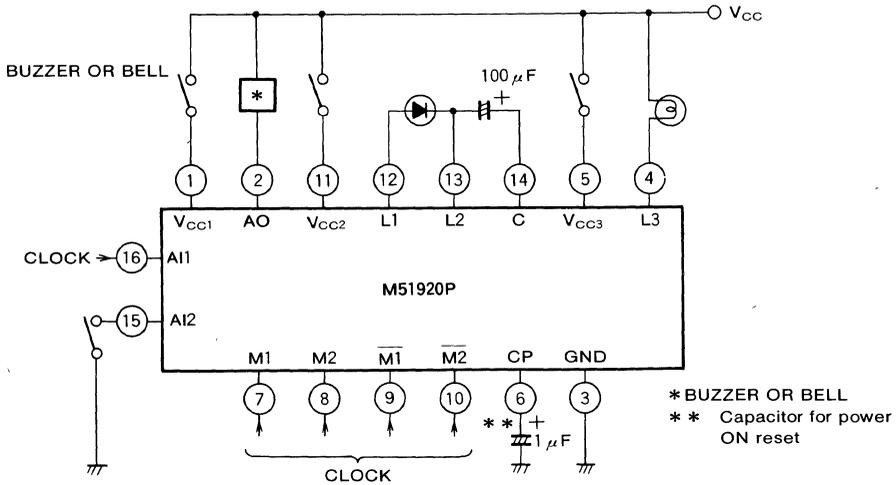
STANDBY CURRENT DRAIN VS. AMBIENT TEMPERATURE



LOW POWER SUPPLY VOLTAGE INDICATOR, WARNING DRIVER



APPLICATION EXAMPLE



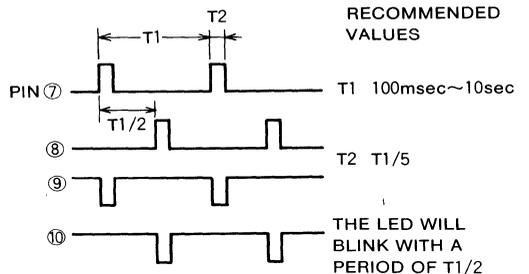
The power supply system is classified as follows: V_{CC1} for buzzer or bell, V_{CC2} for LED drive, and V_{CC3} for timer drive.

FUNCTIONAL DESCRIPTION

A clock signal applied to pin ⑩ will drive the Pin ② buzzer.

A low-level state of pin ⑮ will drive the Pin ② bell.

Applying a clock signal to pins ⑦ and ⑧, or ⑨ and ⑩ will cause the LED to blink on and off.



By applying a clock signal to pins ⑦ and ⑧, or ⑨ and ⑩, pin ④ will go "ON" for only a period of $8 \cdot T1$ from V_{CC3} ON; thereafter, pin ④ will be "OFF" until V_{CC3} is put on again.

M51950L

CAR AIR CONDITIONER CONTROLLER

DESCRIPTION

The M51950L is a semiconductor integrated circuit designed for use in automobile air conditioner controls.

This device receives a pulse signal (ignition pulse) of a frequency proportional to that of the number of revolutions of the engine, and a temperature detection signal from the thermistor; it generates a control signal that switches the air conditioner on and off.

The M51950L uses supply voltage and a detection circuit, which have little dependency on temperature, permitting highly accurate control.

In addition, housing the device in an SIP has made high-density mounting possible.

FEATURES

- Small number of external parts
- (Number of revolutions) detection supply voltage characteristics 0.1%/V (typ.)
- (Number of revolutions) detection temperature characteristics 100ppm/°C (typ.)
- Thermistor resistance detection supply voltage characteristics 0.02%/V (typ.)

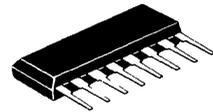
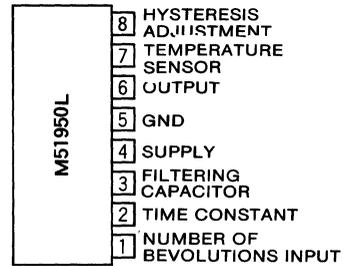
APPLICATION

Automobile air conditioners.

RECOMMENDED OPERATING CONDITIONS

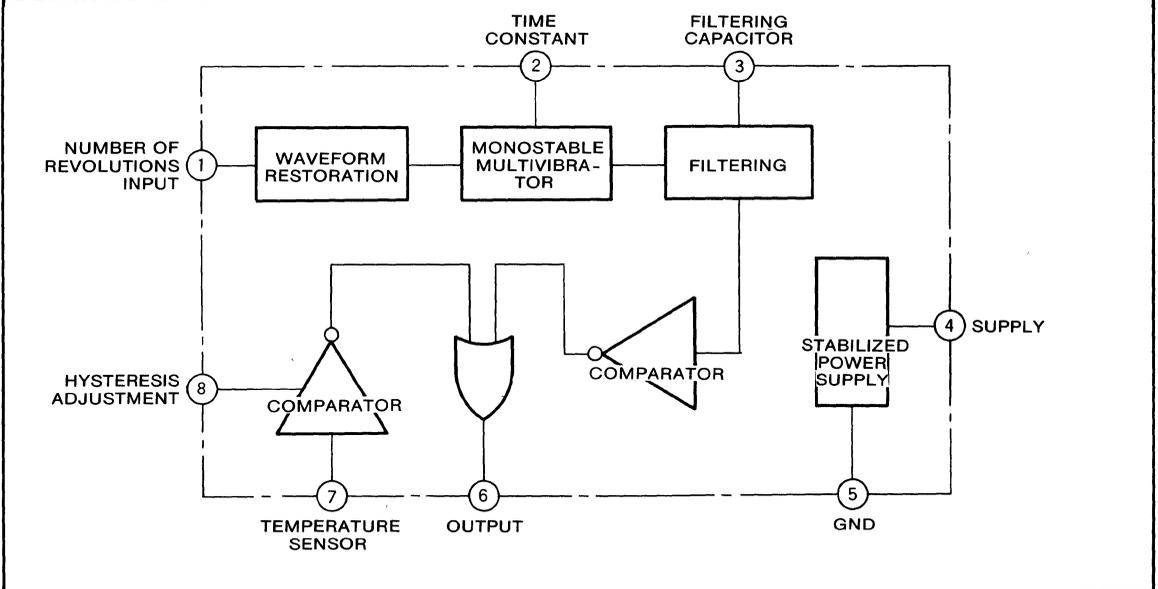
- Supply voltage range 10~17V
- Rated supply voltage 13.2V

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic SIP

BLOCK DIAGRAM



CAR AIR CONDITIONER CONTROLLER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^{\circ}\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
$V_{\text{⑥}}$	Pin ⑥ voltage	Output OFF	27	V
$I_{\text{⑥}}$	Pin ⑥ sink current	Output ON	30	mA
P_d	Power dissipation		360	mW
K_{θ}	Thermal derating	$T_a \geq 25^{\circ}\text{C}$	3.6	mW/ $^{\circ}\text{C}$
T_{opr}	Operating temperature range		$-20 \sim +70$	$^{\circ}\text{C}$
T_{stg}	Storage temperature range		$-40 \sim +125$	$^{\circ}\text{C}$

ELECTRICAL CHARACTERISTICS 1 ($V_{CC}=13.2\text{V}$, $T_a=25^{\circ}\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I_{CC}	Circuit current			3.5	8.0	mA
f_{ON}	Revolution detector ON frequency		330	376	430	Hz
Δf	Revolution detector ON, OFF frequency difference		40	65	90	Hz
R_{thON}	Thermo circuit ON resistance value		5.95	6.45	6.90	k Ω
ΔR_{th}	Thermo circuit ON, OFF resistance value difference		200	230	270	Ω
$V_{\text{⑥ON}}$	Pin ⑥ ON voltage			0.10	0.50	V
$V_{\text{⑥OFF}}$	Pin ⑥ OFF voltage		12.8	13.2		V
$I_{\text{②in}}$	Pin ② sink current		-100	0.02	100	nA
$I_{\text{③in}}$	Pin ③ sink current		75	130	200	μA
$I_{\text{⑦in}}$	Pin ⑦ sink current			50	500	nA
ΔV_{CH}	Chattering prevention circuit hysteresis width		180	215		mV

ELECTRICAL CHARACTERISTICS 2 ($V_{CC}=10 \sim 17\text{V}$, $T_a=25^{\circ}\text{C}$, unless otherwise noted)

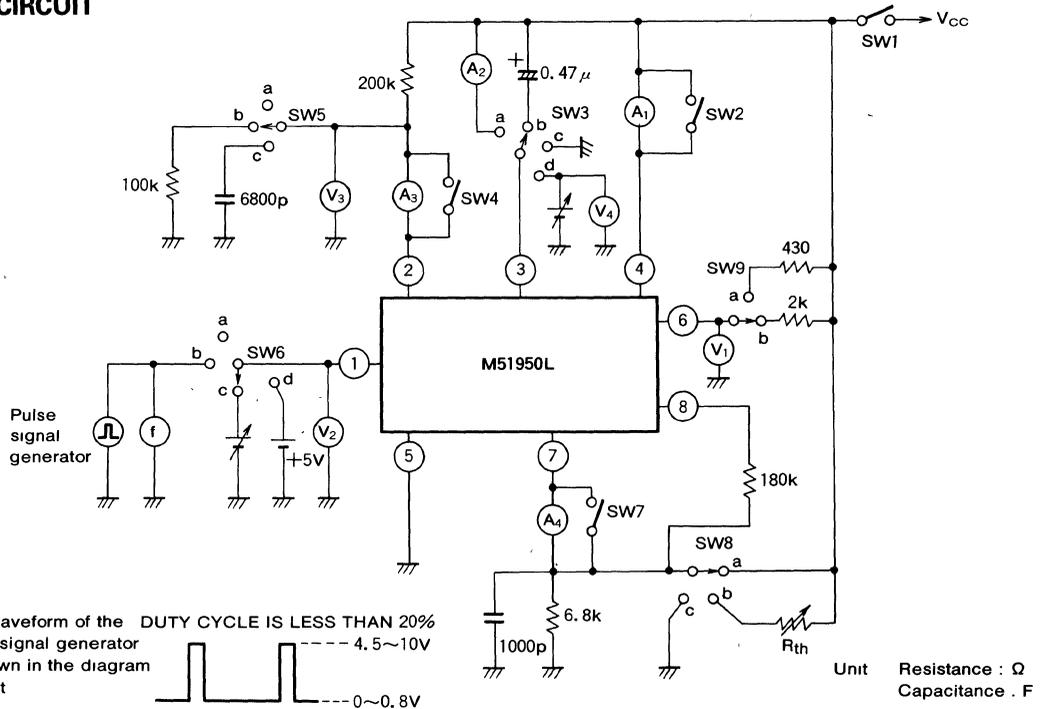
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{\text{①ON}}$	Pin ① ON level			3.0	4.0	V
$V_{\text{①OFF}}$	Pin ① OFF level		1.0	2.0		V
f_{ON}	Revolution detector ON frequency		330	376	430	Hz
Δf	Revolution detector ON, OFF frequency difference		40	65	90	Hz
R_{thON}	Thermo circuit ON resistance value		5.80	6.45	7.15	k Ω
ΔR_{th}	Thermo circuit ON, OFF resistance value difference		190	230	280	Ω

ELECTRICAL CHARACTERISTICS 3 ($V_{CC}=13.2\text{V}$, $T_a=10 \sim 60^{\circ}\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$V_{\text{①ON}}$	Pin ① ON level			3.0	4.5	V
$V_{\text{①OFF}}$	Pin ① OFF level		0.8	2.0		V
f_{ON}	Revolution detector ON frequency		320	377.5	440	Hz
Δf	Revolution detector ON, OFF frequency difference		35	65	95	Hz
R_{thON}	Thermo circuit ON resistance value		5.80	6.45	7.15	k Ω
ΔR_{th}	Thermo circuit ON, OFF resistance value difference		185	230	280	Ω

CAR AIR CONDITIONER CONTROLLER

TEST CIRCUIT



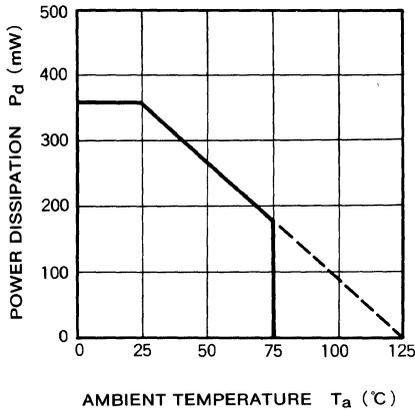
TEST METHODS

Symbol	Switch conditions									Measurement instrument	Method
	SW1	SW2	SW3	SW4	SW5	SW6	SW7	SW8	SW9		
I_{CC}	ON	OFF	b	ON	a	a	ON	c	b	A_1	
f_{ON}	ON	ON	b	ON	c	b	ON	a	b	f	While raising the pin ① input signal frequency, the value of f when V_1 changes from "H" to "L"
Δf	ON	ON	b	ON	c	b	ON	a	b	f	While lowering the pin ① input signal frequency, the difference between the values of f and f_{ON} when V_1 changes from "L" to "H"
R_{thON}	ON	ON	c	ON	a	a	ON	b	b		While decreasing R_{th} , the value of R_{th} when V_1 changes from "H" to "L"
ΔR_{th}	ON	ON	c	ON	a	a	ON	b	b		While increasing R_{th} , the difference between the values of R_{th} and R_{thON} when V_1 changes from "L" to "H"
$V_{⑥ON}$	ON	ON	c	ON	a	a	ON	a	a	V_1	
$V_{⑥OFF}$	ON	ON	c	ON	a	a	ON	c	a	V_1	
$I_{②in}$	ON	ON	c	OFF	b	d	ON	a	b	A_3	
$I_{③in}$	ON	ON	a	ON	b	d	ON	a	b	A_2	
$I_{⑦in}$	ON	ON	c	ON	a	a	OFF	a	b	A_4	
$V_{①ON}$	ON	ON	b	ON	c	c	ON	a	b	V_2	While raising V_2 from a low voltage, the value of V_2 when V_3 changes from "L" to "H"
$V_{①OFF}$	ON	ON	b	ON	c	c	ON	a	b	V_2	While lowering V_2 from a high voltage, the value of V_2 when V_3 changes from "H" to "L"
ΔV_{CH}	ON	ON	d	ON	a	a	ON	a	b	V_4	The difference between the values of V_4 under the following two conditions (i) While decreasing V_4 , the value of V_4 when V_1 changes from "H" to "L", and (ii) While increasing V_4 , the value of V_4 when V_1 changes from "L" to "H"

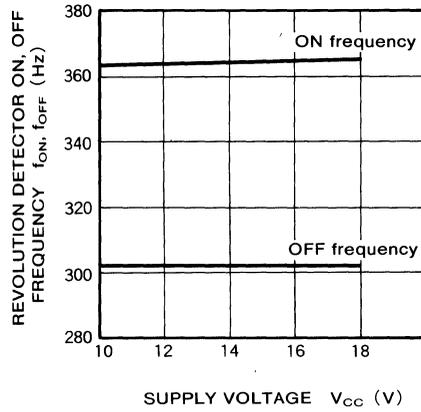
CAR AIR CONDITIONER CONTROLLER

TYPICAL CHARACTERISTICS ($V_{CC}=13.2V$, $T_a=25^{\circ}C$, unless otherwise noted)

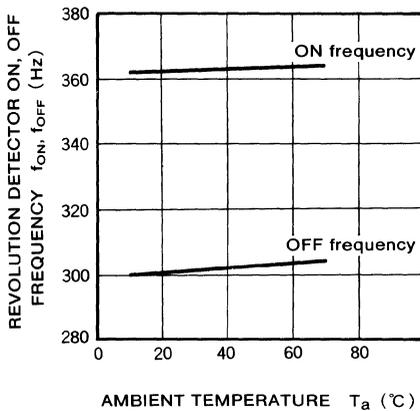
THERMAL DERATING (MAXIMUM RATING)



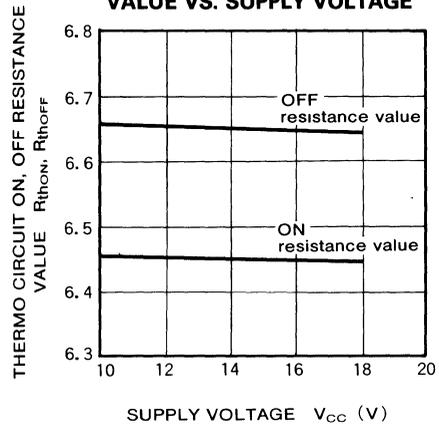
REVOLUTION DETECTOR ON, OFF FREQUENCY VS. SUPPLY VOLTAGE



REVOLUTION DETECTOR ON, OFF FREQUENCY VS. AMBIENT TEMPERATURE

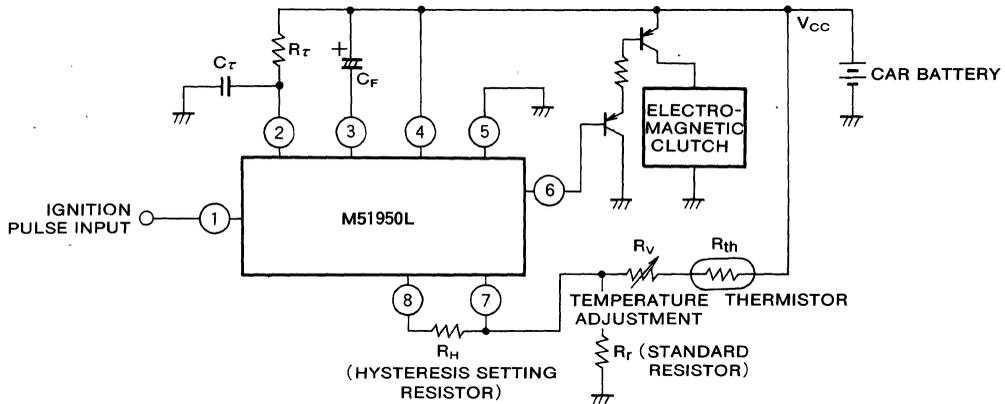


THERMO CIRCUIT ON, OFF RESISTANCE VALUE VS. SUPPLY VOLTAGE



CAR AIR CONDITIONER CONTROLLER

APPLICATION EXAMPLE



NOTES

1. The impedance of Pin ① is over 10kΩ at 5V or greater, and over 3kΩ at 6V or greater.

Refer to the Limits section of ELECTRICAL CHARACTERISTICS for the input sensitivity.

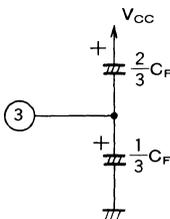
2. When changing the revolution detector ON frequency, the values of R_{τ} and C_{τ} should be changed. Note that the value of R_{τ} should be between 100kΩ and 300kΩ.

When the revolution detector ON frequency (f_{ON}) is changed, the revolution detector ON, OFF frequency difference (Δf) will change so that $\Delta f/f_{ON}$ becomes fixed.

3. Determine the value of C_F as follows:

$$C_F \approx \frac{177}{f_{ON}(\text{Hz})} (\mu\text{F})$$

4. If the circuit operates erroneously due to power supply ripple, distribute C_F between V_{CC} and GND as shown in the diagram below.



5. When changing the thermo circuit ON resistance value, change the value of the standard resistor proportionately.
6. There is an inversely proportional relationship between the hysteresis setting resistance R_H and the thermo circuit ON, OFF resistance value difference (ΔR_{th}).

M51961L

FLASHER CONTROL (WITH DISCONTINUITY DETECTION FUNCTION)

DESCRIPTION

The M51961L is a semiconductor integrated circuit designed for use in flashers, especially automobile flasher systems.

This device has been designed so that when it is set at a flashing frequency of 85cpm for turn signals or hazard indication, the flashing frequency will increase to 192cpm when the line is broken.

Discontinuity detection is accomplished by a discontinuity detection resistor which is connected between the battery and the lamp.

The threshold value for discontinuity detection is 90mV when $V_{CC}=12.8V$ and is designed to compensate for power supply voltage and temperature characteristics affecting lamp current.

FEATURES

- Short time between the turn signal switch turn-on and lamp lighting 25ms(max.)
- Discontinuity detection threshold value compensates for lamp characteristics
- Small dependency of flashing frequency on power supply voltage and temperature 1%(typ.) ($V_{CC}=7\sim 17V$, $T_a=-20\sim 60^{\circ}C$)
- Built-in stabilized power supply eliminates effect of power supply ripple on flashing frequency
- Built-in Zener diode for back absorption of relay

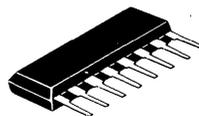
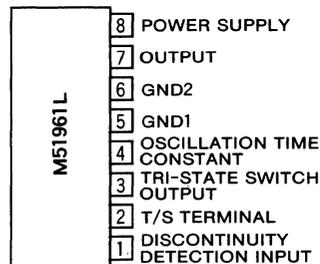
APPLICATION

Automobile flashers and lamp discontinuity detection.

RECOMMENDED OPERATING CONDITIONS

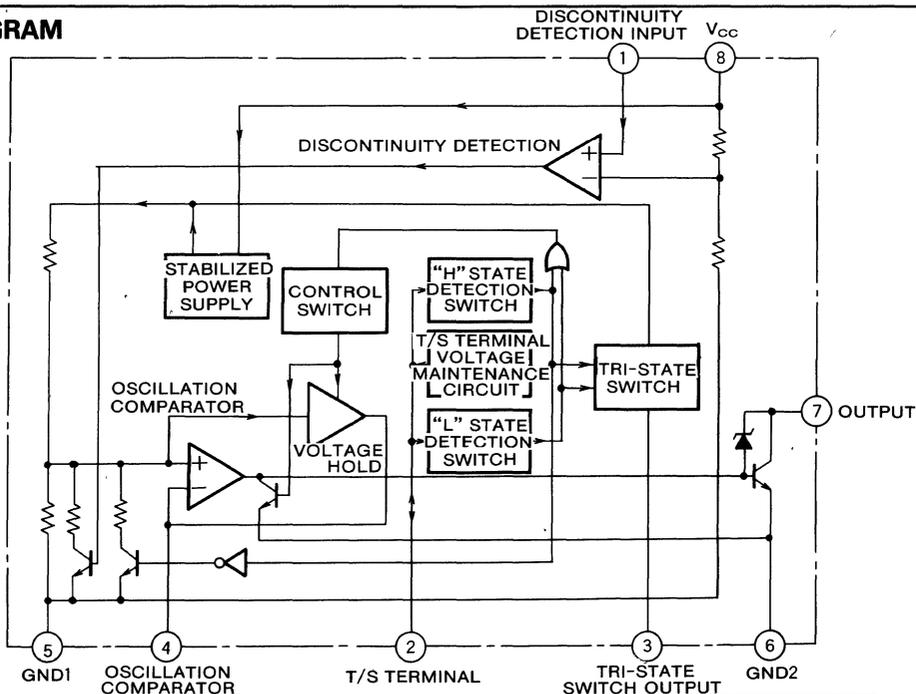
- Supply voltage range 9~16V
- Rated supply voltage 12.8V

PIN CONFIGURATION (TOP VIEW)



8-pin molded plastic SIP

BLOCK DIAGRAM



FLASHER CONTROL (WITH DISCONTINUITY DETECTION FUNCTION)

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC1}	Supply voltage 1		18	V
V _{CC2}	Supply voltage 2	Application limited to within 1 minute	24.5	V
BV _O	Output withstand voltage		24.5	V
I _O	Output current		150	mA
V _②	Applied voltage on pin ②		V _{CC} -36V~V _{CC}	V
V _①	Applied voltage on pin ①		0~V _{CC}	V
P _d	Power dissipation		900	mW
K _θ	Thermal derating	T _a ≥ 25°C	7.2	mW/°C
T _{opr}	Operating temperature range		-40~+85	°C
T _{stg}	Storage temperature range		-50~+125	°C

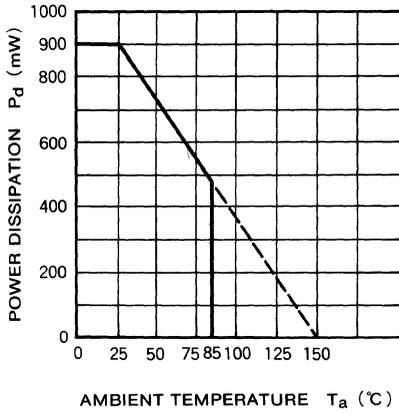
ELECTRICAL CHARACTERISTICS (R_τ = 55kΩ, C_τ = 10μF, T_a = 25°C, V_{CC} = 12.8V, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage range		7	12.8	17	V
T _S	Lamp light-up starting time	From turn signal switch ON to lamp ON		5	25	ms
N ₁	Flashing frequency 1	At turn signal or hazard	81	85	89	cpm
N ₂	Flashing frequency 2	At discontinuity	178	192	207	cpm
D ₁	Lamp ON duty 1	At turn signal or hazard	41	45	49	%
D ₂	Lamp ON duty 2	At discontinuity	33	37	41	%
V _{①TH}	Discontinuity detection threshold value		84	91.3	98.5	mV
I _{CC1}	Circuit current 1	At output OFF		3.4	5.3	mA
I _{CC2}	Circuit current 2	At output ON		8.9	15	mA
V _②	Pin ② voltage	Pin ② open	6.0	7.4	9.0	V
V _{②THH}	Pin ② "H" threshold		V _② +1	9.2	V _② +2.5	V
V _{②THL}	Pin ② "L" threshold		V _② -2.5	5.6	V _② -1	V
I _{②IN1}	Pin ② sink current	When V _② = V _{CC}	0.6	0.96	1.6	mA
I _{②IN2}	Pin ② outflow current	When V _② = GND	-2.2	-1.36	-0.9	mA
I _{①IN}	Pin ① input current	When V _① = V _{CC}		1.6	20	μA
V _{OS}	Output saturation voltage	R _L = 120Ω		180	500	mV
I _{OL}	Output leakage current				100	nA
V _{OZ}	Output Zener voltage		26	30	36	V

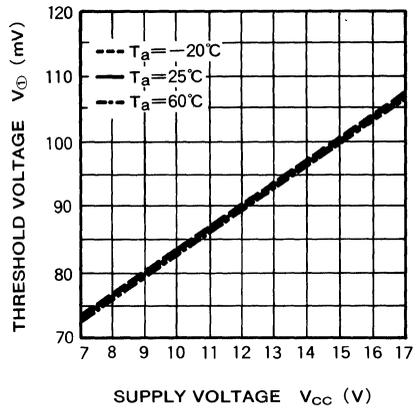
FLASHER CONTROL (WITH DISCONTINUITY DETECTION FUNCTION)

TYPICAL CHARACTERISTICS ($V_{CC}=12.8$, $T_a=25^\circ\text{C}$, unless otherwise noted)

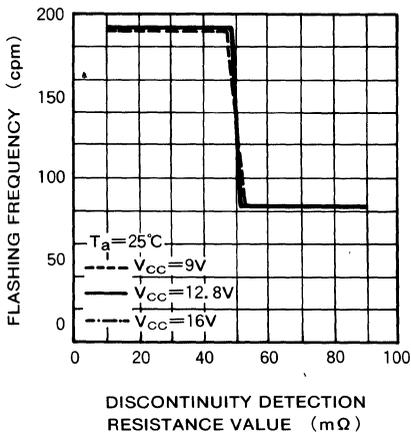
THERMAL DERATING (MAXIMUM RATING)



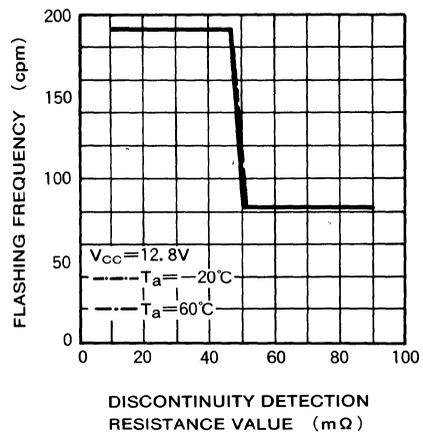
DISCONTINUITY DETECTION THRESHOLD VALUE CHARACTERISTICS



FLASHING FREQUENCY CHARACTERISTICS ①



FLASHING FREQUENCY CHARACTERISTICS ②



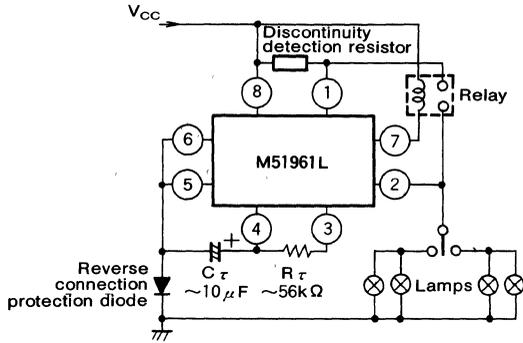
Note 1 For the flashing frequency characteristics the discontinuity detection resistance value has been measured instead of the lamp resistance value.

Note 2 In flashing frequency characteristics ②, C_τ and R_τ are measured outside of a constant-temperature tank.

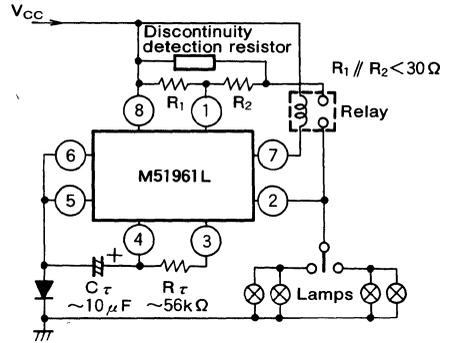
FLASHER CONTROL (WITH DISCONTINUITY DETECTION FUNCTION)

APPLICATION EXAMPLES

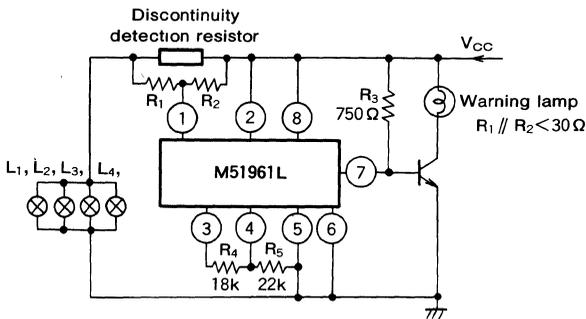
(1) Typical application circuit



(2) Adjustment of the discontinuity threshold value



(3) Discontinuity warning resistor



Unit Resistance : Ω
 Capacitance : F

The warning lamp will light when one or more of the four lamps become discontinuous.

MITSUBISHI LINEAR ICs M52670P,FP

4-BIT A-D CONVERTER

DESCRIPTION

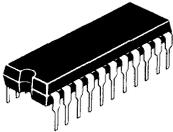
The M52670P,FP are semiconductor integrated circuits developed for use as interfaces to convert and process analog signals into digital signals. They are 4-bit converters for video band signal applications.

These devices have the following functions when used as parallel-comparison 4-bit A-D converters for video band signals:

1. 4-bit comparators (16)
2. Grey code converter
3. ECL to TTL converter
4. Latch circuit
5. Grey to binary code
6. Sampling clock restoration circuit
7. Underflow output switching circuit

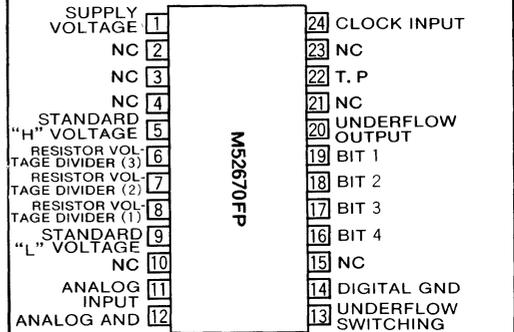
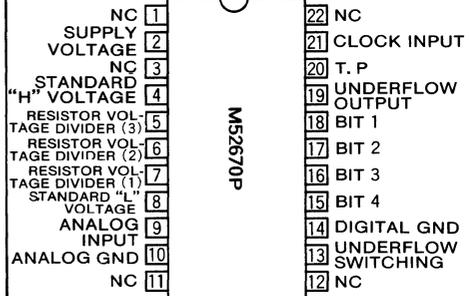
FEATURES

- Sampling rate: 11M samples/sec (max.)
- Output form: Open collector (TTL compatible)
- Expansion to a 5-bit A-D converter:
Easily accomplished by joining two units together
- Non-linear error $\pm 1/2\text{LSB}$



22-pin molded plastic DIP 24-pin molded plastic shrick
FLAT

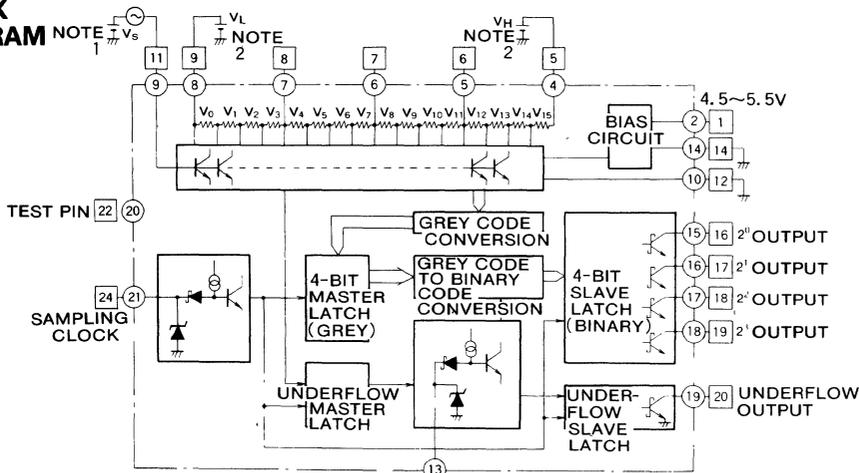
PIN CONFIGURATION (TOP VIEW)



T. P TEST PIN

NC · NO CONNECTION

BLOCK DIAGRAM



□ number designates M52670FP
○ number designates M52670P

UNDERFLOW INPUT (GHD or V_{CC})

Note 1 Set so that the impedance of the drive source is less than 100Ω
Note 2 Set so that the impedance of the drive source is less than 50Ω

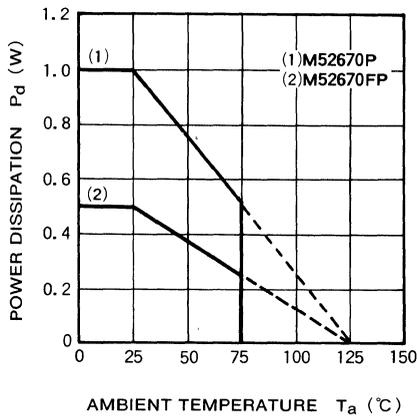
ABSOLUTE MAXIMUM RATINGS (T_a=25°C)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		7.0	V
V _{IN}	Input voltage ^{⑬, ⑰} (^{⑬, ⑰})		7.0	V
V _O	Output voltage ^{⑮~⑲} (^{⑮~⑲})		7.0	V
V _A	Operating supply voltage		4.5~5.5	V
P _d	Power dissipation	() designate FP	1000(500)	mW
T _{opr}	Operating temperature range		-10~+75	°C
T _{stg}	Storage temperature range		-40~+125	°C

Note: Numbers inside □ designate M52670FP and numbers inside ○ designate M52670P

TYPICAL CHARACTERISTICS

THERMAL DERATING (MAXIMUM RATING)



ELECTRICAL CHARACTERISTICS (T_a=25°C)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I _{CC}	Circuit current 1	V _{CC} =5.0V, Clock 11MHz		33	48	mA
f _{SC}	Maximum sampling frequency	V _{CC} =5.0V, Duty 50%			11	MHz
V _{IN⑧⑨}	⑧(⑨) Minimum input voltage	V _{CC} =5.0V	1.1			V
V _{IN④⑤}	④(⑤) Maximum input voltage				V _{CC} -0.7	V
V _{IN⑨⑩⑪}	⑨(⑩) Maximum input voltage				V _{CC} -0.7	V
ΔV _{④⑤-⑧⑨}	Minimum input voltage between ④-⑧(⑤-⑨)				1.0	V
V _{IH}	⑬⑬, ⑰⑰ Input "H" voltage		2.0			V
V _{IL}	⑬⑬, ⑰⑰ Input "L" voltage				0.8	V
I _{IN1}	⑬⑬, ⑰⑰ Input current 1	V _{CC} =5.25V, V _I =2.7V			20	μA
I _{IN2}	⑬⑬, ⑰⑰ Input current 2	V _{CC} =5.25V, V _I =0.4V			(*)-0.4	mA
V _{IK}	⑰⑰ Input clamp voltage	V _{CC} =4.75V, I _{IN} =-18mA			(*)-0.5	V
V _{OL1}	⑮⑮-⑲⑲ Output voltage 1	When V _{CC} =4.75V, and output "L", I _{OL} =6.0mA			0.5	V
V _{OL2}	⑮⑮-⑲⑲ Output voltage 2	When V _{CC} =4.75V, and output "L", I _{OL} =3.0mA			0.4	V
R _{⑦⑧, ⑧⑨}	Resistance between ⑦-⑧, ⑧-⑨		300	400	500	Ω
R _{⑥⑦, ⑦⑧}	Resistance between ⑥-⑦, ⑦-⑧		300	400	500	Ω
R _{⑤⑥, ⑥⑦}	Resistance between ⑤-⑥, ⑥-⑦		300	400	500	Ω
R _{④⑤, ⑤⑥}	Resistance between ④-⑤, ⑤-⑥		300	400	500	Ω

Note 1. Current flowing to the IC is taken as "+", and current flowing out as "-".

2. IC ground is taken as a reference for voltage, while "-" indicates voltage values that are lower than IC ground.

3. Numbers inside □ designate M52670FP and numbers inside ○ designate M52670P

4-BIT A-D CONVERTER

5-BIT EXPANSION METHOD

1. The output form (4 bit) of ⑮-⑲(⑳-㉔) with respect to V_S .

a) ⑬⑭: at GND

		V_S																
		V_0	V_1	V_2	V_3	V_4	V_5	V_6	V_7	V_8	V_9	V_{10}	V_{11}	V_{12}	V_{13}	V_{14}	V_{15}	
⑮⑯	⑰⑱	L	L	H	L	H	L	H	L	H	L	H	L	H	L	H	L	H
⑯⑰	⑲⑳	L	L	L	H	H	L	L	H	H	L	L	H	H	L	L	H	H
⑰⑲	㉑⑳	L	L	L	L	L	H	H	H	H	L	L	L	L	H	H	H	H
⑲⑳	㉓㉒	L	L	L	L	L	L	L	L	L	H	H	H	H	H	H	H	H
㉑⑳	㉓㉒	H	H	H	H	H	H	H	H	H	H	H	H	H	H	H	H	H

b) ⑬⑭: at V_{CC}

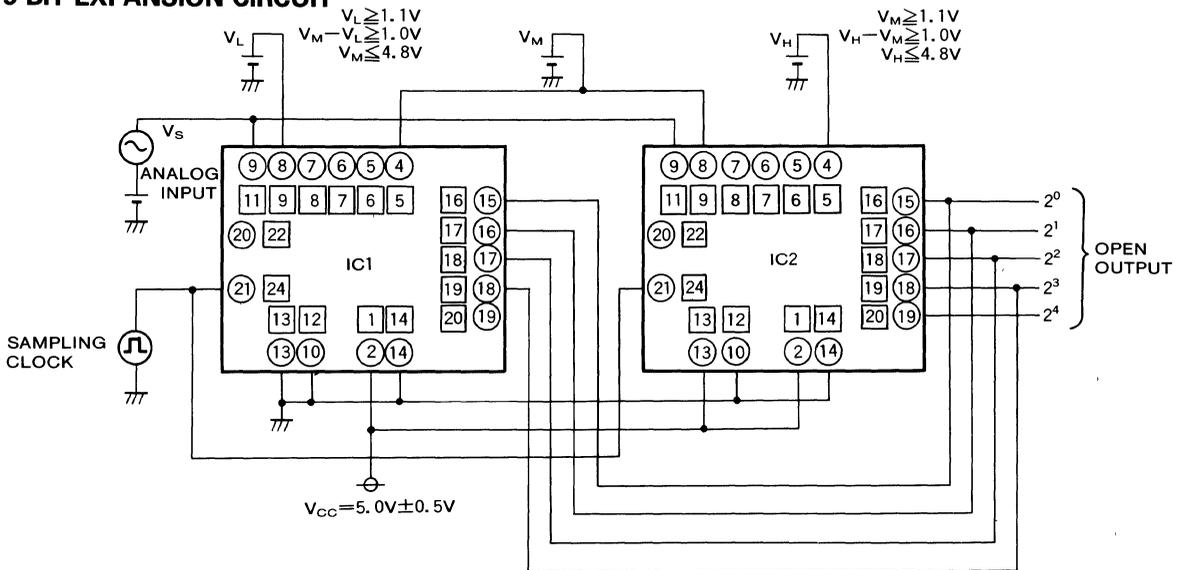
		V_S																
		V_0	V_1	V_2	V_3	V_4	V_5	V_6	V_7	V_8	V_9	V_{10}	V_{11}	V_{12}	V_{13}	V_{14}	V_{15}	
⑮⑯	⑰⑱	H	L	H	L	H	L	H	L	H	L	H	L	H	L	H	L	H
⑯⑰	⑲⑳	H	L	L	H	H	L	L	H	H	L	L	H	H	L	L	H	H
⑰⑲	㉑⑳	H	L	L	L	L	H	H	H	H	L	L	L	L	H	H	H	H
⑲⑳	㉓㉒	H	L	L	L	L	L	L	L	L	L	L	L	L	H	H	H	H
㉑⑳	㉓㉒	L	H	H	H	H	H	H	H	H	H	H	H	H	H	H	H	H

2. Output form (5 bit)

		V_S																
		V_M																
		V_L	V_1	V_2	V_3	V_4	V_5	V_6	V_7	V_8	V_9	V_{10}	V_{11}	V_{12}	V_{13}	V_{14}	V_{15}	
2^0		L	L	H	L	H	L	H	L	H	L	H	L	H	L	H	L	H
2^1		L	L	L	H	H	L	L	H	H	L	L	H	H	L	L	H	H
2^2		L	L	L	L	L	H	H	H	H	L	L	L	L	H	H	H	H
2^3		L	L	L	L	L	L	L	L	L	H	H	H	H	H	H	H	H
2^4		L	L	L	L	L	L	L	L	L	L	L	L	L	L	L	L	L

Note. Numbers inside □ designate M52670FP and numbers inside ○ designate M52670P

5-BIT EXPANSION CIRCUIT

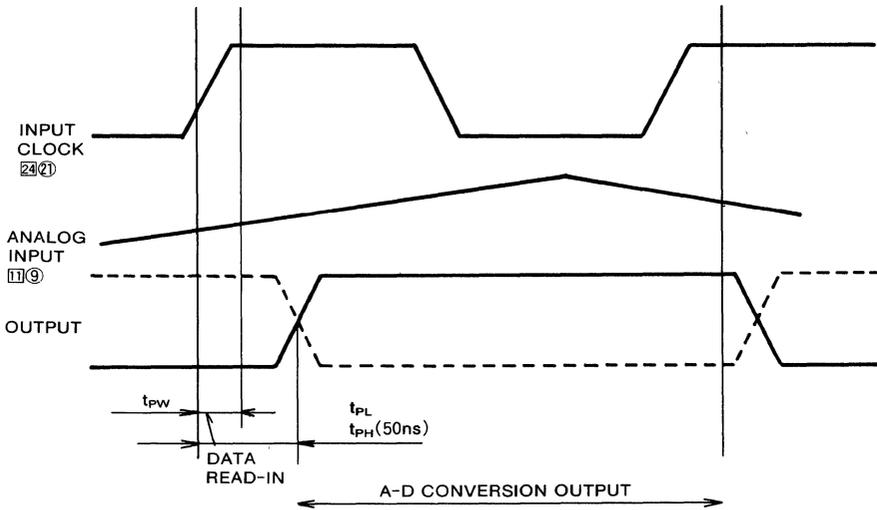


□ Numbers inside designate M52670FP
 ○ Numbers inside designate M52670P

4-BIT A-D CONVERTER

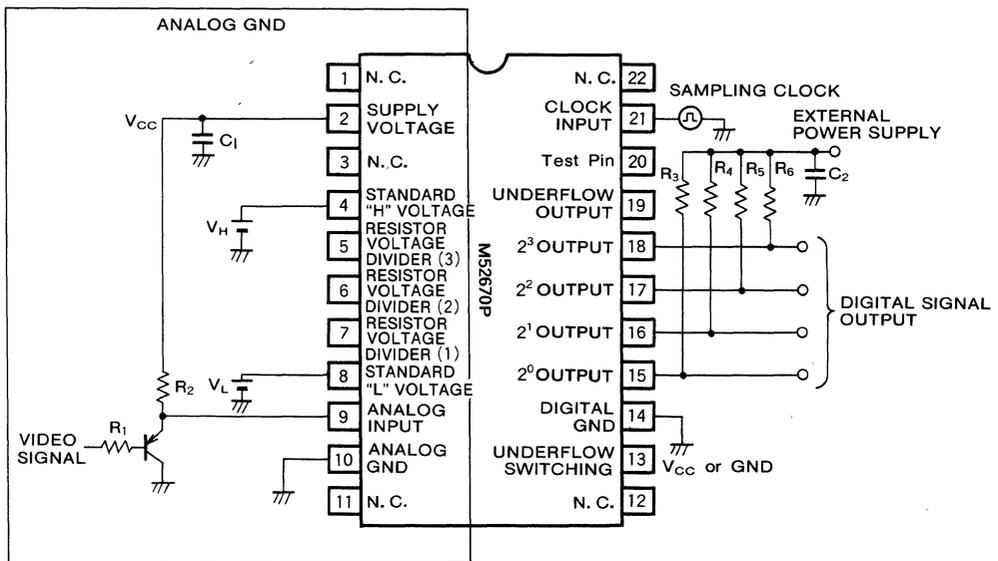
TIMING CHART

TIMING OF SAMPLING CLOCK AND OUTPUT



□ Numbers inside designate M52670FP
 ○ Numbers inside designate M52670P

APPLICATION EXAMPLE



M5109P

DUAL DIFFERENTIAL AMPLIFIER

DESCRIPTION

The M5109P is a semiconductor integrated circuit consisting of two differential amplifiers, fabricated by making use of complementary symmetry.

Since the two differential amplifiers are part of the same structure and have closely matched characteristics, this device is suitable for use in applications requiring such matched characteristics. A bias diode is built into the device for the convenience of applications.

The high reliability of this device makes it useful in applications such as audio, communications, and control equipment.

FEATURES

- Two differential amplifiers with closely matched characteristics.
- Small input offset voltage5mV(max)
- Small input offset current2 μ A(max)
- Built-in bias diode

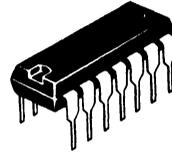
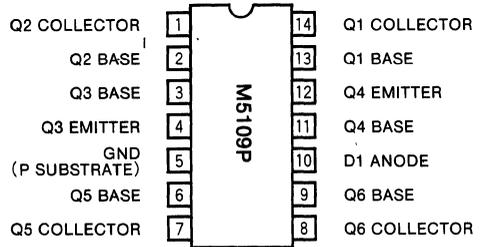
APPLICATION

RF/IF amplifiers, double balanced mixer voltage comparators, balanced dual differential amplifiers, and detectors.

RECOMMENDED OPERATING CONDITIONS

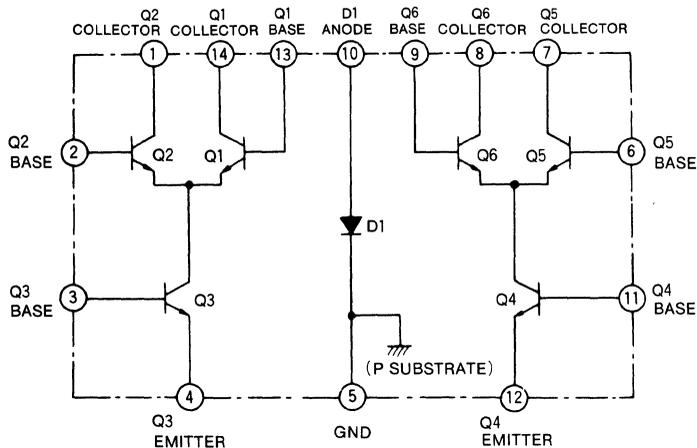
- Supply voltage range2~15V
- Rated supply voltage 12V

PIN CONFIGURATION (TOP VIEW)



14-pin molded plastic DIP

EQUIVALENT CIRCUIT



DUAL DIFFERENTIAL AMPLIFIER

ABSOLUTE MAXIMUM RATINGS ($T_a=25^\circ\text{C}$, unless otherwise noted)

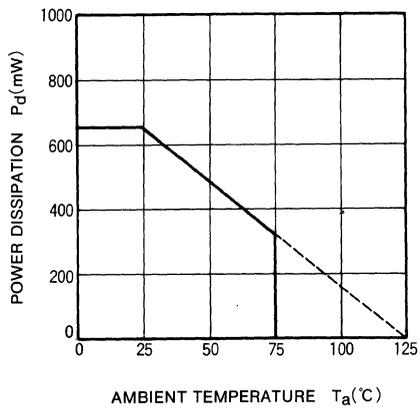
Symbol	Parameter	Conditions	Ratings	Unit
V_{CE0}	Collector-emitter voltage		15	V
V_{CBO}	Collector-base voltage		20	V
V_{EBO}	Emitter-base voltage		5	V
I_C	Collector current		50	mA
P_d	Power dissipation		650	mW
K_θ	Thermal derating	$T_a \geq 25^\circ\text{C}$	6.5	mW/ $^\circ\text{C}$
Topg	Operating temperature range		-20~+75	$^\circ\text{C}$
Tstg	Storage temperature range		-40~+125	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V_{IO}	Input offset voltage	$V_{CE}=3\text{V}, I_E=2\text{mA}$			5	mV
I_{IO}	Input offset current				2	μA
I_{IB}	Input bias current				24	μA
$\frac{I_{C(O1)}}{I_{C(O2)}} \text{ or } \frac{I_{C(O5)}}{I_{C(O6)}}$	Differential stage current ratio			1.0	—	
I_{CBO}	Collector cutoff current	$V_{CB}=18\text{V}, I_E=0$			1	μA

TYPICAL CHARACTERISTICS ($T_a=25^\circ\text{C}$, unless otherwise noted)

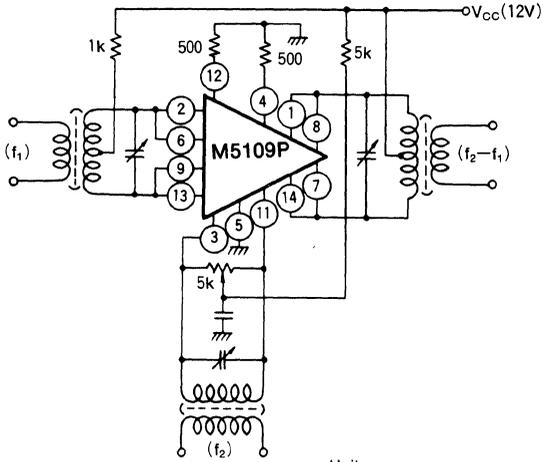
**THERMAL DERATING
(MAXIMUM RATING)**



DUAL DIFFERENTIAL AMPLIFIER

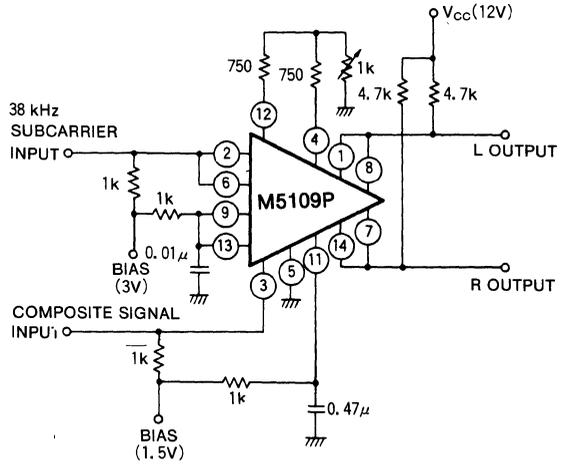
APPLICATION EXAMPLES

(1) Frequency mixer



Unit
 Resistance : Ω
 Capacitance : F

(2) FM Stereo demodulator



M54940P

8-DIGIT FLUORESCENT DISPLAY DRIVER FOR MICROCOMPUTER

DESCRIPTION

The M54940P, a monolithic integrated circuit fabricated with using an IIL technology, is designed for driving an 8-digit, 7-segment fluorescent display.

FEATURES

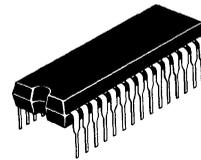
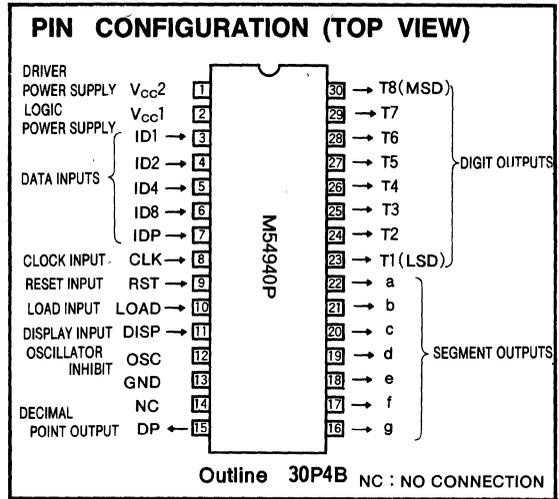
- Separated power supplies; 5V (Logic circuit), and 35V (Output circuit)
- Integral scanning oscillator circuit for display

APPLICATIONS

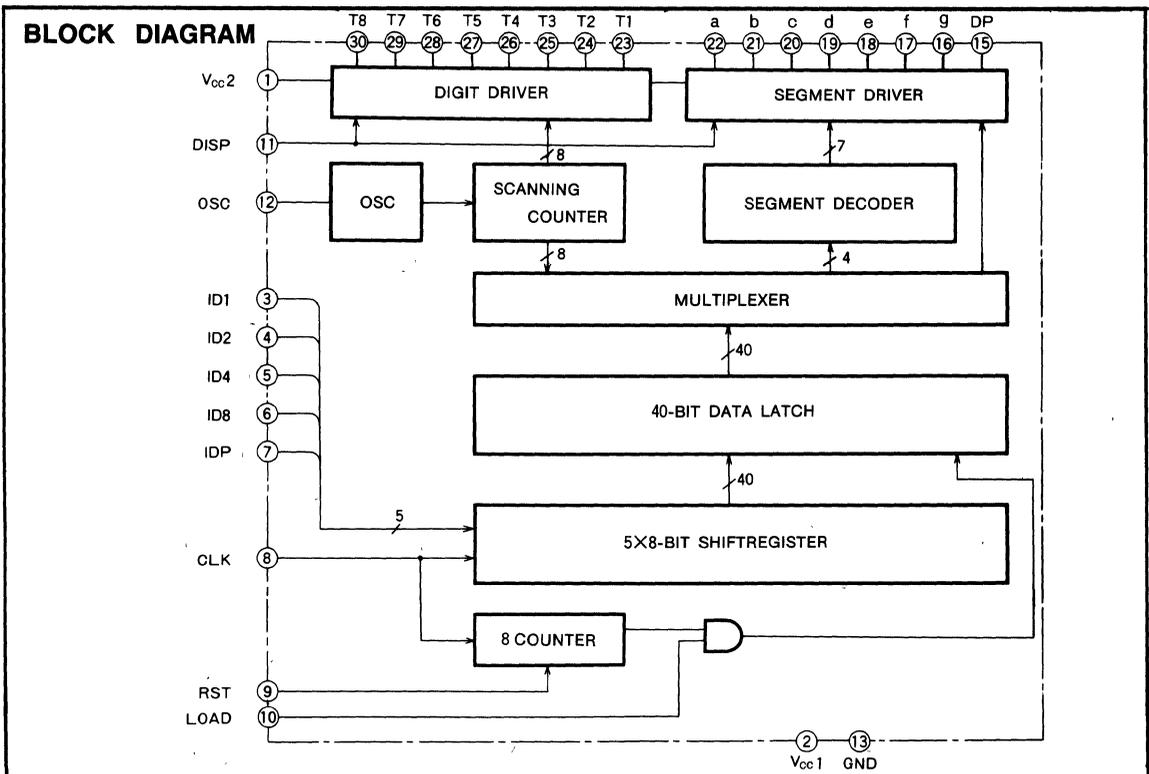
- Micro computer display
- Digital equipment for consumer and Industrial use.

FUNCTION

The M54940P, having a 5-bit X 8-digit memory, is a decoder driver for dynamic displaying of a vacuum fluorescent tubes. The data for one digit section is organized into a 4-bit BCD and an 1-bit decimal point. The data memory consists of a shift register and a latch, and is capable of displaying the previous data while the data is being transported.



30-pin molded plastic DIP (SHRINK)



8-DIGIT FLUORESCENT DISPLAY DRIVER FOR MICROCOMPUTER

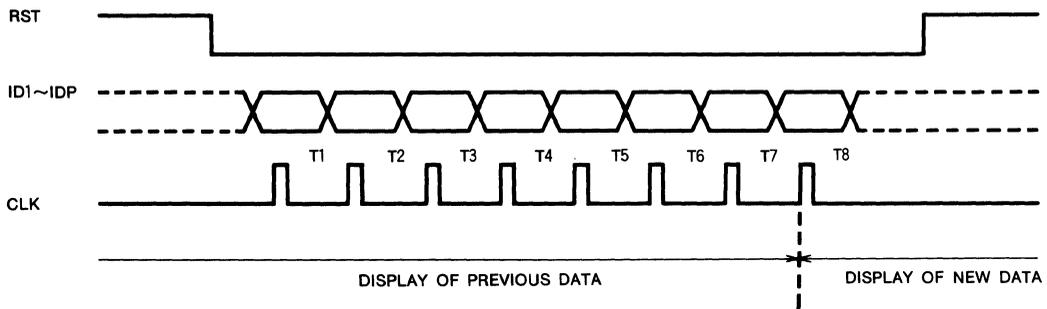
INPUT PIN FUNCTION

- 1) OSC : External capacitor connecting terminal for the oscillator circuit.
- 2) ID 1 } : BCD Data Input; refer to the numerical
ID 2 }
ID 4 } Designations-resultant displays for the relation
ID 8 } of the input data to the display.
- 3) IDP : decimal point data Input
- 4) CLK : Data transport clock Input: the data can be input at a positive-going edge of the CLK
- 5) RST : Reset Input : the CLK input counter is reset at "H".
- 6) LOAD : Signal Input to load the data latch with the data of the shift register. The input LOAD will not be accepted until the 8th CLK Input has been received.
- 7) DISP : When it is set to "H" it displays. When it is set to "L", the display is inhibited. During the display inhibition period, both the segment and digit outputs will be at "L".

TIMING CHART

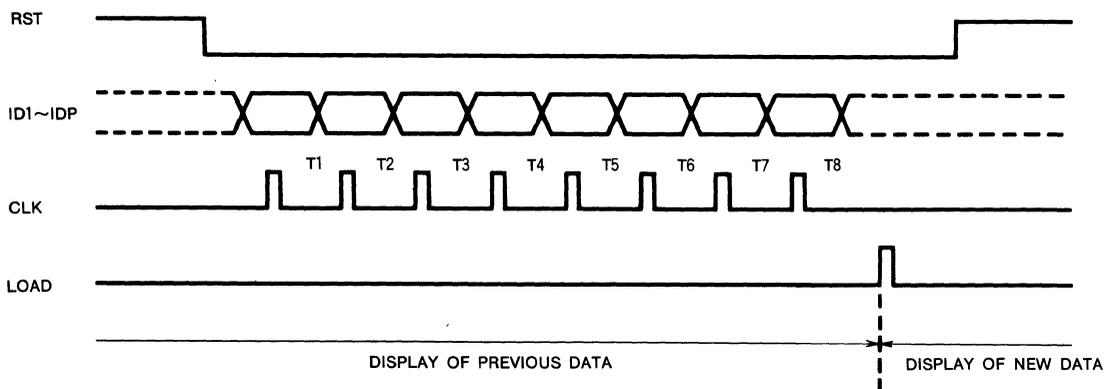
DATA PROGRAMMING

(1) USING CLK AND RST inputs with LOAD="H".



When LOAD is kept at "H", LOAD is automatically done at the 8th CLK input when RST="L". However, while RST="L", if there is a 9th CLK input, the 9th data will be loaded and displayed

(2) Using CLK, RST and LOAD inputs.

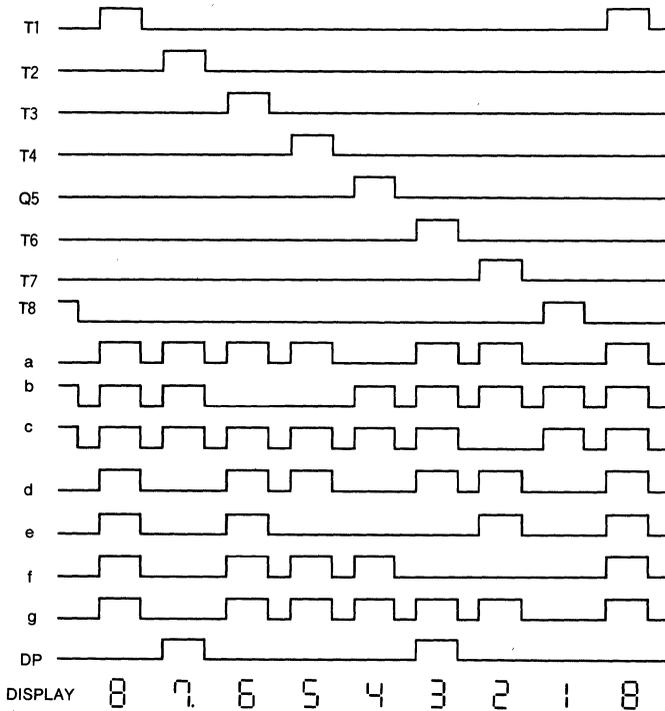


After the 8th clock input, the LOAD is valid only in the period while RST="L". Furthermore, if there is 9th CLK input before the LOAD input, the LOAD input is ignored.

M54940P

8-DIGIT FLUORESCENT DISPLAY DRIVER FOR MICROCOMPUTER

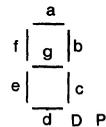
OUTPUT TIMING CHART



NUMERICAL DESIGNATIONS-RESULTANT DISPLAYS

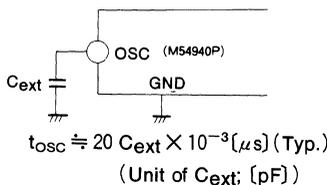
		0	1	2	3	4	5	6	7	8	9	A	B	C	D	E	F
BCD	ID1	L	H	L	H	L	H	L	H	L	H	L	H	L	H	L	H
	ID2	L	L	H	H	L	L	H	H	L	L	H	H	L	L	H	H
DATA	ID4	L	L	L	L	H	H	H	H	L	L	L	L	L	L	H	H
	ID8	L	L	L	L	L	L	L	L	L	H	H	H	H	H	H	H
Display		0	1	2	3	4	5	6	7	8	9	-	E	P	L	7	Blank

* The decimal point, independent of BCD data, is output when the decimal bit of the corresponding digit is at "H". Furthermore, when the decimal point bit is set at "H" at plural digits, plural decimal points are displayed.



OSCILLATOR CIRCUIT

1) External connection



2) Oscillation period

DISPLAYS IMMEDIATELY "AFTER POWER ON."

The display which appears immediately after "power-on" is indefinable. During the period before the regular data is transported the display can be erased if DISP input is set at "L".

8-DIGIT FLUORESCENT DISPLAY DRIVER FOR MICROCOMPUTER

ABSOLUTE MAXIMUM RATINGS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC1}	Logic supply voltage		-0.3~+9	V
V_{CC2}	Driver supply voltage		-0.3~+38	V
V_I	Input voltage		-0.3~ V_{CC1}	V
V_O	Output voltage		0~ V_{CC2}	V
T_{stg}	Storage temperature range		-55~+150	$^\circ\text{C}$
T_{opr}	Operating temperature range		-20~+75	$^\circ\text{C}$
P_d	Power dissipation		600	mW

RECOMMENDED OPERATING CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{CC1}	Logic supply voltage	4	5	7	V
V_{CC2}	Driver supply voltage	10	30	35	V

ELECTRICAL CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, $V_{CC1} = 5\text{V}$, $V_{CC2} = 35\text{V}$, unless otherwise noted)

Symbol	Parameter	Test Conditions	Limits			Unit
			Min	Typ	Max	
V_{IH}	"H" Input voltage	$V_{CC1} = 4 \sim 7\text{V}$	2.7		V_{CC1}	V
V_{IL}	"L" Input voltage	$V_{CC1} = 4 \sim 7\text{V}$	0		0.7	V
I_{IH}	"H" Input current	$V_{IH} = 5\text{V}$		0	20	μA
I_{IL}	"L" Input current	$V_{IL} = 0\text{V}$		-0.25	-0.4	mA
V_{OH}	"H" Output voltage	Digit output	$I_{OH} = -10\text{mA}$	33	33.8	V
		Segment output	$I_{OH} = -2\text{mA}$	33	34	
V_{OL}	"L" Output voltage	$I_{OL} = 0\text{mA}$		0	2	V
I_{CC1}	Logic circuit current	Input : open All segment outputs, ON		12	22	mA
I_{CC2}	Driver circuit current	Output : Open All segment outputs : ON		8	14	mA
t_{OSC}	Oscillation Period	$C_{ext} = 1000\text{pF}$	10	20	40	μs

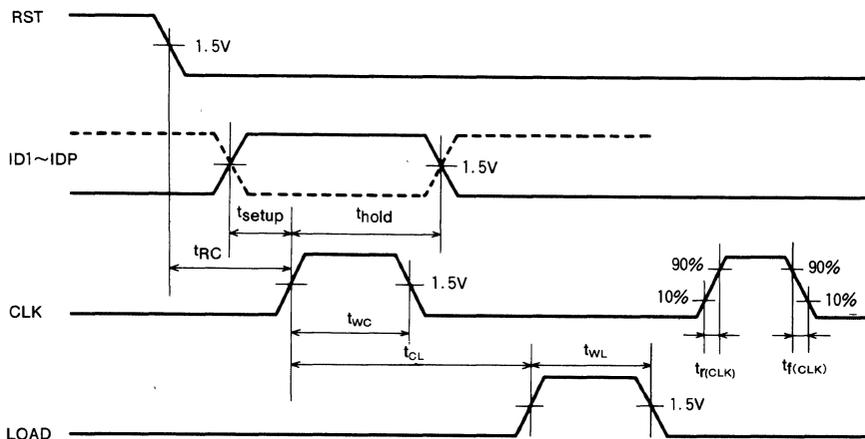
*Typical values are measured at 25 $^\circ\text{C}$

TIMING CHARACTERISTICS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

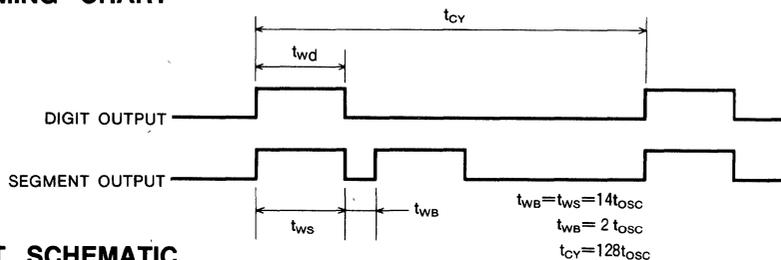
Symbol	Parameter	Test Conditions	Limits			Unit
			Min	Typ	Max	
f_{CLK}	Clock Frequency				100	kHz
f_{OSC}	Oscillation frequency		10		100	kHz
t_{WC}	Clock pulse width		2			μs
t_{WL}	Load Pulse width		2			μs
t_{setup}	Data setup time (DATA \rightarrow CLK)		4			μs
t_{hold}	Data Hold time (CLK \rightarrow DATA)		2			μs
t_{RC}	Reset-clock time (RST \rightarrow CLK)		4			μs
t_{CL}	Clock-load, time (CLK \rightarrow LOAD)		4			μs
$t_{r(CLK)}$	Clock pulse rise time				10	μs
$t_{f(CLK)}$	Clock pulse fall time				10	μs

8-DIGIT FLUORESCENT DISPLAY DRIVER FOR MICROCOMPUTER

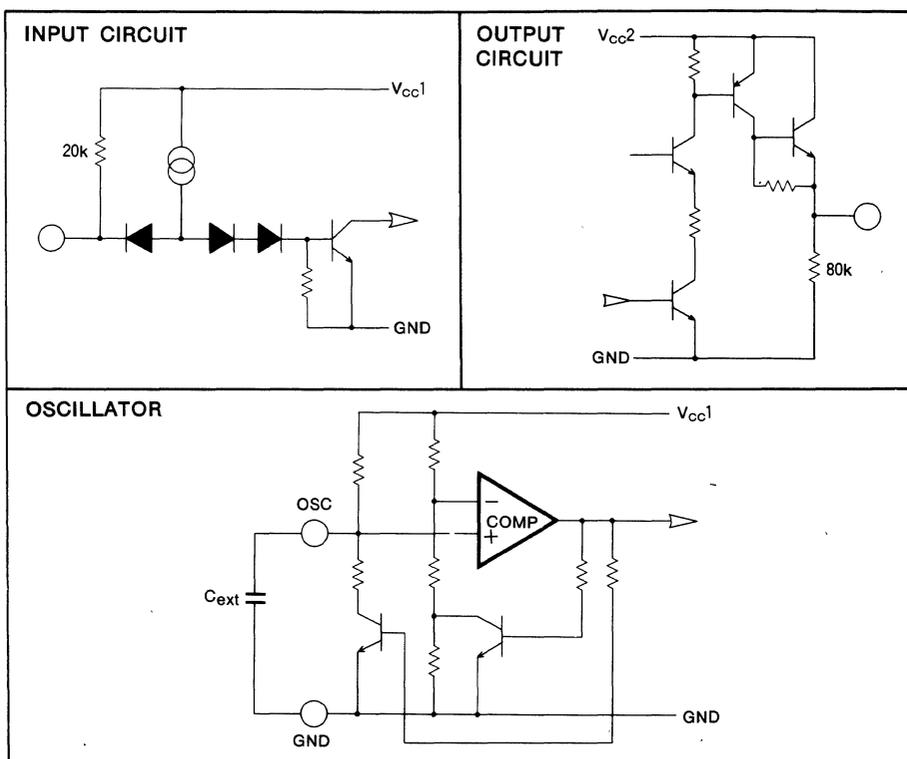
INPUT TIMING CHART



OUTPUT TIMING CHART



I/O CIRCUIT SCHEMATIC



M54570L

TUNER BAND DECODER/DRIVER

DESCRIPTION

The M54570L is a semiconductor integrated circuit capable of switching four bands in TV and VTR tuners.

FEATURES

- Low output saturation voltage ($V_{CE(sat)} \leq 0.5V$ at $I_o = -35mA$).
- High output sustaining voltage ($BV_{CEO} \geq 26V$)
- Four-bands switching

APPLICATION

Switching bands in TV and VTR tuners

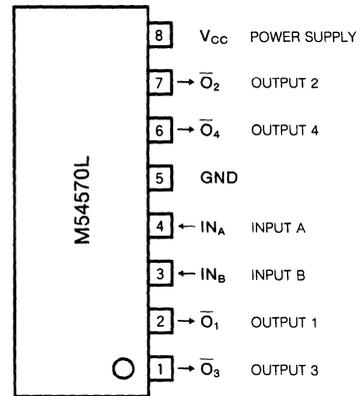
FUNCTIONAL DESCRIPTION

The M54570L is an IC suitable for four-band switching in TV and VTR tuners. Since the output drives the power supply of each tuner band, a low saturation voltage ($V_{CC} - V_o$) becomes necessary. This need is satisfied through a first stage configured of PNP transistors.

The input, being three-valued logic input, can be switched into 6 output modes as shown in the truth table.

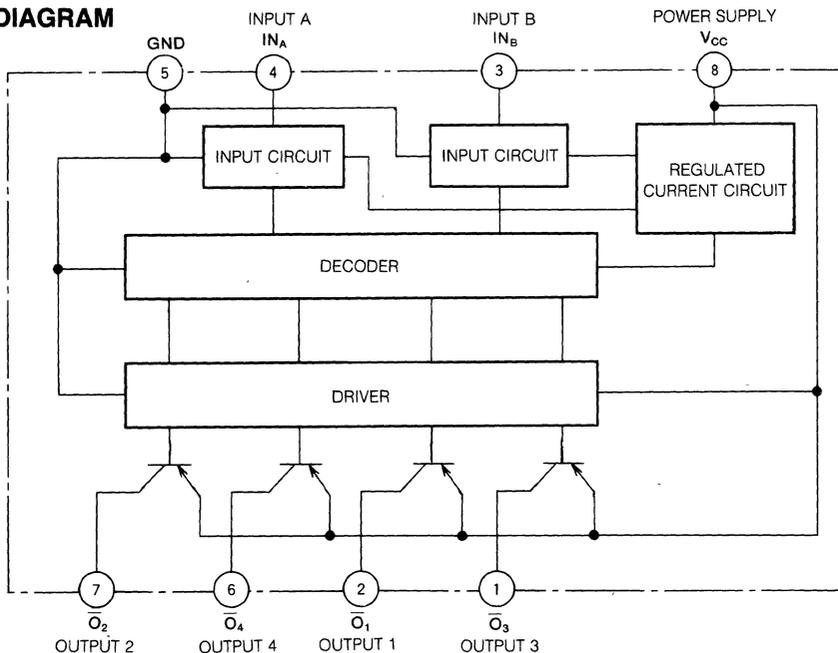
The selection mode can be altered by making a wired OR connection on the outputs when used as a three-band device.

PIN CONFIGURATIONS (TOP VIEW)



Outline 8P5

BLOCK DIAGRAM



TUNER BAND DECODER/DRIVER

TRUTH TABLE

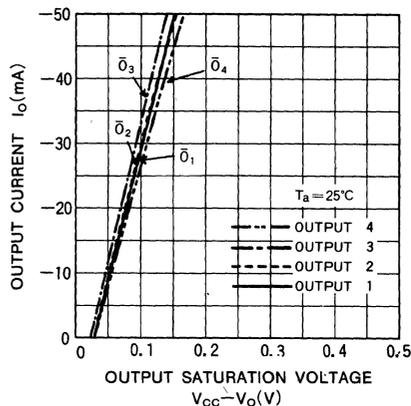
Input		Output			
IN _A	IN _B	\bar{O}_1	\bar{O}_2	\bar{O}_3	\bar{O}_4
0	0	1	1	0	1
0	1	0	0	1	0
1	0	1	1	0	0
1	1	1	0	0	0
1*	1*	1	0	0	1
1*	0	1	1	0	0

Input "0" = 0.4V (max.)
 "1" = 4V (min.), 6V (max.)
 "1*" = 10V (min.), V_{CC} (max.)

Output "0" = output transistor off-state
 "1" = output transistor on-state

TYPICAL CHARACTERISTICS

SOURCE OUTPUT SATURATION CHARACTERISTICS



ABSOLUTE MAXIMUM RATINGS (T_a = 25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		15	V
V _{CEO}	Output sustaining voltage		-0.5 ~ +26	V
V _I	Input voltage		15	V
I _O	Output current		-40	mA
T _{opr}	Operating ambient temperature range		-10 ~ +60	°C
T _{stg}	Storage temperature range		-55 ~ +125	°C

RECOMMENDED OPERATING CONDITIONS (T_a = 25°C, unless otherwise noted)

Symbol	Parameter	Limits			Unit	
		Min	Typ	Max		
V _{CEO}	Output sustaining voltage	0		24	V	
I _O	Output current	Outputs 1 and 3	0	-35	-40	mA
		Outputs 2 and 4	0	-20	-25	
V _{IH}	"H" input voltage	4		6	V	
V _{IL}	"L" input voltage	0		0.4	V	
V _{IH*}	"H*" input voltage	10		V _{CC}	V	

ELECTRICAL CHARACTERISTICS (T_a = 25°C, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
I _{O(leak)}	Output leakage current	V _{CC} =12V, V _O =-12V, output opened			-100	μA
V _{OH}	"H" output voltage	V _{CC} =12V	I _O =-20mA	11.7	11.9	V
			I _O =-35mA (output 1, 3)	11.5	11.9	
I _{IH}	"H" input current	V _{CC} =12V, V _I =4V			10	μA
I _{IH*A}	"H*" input current (input A)	V _{CC} =12V, V _I *=10V		0.63	1.3	mA
I _{IH*B}	"H*" input current (input B)	V _{CC} =12V, V _I *=10V			20	μA
I _{IL}	"L" input current	V _{CC} =12V, V _I =0.4V			-100	μA
I _{CC}	Supply current	V _{CC} =13V, V _{IA} =0V, V _{IB} =4V, output opened		17	28	mA

M54572L

TUNER BAND DECODER/DRIVER

DESCRIPTION

The M54572L is a semiconductor integrated circuit capable of switching four bands in TV and VTR tuners.

FEATURES

- Low output saturation voltage ($V_{CE(sat)} \leq 0.5V$ at $I_O = -30mA$).
- High output sustaining voltage ($BV_{CEO} \geq 28V$)
- Four-bands switching

APPLICATION

Switching bands in TV and VTR tuners

FUNCTIONAL DESCRIPTION

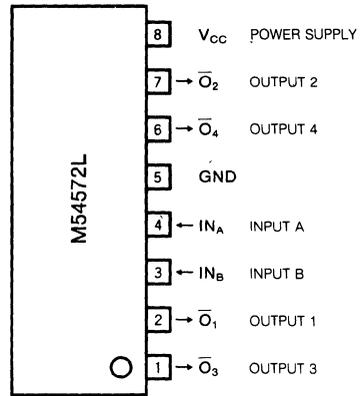
The M54572L is an IC suitable for four-band switching in TV and VTR tuners. Since the output (outputs 1~3) drives the power supply of each tuner band, a low saturation voltage ($V_{CC} - V_O$) becomes necessary. This need is satisfied through a first stage configured of PNP transistors.

Output 4 can be used for changing modes with the same power supply as the NPN transistor has an open collector output.

The input mode can be switched between four modes as shown in the truth table.

The selection mode can be altered by making a wired OR connection to outputs 1 ~ 3 when used as a three-band device.

PIN CONFIGURATIONS (TOP VIEW)



Outline 8P5

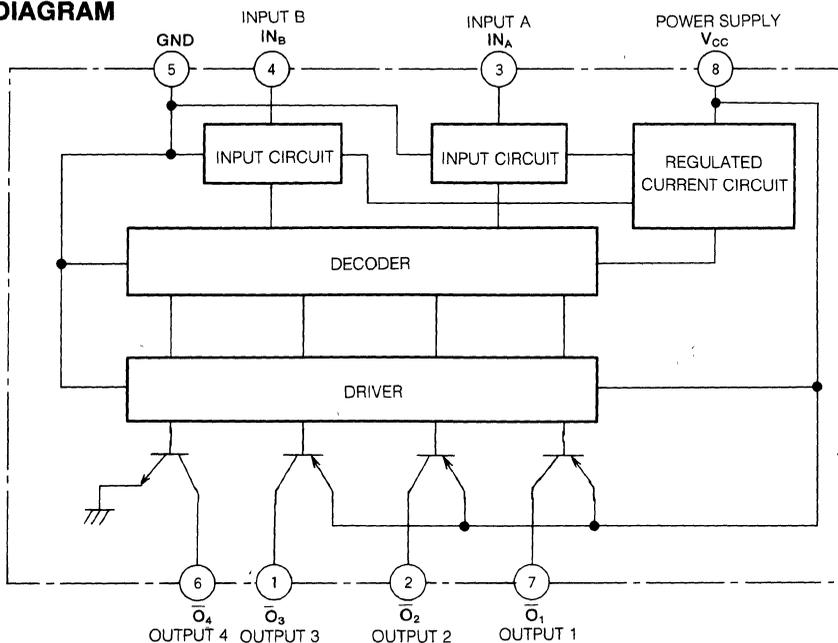
TRUTH TABLE

Input		Output			
IN _A	IN _B	O ₁	O ₂	O ₃	O ₄
0	0	1	Z	Z	Z
0	1	Z	Z	1	Z
1	0	Z	1	Z	Z
1	1	Z	Z	1	0

Input "0" = 1V (max.)
"1" = 3V (min)

Output "0" = current sink
"1" = current source
"Z" = high impedance

BLOCK DIAGRAM



TUNER BAND DECODER/DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a = 25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
V_{CEO}	Output sustaining voltage		-0.5~+28	V
V_I	Input voltage		18	V
I_{SO}	Output source current	$\bar{O}1 \sim \bar{O}3$	-40	mA
I_{SI}	Output sink current	$\bar{O}4$	40	mA
T_{opr}	Operating ambient temperature range		-10~+60	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a = 25^\circ\text{C}$, unless otherwise noted)

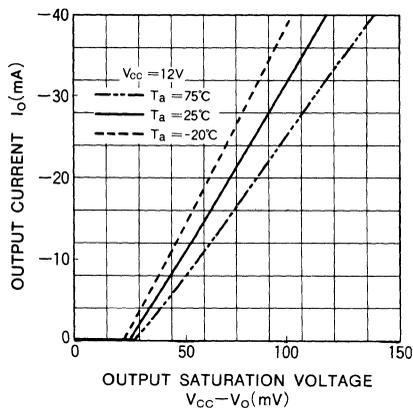
Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{CC}	Supply voltage	12		15	V
V_{CEO}	Output sustaining voltage	0		25	V
I_{SO}	Output source current	$\bar{O}1 \sim \bar{O}3$	0	-30	mA
I_{SI}	Output sink current	$\bar{O}4$	0	30	mA
V_{IH}	"H" input voltage	3		V_{CC}	V
V_{IL}	"L" input voltage	0		1	V

ELECTRICAL CHARACTERISTICS ($T_a = 25^\circ\text{C}$, $V_{CC} = 12\text{V}$, unless otherwise noted)

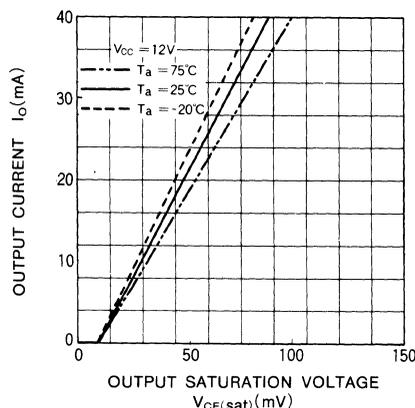
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{SO}(\text{leak})$	Source output leakage current	$V_{SO} = -12\text{V}$			-100	μA
$I_{SI}(\text{leak})$	Sink output leakage current	$V_{SI} = 25\text{V}$			100	μA
V_{SOH}	Source output "H" voltage	$I_{SO} = -30\text{mA}$	11.5	11.8		V
V_{SIL}	Sink output "L" voltage	$I_{SI} = 30\text{mA}$		0.2	0.5	V
I_{IH}	"H" input current	$V_I = 3\text{V}$			10	μA
I_{IL}	"L" input current	$V_I = 1\text{V}$			-100	μA
I_{CC}	Supply current	$V_{CC} = 13\text{V}$, $V_{IA} = 3\text{V}$, $V_{IB} = 0\text{V}$, output opened			28	mA

TYPICAL CHARACTERISTICS

SOURCE OUTPUT SATURATION CHARACTERISTICS



SINK OUTPUT SATURATION CHARACTERISTICS



DESCRIPTION

The M54573L is a semiconductor integrated circuit capable of switching three bands in TV and VTR tuners.

FEATURES

- Low output saturation voltage ($V_{CE(sat)} \leq 0.5V$ at $I_O = -30mA$).
- High output sustaining voltage ($BV_{CEO} \geq 28V$)

APPLICATION

Switching bands in TV and VTR tuners

FUNCTIONAL DESCRIPTION

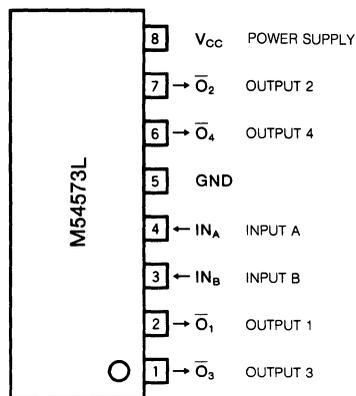
The M54573L is an IC suitable for three-band switching in TV and VTR tuners. Since the output (outputs 1~3) drives the power supply of each tuner band, a low saturation voltage ($V_{CC} - V_O$) becomes necessary. This need is satisfied through a first stage configured of PNP transistors.

Output 4 can be used for changing modes with the same power supply as the NPN transistor has an open collector output.

The input mode can be switched between three modes as shown in the truth table. The "0", "1" mode and the "1", "1" mode are the same modes.

The selection mode can be altered by making an OR connection on outputs 1~3.

PIN CONFIGURATIONS (TOP VIEW)



Outline 8P5

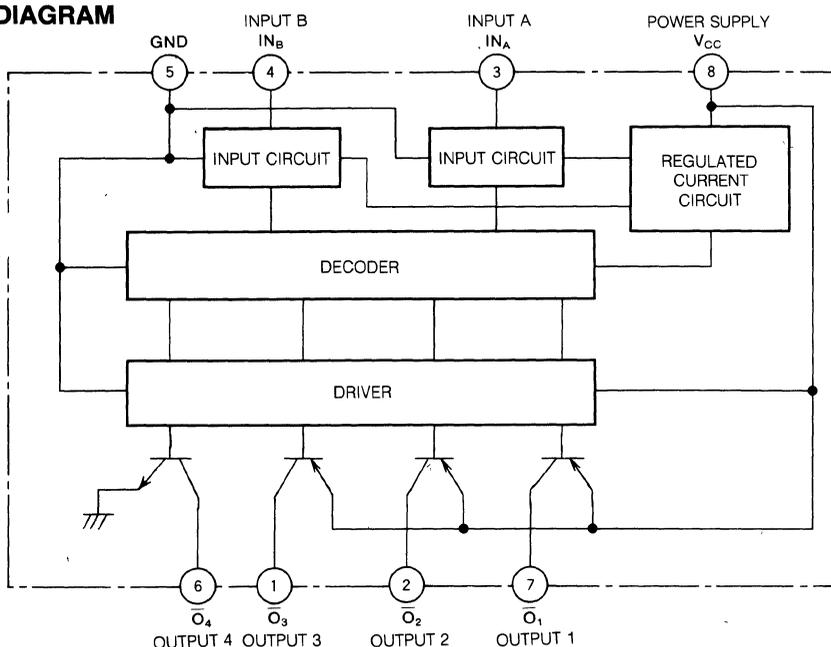
TRUTH TABLE

Input		Output			
IN _A	IN _B	\bar{O}_1	\bar{O}_2	\bar{O}_3	\bar{O}_4
0	0	1	Z	Z	0
0	1	Z	Z	1	0
1	0	Z	1	Z	Z
1	1	Z	Z	1	0

Input "0" = 1V (max)
"1" = 3V (min.)

Output "0" = current sink
"1" = current source
"Z" = high impedance

BLOCK DIAGRAM



TUNER BAND DECODER/DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a = 25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		18	V
V_{CEO}	Output sustaining voltage		-0.5~+28	V
V_i	Input voltage		18	V
I_{SO}	Output source current	$\bar{O}1\sim\bar{O}3$	-40	mA
I_{SI}	Output sink current	$\bar{O}4$	40	mA
T_{opr}	Operating ambient temperature range		-10~+60	$^\circ\text{C}$
T_{stg}	Storage temperature range		-55~+125	$^\circ\text{C}$

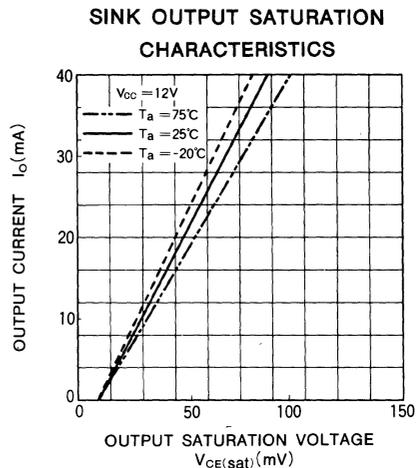
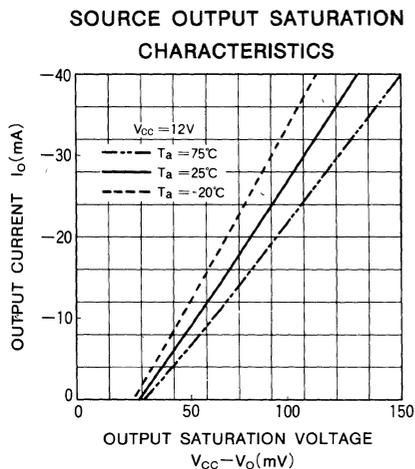
RECOMMENDED OPERATING CONDITIONS ($T_a = 25^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{CC}	Supply voltage	12		15	V
V_{CEO}	Output sustaining voltage	0		25	V
I_{SO}	Output source current	$\bar{O}1\sim\bar{O}3$	0	-30	mA
I_{SI}	Output sink current	$\bar{O}4$	0	30	mA
V_{IH}	"H" input voltage		3	V_{CC}	V
V_{IL}	"L" input voltage		0	1	V

ELECTRICAL CHARACTERISTICS ($T_a = 25^\circ\text{C}$, $V_{CC} = 12\text{V}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{SO}(\text{leak})$	Source output leakage current	$V_{SO} = -12\text{V}$			-100	μA
$I_{SI}(\text{leak})$	Sink output leakage current	$V_{SI} = 25\text{V}$			100	μA
V_{SOH}	Source output "H" voltage	$I_{SO} = -30\text{mA}$	11.5	11.8		V
V_{SIL}	Sink output "L" voltage	$I_{SI} = 30\text{mA}$		0.2	0.5	V
I_{IH}	"H" input current	$V_i = 3\text{V}$			10	μA
I_{IL}	"L" input current	$V_i = 1\text{V}$			-100	μA
I_{CC}	Supply current	$V_{CC} = 13\text{V}$, $V_{IA} = 3\text{V}$, $V_B = 0\text{V}$, output opened			28	mA

TYPICAL CHARACTERISTICS



M54847AP

2-DIGIT BCD-7SEGMENT DECODER/DRIVER

DESCRIPTION

The M54847AP is a semiconductor integrated circuit consisting of an IIL 2 digit BCD-7 segment decoder/driver.

FEATURES

- Direct drive of LEDs (common cathode type. No need for current limiting resistors, segment current: 10mA max.)
- Direct drive of fluorescent character displays (Segment withstand output is -25V max at $V_{CC}=5V$.)
- Data input in both serial and parallel formats
- Brightness control input enables continuous LED brightness adjustment.

APPLICATION

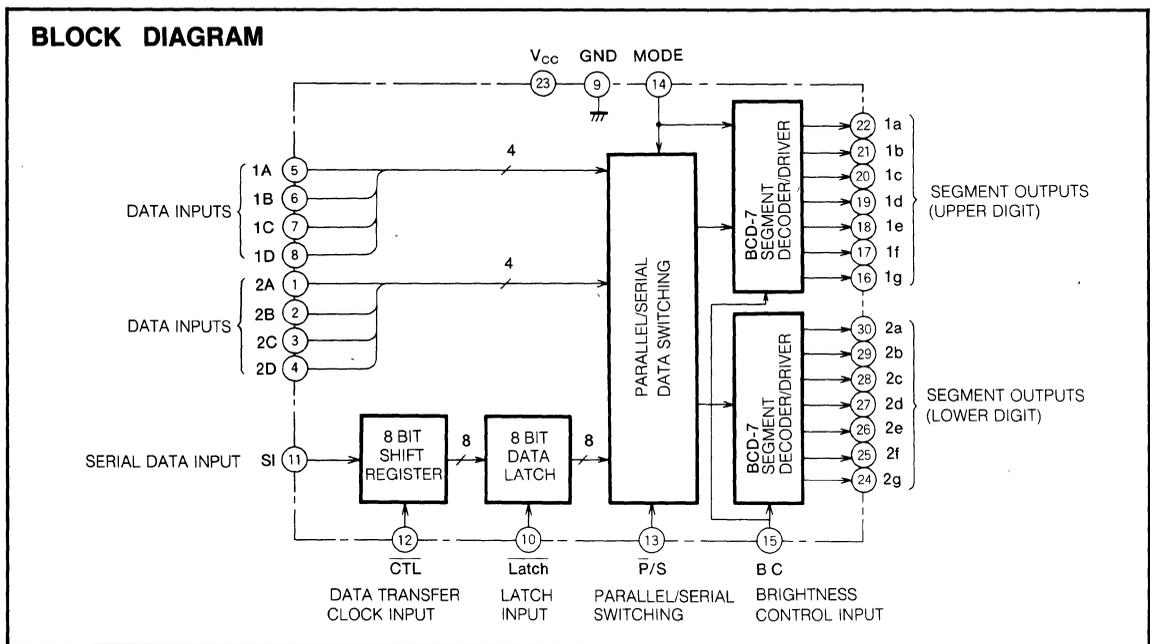
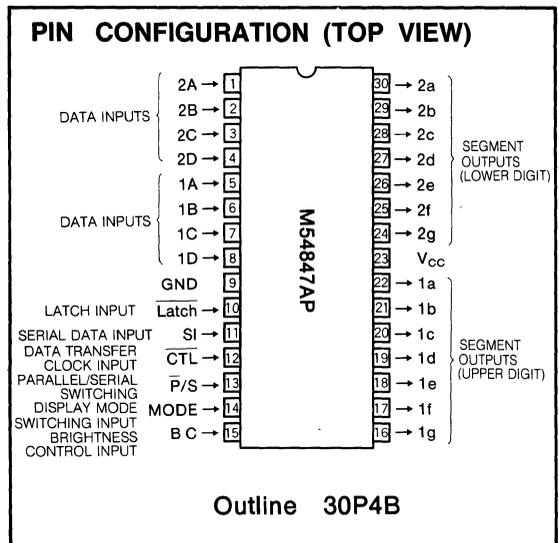
TV channel display

FUNCTIONAL DESCRIPTION

The M54847AP is a 2 digit BCD-7 segment decoder/driver for static drive of LED and fluorescent character displays.

The following display modes are possible.

- MODE I Numerical display of 00 ~ 99
 MODE II Numerical display of 0 ~ 39, and
 AU, CA, --



2-DIGIT BCD-7SEGMENT DECODER/DRIVER

OPERATING DESCRIPTION

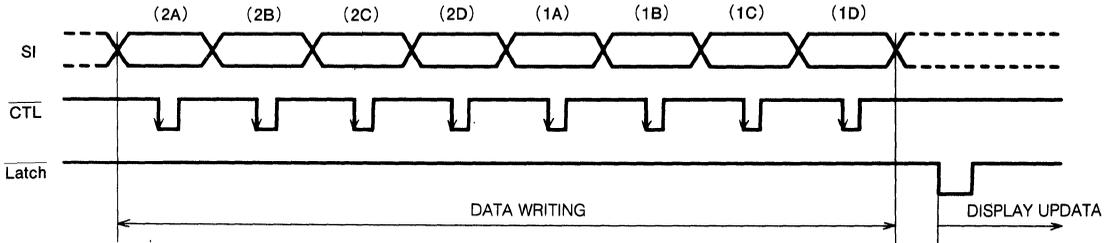
● Display mode

- (1) When the mode switching input is high, both digits are driven in accordance with Function Table I.
- (2) When the mode switching input is low, input 1C and 1D become the character data inputs, driving the display in accordance with Function Table II.

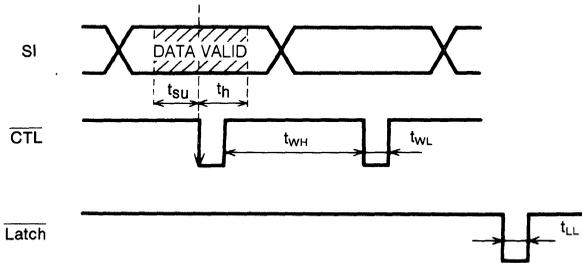
● Serial input data writing

Data 2A→2D and 1A→1D is read sequentially by the low edge of CTL. After all 8 bits are loaded in the shift register, the display is updated by switching Latch input to low.

SERIAL DATA WRITING



INPUT TIMING DIAGRAM



MINIMUM VALUES

- $t_{su} = 20\mu s$
- $t_h = 20\mu s$
- $t_{WH} = 30\mu s$
- $t_{WL} = 10\mu s$
- $t_{LL} = 10\mu s$

FUNCTION TABLE I

Data input				Segment output							Display
A	B	C	D	a	b	c	d	e	f	g	
L	L	L	L	H	H	H	H	H	H	L	0
H	L	L	L	L	H	H	L	L	L	L	1
L	H	L	L	H	H	L	H	H	L	H	2
H	H	L	L	H	H	H	H	L	L	H	3
L	L	H	L	L	H	H	L	L	H	H	4
H	L	H	L	H	L	H	H	L	H	H	5
L	H	H	L	H	L	H	H	H	H	H	6
H	H	H	L	H	H	H	L	L	L	L	7
L	L	L	H	H	H	H	H	H	H	H	8
H	L	L	H	H	H	H	H	L	H	H	9
L	H	L	H	L	L	L	L	L	L	H	-
H	H	L	H	L	L	L	L	H	H	H	E
L	L	H	H	H	L	L	H	H	H	L	F
H	L	H	H	L	L	L	L	L	L	L	Blank
L	H	H	H	L	L	H	H	H	L	H	o
H	H	H	H	L	L	L	L	L	L	L	Blank

FUNCTION TABLE II

Data input		Other data 1A, 1B 2A~2D	Display	
1C	1D		First digit	Second digit
L	L	X	-(Note 1)	-(Note 1)
H	L	X	E	R
L	H	X	R	U
H	H	-	(Note 2)	(Note 3)

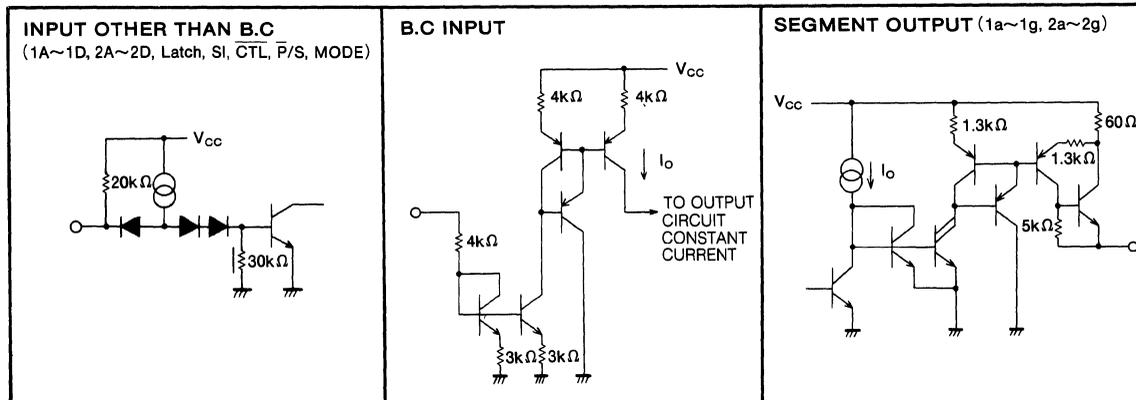
- Note 1 : Only segment g lights.
 Note 2 : When both 1C and 1D inputs are high, first digit display blanking or numerical display of 1, 2 or 3 is determined by 1A, 1B input state.

Data input				Segment output							Display
1A	1B	1C	1D	1a	1b	1c	1d	1e	1f	1g	
L	L	H	H	L	L	L	L	L	L	L	Blank
H	L	H	H	L	H	H	L	L	L	L	1
L	H	H	H	H	H	L	H	H	L	H	2
H	H	H	H	H	H	H	H	L	L	H	3

Note 3 : Other digit codes are identical to those in function table I.

2-DIGIT BCD-7SEGMENT DECODER/DRIVER

I/O CIRCUIT DIAGRAM



ABSOLUTE MAXIMUM RATINGS (T_a = -10~+60°C, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage		-0.5~+7	V
V _I	Input voltage		-0.5~+V _{CC}	V
V _{CC} -V _O	Voltage between supply and output		-0.5~+35	V
T _{opr}	Operating temperature		-10~+60	°C
T _{stg}	Storage temperature		-40~+125	°C
P _d	Power dissipation	T _a = 60°C	800	mW

RECOMMENDED OPERATING CONDITIONS (T_a = -10~+60°C, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V _{CC}	Supply voltage		4.5	5	6	V
I _{seg}	Segment current				-10	mA
V _O	Output withstand voltage when output is off				-25	V

ELECTRICAL CHARACTERISTICS (T_a = -10~+60°C, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{IH}	High input voltage	V _{CC} = 4.5~6V	2		V _{CC}	V
V _{IL}	Low input voltage	V _{CC} = 4.5~6V	0		0.6	V
I _{IH}	High input current	V _{CC} = 6V	0.5	0.75	1.2	mA
		V _{IH} = 6V				
I _{IL}	Low input current	V _{CC} = 6V			50	μA
		V _{IL} = 0V				
I _{seg}	Segment output current	V _{CC} = 5V, V _O = 2V, B.C pin is connected to V _{CC} .	-10			mA
I _{sIk}	Segment leak current	V _{CC} = 5V, V _O = -25V			-50	μA
I _{CC1}	Supply voltage	V _{CC} = 6V, All inputs and outputs are open		4	8	mA

Note 4 : All typical values are at V_{CC} = 5V, T_a = 25°C.

M54970P

9-BIT SERIAL-INPUT, LATCHED DRIVER

DESCRIPTION

The M54970P is a semiconductor integrated circuit of I^2L structure containing a serial input to serial/parallel output 9-bit shift register and latch as well as an output driver.

FEATURES

- Serial input to serial/parallel output
- Cascade connections possible through serial output
- Enable input for output control
- Power-cut input
- Driver : Withstand voltage $BV_{CEO} \geq 20V$
Large drive current ($I_{O(max)} = 300mA$)
- Wide operating temperature range $T_a = -20 \sim +75^\circ C$

APPLICATION

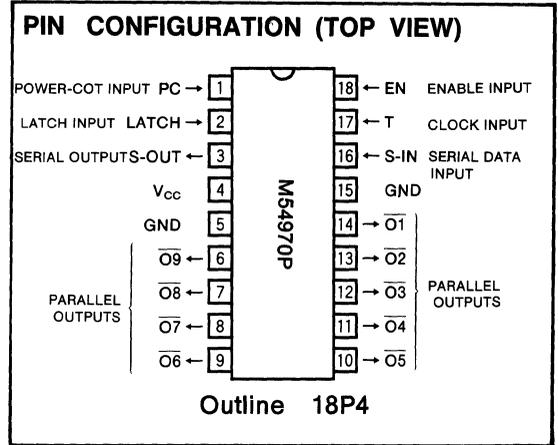
Thermal printer head dot driver Serial-to-parallel conversion

FUNCTIONAL DESCRIPTION

The M54970P consists of a 9bit D-type flip-flop, the output of which is connected to 9 latches.

When data is applied to the serial data input (S-IN) and a clock pulse is applied to clock input (T), an "L" to "H" change of the clock will cause the data input signals to enter the internal shift registers and the data in the shift registers will be shifted in order.

Using a number of M54970P units for bit expansion in

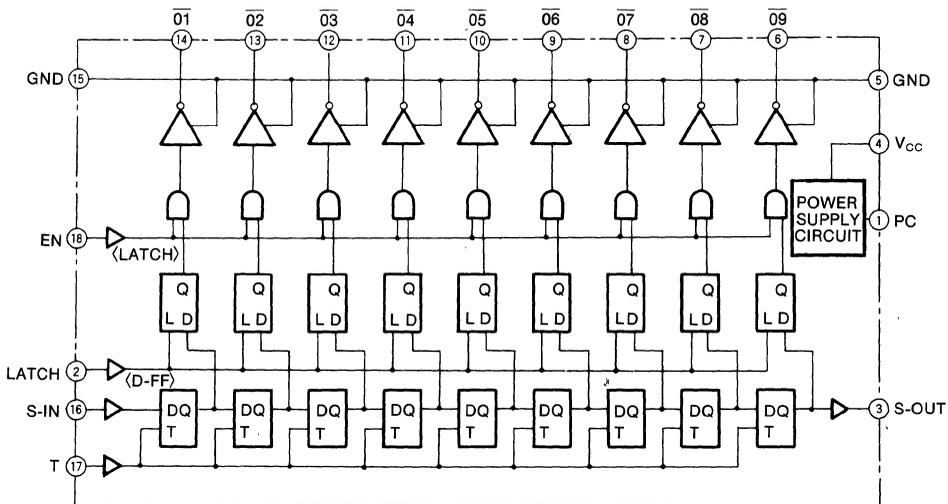


series will entail connecting serial output (S-OUT) to S-IN of the next-stage M54970P.

In parallel output, when the power-cut input and latch input are set to "H" and the output-control input (enable input EN) is "H", a clock pulse changing from "L" to "H" will cause the serial data input signal to appear at output $\bar{O}1$, and the data will be shifted in order at outputs $\bar{O}2 \sim \bar{O}9$.

The parallel output will yield a signal that is inverted with respect to the serial data input.

BLOCK DIAGRAM



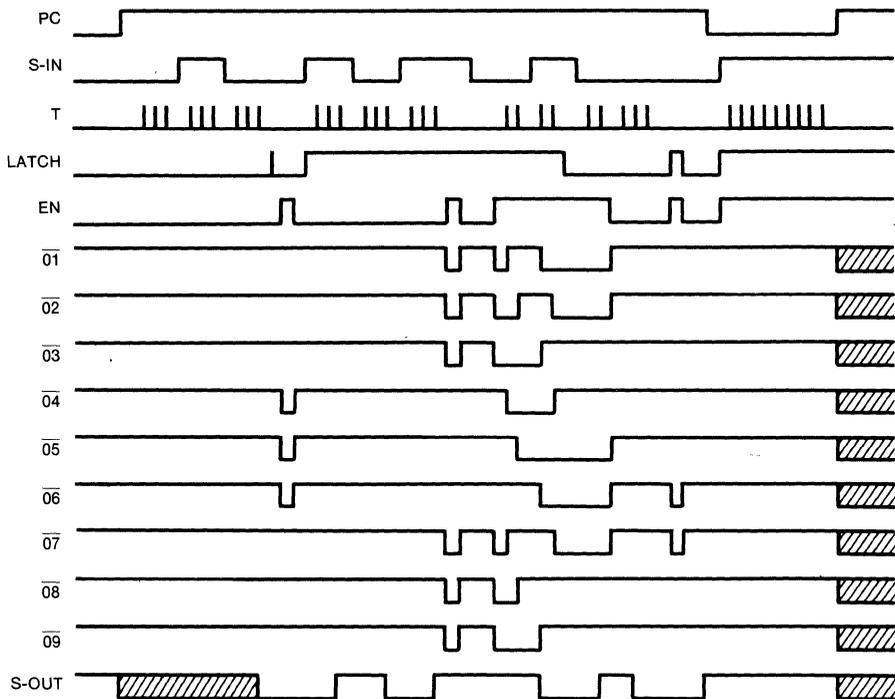
9-BIT SERIAL-INPUT, LATCHED DRIVER

Setting the LATCH input to "L" will prevent data from entering the latch.

When the EN input is set to "L", all outputs ($\overline{01} \sim \overline{09}$) will be set to OFF. Since the internal logic state of the IC is uncertain at power-on time, set the EN input to "L" (and outputs $\overline{01} \sim \overline{09}$ will be set to OFF) until the input data is set and

the internal logic state has been determined.

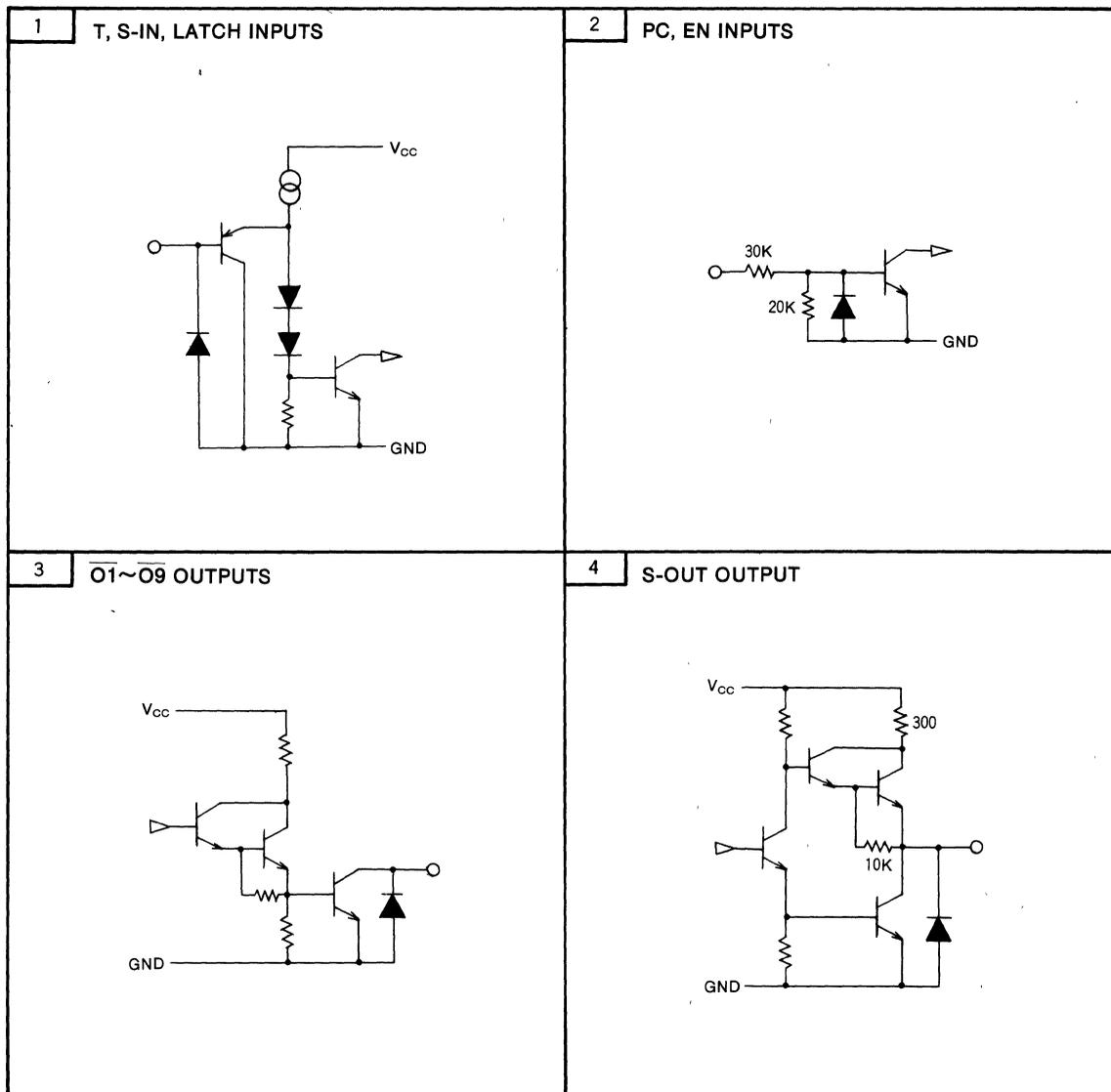
The power will be cut when the power-cut input is set to "L", and since the data of the shift registers and latches are not maintained in this state, it will be necessary to input data again in order to set the output following a change of PC input from "L" to "H".

TIMING CHART

*The state of the shaded areas is uncertain

9-BIT SERIAL-INPUT, LATCHED DRIVER

INPUT/OUTPUT EQUIVALENT CIRCUIT SCHEMATICS



9-BIT SERIAL-INPUT, LATCHED DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a = -20^\circ\text{C} \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		$-0.5 \sim +8$	V
V_I	Input voltage		$-0.5 \sim +10$	V
V_O	Output voltage	Output is OFF	$-0.5 \sim +20$	V
I_O	Output current		350	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.25	W
T_{opr}	Operating temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4.5	5.0	5.5	V
V_O	Applied output voltage	When output is OFF			20	V
I_O	Output current (per circuit)	All outputs ON simultaneously Duty cycle less than 30%			300	mA

ELECTRICAL CHARACTERISTICS ($T_a = +25^\circ\text{C}$, unless otherwise noted)

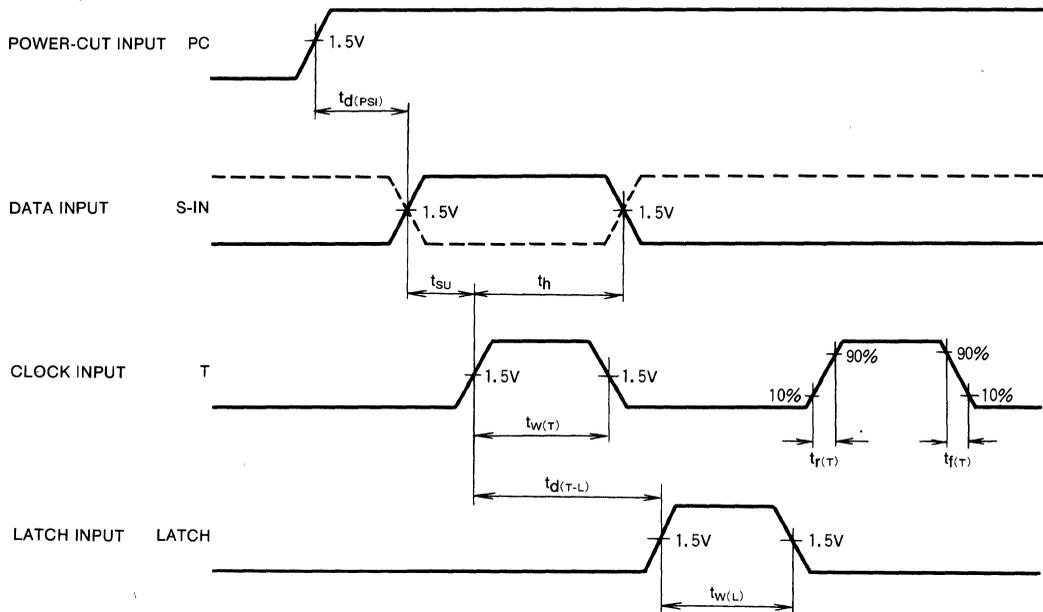
Symbol	Parameter	Pin	Test conditions	Limits			Unit
				Min	Typ	Max	
V_{IH}	"H" input voltage	2, 16, 17	$T_a = -20 \sim +75^\circ\text{C}$	2.2		V_{CC}	V
V_{IL}	"L" input voltage			0		0.8	V
V_{IH}	"H" input voltage	1, 18	$T_a = -20 \sim +75^\circ\text{C}$	2.2		V_{CC}	V
V_{IL}	"L" input voltage			0		0.8	V
I_{IH}	"H" input current	2, 16, 17	$V_{CC} = 5.5\text{V}, V_{IH} = 2.4\text{V}$			10	μA
I_{IL}	"L" input current		$V_{CC} = 5.5\text{V}, V_{IL} = 0.4\text{V}$			-50	μA
I_{IH}	"H" input current	1, 18	$V_{CC} = 5.5\text{V}, V_{IH} = 5.5\text{V}$			250	μA
I_{IL}	"L" input current		$V_{CC} = 5.5\text{V}, V_{IL} = 2.4\text{V}$			100	μA
			$V_{CC} = 5.5\text{V}, V_{IL} = 0\text{V}$			-10	μA
V_{OH}	"H" output voltage	3	$V_{CC} = 4.5\text{V}, I_{OH} = -400\mu\text{A}$	2.4			V
V_{OL}	"L" output voltage		$V_{CC} = 4.5\text{V}, I_{OL} = 8\text{mA}$			0.4	V
V_{OL}	"L" output voltage	6~14	$V_{CC} = 4.5\text{V}, I_{OL} = 300\text{mA}$			0.6	V
I_{CC1}	Supply current	4	$V_{CC} = 5.5\text{V}$, power-cut is ON			10	μA
I_{CC2}			$V_{CC} = 5.5\text{V}$, EN is "L"		10	15	mA
I_{CC3}			$V_{CC} = 5.5\text{V}$, all outputs are ON		90	130	mA
$I_{O(leak)}$	Output leakage current	6~14	$V_{CC} = 5.5\text{V}, V_{OH} = 20\text{V}$			100	μA

9-BIT SERIAL-INPUT, LATCHED DRIVER

REQUIRED TIMING CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
$f_{(T)}$	Clock frequency	Input duty cycle 40~60%			1	MHz
$t_{W(T)}$	Clock pulse width		0.4			μS
$t_{W(L)}$	Latch pulse width		0.4			μS
t_{SU}	Data setup time		0.2			μS
t_H	Data hold time		0.3			μS
$t_{d(T-L)}$	Clock-latch time		1			μS
$t_{r(T)}$	Clock pulse rise time				0.5	μS
$t_{f(T)}$	Clock pulse fall time				0.5	μS
$t_{d(P-SI)}$	Power-cut input \rightarrow data input setting time	Hold EN input at "L" when PC input is changed from "L" to "H"	2			μS

VOLTAGE WAVEFORM

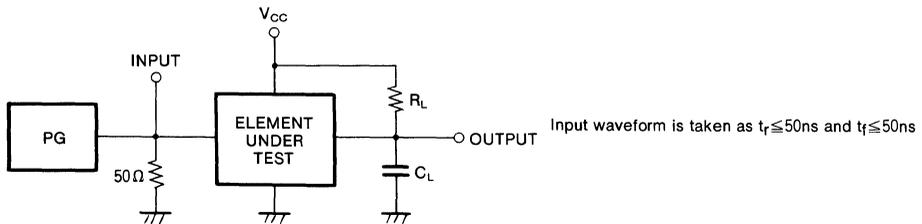


9-BIT SERIAL-INPUT, LATCHED DRIVER

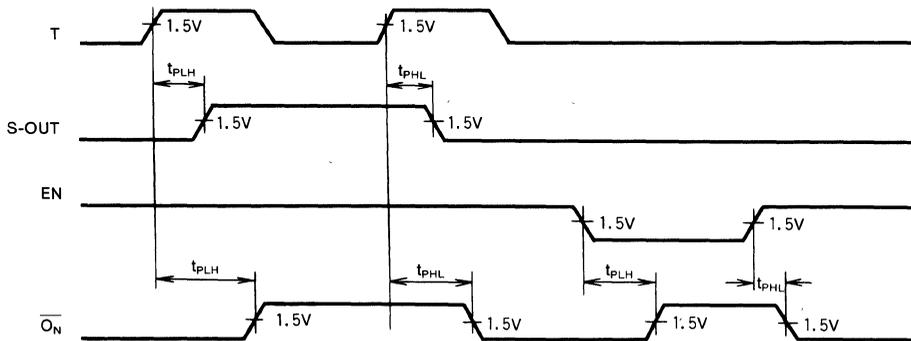
SWITCHING CHARACTERISTICS ($T_a=25^\circ\text{C}$, $V_{CC}=5\text{V}$)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
t_{PLH}	"L-H" output propagation time (Input T to output S-OUT)	$V_{IH}=3\text{V}$ $V_{IL}=0\text{V}$ $R_L: S\text{-OUT}=2\text{K}\Omega$ $R_L: \overline{O_N}=100\Omega$ (N=1~9) $C_L=15\text{pF}$ (Note 1)			0.7	μS
t_{PHL}	"H-L" output propagation time (Input T to output S-OUT)				0.8	μS
t_{PLH}	"L-H" output propagation time (Input T to output $\overline{O_N}$)				5	μS
t_{PHL}	"H-L" output propagation time (Input T to output $\overline{O_N}$)				1	μS
t_{PLH}	"L-H" output propagation time (Input EN to output $\overline{O_N}$)				10	μS
t_{PHL}	"H-L" output propagation time (Input EN to output $\overline{O_N}$)				1	μS

(Note 1) TEST CIRCUIT



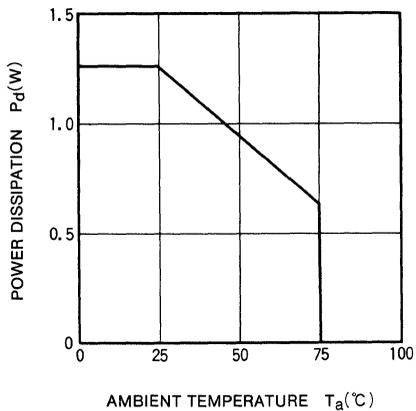
(VOLTAGE WAVEFORM)



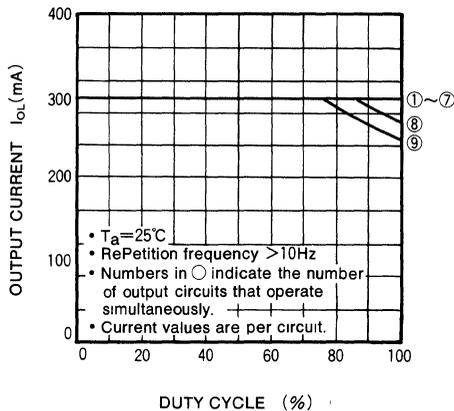
9-BIT SERIAL-INPUT, LATCHED DRIVER

TYPICAL CHARACTERISTICS

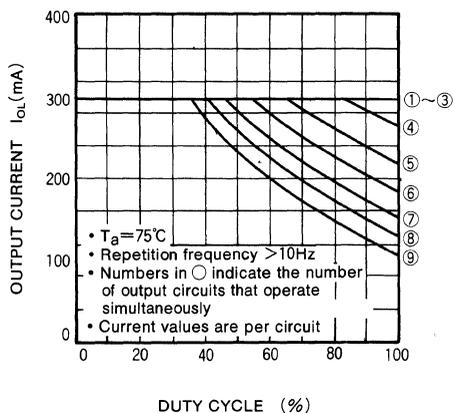
THERMAL DERATING



DUTY CYCLE VS PERMISSIBLE OUTPUT CURRENT



DUTY CYCLE VS PERMISSIBLE OUTPUT CURRENT



M54975P

BI-CMOS 8-BIT SERIAL-INPUT, LATCHED DRIVER

DESCRIPTION

The M54975P is a semiconductor integrated circuit fabricated using Bi-CMOS technology. It contains a serial input to serial/parallel output 8-bit shift register and latch as well as a bipolar 8-bit parallel-output driver.

FEATURES

- Serial input to serial/parallel output
- Cascade connections possible through serial output
- Latch circuit included for each stage
- Enable input for output control
- Low supply current $I_{CC} \geq 10\mu A$ at standby
- Serial input/output level is compatible with standard CMOS
- Driver : Withstand voltage $BV_{CEO} \geq 30$
Large drive current ($I_{O(max)} = 300mA$)
- Wide operating temperature range $T_a = -20 \sim +75^\circ C$

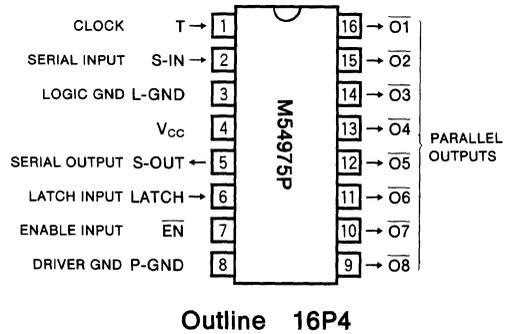
APPLICATION

- Thermal printer head dot driver
- Serial-to parallel conversion
- Relay, solenoid driver

FUNCTION DESCRIPTION

The M54975P consists of an 8-bit D-type flip-flop, the output of which is connected to 8 latches. When data is applied to the serial data input (S-IN) and a clock pulse is applied to clock input (T), an "L" to "H" change of the clock will cause the data input signals to enter the internal shift registers and the data in the shift registers will be shifted in order.

PIN CONFIGURATION (TOP VIEW)



Using a number of M54975P units for bit expansion in series will entail connecting serial output (S-OUT) to S-IN of the next-stage M54975P.

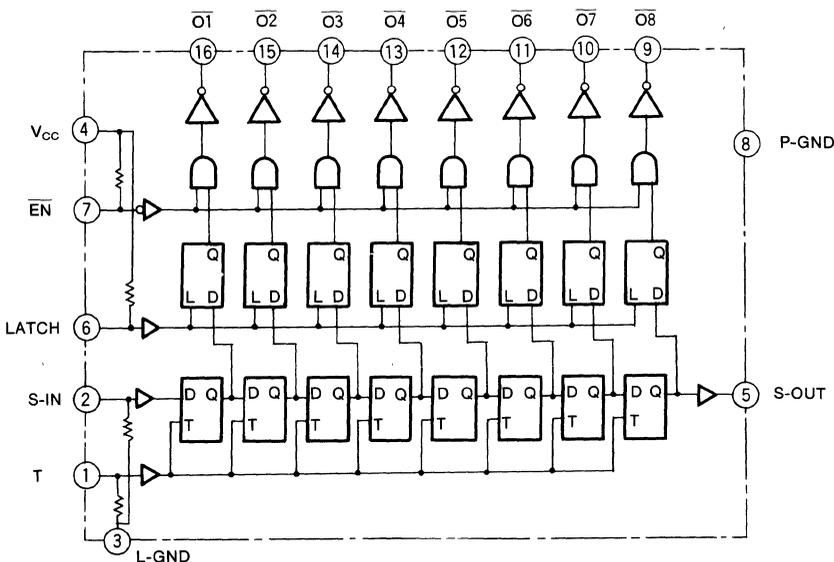
In parallel output, when the latch input is set to "H" and the output-control input (enable input EN) is "L", a clock pulse changing from "L" to "H" will cause the serial data input signal to appear at output $\overline{O1}$, and the data will be shifted in order at outputs $\overline{O2} \sim \overline{O8}$.

The parallel output will yield a signal that is inverted with respect to the serial data input.

Setting the LATCH input to "L" will prevent data from entering the latch.

When the EN input is set to "H", all outputs ($\overline{O1} \sim \overline{O8}$) will be set to OFF. Since the internal logic state of the IC is uncertain at power-on time, set the \overline{EN} input to "H" (and out-

BLOCK DIAGRAM

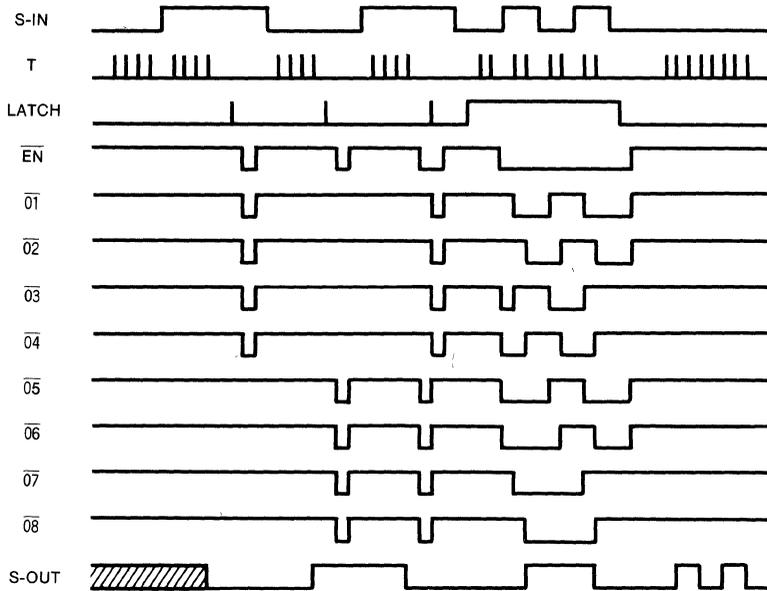


BI-CMOS 8-BIT SERIAL-INPUT, LATCHED DRIVER

puts $\overline{O1} \sim \overline{O8}$ will set to OFF) until the input data is set and the internal logic state has been determined.
 L-GND is the ground of the CMOS logic circuit section and

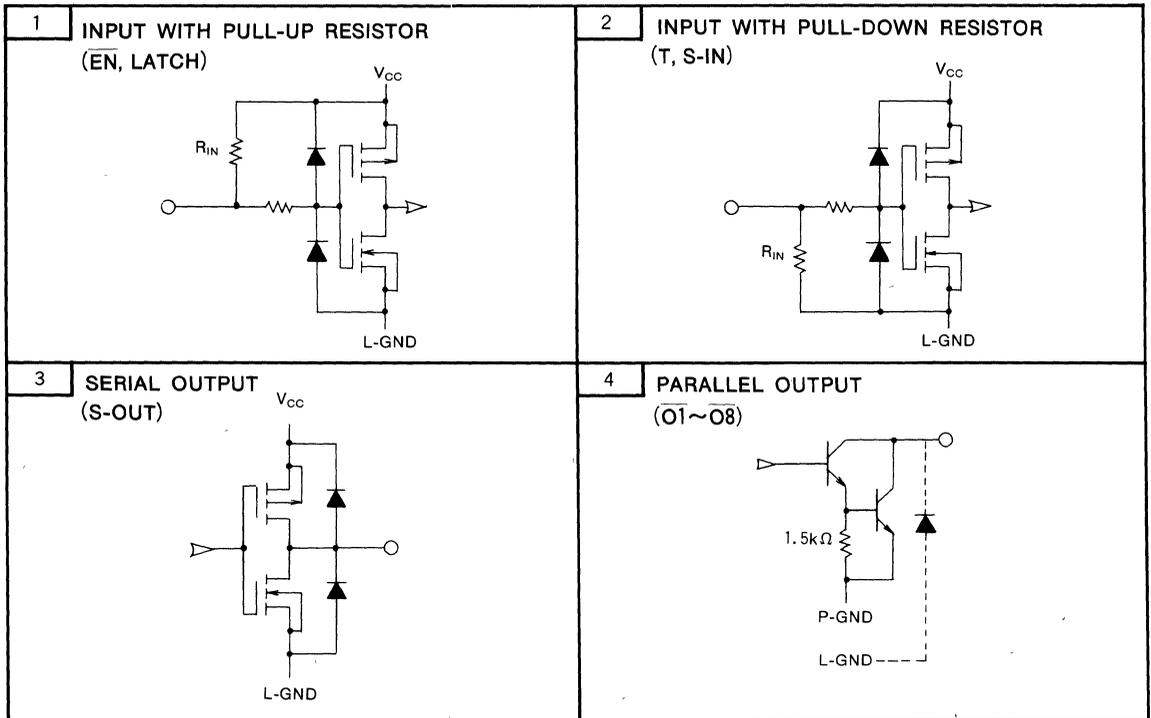
P-GND is the ground for the output driver section ($\overline{O1} \sim \overline{O8}$), which is made up of bipolar transistors that are capable of driving large currents.

TIMING CHART



*The stage of the shaded areas is uncertain.

INPUT/OUTPUT CIRCUIT SCHEMATICS



BI-CMOS 8-BIT SERIAL-INPUT, LATCHED DRIVER

ABSOLUTE MAXIMUM RATING ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		$-0.5 \sim +8$	V
V_I	Input voltage		$-0.5 \sim V_{CC} + 0.5$	V
V_O	Output voltage	S-OUT	$-0.5 \sim V_{CC} + 0.5$	V
		$O1 \sim O8$: Output is OFF	$-0.5 \sim +30$	
I_O	Output Current	$O1 \sim O8$: Output is ON	350	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.25	W
T_{opr}	Operating temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$ unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	5	6	V
V_O	Applied output voltage	$O1 \sim O8$: When output is OFF			30	V
I_O	Output current (per circuit)	$O1 \sim O8$: All outputs ON simultaneously Duty cycle less than 15%			300	mA

ELECTRICAL CHARACTERISTICS ($T_a = +25^\circ\text{C}$, $V_{CC} = 5\text{V}$, unless otherwise noted)

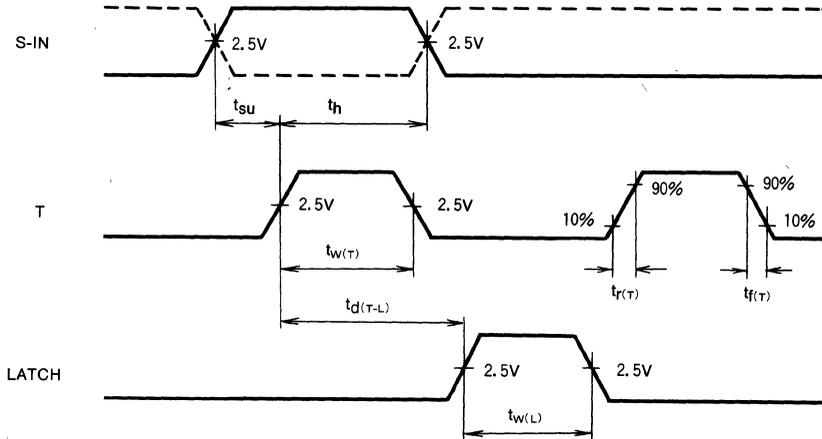
Symbol	Parameter	Pin	Conditions	Limits			Unit
				Min	Typ	Max	
V_{IH}	"H" input voltage	1, 2	$T_a = -20 \sim +75^\circ\text{C}$	$0.7V_{CC}$		V_{CC}	V
V_{IL}	"L" input voltage	6, 7		0		$0.3V_{CC}$	V
R_{IN}	input resistance			50	—	—	$k\Omega$
V_{OH}	"H" output voltage	5	$ I_O \leq 1\mu\text{A}$	4.9	—	—	V
V_{OL}	"L" output voltage			—	—	0.1	—
I_{OH}	"H" output current	5	$V_{OH} = 4.5\text{V}$	-100	—	—	μA
I_{OL}	"L" output current		$V_{OL} = 0.4\text{V}$	400	—	—	μA
V_{OL1}	"L" output voltage	9~16	$I_{OL} = 100\text{mA}$	—	—	1.2	V
V_{OL2}			$I_{OL} = 200\text{mA}$	—	—	1.4	V
V_{OL3}			$I_{OL} = 300\text{mA}$	—	—	1.6	V
I_{OLK}	Output leakage current		$V_O = 30\text{V}$	—	—	50	μA
I_{CC1}	Supply current	4	Inputs free, all driver outputs OFF	—	—	10	μA
I_{CC2}			Driver output : 1 circuit ON	—	—	1.7	mA

REQUIRED TIMING CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$f_{(T)}$	Clock frequency	Input duty cycle 40~60%			2	MHz
$t_{W(T)}$	Clock pulse width		200			nS
$t_{W(L)}$	Latch pulse width		200			nS
t_{su}	Data setup time		100			nS
t_h	Data hold time		100			nS
$t_{d(T-L)}$	Clock-latch time		400			nS
$t_{r(T)}$	Clock pulse rise time				500	nS
$t_{f(T)}$	Clock pulse fall time				500	nS

BI-CMOS 8-BIT SERIAL-INPUT, LATCHED DRIVER

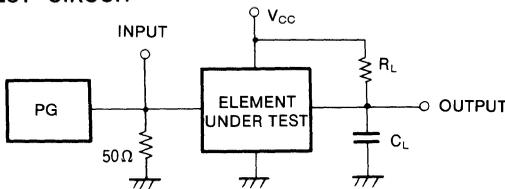
VOLTAGE WAVEFORM



SWITCHING CHARACTERISTICS ($T_a = +25^\circ\text{C}$, $V_{CC} = 5\text{V}$)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
t_{PLH}	"L-H" output propagation time (Input T to output S-OUT)	$V_{IH} = 5\text{V}$ $V_{IL} = 0\text{V}$ $R_L : S\text{-OUT} = \infty$ $R_L : \overline{O}_N = 100\Omega$ $(N=1\sim 8)$ $C_L = 15\text{pF}$ (Note 1)			0.3	μS
t_{PHL}	"H-L" output propagation time (Input T to output S-OUT)				0.3	μS
t_{PLH}	"L-H" output propagation time (Input T to output \overline{O}_N)				10	μS
t_{PHL}	"H-L" output propagation time (Input T to output \overline{O}_N)				2	μS
t_{PLH}	"L-H" output propagation time (Input EN to output \overline{O}_N)				10	μS
t_{PHL}	"H-L" output propagation time (Input EN to output \overline{O}_N)				2	μS

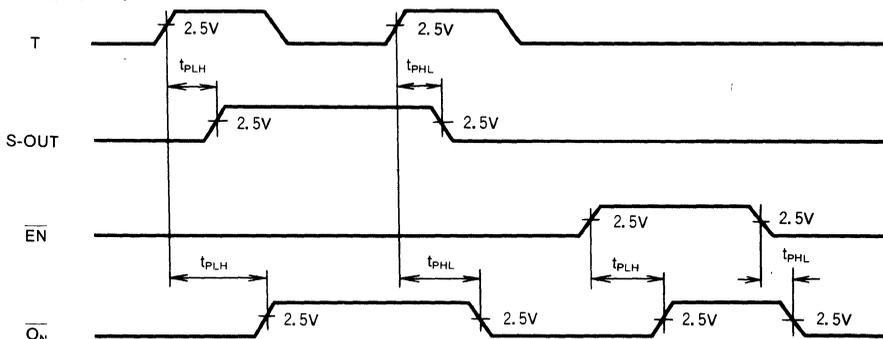
(Note 1) TEST CIRCUIT



Input waveform is taken as $t_r \leq 20\text{ns}$ and $t_f \leq 20\text{ns}$

C_L includes wiring stray capacitance and probe input capacitance.

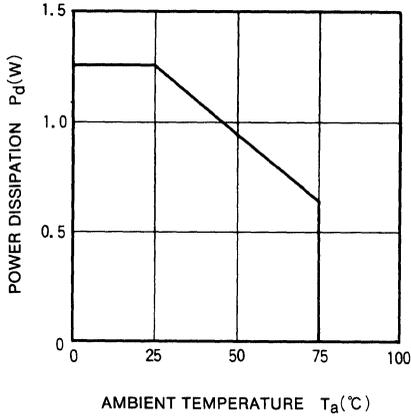
(VOLTAGE WAVEFORM)



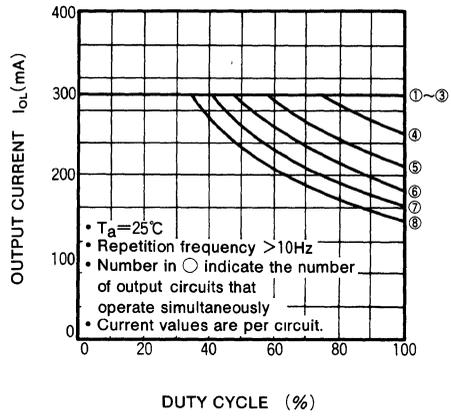
BI-CMOS 8-BIT SERIAL-INPUT, LATCHED DRIVER

TYPICAL CHARACTERISTICS

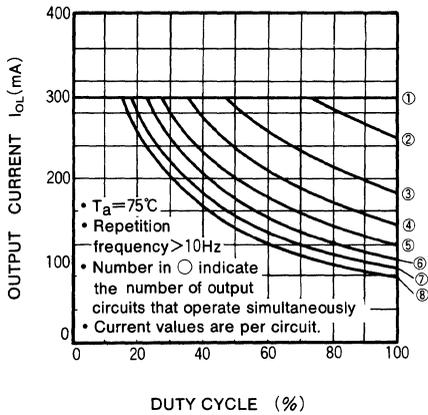
THERMAL DRAINING



DUTY CYCLE VS PERMISSIBLE OUTPUT CURRENT



DUTY CYCLE VS PERMISSIBLE OUTPUT CURRENT



M54976P

BI-CMOS 8-BIT PARALLEL-INPUT LATCHED DRIVER

DESCRIPTION

The M54976P is a semiconductor integrated circuit fabricated using Bi-CMOS technology. It contains 8 output drivers of latch and bipolar structure.

FEATURES

- Enable input for output control
- Low supply current..... $I_{CC} \leq 10\mu A$ at standby
- Input level is compatible with standard CMOS
- Driver : Withstand voltage..... $BV_{CEO} \geq 30$
Large drive current..... ($I_{O(max)} = 300mA$)
- Wide operating temperature range..... $T_a = -20 \sim +75^\circ C$

APPLICATION

- Printer head dot driver
- Relay, poleoid driver

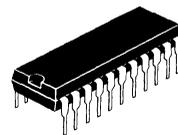
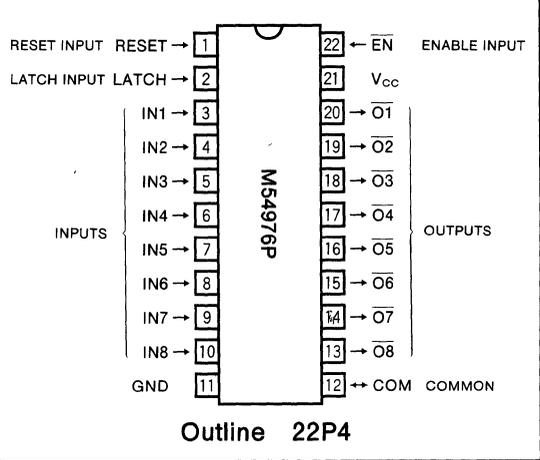
FUNCTIONAL DESCRIPTION

When data is applied to inputs IN1-IN8 and LATCH input is set to "H", the data will be latched with the truth table. Note that when an "H" signal is applied to the RESET input, the latch will maintain the reset state.

When the EN input is set to "L" and the data maintained in the latch are "H", the corresponding output will be ON and become "L".

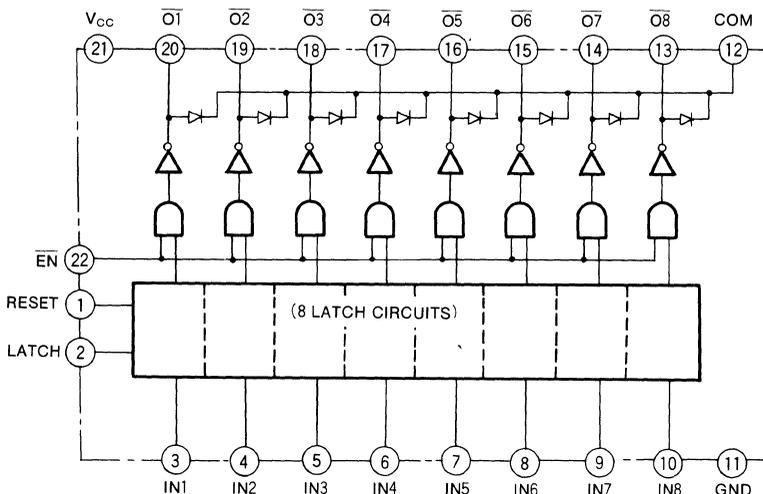
When both the LATCH and RESET inputs are "L", the latch will maintain the prior state irrespective of input signals IN1-IN8.

PIN CONFIGURATION (TOP VIEW)



22-pin molded plastic DIP

BLOCK DIAGRAM



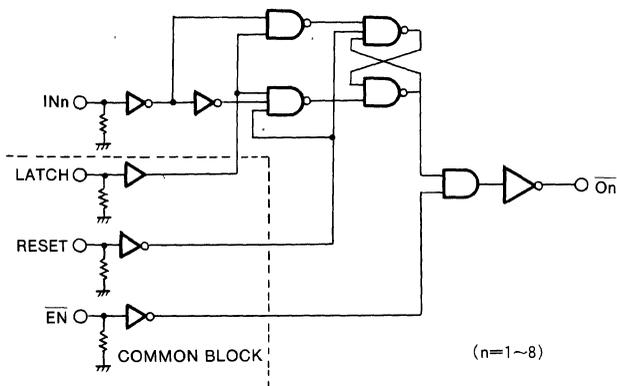
BI-CMOS 8-BIT PARALLEL-INPUT LATCHED DRIVER

TRUTH TABLE

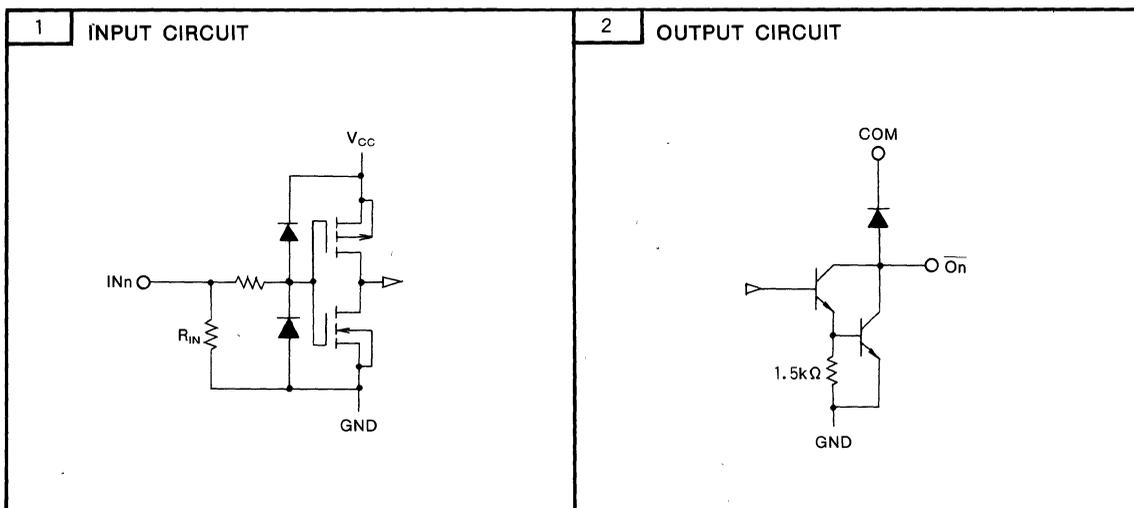
INPUTS				OUTPUT $\overline{O_n}$	
$\overline{IN_n}$	LATCH	RESET	\overline{EN}	t-1	t
L	H	L	L	X	H
H	H	L	L	X	L
X	X	H	X	X	H
X	X	X	H	X	H
X	L	L	L	L	L
X	L	L	L	H	H

L : "L" level
 H : "H" level
 X : Irrelevant
 t-1 : Previous state
 t : Present state
 Output H is in the OFF state
 Output L is in the ON state

LOGIC DIAGRAM (1 CIRCUIT)



INPUT/OUTPUT EQUIVALENT CIRCUIT SCHEMATICS



BI-CMOS 8-BIT PARALLEL-INPUT LATCHED DRIVER

ABSOLUTE MAXIMUM RATINGS ($T_a = -20^\circ\text{C} \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		$-0.5 \sim +8$	V
V_I	Input voltage		$-0.5 \sim V_{CC} + 0.5$	V
V_O	Output voltage	Output is OFF	$-0.5 \sim +30$	V
I_O	Output current	Output is on	350	mA
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.42	W
T_{opr}	Operating temperature range		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	5	6	V
V_O	Applied output voltage	When output is OFF			30	V
I_O	Output current (per circuit)	All outputs ON simultaneously Duty cycle less than 25%			300	mA

ELECTRICAL CHARACTERISTICS ($T_a = +25^\circ\text{C}$, $V_{CC} = 5\text{V}$, unless otherwise noted)

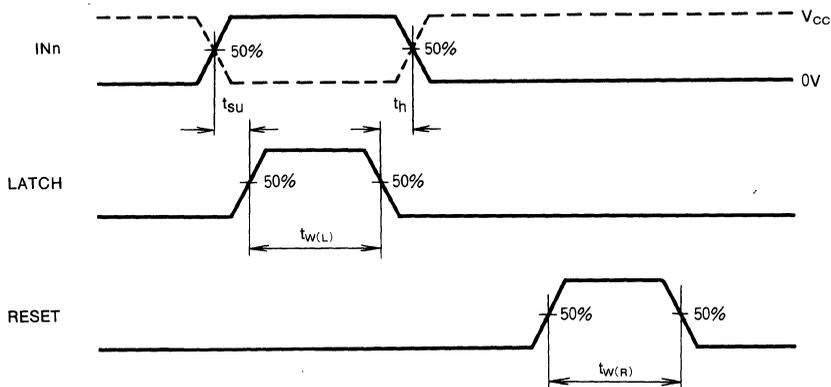
Symbol	Parameter	Pin	Test conditions	Limits			Unit
				Min	Typ	Max	
V_{IH}	"H" input voltage	1~10	$T_a = -20 \sim +75^\circ\text{C}$	$0.7V_{CC}$		V_{CC}	V
V_{IL}	"L" input voltage	22		0		$0.3V_{CC}$	V
R_{IN}	Input resistance			50			k Ω
V_{OL1}	"L" output voltage	13~20	$I_{OL} = 100\text{mA}$			1.2	V
V_{OL2}			$I_{OL} = 200\text{mA}$			1.4	V
V_{OL3}			$I_{OL} = 300\text{mA}$			1.6	V
I_{OLK}	Output leakage current	13~20	$V_O = 30\text{V}$			50	μA
I_{VF}	Clamp diode forward current	13~20	$I_F = 300\text{mA}$			2	V
I_{VR}	Clamp diode reverse current		$V_R = 30\text{V}$			50	μA
I_{CC1}	Supply current	21	All inputs are 0V, all outputs OFF			10	μA
I_{CC2}			Output : 1 circuit ON			1.4	mA

BI-CMOS 8-BIT PARALLEL-INPUT LATCHED DRIVER

REQUIRED TIMING CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$t_{w(L)}$	Latch pulse width		0.1			μS
$t_{w(R)}$	Reset pulse width		0.1			μS
t_{su}	Data setup time		0			μS
t_h	Data hold time		0.1			μS

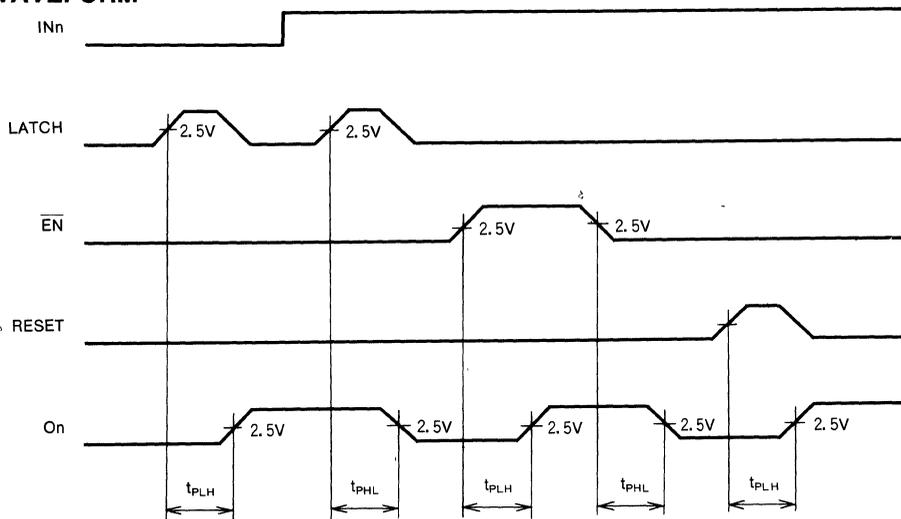
VOLTAGE WAVEFORM



SWITCHING CHARACTERISTICS ($T_a = +25^\circ\text{C}$, $V_{cc} = 5\text{V}$)

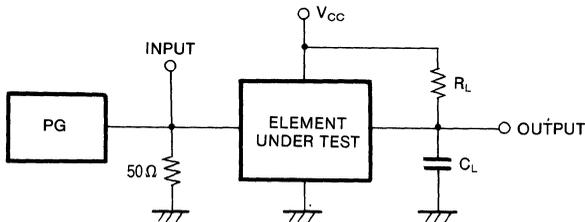
Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
t_{PLH}	"L-H" output propagation time (Input LATCH to output $\bar{O}n$)	$V_{IH} = 5\text{V}$ $V_{IL} = 0\text{V}$ $R_L = 100\Omega$ $C_L = 15\text{pF}$ (Note 1)		(0.6)	2	μS
t_{PHL}	"H-L" output propagation time (Input LATCH to output $\bar{O}n$)			(0.1)	0.5	μS
t_{PLH}	"L-H" output propagation time (Input $\bar{E}N$ to output $\bar{O}n$)			(0.6)	2	μS
t_{PHL}	"H-L" output propagation time (Input $\bar{E}N$ to output $\bar{O}n$)			(0.1)	0.5	μS
t_{PLH}	"L-H" output propagation time (Input RESET to output $\bar{O}n$)			(0.6)	2	μS

VOLTAGE WAVEFORM



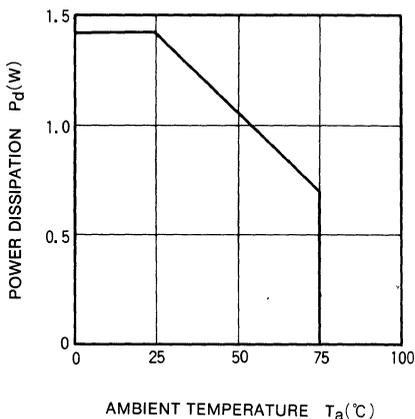
BI-CMOS 8-BIT PARALLEL-INPUT LATCHED DRIVER

(Note 1) TEST CIRCUIT

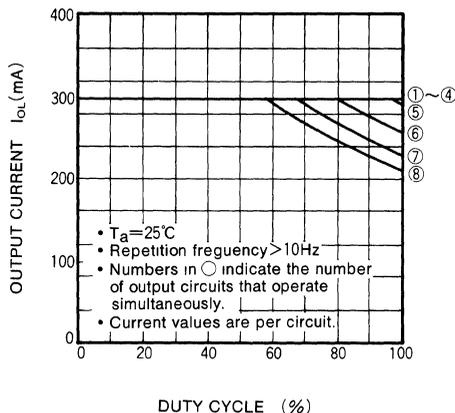


- Input waveform is taken as $t_r \leq 20\text{ns}$ and $t_f \leq 20\text{ns}$
- C_L includes wiring stray capacitance and probe input capacitance.

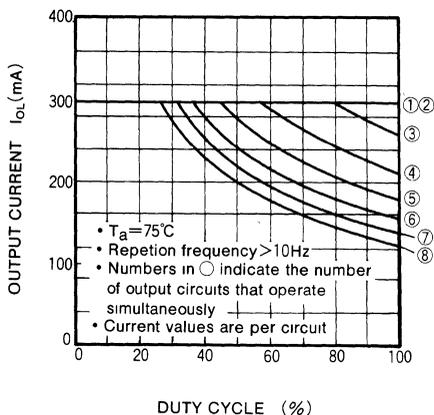
THERMAL DERATING



DUTY CYCLE VS PERMISSIBLE OUTPUT CURRENT



DUTY CYCLE VS PERMISSIBLE OUTPUT CURRENT



M54977P

BI-CMOS 12-BIT SERIAL-INPUT, LATCHED DRIVER

DESCRIPTION

The M54977P is a semiconductor integrated circuit fabricated using Bi-CMOS technology. It contains a serial input to serial/parallel output 12-bit shift register and latch as well as a bipolar 12-bit parallel-output driver.

FEATURES

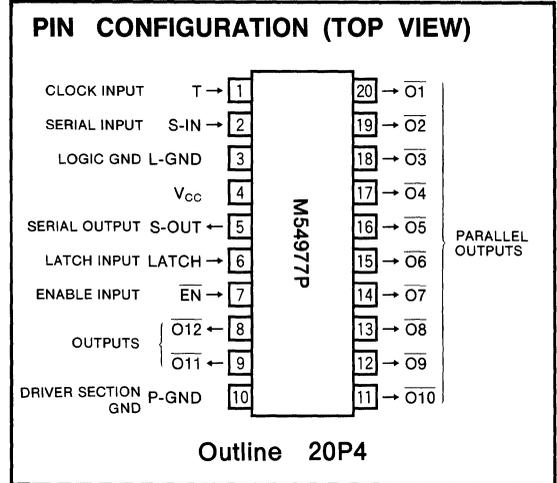
- Serial input to serial/parallel output
- Cascade connections possible through serial output
- Latch circuit included for each stage
- Enable input for output control
- Low supply current $I_{CC} \geq 10\mu A$ at standby
- Serial input/output level is compatible with standard CMOS
- Driver : Withstand voltage $BV_{CE0} \geq 30$
Large drive current ($I_{O(max)} = 200mA$)
- Wide operating temperature range $T_a = -20 \sim +75^\circ C$

APPLICATION

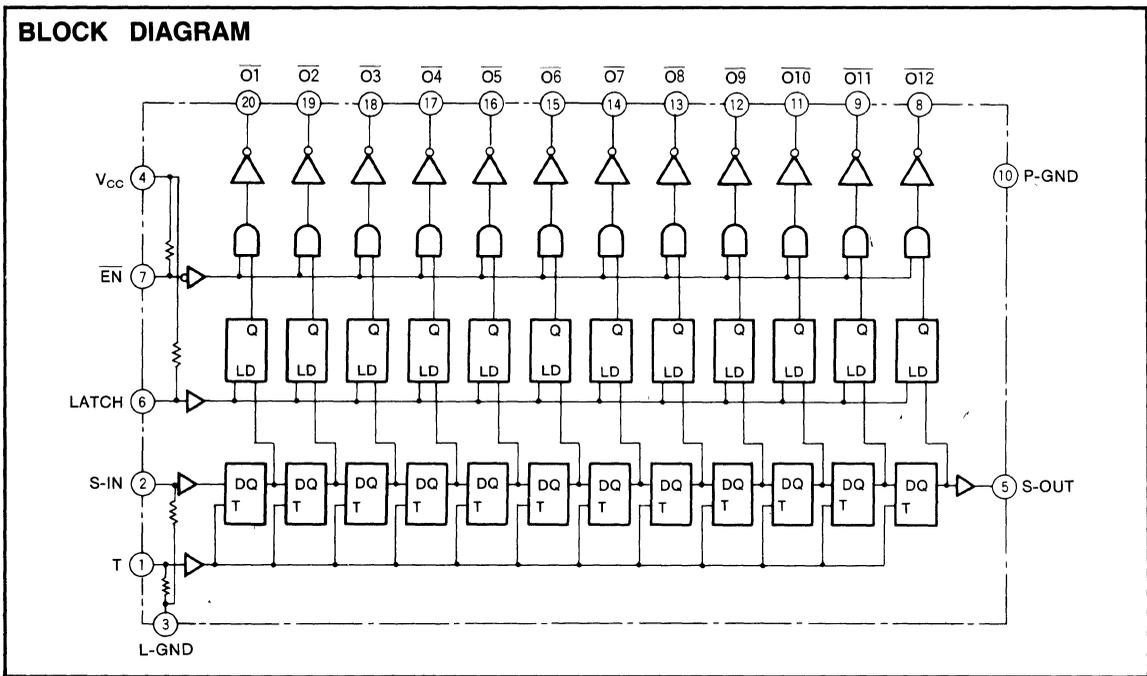
- Thermal printer head dot driver
- Serial-to parallel conversion
- Relay, solenoid driver

FUNCTION DESCRIPTION

The M54977P consists of an 12-bit D-type flip-flop, the output of which is connected to 12 latches. When data is applied to the serial data input (S-IN) and a clock pulse is applied to clock input (T), an "L" to "H" change of the clock will cause the data input signals to en-



ter the internal shift registers and the data in the shift registers will be shifted in order. Using a number of M54977P units for bit expansion in series will entail connecting serial output (S-OUT) to S-IN of the next-stage M54977P. In parallel output, when the latch input is set to "H" and the output-control input (enable input EN) is "L", a clock pulse changing from "L" to "H" will cause the serial data input signal to appear at output O1, and the data will be shifted in order at outputs O2~O12.



BI-CMOS 12-BIT SERIAL-INPUT, LATCHED DRIVER

The parallel output will yield a signal that is inverted with respect to the serial data input.

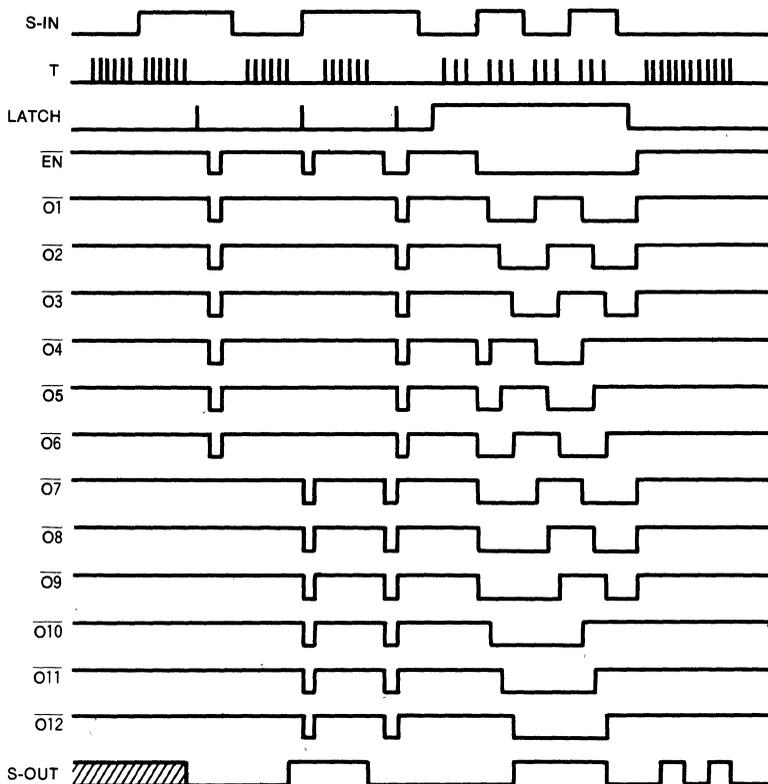
Setting the LATCH input to "L" will prevent data from entering the latch.

When the \overline{EN} input is set to "H", all outputs ($\overline{O1} \sim \overline{O12}$) will be set to OFF. Since the internal logic state of the IC is uncertain at power-on time, set the \overline{EN} input to "H" (and out-

puts $\overline{O1} \sim \overline{O12}$ will set to OFF) until the input data is set and the internal logic state has been determined.

L-GND is the ground of the CMOS logic circuit section and P-GND is the ground for the output driver section ($\overline{O1} \sim \overline{O12}$), which is made up of bipolar transistors that are capable of driving large currents.

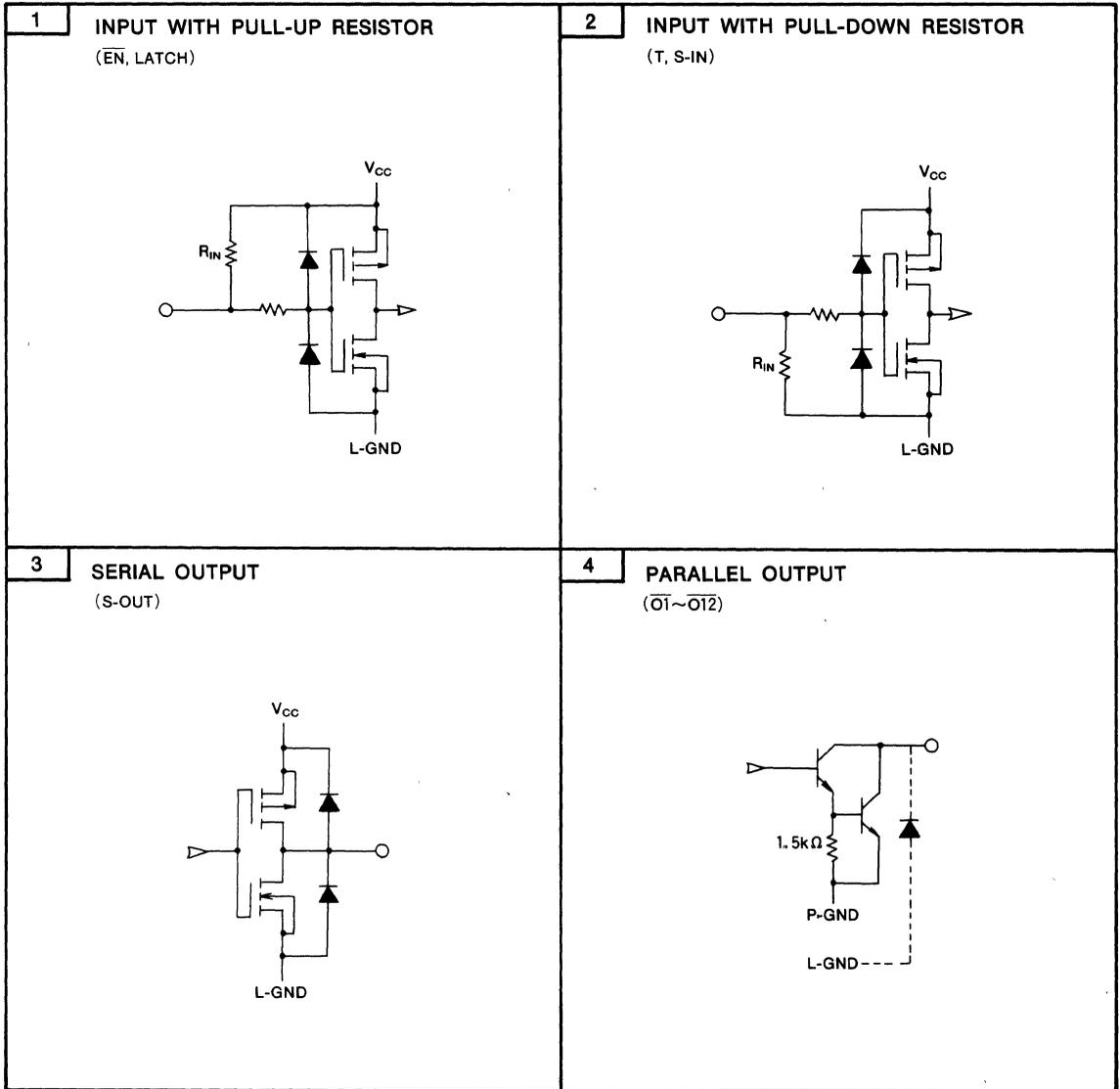
TIMING CHART



*The stage of the shaded areas is uncertain.

BI-CMOS 12-BIT SERIAL-INPUT, LATCHED DRIVER

INPUT/OUTPUT CIRCUIT SCHEMATICS



BI-CMOS 12-BIT SERIAL-INPUT, LATCHED DRIVER

ABSOLUTE MAXIMUM RATING ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage		$-0.5 \sim +8$	V
V_I	Input voltage		$-0.5 \sim V_{CC} + 0.5$	V
V_O	Output voltage	S-OUT	$-0.5 \sim V_{CC} + 0.5$	V
		$\bar{O}1 \sim \bar{O}12$: Output is OFF	$-0.5 \sim +30$	
I_O	Output Current	$\bar{O}1 \sim \bar{O}8$: Output is ON	250	mA
		$T_a = 25^\circ\text{C}$	1.25	
P_d	Power dissipation	$T_a = 25^\circ\text{C}$	1.25	W
T_{opr}	Operating temperature		$-20 \sim +75$	$^\circ\text{C}$
T_{stg}	Storage temperature range		$-55 \sim +125$	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$ unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{CC}	Supply voltage		4	5	6	V
V_O	Applied output voltage	$\bar{O}1 \sim \bar{O}12$: When output is OFF			30	V
I_O	Output current (per circuit)	All outputs ON simultaneously Duty cycle less than 20%			200	mA

ELECTRICAL CHARACTERISTICS ($T_a = +25^\circ\text{C}$, $V_{CC} = 5\text{V}$, unless otherwise noted)

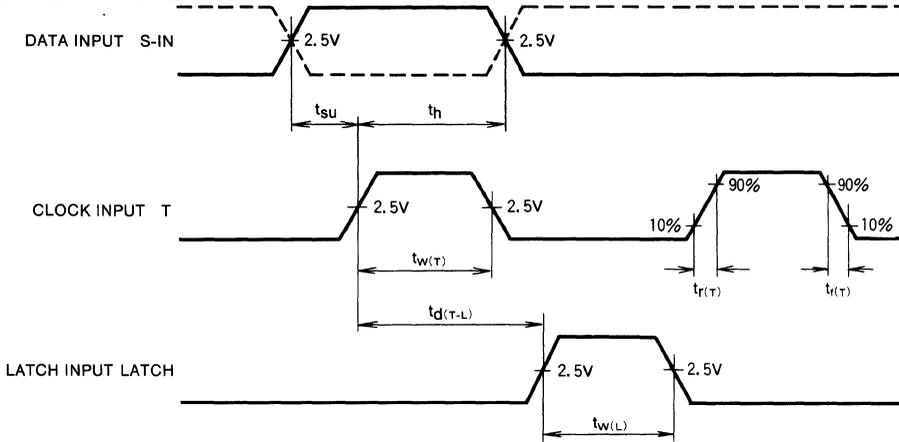
Symbol	Parameter	Pin	Conditions	Limits			Unit
				Min	Typ	Max	
V_{IH}	"H" input voltage	1, 2	$T_a = -20 \sim +75^\circ\text{C}$	$0.7V_{CC}$		V_{CC}	V
V_{IL}	"L" input voltage	6, 7		0		$0.3V_{CC}$	V
R_{IN}	input resistance			50		—	$k\Omega$
V_{OH}	"H" output voltage	5	$ I_O \leq 1\mu\text{A}$	4.9		—	V
V_{OL}	"L" output voltage			—		0.1	V
I_{OH}	"H" output current	5	$V_{OH} = 4.5\text{V}$	-100		—	μA
I_{OL}	"L" output current		$V_{OL} = 0.4\text{V}$	400		—	μA
V_{OL1}	"L" output voltage	8, 9	$I_{OL} = 100\text{mA}$	—		1.2	V
V_{OL2}		11~20	$I_{OL} = 200\text{mA}$	—		1.4	V
I_{OLK}	Output leakage current		$V_O = 30\text{V}$	—		50	μA
I_{CC1}	Supply current	4	Inputs free, all driver outputs OFF	—		10	μA
I_{CC2}			Driver output: 1 circuit ON	—		1.25	mA

BI-CMOS 12-BIT SERIAL-INPUT, LATCHED DRIVER

REQUIRED TIMING CONDITIONS ($T_a = -20 \sim +75^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$f_{(T)}$	Clock frequency	Input duty cycle 40~60%			2	MHz
$t_{w(T)}$	Clock pulse width		200			nS
$t_{w(L)}$	Latch pulse width		200			nS
t_{su}	Data setup time		100			nS
t_h	Data hold time		100			nS
$t_{d(T-L)}$	Clock-latch time		400			nS
$t_{r(T)}$	Clock pulse rise time				500	nS
$t_{f(T)}$	Clock pulse fall time				500	nS

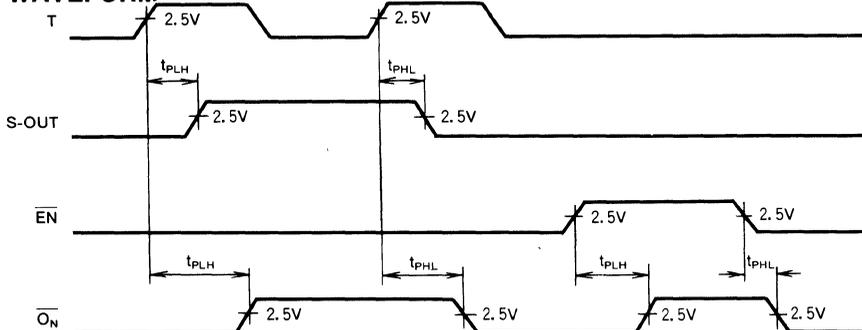
VOLTAGE WAVEFORM



SWITCHING CHARACTERISTICS ($T_a = +25^\circ\text{C}$, $V_{cc} = 5\text{V}$)

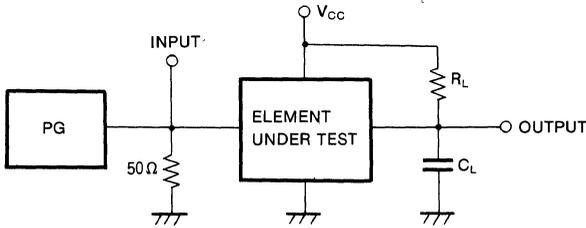
Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
t_{PLH}	"L-H" output propagation time (Input T to output S-OUT)	$V_{IH} = 5\text{V}$ $V_{IL} = 0\text{V}$ $R_L : S\text{-OUT} = \infty$ $R_L : \overline{O}_N = 100\Omega$ ($N=1 \sim 8$) $C_L = 15\text{pF}$ (Note 1)		(0.15)	0.3	μS
t_{PHL}	"H-L" output propagation time (Input T to output S-OUT)			(0.15)	0.3	μS
t_{PLH}	"L-H" output propagation time (Input T to output \overline{O}_N)			(2)	10	μS
t_{PHL}	"H-L" output propagation time (Input T to output \overline{O}_N)			(0.5)	2	μS
t_{PLH}	"L-H" output propagation time (Input EN to output \overline{O}_N)			(2)	10	μS
t_{PHL}	"H-L" output propagation time (Input EN to output \overline{O}_N)			(0.5)	2	μS

VOLTAGE WAVEFORM



BI-CMOS 12-BIT SERIAL-INPUT, LATCHED DRIVER

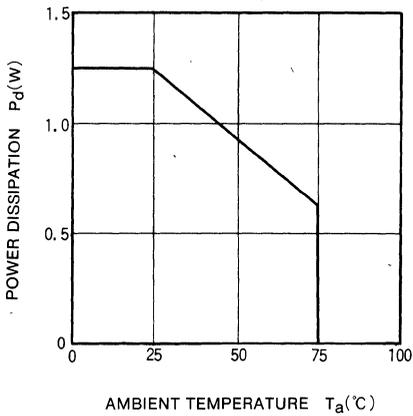
(Note 1) TEST CIRCUIT



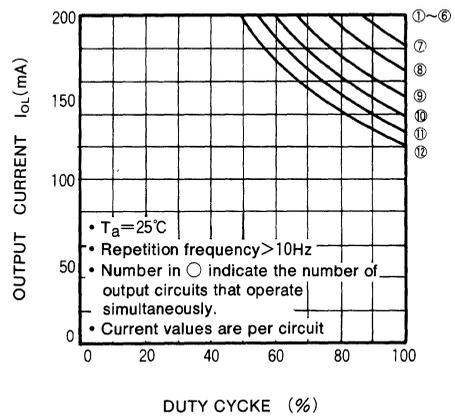
• Input waveform is taken as $t_r \leq 20\text{ns}$ and $t_f \leq 20\text{ns}$
 C_L includes wiring stray capacitance and probe input capacitance.

TYPICAL CHARACTERISTICS

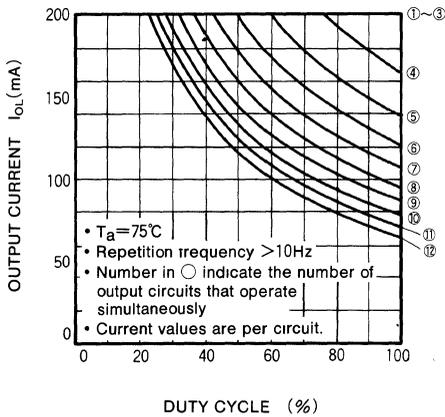
THERMAL DRAINING



DUTY CYCLE VS PERMISSIBLE OUTPUT CURRENT



DUTY CYCLE VS PERMISSIBLE OUTPUT CURRENT



M50800-XXXSP/M50802-XXXP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

DESCRIPTION

The M50800-XXXSP/M50802-XXXP, are single chip CMOS PARCOR speech synthesizers and contain a speech data, ROM, a clock generator, a D-A converter and a speaker drive circuit.

The differences among the M50800-XXXSP/M50802-XXXP are shown in the following table. The M50800-XXXSP will be discussed below.

Type	Number of words selection	Outline
M50800-XXXSP	64 (max)	28-pin, plastic DIP (LEAD PITCH 1.78mm)
M50802-XXXP	32 (max)	24-pin, plastic, DIP flat

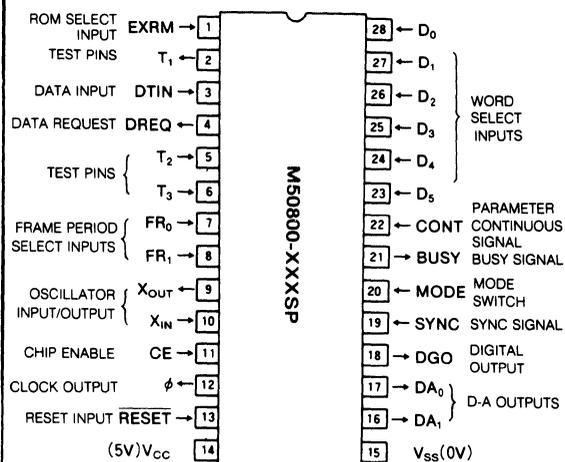
FEATURES

- Speech synthesis time 20 seconds (max)
- Number of words selection 64 words (max)
- Melody generation possible 70 seconds (max)
- Built-in speaker drive circuit
- Built-in D-A converter ± 7 -bits PWM method
- Low power dissipation due to CMOS process
 - Supply current in operating 2mA (max)
 - in stand-by 10 μ A (max)
- Built-in clock generator

APPLICATION

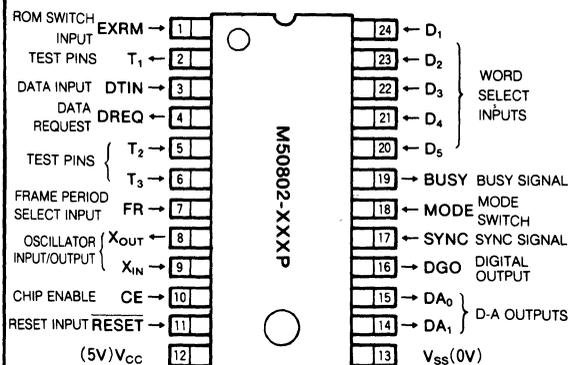
- Home appliances
- Teaching aid
- Watches, cameras, desk-top calculators
- Alarms

PIN CONFIGURATION (TOP VIEW)



Outline 28P4B

PIN CONFIGURATION (TOP VIEW)



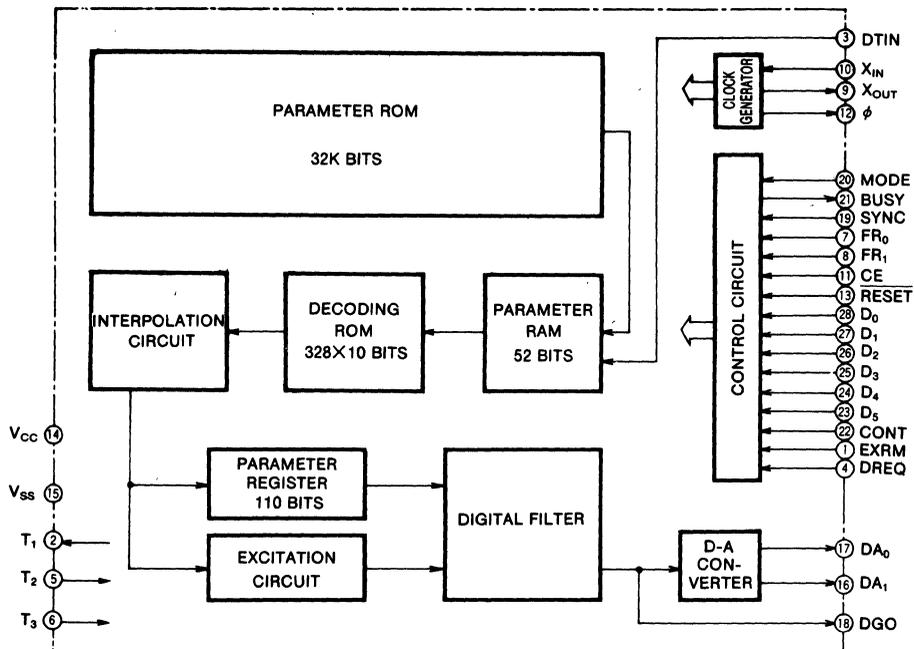
Outline 24P2W

M50800-XXXSP/M50802-XXXP

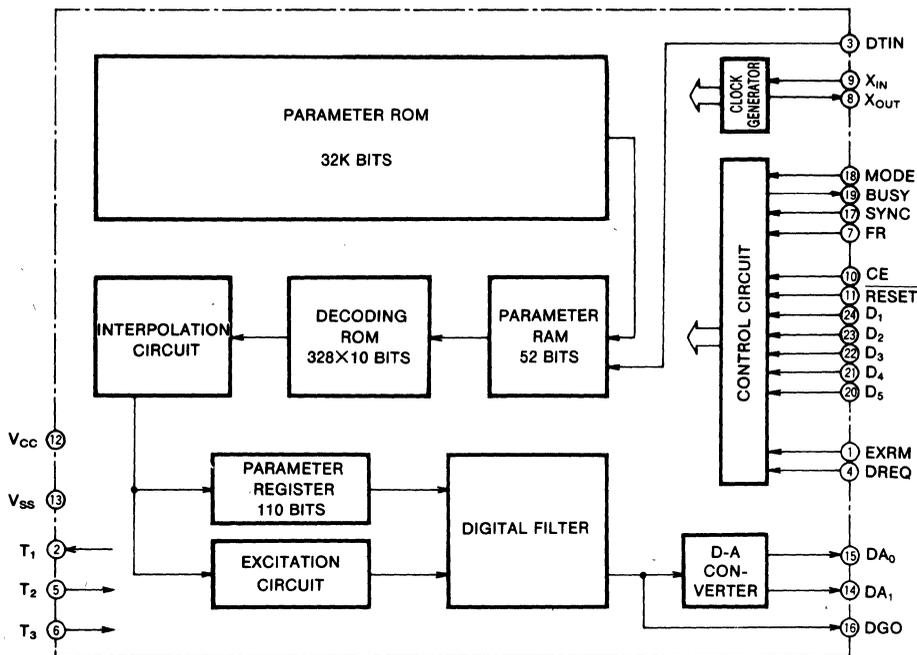
SINGLE-CHIP CMOS SPEECH SYNTHESIZER

BLOCK DIAGRAM

M50800-XXXSP



M50802-XXXP



M50800-XXXSP/M50802-XXXP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

PIN DESCRIPTION

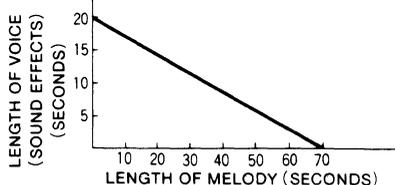
Pin	Name	Input or Output	Pull-down resistor Yes/No	Function
D ₀ ~D ₅	Word select input	Input	Option	In the keyboard input mode, when any of D ₀ ~D ₅ are set to high, speech synthesis begins. A maximum of 63 words can be designated. When using microcomputer control, a maximum of 64 word addresses can be set.
SYNC	Sync signal	Input		When using microcomputer control, the SYNC signal initiates D ₀ ~D ₅ address read, starting speech synthesis.
MODE	Mode switch	Input	Yes	Selects microcomputer or keyboard control modes
BUSY	Busy signal	Output		High during speech synthesis and RESET processes
FR ₀ FR ₁	Frame period select	Input		Used to select frame length (7.5, 10, 15 or 20ms)
CONT	Parameter continuous signal	Input	Yes	In the melody mode, tone production continues while with the CONT pin low and discontinues when it is set to high
RESET	Reset input	Input		Clears internal registers and mutes D/A converter at power on
CE	Chip enable	Input		When set to low disconnects chip power (stand-by mode), when set to high connects chip power (operation mode)
DA ₀ DA ₁	D/A output	Output		±7 bit PWM speech signal output
X _{IN} X _{OUT}	Oscillator	Input Output		Terminals for connecting ceramic resonator
φ	Clock output	Output		200kHz clock output signal
EXRM	Internal/external ROM select input	Input		ROM select input. When high, external ROM is used
DTIN	Data input	Input		Used for speech data input from external ROM
DREQ	Data request	Output		Data request signal for external ROM
DGO	Digital output	Output		The digital speech signal output
T ₁ ~T ₃	Test pins	Input/Output		During operation, T ₁ is open and set T ₂ and T ₃ are set low

BASIC FUNCTION BLOCKS

Parameter ROM

A 4096 word × 8 bits constructed mask ROM, it stores speech parameters.

There is enough memory storage for about 20 seconds of voice and sound effects and in the case of melody about 70 seconds. Voice, sound effects and melody can be stored together. In this case each length of time can be calculated by the following diagram.



Parameter RAM

This has 52bits capacity for temporary storage of one frame of speech parameters (amplitude, pitch, and K-parameters)

Decoding ROM

Decodes the speech parameters that had been coded by the parameter RAM. It has 328 words × 10 bits construction.

Interpolation Circuit

Provides a linear interpolation of K-parameters, pitch and amplitude every 2.5ms.

Exitation Circuit

Having both a pulse generator and a white noise generator, it generates voiced and unvoiced sounds.

M50800-XXXSP/M50802-XXXP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

Parameter Register

A register used to temporarily store data which has been linearly interpolated.

Digital Filter

It is a 14 bit 10 stage lattice digital filter.

Control Circuit

According to the external input, it controls LSI operation.

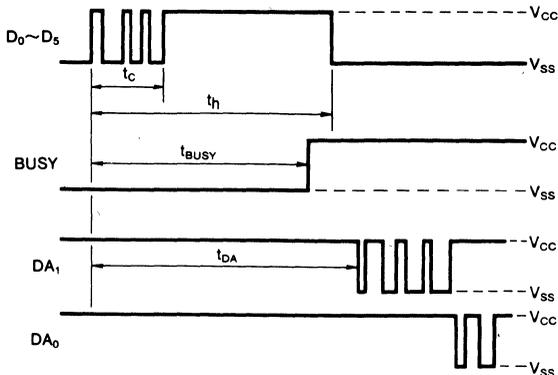
Mode Select

Depending on mode select signal MODE and internal/external ROM switch signal EXRM, three kinds of mode may be selected as shown in the following diagram.

Mode	MODE	EXRM	Use
Key input mode	H	L	Suitable for use when not editing phrases is necessary.
Microcomputer control mode	L	L	Suitable for use when the editing of phrases is necessary
External memory mode	L	H	For use when an external memory such as EPROM and RAM is used.

Pin Conditions of Key Input Mode

- (1) MODE pin Connect with V_{CC} .
- (2) SYNC pin Connect with V_{SS} .
- (3) T_2, T_3 pins
EXRM pin } Connect with V_{SS} .
DTIN pin }
- (4) Pins $D_0 \sim D_5$ When any of these pins are set (word selection inputs) high speech synthesis begins automatically.
- (5) Input/output timing chart

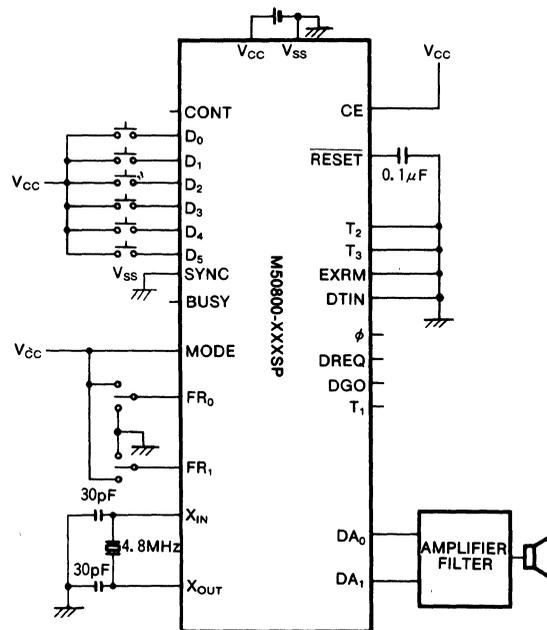


Timing Requirements

Symbol	Parameter	Min	Max	Unit
t_c	Key debounce protection time	5		ms
t_h	Data hold time	30		ms
t_{BUSY}	Busy output delay	25	50	ms
t_{DA}	Speech output delay	70		ms

$f_{osc} = 4.8\text{MHz}$

Application Example in the Key Input Mode



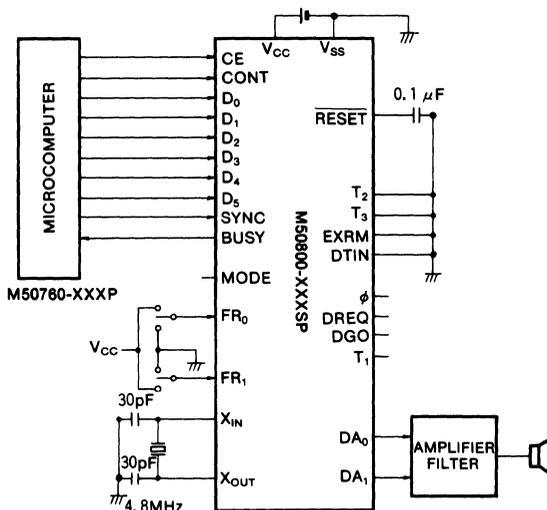
M50800-XXXSP/M50802-XXXP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

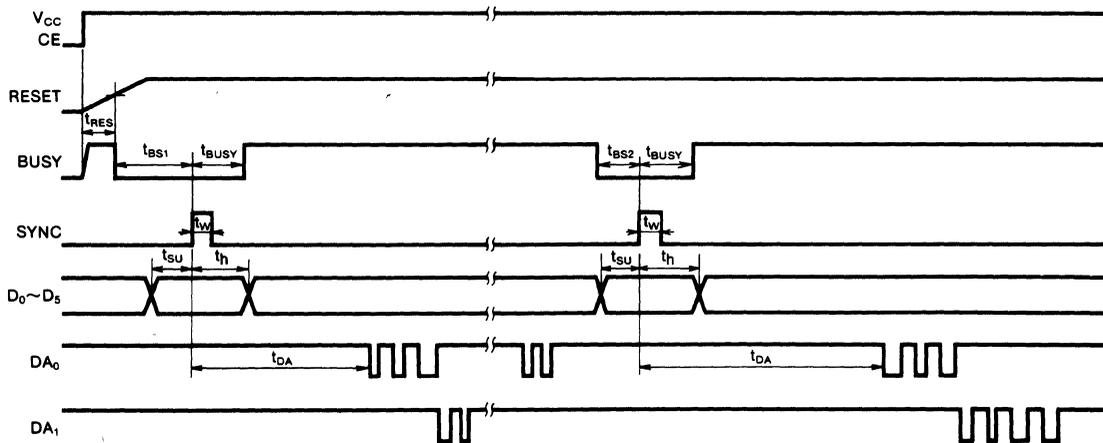
Pin Conditions of Microcomputer Control Mode

- (1) MODEOpen (Pull-down resistance)
- (2) T₂, T₃ pins
EXRM pin }Connect with V_{SS}.
DTIN pin }
- (3) D₀~D₅ pinsSelect 64 kinds of words.
- (4) SYNC pinData of D₀ ~ D₅ is latched and speech generation is commenced in the "High-level" at more than 15μs. If the power supply or CE input is switched on for the next speech output, it is necessary to delay the SYNC input a minimum of 400μs after the BUSY output goes low.

Application Example in the Microcomputer Control Mode



(5) Input/Output timing chart



Timing Requirements

Symbol	Parameter	Min	Max	Unit	Notes
t _{RES}	Reset time	5		ms	
t _{SU}	Data set up time	0		μs	
t _H	Data hold time	25		μs	
t _{w(SYNC)}	SYNC pulse width	15		μs	
t _{BS1}	SYNC pulse output delay	400		μs	After the power supply or CE input is switched.
t _{BS2}		0		μs	Except above condition.
t _{BUSY}	BUSY output delay	10	20	μs	
t _{DA}	Speech output delay	40	60	ms	

Note : Designate the single voice generation as a mask option, when the LSI is used in the microcomputer control mode.

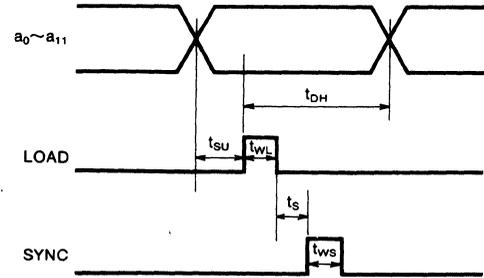
M50800-XXXSP/M50802-XXXP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

Pin Conditions of External Memory Mode

1. MODE pinOpen (microcomputer mode)
2. EXRM pinConnect with V_{CC} .
3. Word select pinOpen $D_0 \sim D_5$ and word select is performed by setting the first address to the external address counter.
4. SYNC pinGenerates speech in the "High-level".
5. DTIN pinReceives serial data from external ROM.
6. DREQ pinOutput read pulse of external ROM data.

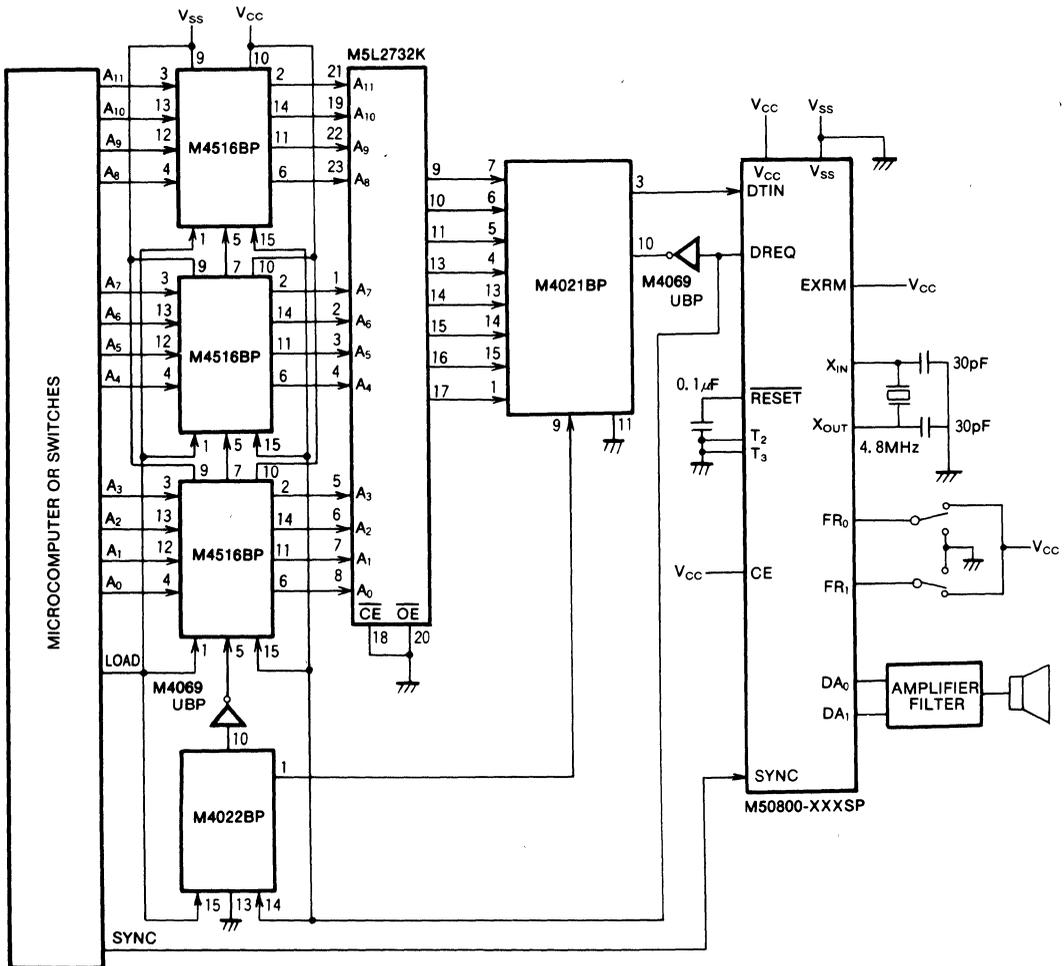
7. Timing chart



Timing Requirement

Symbol	Parameter	Min	Max	Unit
t_{SU}	Data set up time	1		μS
t_{DH}	Data hold time	2		μS
t_{WL}	LOAD pulse width	1		μS
t_s	LOAD SYNC pulse interval	0		μS
t_{ws}	SYNC pulse width	15		μS

Application Example in the External Memory Mode



M50800-XXXSP/M50802-XXXP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

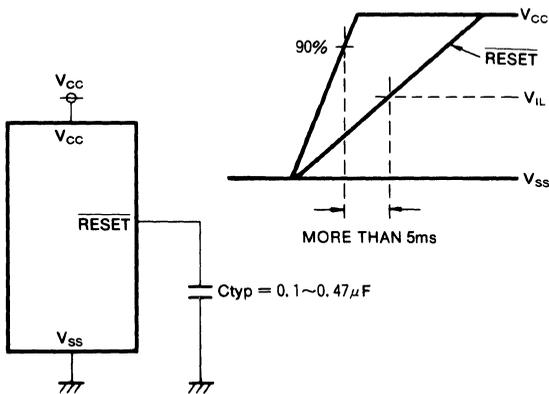
INPUT/OUTPUT PINS

Chip Enable Pin (CE)

CE	Function
L	When input is low-level, internal power supply is turned off and device enters stand-by condition. Under this stand-by mode, $D_0 \sim D_5$ are high impedance, and BUSY, DA_0 , and DA_1 are high.
H	When input is high-level internal power supply is turned on. At this time a reset signal is automatically generated and device enters power-on-initialization sequence.

Reset Pin (RESET)

The reset pin must become low-level in more than 5ms after power has been applied and the supply voltage has reached 90%.

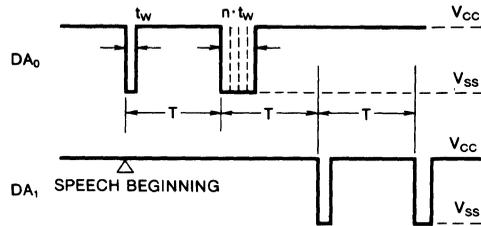


Busy Pin (BUSY)

The output of this pin becomes high-level during each of following. Voice generation, power-on-initialization and stand-by. If the BUSY signal is high-level, the voice generation start signal has no effect.

D-A Output Pin (DA_0 , DA_1)

D-A output is PWM (pulse width modulation) and outputted from pin DA_0 when positive and pin DA_1 when negative.



T : SAMPLING PERIOD $100\mu s$ (Typ)

t_w : MINIMUM PULSE WIDTH $830ns$ (Typ) $f_{osc} = 4.8MHz$

Frame Period Select (FR_0 , FR_1 : M50800-XXXSP) (FR : M50802-XXXP)

M50800-XXXSP

FR_0	FR_1	Frame length	Bit rate b/s (Note 1)
H	H	20ms	2650 (Max.)
L	H	15ms	3533 (Max.)
L	L	10ms	5300 (Max.)
H	L	7.5ms	7067 (Max.)

M50802-XXXP

FR	Frame length	Bit rate b/s (Note 1)
H	20ms	2650 (Max.)
L	10ms	5300 (Max.)

Note 1 : Where $f_{osc} = 4.8MHz$

The standard frame length is 20ms.

Speech speed depends on frame length. For example when frame length is set to 15ms in analysis and frame period select is set to 20ms in synthesis, speech speed is 33% slow. When frame period select is set to 10ms, 33% fast.

M50800-XXXSP/M50802-XXXP**SINGLE-CHIP CMOS SPEECH SYNTHESIZER****Clock Generator**

The device contains a clock generator which can generate a clock signal when a ceramic resonator is connected externally between the clock output and input pins (X_{IN} , X_{OUT}).

When a clock signal from an external source is being used, connect the source of the clock oscillator to pin X_{IN} and open pin X_{OUT} .

Circuit examples are given below.



Externally connected ceramic resonator circuit

External clock input circuit

Mask Option

The following mask options are available, specifiable at the time of ordering maskings.

- (1) Continuous/Single voice generation selecting specification. This option is available when the LSI is used only in key input mode. The single voice generation as a mask option must be designated when the LSI is used in the microcomputer control mode.

Continuous voice generation :

Voice generation continues as long as $D_0 \sim D_5$ is kept continuously high-level in the key input mode.

Single voice generation :

Voice generation occurs only once even if $D_0 \sim D_5$ is kept continuously high-level in the key input mode or SYNC is kept continuously high-level in the microcomputer control mode. To generate the voice again it is necessary to return either the $D_0 \sim D_5$ or the SYNC to low-level one time and then to high-level again.

- (2) Specifications concerning the inclusion or exclusion of pull-down resistors for the word select input pins $D_0 \sim D_5$

M50800-XXXSP/M50802-XXXP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

ELECTRICAL CHARACTERISTICS OF THE M50800-XXXSP/M50802-XXXP ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Conditions	Ratings	Unit
V_{CC}	Supply voltage	With respect to the V_{SS} pin	-0.3~7	V
V_I	Input voltage		-0.3~7	V
P_d	Maximum power dissipation	$T_a = 25^\circ\text{C}$	150	mW
T_{opr}	Operating temperature		-10~70	$^\circ\text{C}$
T_{stg}	Storage temperature		-40~125	$^\circ\text{C}$

RECOMMENDED OPERATING CONDITIONS ($T_a = -10\sim 70^\circ\text{C}$, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V_{CC}	Supply voltage	2.4	5	5.5	V
V_{SS}	Supply voltage	0	0	0	V
f_{osc}	Oscillation frequency		4.8		MHz

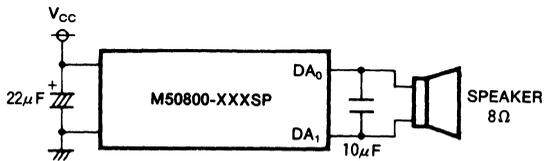
ELECTRICAL CHARACTERISTICS ($T_a = -10\sim 70^\circ\text{C}$, $f = 4.75\sim 4.85\text{MHz}$, $V_{CC} = 4.5\sim 5.5\text{V}$, unless otherwise noted)

Symbol	Parameter	Conditions	Limits			Unit
			Min	Typ	Max	
V_{OH}	High-level output voltage	$I_{OH} = 0.1\text{mA}$	$V_{CC}-0.8$			V
V_{OL}	Low-level output voltage	$I_{OL} = 0.4\text{mA}$			0.8	V
I_{OH}	High-level output current	$DA_0, DA_1, V_{OH} = V_{CC}-0.4\text{V}$	1			mA
I_{OL}	Low-level output current	$DA_0, DA_1, V_{OL} = 0.4\text{V}$	1.6			mA
I_{DA}	D/A output current	Resistor between DA_0 and DA_1 10Ω at $V_{CC} = 5\text{V}$		40		mA
I_{OC}	Supply current (in operation)	Unloaded input/output			2	mA
I_{SC}	Supply current (stand-by)	Unloaded input/output			10	μA
I_{IL}	Input leak current	$V_I = 0\sim V_{CC}$			± 1	μA
R_I	Pull-down resistor	$V_I = V_{CC}$		100		k Ω
V_{IH}	High-level input voltage		$V_{CC}-1$		V_{CC}	V
V_{IL}	Low-level input voltage		0		1	V

Example of Speaker Driving Circuits

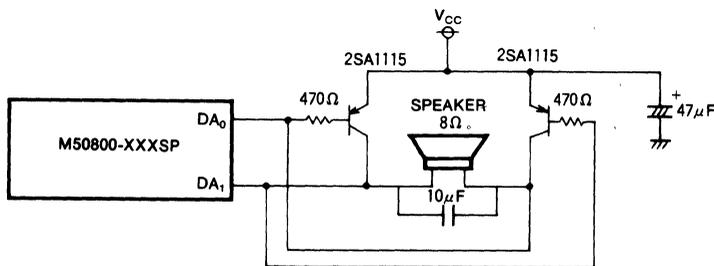
(a) Direct drive

Maximum output power=10mW typ. ($V_{CC}=5V$)



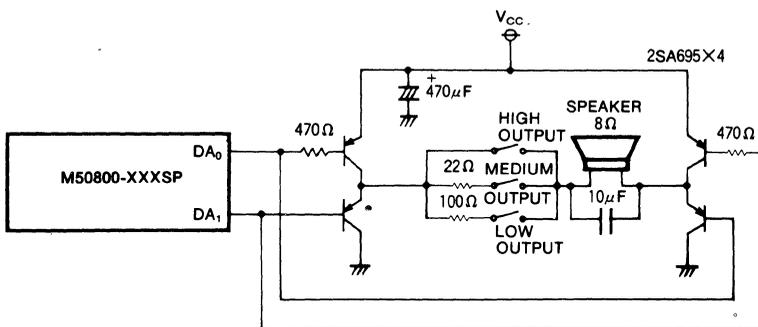
(b) Using an amplifier

Maximum output power=20mW typ. ($V_{CC}=5V$)



Maximum output power=500mW typ. ($V_{CC}=5V$)

With volume control

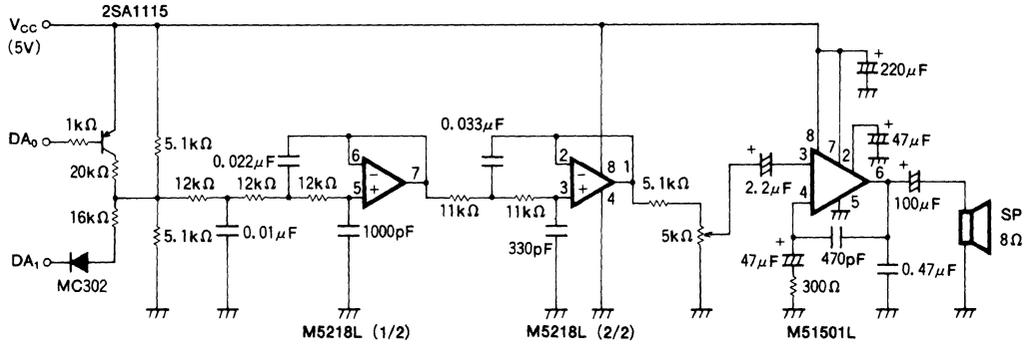


M50800-XXXSP/M50802-XXXP

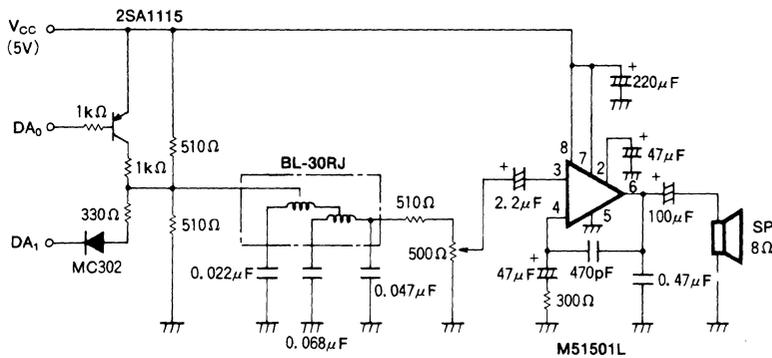
SINGLE-CHIP CMOS SPEECH SYNTHESIZER

(c) Filter amplifier circuit

- (1) Operational amplifier used as a filter amplifier circuit
 Maximum output power=200mW typ. ($V_{CC}=5V$)



- (2) L device used as a filter amplifier circuit
 Maximum output power=200mW typ. ($V_{CC}=5V$)

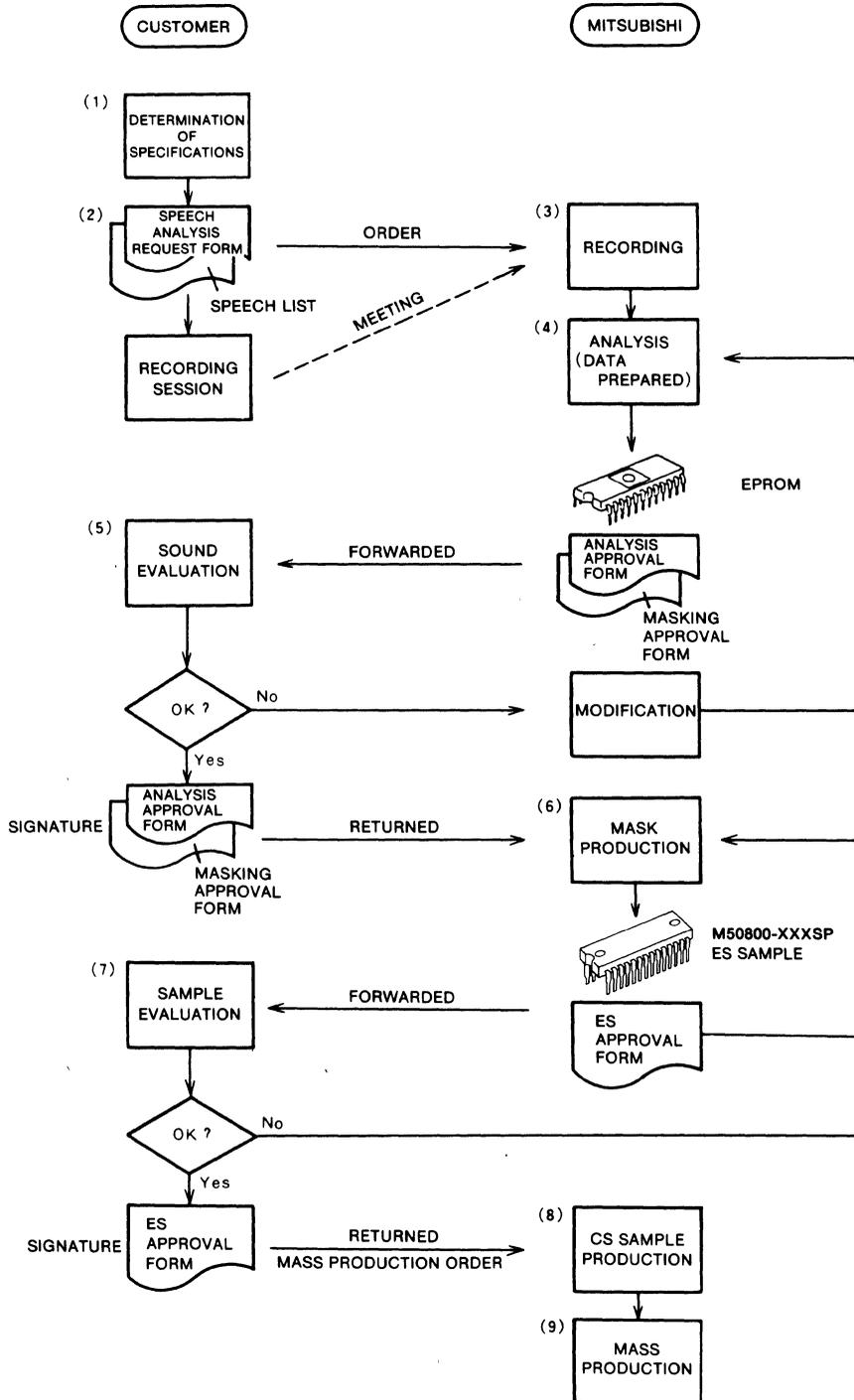


BL-30RJ : Made by KOURINGIKEN CORPORATION

**METHOD OF ORDERING SPEECH
ANALYSIS AND MASK ROM****Steps Taken in Ordering**

- (1) **Decisions concerning specifications**
Decide the specifications concerning the content of the speech, the language, the speaker, etc.
- (2) **Ordering**
When ordering, ensure that the necessary information has been entered in both the speech Analysis Request Form and Speech List. Be sure to also include the musical score if a melody is necessary.
- (3) **Recording**
A recording of the original speech is made in a studio using an experienced narrator. Attendance is permitted for either confirmation or giving speech instructions.
- (4) **Analysis**
Speech parameters of the voice are analyzed through use of a voice analysis system centered on a MELCOM 70/40 minicomputer in the Mitsubishi Speech Center. The data is written on a EPROM, and along with the Analysis Approval Form and Masking Approval Form is forwarded the customer.
- (5) **Sound evaluation**
The delivered EPROM is inserted in either Mitsubishi's voice evaluation board or a similar device and the quality of the synthesized speech is confirmed. If the quality is found to be acceptable, the Analysis Approval Form and the Masking Approval Form are to be signed and returned to Mitsubishi.
- (6) **Mask production**
A MASK ROM is produced from the EPROM data. An ES sample for the confirmation of electrical characteristics is sent along with an ES Approval Form to the customer.
- (7) **Sample evaluation**
If results of ES sample evaluation prove to be acceptable the ES Approval Form is to be signed and returned to Mitsubishi.
- (8) **CS sample production**
This is a sample having the same characteristics as the mass produced device.
- (9) **Mass production**
After production of the CS sample mass production is begun.

ORDERING FLOW CHART



M50800-XXXSP/M50802-XXXP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

**MITSUBISHI SPEECH SYNTHESIZER LSI
M50800-XXXSP
M50802-XXXP
SPEECH ANALYSIS REQUEST FORM**

Customer _____	Signature _____	
Company name _____	Prepared _____	
Company address _____	Tel _____	Approved _____
Company contact _____	Data _____	

1. Specifications

Recording

(necessary/unnecessary)

_____ Language _____ Sex _____

Analysis

(necessary/unnecessary)

_____ Sampling frequency _____ KHz (10KHzTYP)
 _____ Frame period _____ ms (20msTYP)
 _____ Length of speech _____ seconds
 _____ Length of melody _____ seconds
 _____ Voice evaluation board (necessary/unnecessary)

Mask

(necessary/unnecessary)

2. Schedule

Recording → Analysis completion → Analysis and masking approval → ES → CS → MP
 (/) (/) (/) (/) (/) (/)

Disired date
(month/day)

EPROM

1 set

_____ unit(s) _____ unit(s) _____ unit(s)
/month

Lot _____ unit(s)

Mitsubishi Entry Column

Parameter	Type	Order number	Amount	Cost	Order date	Completion date	Transfer of money
Recording	VOICE-SOFT	00-38-					
Analysis	VOICE-SOFT	00-38-					
Voice evaluation board	VOICE-SOFT	00-38-					
Mask	VOICE-MASK						

Notes

M50805-XXXP,FP/M50806-XXXP,FP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

DESCRIPTION

The M50805-XXXP, FP/M50806-XXXP, FP, are single chip C-MOS PARCOR speech synthesizers and contain a speech data, ROM, a clock generator, a D-A converter and a speaker drive circuit.

The differences among the M50805-XXXP, FP/M50806-XXXP, FP are shown in the following table. The M50805-XXXP will be discussed below.

Type	Operating supply voltage range	Outline
M50805-XXXP	$V_{CC}=3.0\sim 5.5V$	22 pin, plastic DIP
M50805-XXXFP	$V_{CC}=3.0\sim 5.5V$	24 pin, plastic DIP
M50806-XXXP	$V_{CC}=2.1\sim 3.5V$	22 pin, plastic DIP
M50806-XXXFP	$V_{CC}=2.1\sim 3.5V$	24 pin, plastic DIP

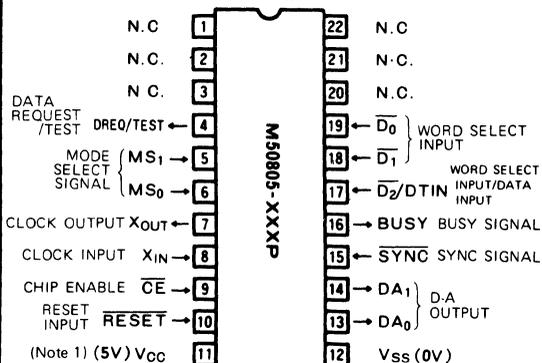
FEATURES

- Speech synthesis time 5 seconds (max)
- Number of words selection 8 (max)
- Melody generation possible 20 seconds (max)
- Automatic power down function
- Built-in speaker drive circuit
- Built-in D-A converter ± 6 bits PWM method
- Low power dissipation due to C-MOS process
 - Supply current in operating . . . 2mA (max)
 - in standby 10 μ A (max)
- Built-in clock generator

APPLICATION

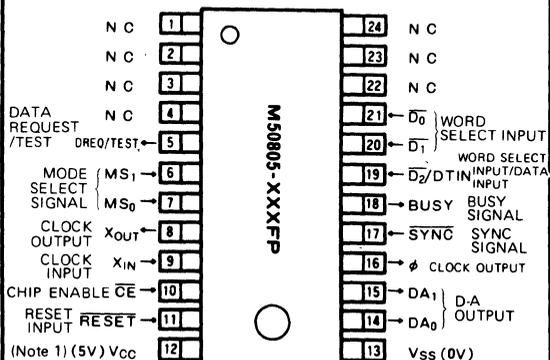
- Home appliances
- Toys
- Watches, cameras, desk-top calculators
- Alarms

PIN CONFIGURATION (TOP VIEW)



Outline 22P4 (M50805-XXXP) (M50806-XXXP)

Note 1 For M50806-XXXP, FP the value is 3V



Outline 24P2W (M50805-XXXFP) (M50806-XXXFP)

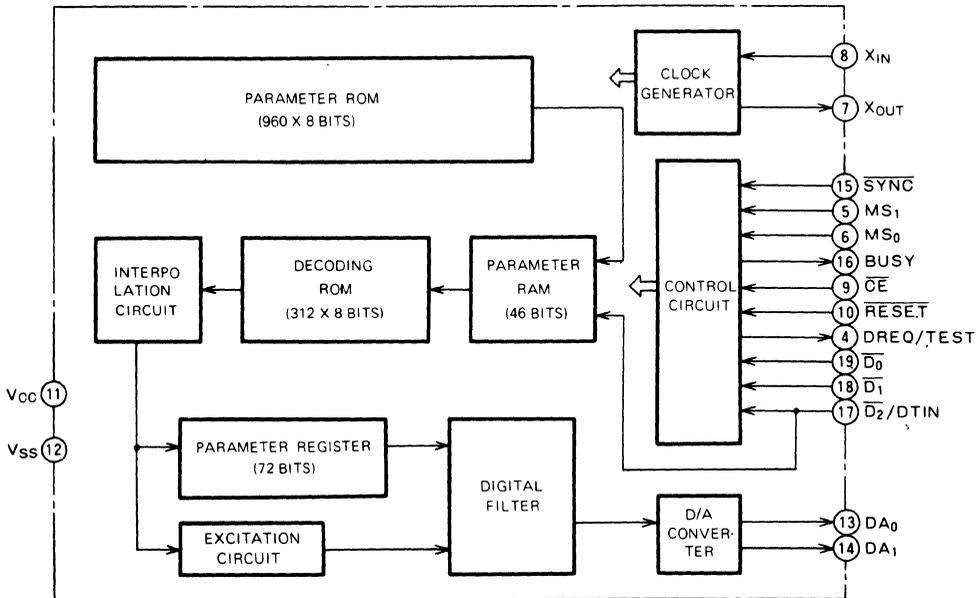
NC NO CONNECTION

M50805-XXXP,FP/M50806-XXXP,FP

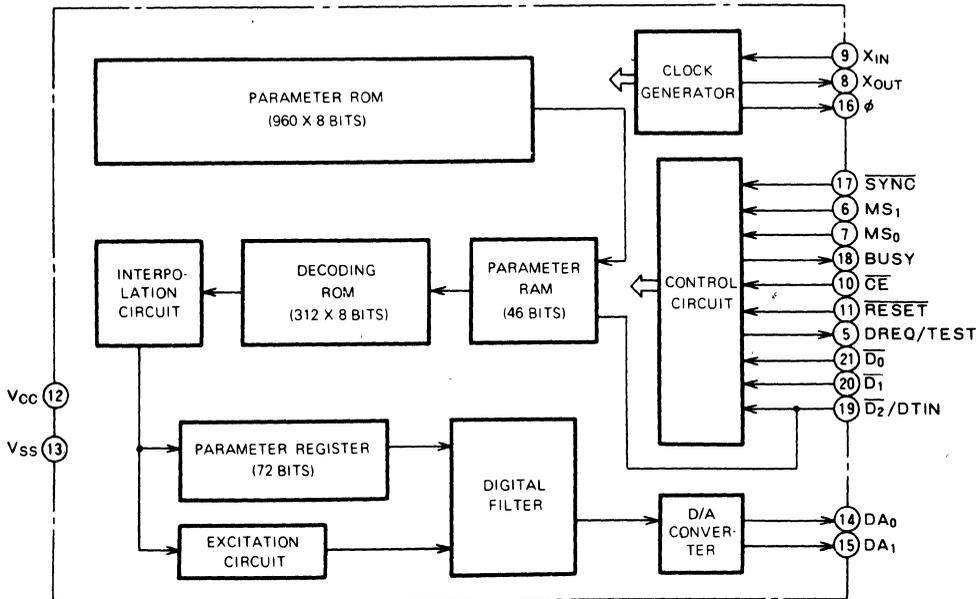
SINGLE-CHIP CMOS SPEECH SYNTHESIZER

BLOCK DIAGRAM

M50805-XXXP/M50806-XXXP



M50805-XXXFP/M50806-XXXFP



M50805-XXXP,FP/M50806-XXXP,FP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

PIN DESCRIPTION

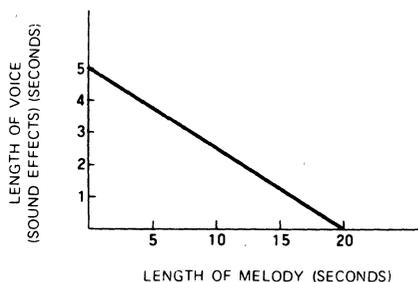
Pin	Name	Input or output	Pull-up resistor Yes/No	Function
V _{CC}	Supply voltage input	Input		Positive voltage supply pin
V _{SS}	Supply voltage input	Input		Ground pin
MS ₀ , MS ₁	Mode select signal	Input		Depending on the MS ₀ , MS ₁ signal there are 4 modes that may be selected, key input mode, microcomputer control mode, external memory mode and test mode
$\overline{D_0}, \overline{D_1}$	Word select input	Input	Option	When in the key input mode, voice generation begins whenever $\overline{D_0}$, $\overline{D_1}$, or $\overline{D_2}$ becomes low-state A maximum of 7 words may be designated
$\overline{D_2}/\overline{DTIN}$	Word select input/data input			When in the microcomputer control mode, a maximum of addresses for 8 words may be designated with $\overline{D_0}$, $\overline{D_1}$, $\overline{D_2}$ When in the external memory mode, speech parameters are inputted via DTIN
\overline{SYNC}	Synchronous signal	Input	Option	When in microcomputer control mode, the SYNC signal initiates $\overline{D_0} \sim \overline{D_2}$ address read, starting voice generation When in external memory mode, voice generation is started
BUSY	Busy signal	Output		Output signal pin for indicating that voice generation is in progress
\overline{CE}	Chip enable	Input		When the pin CE is high-level, the internal power supply is off and, when low-level, the internal power supply is on
DA ₀ , DA ₁	D-A output	Output		±6 bit PWM output
X _{IN}	Clock input	Input		These pins are for connection to a ceramic resonator for the clock generator, or they can be connected to an external resistance R for CR clock generation. When using an external clock input, connect the oscillator source to X _{IN} pin and open X _{OUT} .
X _{OUT}	Clock output	Output		
DREQ/TEST	Data request/test	Output		This is the data request signal output pin for external memory. During test mode this is the test output pin.
\overline{RESET}	Reset input	Input	Yes	When power input is first applied the internal registers are cleared and D-A output is muted
ϕ (M50805-XXXP) ϕ (M50806-XXXP)	Clock output	Output		160kHz clock output pin

BASIC FUNCTION BLOCKS

Parameter ROM

A 960 word x 8 bit constructed mask ROM, it stores speech parameters.

There is enough memory storage for about 5 seconds of voice and sound effects and in the case of melody about 20 seconds. Voice, sound effects and melody can be stored together. In this case each length of time can be calculated by the following diagram.



Parameter RAM

This has 46 bit capacity for temporary storage of one frame of speech parameters (amplitude, pitch, and K-parameters)

Decoding ROM

Decodes the speech parameters that had been coded by the parameter RAM. It has 312 word x 8 bit construction.

Interpolation Circuit

Provides a linear interpolation of K-parameters, pitch and amplitude every 2.5ms.

Excitation Circuit

Having both a pulse generator and a white noise generator, it generates voiced and unvoiced sounds.

Parameter Register

A register used to temporarily store data which has been linearly interpolated.

Digital Filter

It is a 12 bit 8 stage lattice digital filter.

M50805-XXXP,FP/M50806-XXXP,FP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

Control Circuit

According to the external input, it controls LSI operation.

Mode Select Signal MS₀, MS₁

Depending on MS₀, and MS₁, four kinds of mode may be selected as shown in the following diagram.

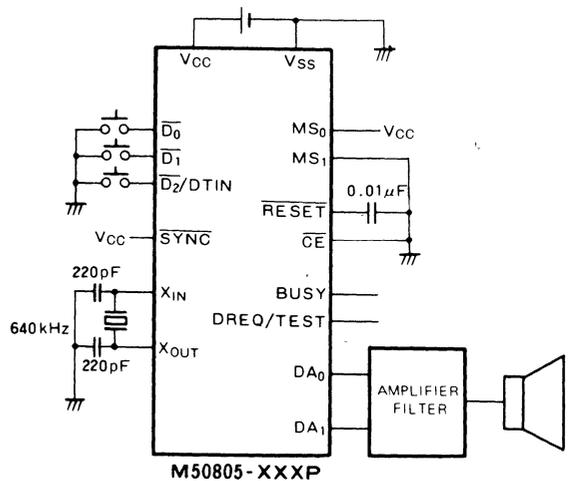
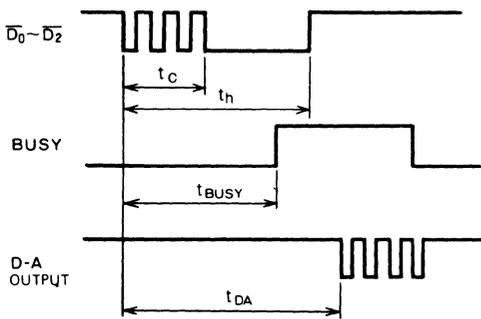
Mode	MS ₀	MS ₁	Use
Key input mode	H	L	Suitable for use when not editing phrases is necessary
Microcomputer control mode	L	L	Suitable for use when the editing of phrases is necessary
External memory mode	L	H	For use when an external memory such as EPROM and RAM is used
Test mode	H	H	For test purposes

(1) Key Input Mode

Pin conditions

- Mode select signal pins MS₀ is connected to V_{CC} and MS₁ to V_{SS}.
- $\overline{\text{SYNC}}$ pin connects to V_{CC}.
- $\overline{\text{CE}}$ pin connects to V_{SS}.
- $\overline{\text{D}}_0 \sim \overline{\text{D}}_2$ pins when any of these pins is low-level, speech generation begins automatically. Seven kinds of phases may be selected.

Input/output timing chart



Symbol	Parameter	Min	Typ	Max	Unit
t _C	Key debounce protection time	5			ms
t _h	Data hold time	50			ms
t _{BUSY}	BUSY output delay	50			ms
t _{DA}	Speech output delay	100			ms

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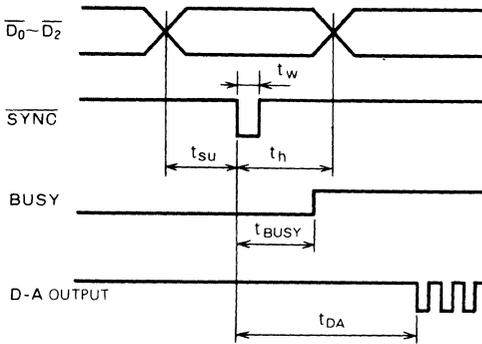
SINGLE-CHIP CMOS SPEECH SYNTHESIZER

(2) Microcomputer Control Mode

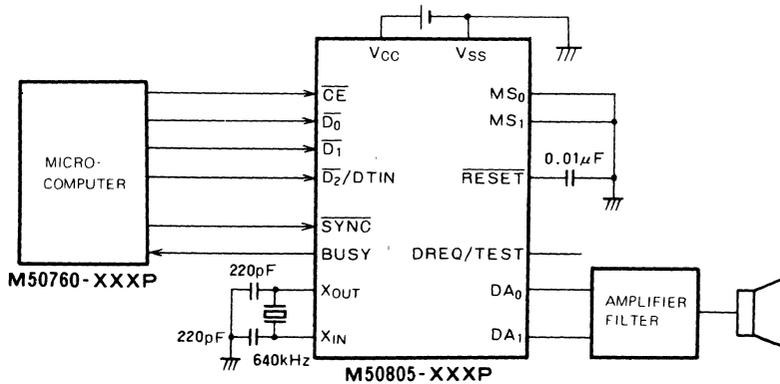
Pin conditions

- Mode select signal pins . . . Both MS₀ and MS₁ are connected to V_{SS}.
- SYNC signal When it becomes low-level for 18μs, data of both D₀ ~ D₂ is latched and speech generation begins.
- D₀ ~ D₂ pins Eight kinds of phrases may be selected.

Input output Timing chart



Symbol	Parameter	Min	Typ	Max	Unit
t_{su}	Data setup time	0			μs
t_h	Data hold time	25			μs
t_w (SYNC)	SYNC pulse width	18			μs
t_{BUSY}	BUSY output delay	12		25	μs
t_{DA}	Speech output delay	40		60	ms



M50805-XXXP,FP/M50806-XXXP,FP

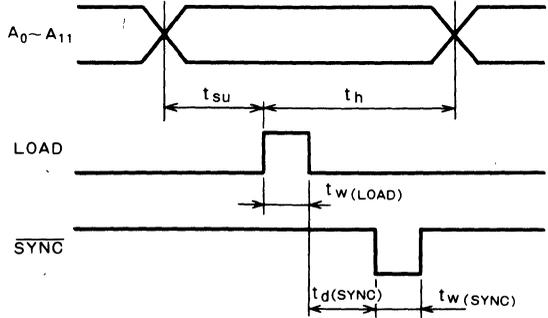
SINGLE-CHIP CMOS SPEECH SYNTHESIZER

(3) External Memory Mode

Pin conditions

- Mode select signal pins . . . MS₀ is connected to V_{SS} and MS₁ to V_{CC}.
- SYNC pin When it becomes low-level for more than 18μs, speech generation begins.
- D₂/DTIN pin Receives serial data from external ROM.
- DREQ/TEST pin Outputs the read pulse for external ROM data
- D₀, D₁ pins Connected to V_{CC}.

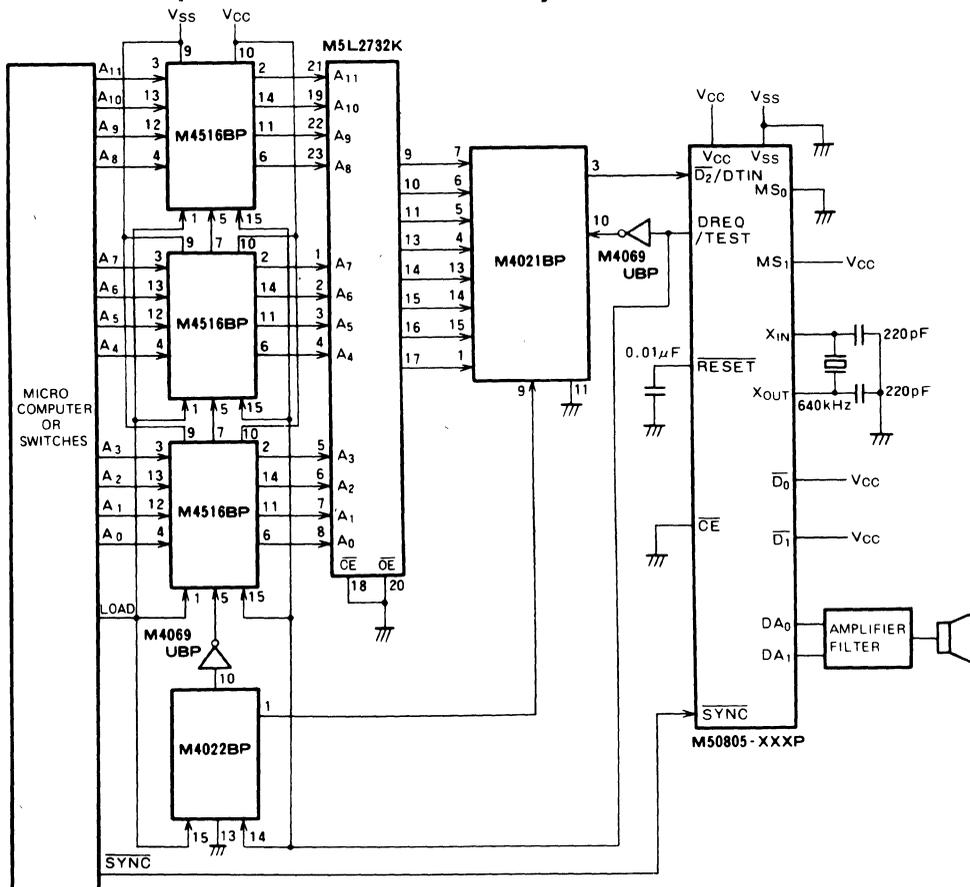
Input timing chart



Timing Requirement

Symbol	Parameter	Min	Typ	Max	Unit
t _{su}	Data set up time	1			μs
t _h	Data hold time	2			μs
t _{w(LOAD)}	LOAD pulse width	1			μs
t _{d(SYNC)}	SYNC delay time	0			μs
t _{w(SYNC)}	SYNC pulse width	18			μs

Application Example in the External Memory Mode



SINGLE-CHIP CMOS SPEECH SYNTHESIZER

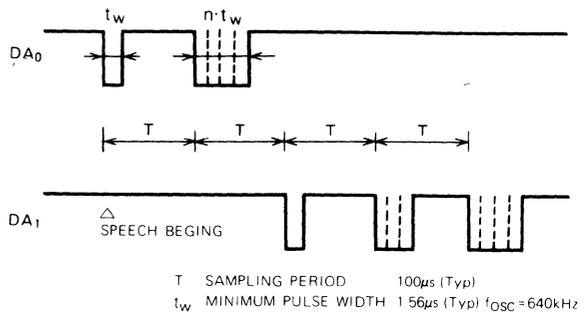
Input/Output Pins

Chip enable pin (CE)

CE	Mode	Function
H	All modes	When input is high-level, internal power supply is turned off and device enters standby condition. In standby, both DA ₀ and DA ₁ become high-level.
L	Microcomputer control mode/external memory mode	When input is low-level, internal power supply is turned on. At this time a reset signal is automatically generated and device enters power-on-initialization sequence.
	Key input mode	The internal power supply is turned on after receiving key input. At this time a reset signal is automatically generated and device enters power-on-initialization sequence. After voice generation has ceased, automatic power down circuit operates and device enters key input waiting state. While in the waiting state, BUSY becomes low-level and both DA ₀ and DA ₁ become high-level.

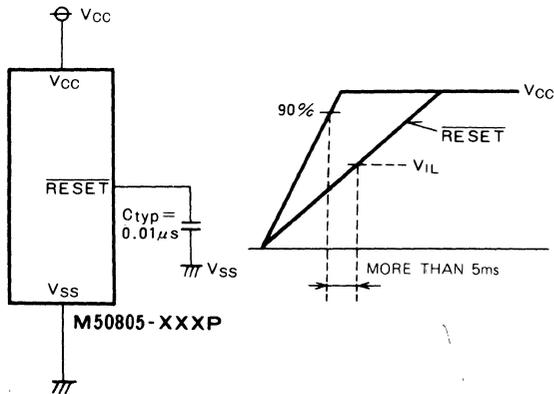
D-A Output Circuit (DA₀, DA₁)

D-A output is PWM (pulse width modulation) and outputted from pin DA₀ when positive and pin DA₁ when negative.



Reset Pin (RESET)

The reset pin must become low-level in more than 5ms after power has been applied and the supply voltage has reached 90%.



Busy Pin (BUSY)

Mode	BUSY output
Microcomputer control mode External memory mode	The output of this pin becomes high-level during each of the following: Voice generation, power-on-initialization and standby. It is necessary to ensure that the BUSY signal is low-level prior to generating the voice start signal. If the BUSY signal is high-level, the voice generation start signal has no effect.
Key input mode	The output of this pin becomes high-level during voice generation and low-level during power-on-initialization and while in the waiting condition. It is necessary to ensure that the BUSY signal is low-level prior to generating the voice start signal. If the BUSY signal is high-state, the voice start signal has no effect.

M50805-XXXP,FP/M50806-XXXP,FP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

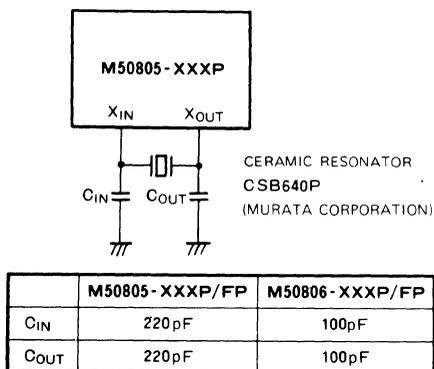
Clock Generator

The device contains a clock generator which can generate a clock signal when either a ceramic resonator or resistor is connected externally between the clock output and input pins (X_{IN} , X_{OUT}).

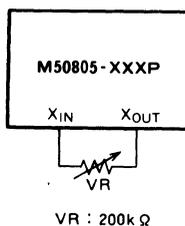
When a clock signal from an external source is being used, connect the source of the clock oscillator to pin X_{IN} and open pin X_{OUT} . Designate whether or not a ceramic resonator or a resistor will be used in the clock oscillator circuit as a mask option.

Circuit examples are given below.

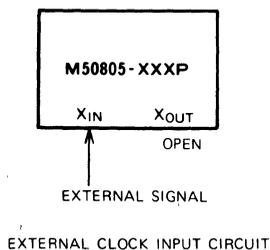
Externally connected ceramic resonator circuit



Externally connected resistor



External clock input circuit



MASK OPTION

The following mask options are available, specifiable at the time of ordering maskings.

(1) Continuous/single voice generation selecting specification

Continuous voice generation:

Voice generation continues as long as $\overline{D_0} \sim \overline{D_2}$ is kept continuously low-level in the key input mode or \overline{SYNC} is kept continuously low-level in the microcomputer control mode.

Single voice generation:

Voice generation occurs only once when the above conditions are met. To generate the voice again it is necessary to return either the $\overline{D_0} \sim \overline{D_2}$ or the \overline{SYNC} to high-level one time and then to low-level again.

(2) Specifications concerning the inclusion or exclusion of pull-up resistors for the word select input pins $\overline{D_0}$, $\overline{D_1}$, $\overline{D_2}$, and the synchronous signal \overline{SYNC} .

(3) Specifications concerning whether or not a ceramic resonator or a resistor will be used in the oscillator circuit.

M50805-XXXP,FP/M50806-XXXP,FP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

ELECTRICAL CHARACTERISTICS OF THE M50805-XXXP/M50805-XXXFP

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Conditions	Ratings	Unit
V _{CC}	Supply voltage	With respect to the V _{SS} Pin	-0.3~7	V
V _I	Input voltage		-0.3~7	V
P _d	Maximum power dissipation	T _a = 25°C	150	mW
T _{opr}	Operating temperature		-10~70	°C
T _{stg}	Storage temperature		-40~125	°C

RECOMMENDED OPERATING CONDITIONS (T_a = -10~70°C, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V _{CC}	Supply voltage	3	5	5.5	V
V _{SS}	Supply voltage	0	0	0	V
f _{OSC}	Oscillation frequency		640		kHz

ELECTRICAL CHARACTERISTICS (T_a = -10~70°C, f_{OSC} = 640 ± 20kHz, V_{CC} = 3~5.5V, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{OH}	High level output voltage BUSY, DREQ/TEST	I _{OH} = 0.1mA	V _{CC} -0.8			V
V _{OL}	Low level output voltage BUSY, DREQ/TEST	I _{OL} = 0.4mA			0.8	V
I _{OH}	High-level output current DA ₀ , DA ₁	V _{OH} = V _{CC} - 0.4V	1			mA
I _{OL}	Low level output current DA ₀ , DA ₁	V _{OL} = 0.4V	1.6			mA
I _{DA}	D/A output current	Resistor between DA ₀ and DA ₁ 10Ω at V _{CC} = 5V		45		mA
I _{OC}	Supply current (in operation)	Unloaded input/output			2	mA
I _{SC}	Supply current (standby)	Unloaded input/output			10	μA
I _{IL}	Input leak current	V _I = 0 ~ V _{CC}			±1	μA
R _I	Pull-up resistor $\overline{D_0} \sim \overline{D_2}$	V _I = V _{SS} , V _{CC} = 5V		180		kΩ
R _I	Pull-up resistor \overline{SYNC}	V _I = V _{SS} , V _{CC} = 5V		300		kΩ
V _{IH}	High-level input voltage	V _{CC} = 4 ~ 5.5V	V _{CC} -1		V _{CC}	V
V _{IL}	Low level input voltage	V _{CC} = 4 ~ 5.5V	0		1	V

M50805-XXXP,FP/M50806-XXXP,FP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

ELECTRICAL CHARACTERISTICS OF THE M50806-XXXP/M50806-XXXP

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Conditions	Limits	Unit
V _{CC}	Supply voltage	With respect to the V _{SS} pin	-0.3~7	V
V _I	Input voltage			
P _d	Maximum power dissipation	T _a = 25°C	150	mW
T _{opr}	Operating temperature		-10~70	°C
T _{stg}	Storage temperature		-40~125	°C

RECOMMENDED OPERATING CONDITIONS (T_a = -10~70°C, unless otherwise noted)

Symbol	Parameter	Limits			Unit
		Min	Typ	Max	
V _{CC}	Supply voltage	2.1	3	3.5	V
V _{SS}	Supply voltage	0	0	0	V
f _{OSC}	Oscillation frequency		640		kHz

ELECTRICAL CHARACTERISTICS (T_a = -10~70°C, f_{OSC} = 640 ± 20kHz, V_{CC} = 2.1~3.5V, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
V _{OH}	High-level output voltage BUSY, DREQ/TEST	I _{OH} = 0.1mA	V _{CC} - 0.8			V
V _{OL}	Low-level output voltage BUSY, DREQ/TEST	I _{OL} = 0.4mA			0.8	V
I _{OH}	High-level output current DA ₀ , DA ₁	V _{OH} = V _{CC} - 0.4V	0.5			mA
I _{OL}	Low-level output current DA ₀ , DA ₁	V _{OL} = 0.4V	0.8			mA
I _{DA}	D/A output current	Resistor between DA ₀ and DA ₁ 10Ω at V _{CC} = 3V		20		mA
I _{OC}	Supply current (in operation)	Unloaded input/output			1	mA
I _{SC}	Supply current (standby)	Unloaded input/output			10	μA
I _{IL}	Input leak current	V _I = 0 ~ V _{CC}			±1	μA
R _I	Pull-up resistor $\overline{D_0} \sim \overline{D_2}$	V _I = V _{SS} , V _{CC} = 3V		320		kΩ
R _I	Pull-up resistor \overline{SYNC}	V _I = V _{SS} , V _{CC} = 3V		560		kΩ
V _{IH}	High-level input voltage		3/4 V _{CC}		V _{CC}	V
V _{IL}	Low-level input voltage		0		1/4 V _{CC}	V

M50805-XXXP,FP/M50806-XXXP,FP

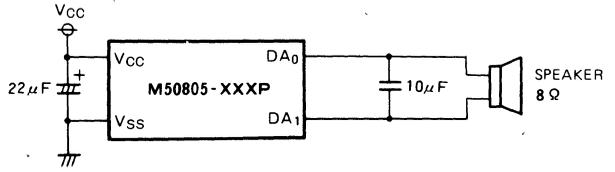
SINGLE-CHIP CMOS SPEECH SYNTHESIZER

APPLICATION CIRCUIT EXAMPLES

Speaker Drive Circuit

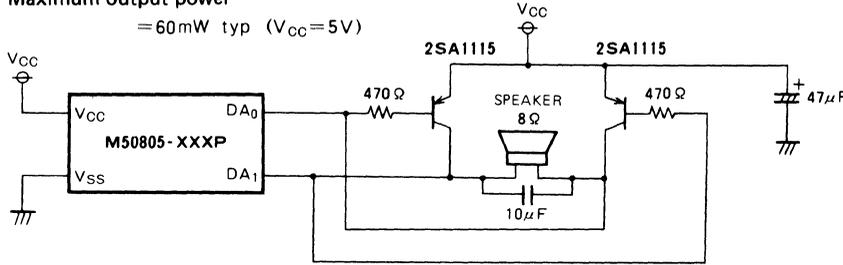
a) Direct drive

Maximum output power
= 10mW typ ($V_{CC}=5V$)

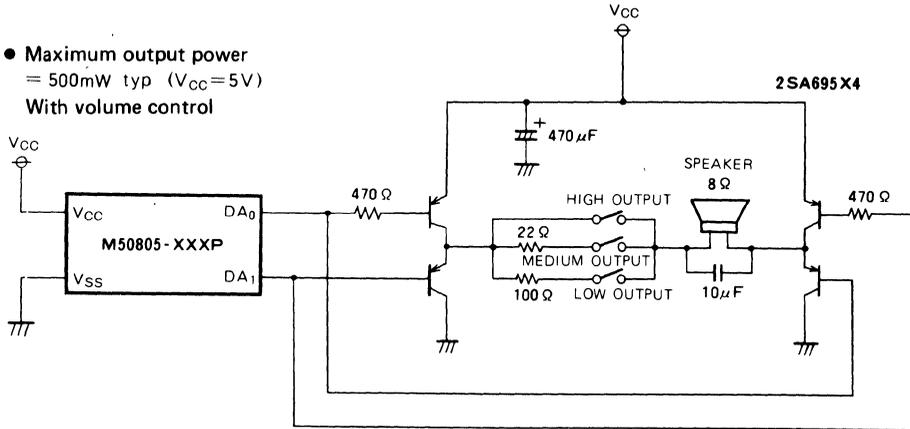


b) Using an amplifier

• Maximum output power
= 60mW typ ($V_{CC}=5V$)



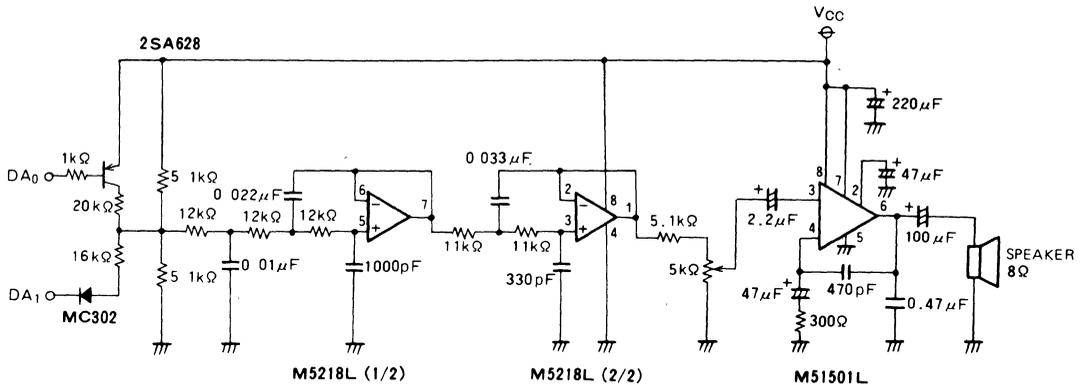
• Maximum output power
= 500mW typ ($V_{CC}=5V$)
With volume control



c) Filter amplifier circuit

Operational amplifier used as a filter amplifier circuit

Maximum output power = 200mW typ. ($V_{CC}=5V$)



METHOD OF ORDERING SPEECH ANALYSIS AND MASK ROM**Steps taken in ordering****(1) Decisions concerning specifications**

Decide the specifications concerning the content of the speech, the language, the speaker, etc.

(2) Ordering

When ordering, ensure that the necessary information has been entered in both the speech Analysis Request Form and Speech List. Be sure to also include the musical score if a melody is necessary.

(3) Recording

A recording of the original speech is made in a studio using an experienced narrator. Attendance is permitted for either confirmation or giving speech instructions.

(4) Analysis

Speech parameters of the voice are analyzed through use of a voice analysis system centered on a MELCOM 70/40 mini-computer in the Mitsubishi Speech Center. The data is written on a EPROM, and along with the Analysis Approval Form and Masking Approval Form is forwarded to the customer.

(5) Sound evaluation

The delivered EPROM is inserted in either Mitsubishi's voice evaluation board or a similar device and the quality of the synthesized speech is confirmed. If the quality is found to be acceptable, the Analysis Approval Form and the Masking Approval Form are to be signed and returned to Mitsubishi.

(6) Mask production

A MASK ROM is produced from the EPROM data. An ES sample for the confirmation of electrical characteristics is sent along with an ES Approval Form to the customer.

(7) Sample evaluation

If results of ES sample evaluation prove to be acceptable the ES Approval Form is to be signed and returned to Mitsubishi.

(8) CS sample production

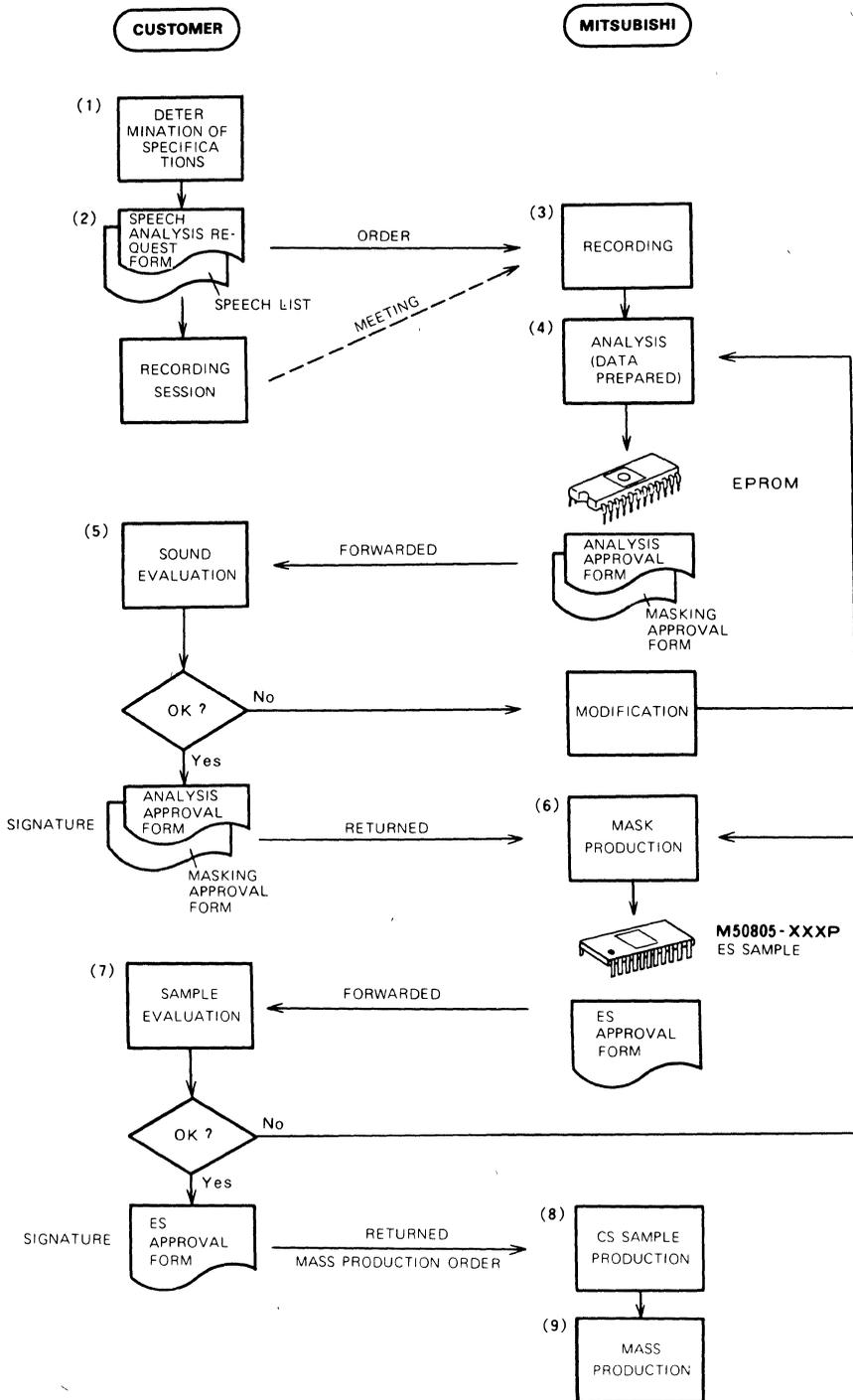
This is a sample having the same characteristics as the mass produced device.

(9) Mass production

After production of the CS sample mass production is begun.

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

ORDERING FLOW CHART



M50805-XXXP,FP/M50806-XXXP,FP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

MITSUBISHI SPEECH SYNTHESIZER LSI

M50805-XXXP, FP

M50806-XXXP, FP

SPEECH ANALYSIS REQUEST FORM

Customer Company name _____ Company address _____ Tel _____ Company contact _____ Date _____	Signature <hr/> Prepared <hr/> Approved
---	---

1. Specifications

Recording

(necessary/unnecessary)

_____ Language _____ Sex _____

Analysis

(necessary/unnecessary)

_____ Length of speech _____ seconds
 _____ Length of melody _____ seconds
 _____ Voice evaluation board (necessary/unnecessary)

Mask

(necessary/unnecessary)

2. Schedule

Recording → Analysis completion → Analysis and masking approval → ES → CS → MP

Desired date (/) (/) (/) (/) (/) (/)

_____ unit(s) _____ unit(s) _____ unit(s)/month

EPROM
 1 set

Lot _____ unit(s)

Mitsubishi Entry Column

Parameter	Type	Order number	Amount	Cost	Order date	Completion date	Transfer of money
Recording	VOICE-SOFT	00-38-					
Analysis	VOICE-SOFT	00-38-					
Voice evaluation board	VOICE-SOFT	00-38-					
Mask	VOICE-MASK						

Notes

M50805-XXXP,FP / M50806-XXXP,FP

SINGLE-CHIP CMOS SPEECH SYNTHESIZER

Mask ROM number	
-----------------	--

M50805-XXXP, FP
 M50806-XXXP, FP

SPEECH LIST

No.	Address			Speech content	Notes
	\overline{D}_2	\overline{D}_1	\overline{D}_0		
1	H	H	L		
2	H	L	H		
3	H	L	L		
4	L	H	H		
5	L	H	L		
6	L	L	H		
7	L	L	L		
8	H	H	H		

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